

SLES006A – SEPTEMBER 2001 – REVISED DECEMBER 2001

TRUE DIGITAL AUDIO AMPLIFIER

TAS5012 DIGITAL AUDIO PWM PROCESSOR

FEATURES

- TAS5012 + TAS5110 TDAA System – High Quality Digital Audio Amplification
- 102-dB Dynamic Range (TAS5012 Device)
- THD+N < 0.06%
- Power Efficiency Is 90% Into 8-Ω Load
- 16-, 20-, or 24-Bit Input Data
- 32-kHz, 44.1-kHz, 48-kHz, 88.2-kHz, 96-kHz, 176.4-kHz, 192-kHz Sampling Rates
- Economical 48-Pin TQFP Package
- Lower-Jitter Internal PLL

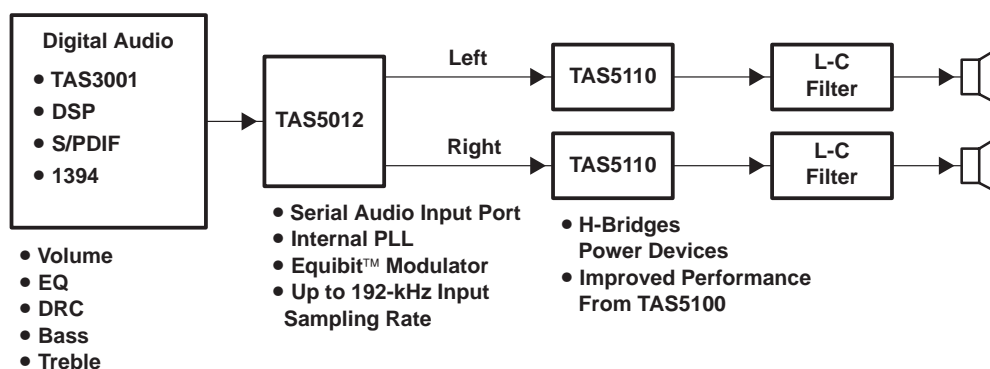
- 3.3-V Power Supply
- Mute
- Clicks and Pops Reduction (Patent Pending)

APPLICATIONS

- DVD-Audio
- Home Theater
- Car Audio Amplifiers and Head Units
- Internet Music Appliance
- Mini/Micro Component Systems

DESCRIPTION

True digital audio amplifier (TDAA) is a new paradigm in digital audio. One TDAA system consists of the TAS5012 PCM-PWM modulator device + TAS5110 PWM power output device. This system accepts a serial PCM digital audio stream and converts it to a 3.3-V PWM audio stream (TAS5012). The TAS5110 device then provides a large-signal PWM output. This digital PWM signal is then demodulated providing power output for driving loudspeakers. This patented technology provides low-cost, high-quality, high-efficiency digital audio applicable to many audio systems developed for the digital age. The TAS5012 is an innovative, cost-effective, high-performance 24-bit stereo PCM-PWM modulator based on Equibit™ technology. It has a wide variety of serial input options including right-justified (16, 20, or 24 bits), IIS (16, 20, or 24 bits), left-justified (16 bits), or DSP (16 bits) data formats. It is fully compatible with AES standard sampling rates of 32 kHz, 44.1 kHz, 48 kHz, 88.2 kHz, 96 kHz, 176.4 kHz, and 192 kHz. The TAS5012 also provides a de-emphasis function for 44.1-kHz and 48-kHz sampling rates.

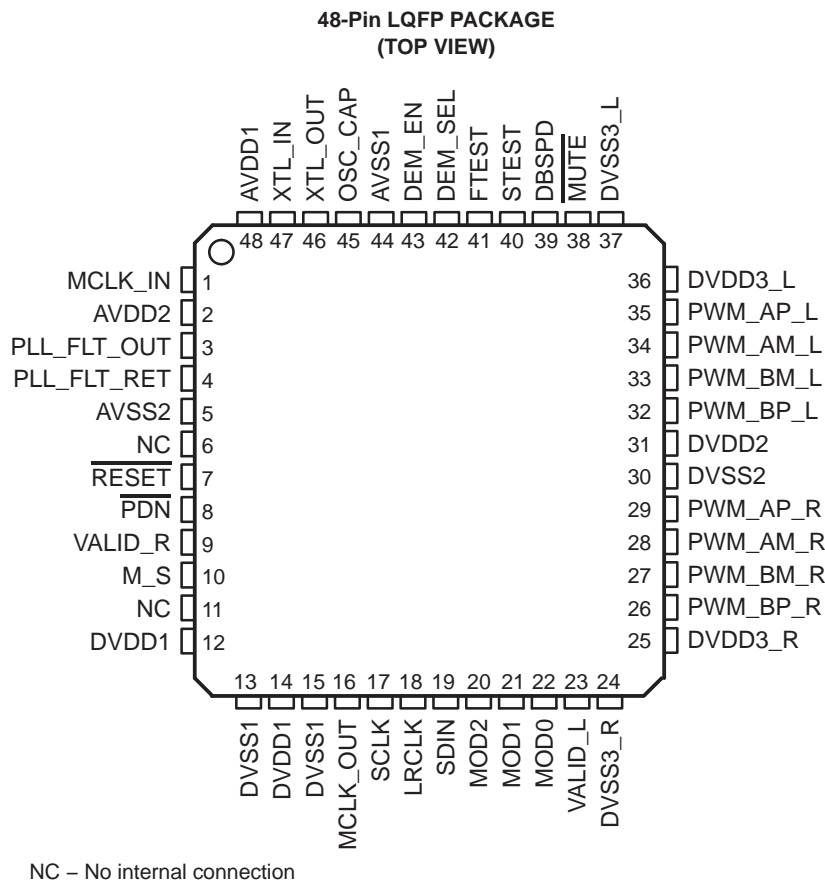


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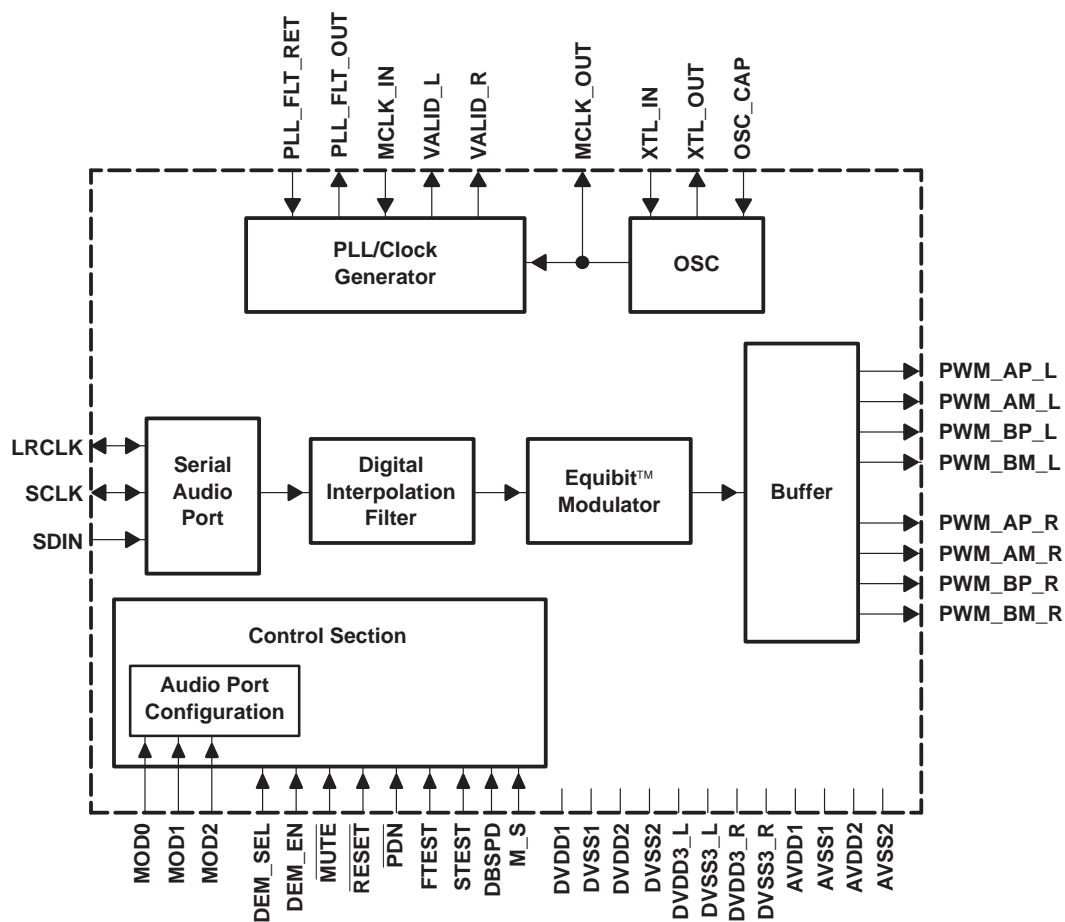
terminal assignments



references

- *True Digital Audio Amplifier TAS5100 PWM Power Output Stage* data sheet – Texas Instruments literature number SLLS419
- *Design Considerations for TAS5000/TAS5110 True Digital Audio Power Amplifiers* data sheet – Texas Instruments literature number SLAA117
- *Digital Audio Measurements* application note – Texas Instruments literature number SLAA114
- *PowerPAD™ Thermally Enhanced Package* application note – Texas Instruments literature number SLMA002

functional block diagram



ordering information

| T _A | PACKAGE |
|----------------|-------------|
| 0°C to 70°C | TAS5012PFB |
| -40°C to 85°C | TAS5012IPFB |

NOTE: These packages are available taped and reeled. Add R suffix to ordering number (e.g., TAS5012PFB R).

Terminal Functions

| TERMINAL NAME | NO. | I/O | DESCRIPTION |
|------------------|--------|-----|---|
| AVDD1 | 48 | I | Analog supply for oscillator |
| AVDD2 | 2 | I | Analog supply for PLL |
| AVSS1 | 44 | I | Analog ground for oscillator |
| AVSS2 | 5 | I | Analog ground for PLL |
| DBSPD | 39 | I | Indicates sample rate is double speed (88.2 kHz or 96 kHz), active high |
| DEM_EN | 43 | I | De-emphasis enable, active high |
| DEM_SEL | 42 | I | De-emphasis select (0 = 44.1 kHz, 1 = 48 kHz) |
| DVDD1 | 12, 14 | I | Digital voltage supply for logic |
| DVDD2 | 31 | I | Digital voltage supply for PWM reclocking |
| DVDD3_L | 36 | I | Digital voltage supply for PWM output (left) |
| DVDD3_R | 25 | I | Digital voltage supply for PWM output (right) |
| DVSS1 | 13, 15 | I | Digital ground for logic |
| DVSS2 | 30 | I | Digital ground for PWM reclocking |
| DVSS3_L | 37 | I | Digital ground for PWM output (left) |
| DVSS3_R | 24 | I | Digital ground for PWM output (right) |
| FTEST | 41 | I | Tied to DVSS1 for normal operation |
| LRCLK | 18 | I/O | Left/right clock (input when M_S = 0; output when M_S = 1) |
| MCLK_IN | 1 | I | MCLK input |
| MCLK_OUT | 16 | O | Buffered system clock output if M_S = 1; otherwise set to 0 |
| MOD0 | 22 | I | Serial interface selection pin, bit 0 |
| MOD1 | 21 | I | Serial interface selection pin, bit 1 |
| MOD2 | 20 | I | Serial interface selection pin, bit 2 (MSB) |
| M_S | 10 | I | Master/slave, master=1, slave=0 |
| MUTE | 38 | I | Muted signal = 0, normal mode = 1 |
| NC | 6, 11 | | No connection |
| OSC_CAP | 45 | I | Oscillator cap return |
| PDN | 8 | I | Power down, active low |
| PLL_FLT_OUT | 3 | O | Output terminal for external PLL filter |
| PLL_FLT_RET | 4 | I | Return for external PLL filter |
| PWM_AM_L | 34 | O | PWM left output A (differential –) |
| PWM_AM_R | 28 | O | PWM right output A (differential –) |
| PWM_AP_L | 35 | O | PWM left output A (differential +) |
| PWM_AP_R | 29 | O | PWM right output A (differential +) |
| PWM_BM_L | 33 | O | PWM left output B (differential –) |
| PWM_BM_R | 27 | O | PWM right output B (differential –) |
| PWM_BP_L | 32 | O | PWM left output B (differential +) |
| PWM_BP_R | 26 | O | PWM right output B (differential +) |
| RESET | 7 | I | Reset (active low) |
| SCLK | 17 | I/O | Shift clock (input when M_S = 0, output when M_S = 1) |
| SDIN | 19 | I | Stereo serial audio data input |
| STEST | 40 | I | Tied to DVSS1 for normal operation |
| VALID_L | 23 | O | PWM left outputs valid (active high) |
| VALID_R | 9 | O | PWM right outputs valid (active high) |
| XTL_IN | 47 | I | Crystal or clock input (MCLK input) |
| XTL_OUT | 46 | O | Crystal output (not for external usage). NC when XTL_IN is MCLK input |

functional description

serial audio port

The serial audio port consists of a shift clock (SCLK pin), a left/right frame synchronization clock (LRCLK pin), and a data input (SDIN pin). The serial audio port supports standard serial PCM formats ($F_s = 32$ kHz, 44.1 kHz, 48 kHz, 88.2 kHz, 96 kHz, 176.4 kHz, or 192 kHz) stereo. See the *serial interface formats* section.

system clocks—master mode and slave mode

The TAS5012 allows multiple system clocking schemes. In this document, master mode indicates that the TAS5012 provides system clocks to other parts of the system ($M_S=1$). Audio system clocks of frequency 256 F_s MCLK_OUT, 64 F_s SCLK, and F_s LRCLK are output from this device when it is configured in master mode. Slave mode indicates that a system master other than the TAS5012 provides system clocks (LRCLK, SCLK, and MCLK_IN) to the TAS5012 ($M_S = 0$). The TAS5012 operates with LRCLK and SCLK synchronized to MCLK. TAS5012 does not require any specific phase relationship between LRCLK and MCLK, but there must be synchronization. In the slave mode MCLK_OUT is driven low. Table 1 shows all the possible master and slave modes. When operating in quad mode ($F_s = 176.4$ kHz or 192 kHz), the device works in slave mode only with MCLK_IN = 128 F_s .

oscillator/sampling frequency

The sampling frequency is determined by the crystal (master mode) or master clock in (slave mode) which should be either 8.192 MHz ($F_s = 32$ kHz), 11.2896 MHz ($F_s = 44.1$ kHz), or 12.288 MHz ($F_s = 48$ kHz). Twice the normal sampling frequency can be selected by using the DBSPD pin which allows usage of $F_s = 88.2$ kHz or $F_s = 96$ kHz. In the double-speed slave mode (DBSPD = 1, $M_S = 0$), the external clock input is either 22.5796 MHz ($F_s = 88.2$ kHz) or 24.576 MHz ($F_s = 96$ kHz). Note that 32-kHz sampling is supported in the normal speed modes. Table 1 explains the proper clock selection.

Table 1. Oscillator, External Clock, and PLL Functions

| DESCRIPTION | M_S | DBSPD | XTL_IN (MHz) [†] | MCLK_IN (MHz) [‡] | SCLK (MHz) [¶] | LRCLK (kHz) [¶] | MCLK_OUT (MHz) [#] |
|---------------------------------|-----|-------|------------------------------|-------------------------------|----------------------------|-----------------------------|--------------------------------|
| Master, normal speed | 1 | 0 | 8.192 | — | 2.048 | 32 | 8.192 |
| Master, normal speed | 1 | 0 | 11.2896 | — | 2.8224 | 44.1 | 11.2896 |
| Master, normal speed | 1 | 0 | 12.288 | — | 3.072 | 48 | 12.288 |
| Master, double speed | 1 | 1 | — | 22.5792 [§] | 5.6448 | 88.2 | 22.5792 |
| Master, double speed | 1 | 1 | — | 24.576 [§] | 6.144 | 96 | 24.576 |
| Slave, normal speed | 0 | 0 | — | 8.192 [§] | 2.048 | 32 | Digital GND |
| Slave, normal speed | 0 | 0 | — | 11.2896 [§] | 2.8224 | 44.1 | Digital GND |
| Slave, normal speed | 0 | 0 | — | 12.288 [§] | 3.072 | 48 | Digital GND |
| Slave, double speed | 0 | 1 | — | 22.5792 [§] | 5.6448 | 88.2 | Digital GND |
| Slave, double speed | 0 | 1 | — | 24.576 [§] | 6.144 | 96 | Digital GND |
| Slave, quad speed | 0 | 0 | — | 22.5792 [§] | 11.2896 | 176.4 | Digital GND |
| Slave, quad speed | 0 | 0 | — | 24.576 [§] | 12.288 | 192 | Digital GND |

[†] Either a crystal oscillator or an external clock of the specified frequency can be connected to XTL_IN.

[‡] MCLK_IN tied low when input to XTL_IN is provided; XTL_IN tied low when MCLK_IN is provided.

[§] External MCLK connected to MCLK_IN input

[¶] SCLK and LRCLK are outputs when $M_S = 1$, inputs when $M_S = 0$.

[#] MCLK_OUT is driven low when $M_S = 0$.

^{||} Quad speed mode is detected automatically. when DBSPD = 0.

phase-locked loop (PLL)/clock generation

A low jitter PLL is incorporated for internal use. Connections for the PLL external loop filter are provided as PLL_FLT_RET and PLL_FLT_OUT. If the PLL loses lock, the PWM output status pins (VALID_L and VALID_R) go low. Note that VALID_L and VALID_R can go low for other conditions as well. See the *error status reporting* section.

functional description (continued)

digital interpolation filter

The 24-bit high-performance linear phase FIR interpolation filter up-samples the input digital data at a rate of two times (quad speed mode = 176.4 kHz or 192 kHz), four times (double speed mode = 88.2 kHz or 96 kHz), or eight times (normal mode = 32 kHz, 44.1 kHz, or 48 kHz) the incoming sample rate. This filter provides very low pass-band ripple and optimized time domain transient response for accurate music reproduction.

digital PWM modulator

The interpolation filter output is sent to the modulator. This modulator consists of a high performance fourth order digital noise shaper and a PCM-to-PWM converter. Following the noise shaper, the PCM signal is fed into a very low distortion PCM-to-PWM conversion block, buffered, and output from the chip. The modulation scheme is based on a 2-state control of the H-bridge output.

control, status, and operational modes

The TAS5012 control section consists of several control-input pins. Three serial mode pins (MOD0, MOD1, and MOD2) are provided to select various serial data formats. During normal operating conditions if any of the MOD0, MOD1, or MOD2 pins changes state, a reset sequence is initiated. Also provided are separate power-down ($\overline{\text{PDN}}$), reset ($\overline{\text{RESET}}$), and mute ($\overline{\text{MUTE}}$) pins.

power up

At power up the $\overline{\text{VALID_L}}$ and $\overline{\text{VALID_R}}$ pins are asserted low and the PWM outputs go to the hard mute state in which the P outputs are held low and the M outputs are held high. Following initialization, the TAS5012 comes up in the operational state (differential PWM audio). There are two cases of power-up timing. The first case is shown in Figure 1 with $\overline{\text{RESET}}$ preceding $\overline{\text{PDN}}$. The second case is shown in Figure 2 with $\overline{\text{PDN}}$ preceding $\overline{\text{RESET}}$.

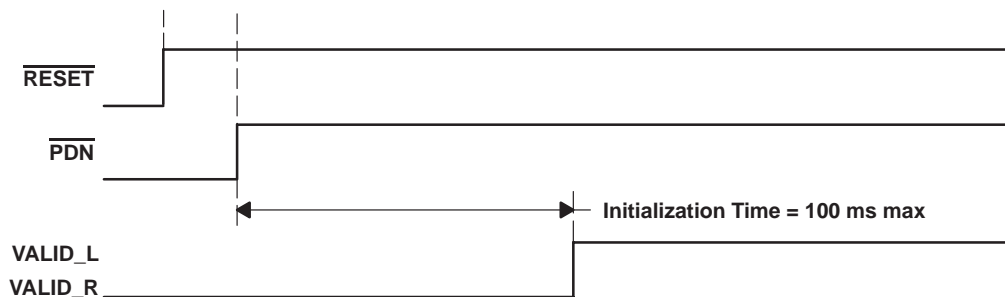


Figure 1. Power-Up Timing ($\overline{\text{RESET}}$ Preceding $\overline{\text{PDN}}$)

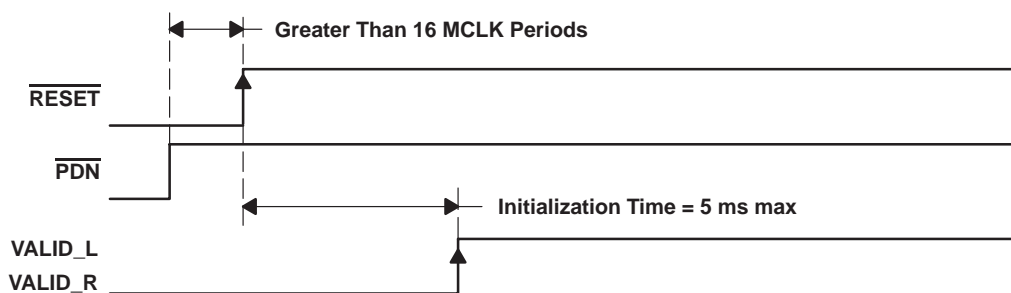


Figure 2. Power-Up Timing ($\overline{\text{PDN}}$ Preceding $\overline{\text{RESET}}$)

functional description (continued)

reset

The reset signal for the TAS5012 must be applied whenever toggling the $\overline{\text{M_S}}$, DBSPD signal. This reset is asynchronous. See Figure 3 for reset timing. To initiate the reset sequence the $\overline{\text{RESET}}$ pin is asserted low. As long as the pin is held low, the chip is in the reset state. During this reset time the PWM outputs are hard-muted (P-outputs held low and M-outputs held high) and the PWM outputs valid pins ($\overline{\text{VALID_L}}$, $\overline{\text{VALID_R}}$) are held low. Assuming $\overline{\text{PDN}}$ is high, the rising edge of the reset pulse begins chip initialization. After the initialization time, the TAS5012 begins normal operation.

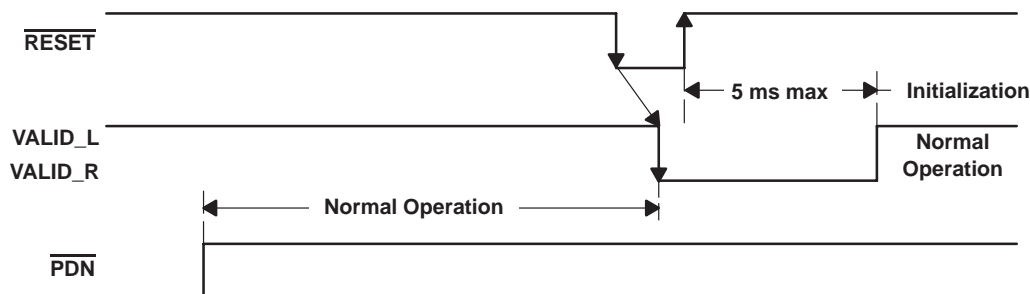


Figure 3. Reset Timing

power down

When $\overline{\text{PDN}}$ is low (see Figure 4), both the PLL and the oscillator are shut down. Note that power down is an asynchronous operation. To place the device in total power-down mode, both $\overline{\text{RESET}}$ and $\overline{\text{PDN}}$ must be held low. As long as these pins are held low, the chip is in the power-down state and the PWM outputs are hard muted with the P outputs held low and the M outputs held high. To place the device back into normal mode, see the *power up* section.

NOTE: In order for the dynamic logic to be properly powered down, the clocks should not be stopped before the $\overline{\text{PDN}}$ pin goes low. Otherwise, the device may drain additional supply current.

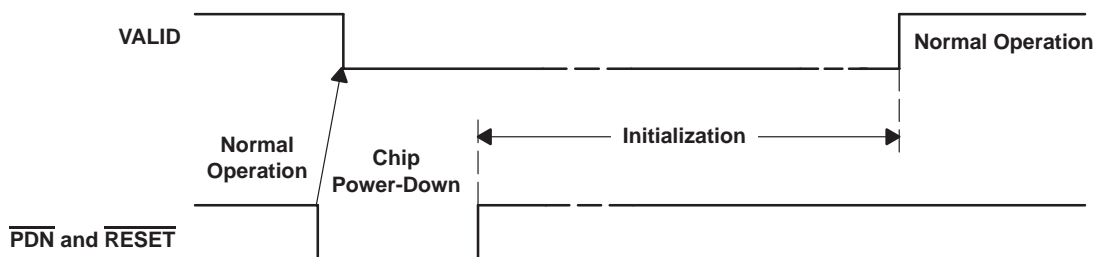


Figure 4. Power-Down Timing

mute

The TAS5012 provides a mute function that is used when the $\overline{\text{MUTE}}$ pin is asserted low. See Table 2 for mute description. This mute is a quiet mute; that is, the mute is accomplished by outputting a zero value waveform in which both sides of the differential PWM outputs have a 50% duty cycle (see Figure 5 for mute timing).

Table 2. Mute Description

| $\overline{\text{MUTE}}$ | P OUTPUTS | M OUTPUTS | DESCRIPTION |
|--------------------------|----------------|--------------------------|------------------|
| 0 | 50% duty cycle | 50% duty cycle | Mute |
| 1 | DATA | $\overline{\text{DATA}}$ | Normal operation |

functional description (continued)

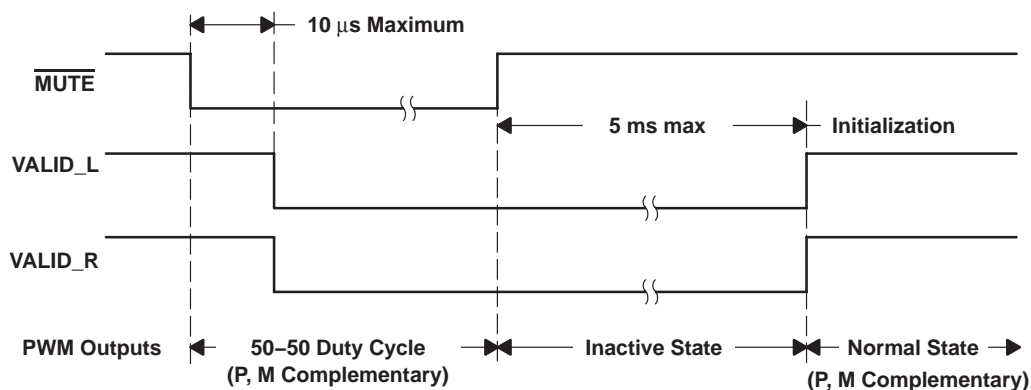


Figure 5. Mute Timing

double speed

Double-speed mode is used to support sampling rates of 88.2 kHz and 96 kHz. In order to put the TAS5012 in double-speed mode with the device in normal operating conditions, the $\overline{\text{RESET}}$ pin must be held low while switching the DBSPD pin high. After the $\overline{\text{RESET}}$ pin is brought high again, a reset sequence takes place. If the change is at power up, a power-up sequence is originated.

quad speed

Quad-speed mode is used to support sampling rates of 176.4 kHz and 192 kHz. It is supported in slave mode only. In order to put the TAS5012 in quad-speed mode, M_S and DBPSB pins are brought low. Quad-speed mode is then automatically detected due to the fact that it is the only mode in which MCLK_IN is 128Fs. DEM_SEL must be set to low when operating in the quad-speed mode.

de-emphasis filter

For audio sources that have been preemphasized, a precision 50-μs/15-μs de-emphasis filter is provided to support the sampling rates of 44.1 kHz and 48 kHz. Pins DEM_SEL and DEM_EN select the de-emphasis functions. See Figure 6 for a graph showing the de-emphasis filtering characteristics. See Table 3 for de-emphasis selection.

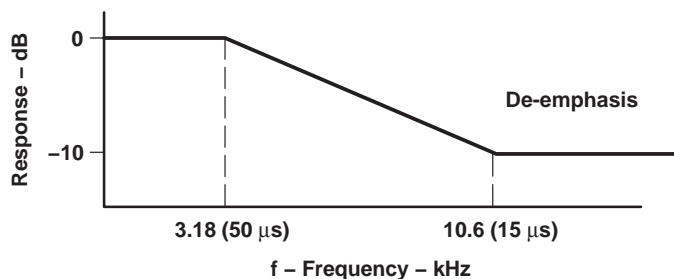


Figure 6. De-Emphasis Filter Characteristics

functional description (continued)**de-emphasis selection**

De-emphasis selection is accomplished by using the DEM_SEL and DEM_EN pins. See Table 3 for de-emphasis selection description.

Table 3. De-Emphasis Selection

| DEM_SEL | DEM_EN | DESCRIPTION |
|---------|--------|--|
| 0 | 0 | De-emphasis disabled |
| 0 | 1 | De-emphasis enabled for $F_s = 44.1$ kHz |
| 1 | 1 | De-emphasis enabled for $F_s = 48$ kHz |
| 1 | 0 | Forbidden state. Do not use. |

error status reporting (VALID_L and VALID_R)

The following is a list of the error conditions that will cause the VALID_L and VALID_R pins to be asserted low:

- No clocks
- Clock phase errors

When either of the above conditions is met, the VALID_L and VALID_R goes low and the PWM outputs go to the hard mute state. If the error condition is removed, the TAS5012 is reinitialized and the VALID_L and VALID_R pins are asserted high.

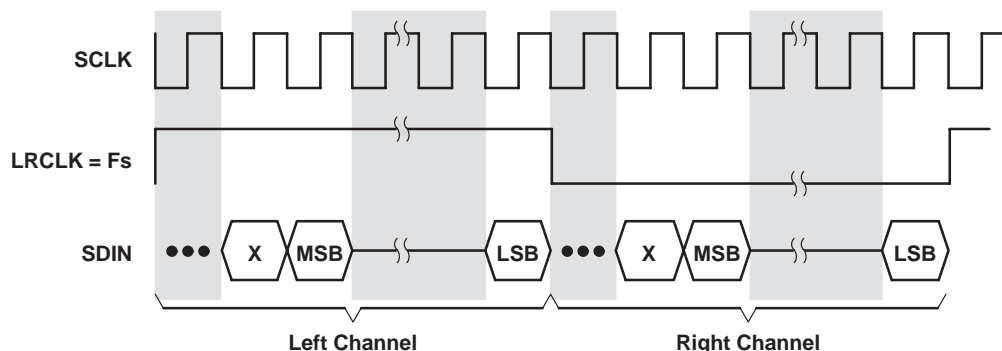
serial interface formats

The TAS5012 is compatible with eight different serial interfaces. Available interface options are IIS, right justified, left justified, and DSP frame. Table 4 indicates how these options are selected using the MOD0, MOD1, and MOD2 pins.

Table 4. Hardware Selection of Serial Audio Modes

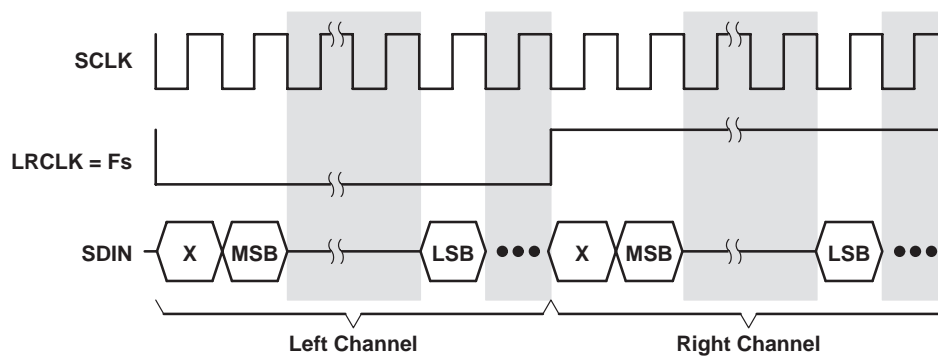
| MODE | MOD2 PIN | MOD1 PIN | MOD0 PIN | SERIAL INTERFACE SDIN |
|------|----------|----------|----------|------------------------------------|
| 0 | 0 | 0 | 0 | 16 bit, MSB first; right justified |
| 1 | 0 | 0 | 1 | 20 bit, MSB first; right justified |
| 2 | 0 | 1 | 0 | 24 bit, MSB first; right justified |
| 3 | 0 | 1 | 1 | 16 bit IIS |
| 4 | 1 | 0 | 0 | 20 bit IIS |
| 5 | 1 | 0 | 1 | 24 bit IIS |
| 6 | 1 | 1 | 0 | 16 bit MSB first, left justified |
| 7 | 1 | 1 | 1 | 16 bit DSP frame |

The following figures illustrate the relationship between the SCLK, LRCLK and the serial data I/O for the different interface protocols. Note that there are always 64 SCLKs per LRCLK. The nondata bits are padded with binary 0s.

functional description (continued)**MSB first, right-justified (for 16, 20, 24 bits)****Figure 7. MSB First Right Justified**

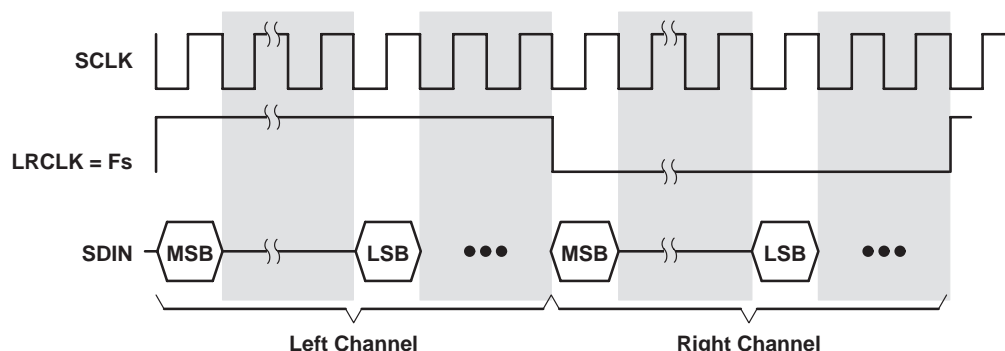
Note the following characteristics of this protocol:

- Left channel is received when LRCLK is high.
- Right channel is received when LRCLK is low.
- SDIN is sampled at the rising edge of SCLK.

IIS compatible serial format (for 16, 20, 24 bits)**Figure 8. IIS Compatible Serial Format**

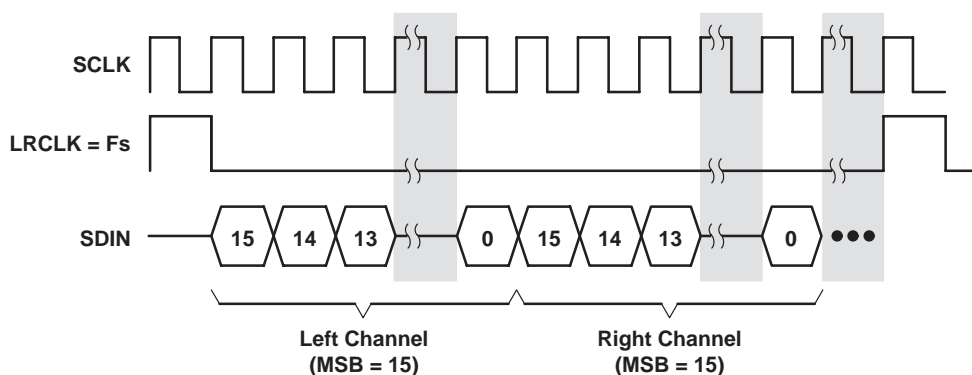
Note the following characteristics of this protocol:

- Left channel is received when LRCLK is low.
- Right channel is received when LRCLK is high.
- SDIN is sampled with the rising edge of the SCLK.

functional description (continued)**MSB left-justified serial interface format (for 16 bits)****Figure 9. MSB Left-Justified Serial Interface Format**

Note the following characteristics of this protocol:

- Left channel is received when LRCLK is high.
- Right channel is received when LRCLK is low.
- SDIN is sampled at the rising edge of SCLK.

DSP compatible serial interface format (for 16 bits)**Figure 10. DSP Compatible Serial Interface Format**

Note the following characteristic of this protocol:

- Serial data is sampled with the falling edge of SCLK.

PWM Outputs

Designed to be used with the TAS5110 family of H-bridges, the PWM outputs provide differential 3.3-V square-wave signals. During normal operation these outputs represent the input PCM audio in the pulse-width modulation scheme. In the hard-mute state the P outputs (PWM_AP_L, PWM_BP_L, PWM_AP_R, and PWM_BP_R) are held low and the M outputs (PWM_AM_L, PWM_BM_L, PWM_AM_R, and PWM_BM_R) are held high. In the quiet-mute state the differential PWM outputs have a 50% duty cycle.

absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

| | |
|--|-------------------------|
| Analog supply voltage range, AVDD1, AVDD2 | –0.3 V to 4.2 V |
| Digital power supply voltage, DVDD1, DVDD2, DVDD3_L, DVDD3_R | –0.3 V to 4.2 V |
| Digital input voltage, V _I (see Note 1) | –0.3 V to DVDDX + 0.3 V |
| Operating free-air temperature, T _A | 0°C to 70°C |
| Storage temperature, T _{stg} | –65°C to 150°C |
| ESD | 2000 V |

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: DVDD1, DVDD2, DVDD3_L, DVDD3_R.

recommended operating conditions, T_A = 25°C, DVDD1 = DVDD2 = DVDD3_L = DVDD3_R = 3.3 V ±10%, AVDD1 = AVDD2 = 3.3 V ±10%, F_s = 44.1 kHz

| | | | MIN | TYP | MAX | UNIT |
|-------------------|---------|-------------|-----|------|-----|------|
| Supply voltage | Digital | DVDDX‡ | 3 | 3.3 | 3.6 | V |
| Supply current | Digital | Operating | | 22 | | mA |
| | | Power down§ | | 10 | 20 | µA |
| Power dissipation | Digital | Operating | | 59.4 | | mW |
| | | Power down§ | | 6.6 | 72 | µW |
| Supply voltage | Analog | AVDDX¶ | 3 | 3.3 | 3.6 | V |
| Supply current | Analog | Operating | | 8 | | mA |
| | | Power down§ | | 10 | 100 | µA |
| Power dissipation | Analog | Operating | | 26.4 | | mW |
| | | Power down§ | | 33 | 360 | µW |

‡ DVDD1, DVDD2, DVDD3_L, DVDD3_R

§ If the clocks are turned off

¶ AVDD1, AVDD2

electrical characteristics, T_A = 25°C, DVDD1 = DVDD2 = DVDD3_L = DVDD3_R = 3.3 V ±10%, AVDD1 = AVDD2 = 3.3 V ±10%**static digital specifications**

| | MIN | MAX | UNIT |
|---|-----|-------|------|
| V _{IH} High-level input voltage | 2 | DVDD1 | V |
| V _{IL} Low-level input voltage | 0 | 0.8 | V |
| V _{OH} High-level output voltage, (I _O = –1 mA) | 2.4 | | V |
| V _{OL} Low-level output voltage, (I _O = 4 mA) | | 0.4 | V |
| Input leakage current | –10 | 10 | µA |

digital interpolation filter and PWM modulator, F_s = 44.1 kHz

| | MIN | TYP | MAX | UNIT |
|---|-----|--------|-----|------|
| Pass band | 0 | | 20 | kHz |
| Pass-band ripple | | ±0.012 | | dB |
| Stop band | | 24.1 | | kHz |
| Stop-band attenuation (24.1 kHz to 152.3 kHz) | 50 | | | dB |
| Group delay | | 700 | | µS |
| PWM modulation index (gain) | | 0.93 | | |

TAS5012/TAS5110 system performance measured at the speaker terminals

See application note SLAA117.

switching characteristics,

$T_A = 25^\circ\text{C}$, $DVDD1 = DVDD2 = DVDD3_L = DVDD3_R = AVDD1 = AVDD2 = 10\% \ 3.3 \text{ V} \pm 10\%$

serial audio ports slave mode

| PARAMETER | | MIN | TYP | MAX | UNIT |
|-------------------------------|--|-----|-----|--------|------|
| $f(\text{SCLK})$ | SCLK frequency | | | 12.288 | MHz |
| $t_{\text{su}}(\text{SDIN})$ | SDIN setup time before SCLK rising edge | 20 | | | ns |
| $t_{\text{h}}(\text{SDIN})$ | SDIN hold time from SCLK rising edge | 10 | | | ns |
| $f(\text{LRCLK})$ | LRCLK frequency | 32 | 48 | 192 | kHz |
| | MCLK duty cycle | | 50% | | |
| | SCLK duty cycle | | 50% | | |
| | LRCLK duty cycle | | 50% | | |
| $t_{\text{su}}(\text{LRCLK})$ | LRCLK edge setup before SCLK rising edge | 20 | | | ns |

serial audio ports master mode, load conditions = 50 pF

| PARAMETER | | MIN | TYP | MAX | UNIT |
|------------------|---------------|-----|-----|-----|------|
| $t(\text{MSD})$ | MCLK to SCLK | 0 | | 5 | ns |
| $t(\text{MLRD})$ | MLCK to LRCLK | 0 | | 5 | ns |

DSP serial interface mode

| PARAMETER | | MIN | TYP | MAX | UNIT |
|--|--|-----|---------------------|--------|------|
| $f(\text{SCLK})$ | SCLK frequency | | | 12.288 | MHz |
| $t_{\text{w}}(\text{FSHIGH})$ | Pulse duration, sync | | $1/(64 \times F_s)$ | | ns |
| $t_{\text{su}}(\text{SDIN}),$ $t_{\text{su}}(\text{LRCLK})$ | SDIN and LRCLK setup time before SCLK falling edge | 20 | | | ns |
| $t_{\text{h}}(\text{SDIN}),$ $t_{\text{h}}(\text{LRCLK})$ | SDIN and LRCLK hold time from SCLK falling edge | 10 | | | ns |
| | SCLK duty cycle | | 50% | | |

PARAMETER MEASUREMENT INFORMATION

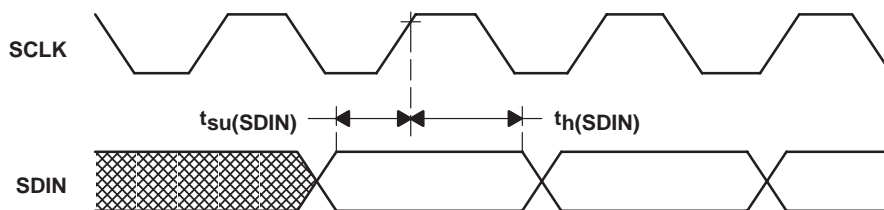
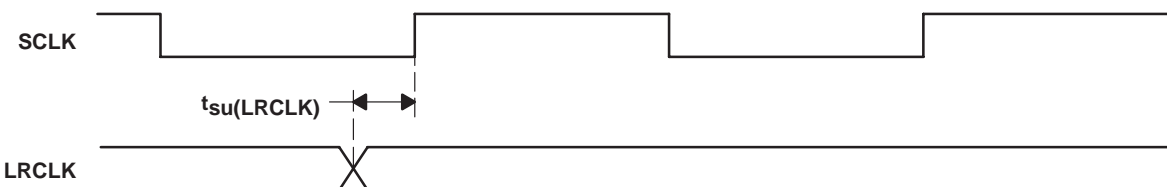


Figure 11. Right-Justified, IIS, Left-Justified Serial Protocol Timing



NOTE: Serial data is sampled with the rising edge of SCLK (setup time = 20 ns and hold time = 10 ns)

Figure 12. Right, Left, and IIS Serial Mode Timing Requirement

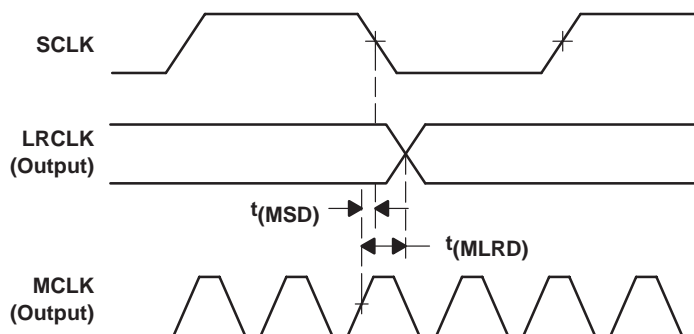


Figure 13. Serial Audio Ports Master Mode Timing

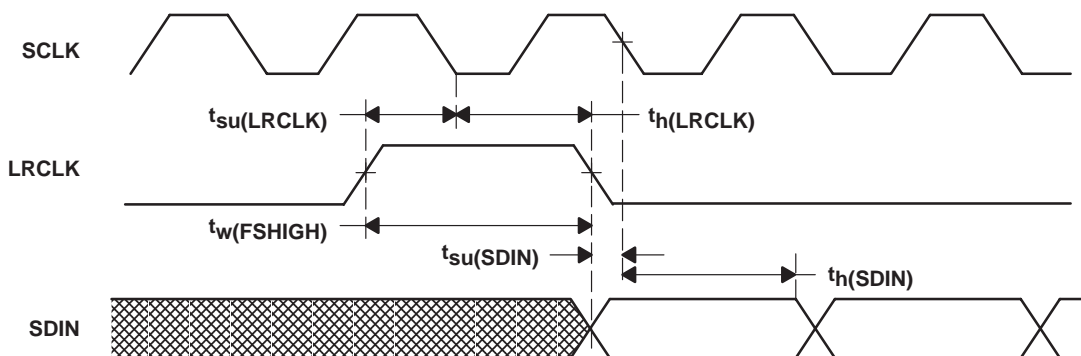


Figure 14. DSP Serial Port Timing

PARAMETER MEASUREMENT INFORMATION

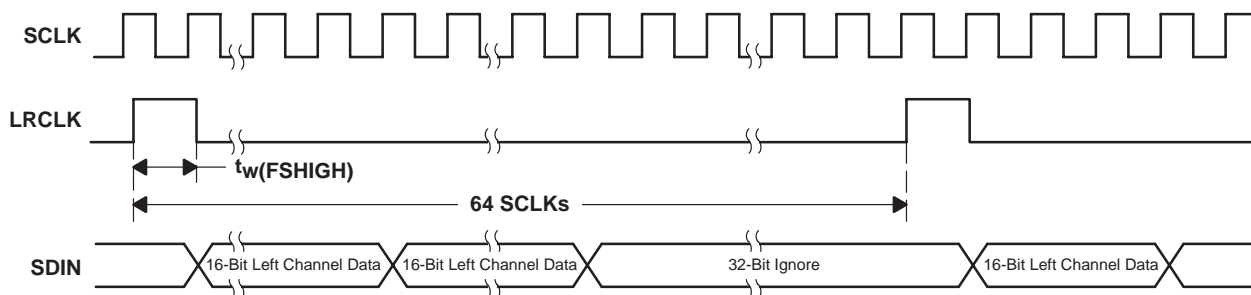
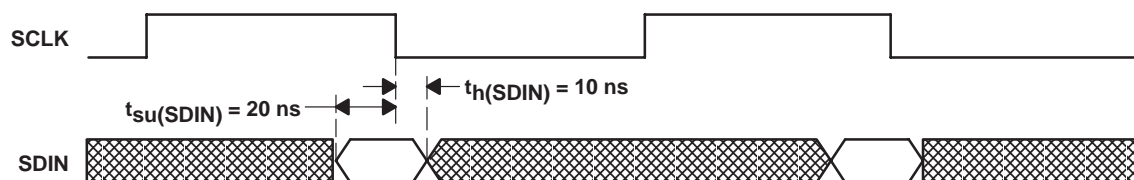


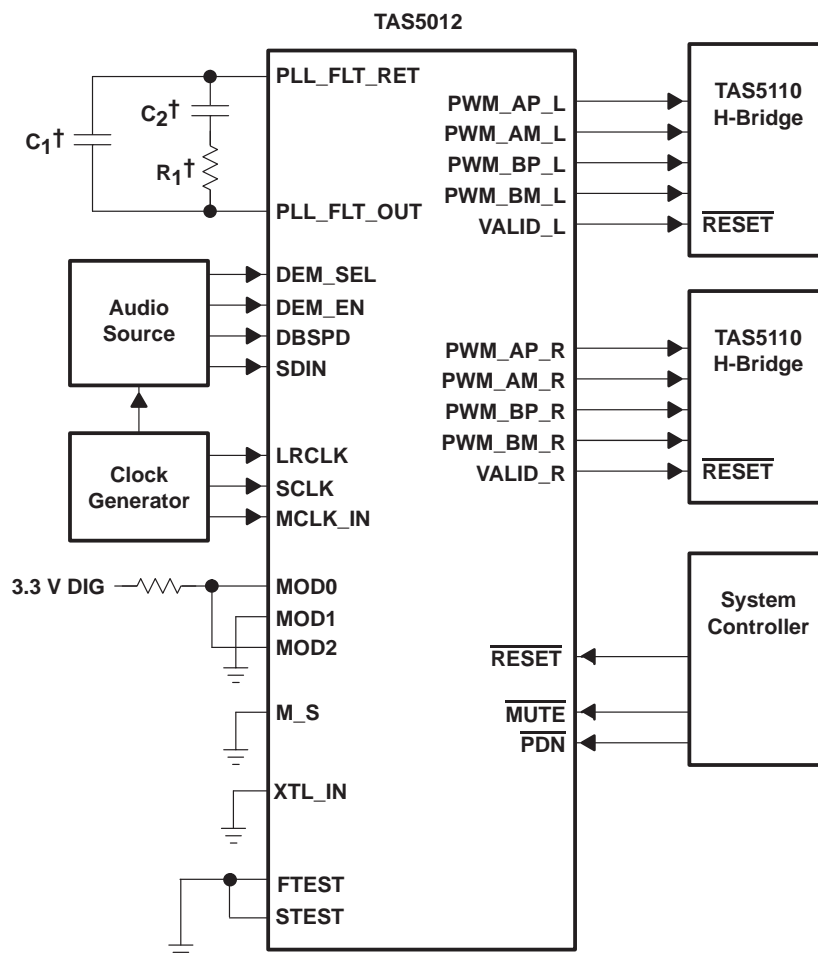
Figure 15. DSP Serial Port Expanded Timing



NOTE: Serial data is sampled with the falling edge of SCLK (setup time = 20 ns and hold time = 10 ns)

Figure 16. DSP Absolute Timing Requirement

APPLICATION INFORMATION

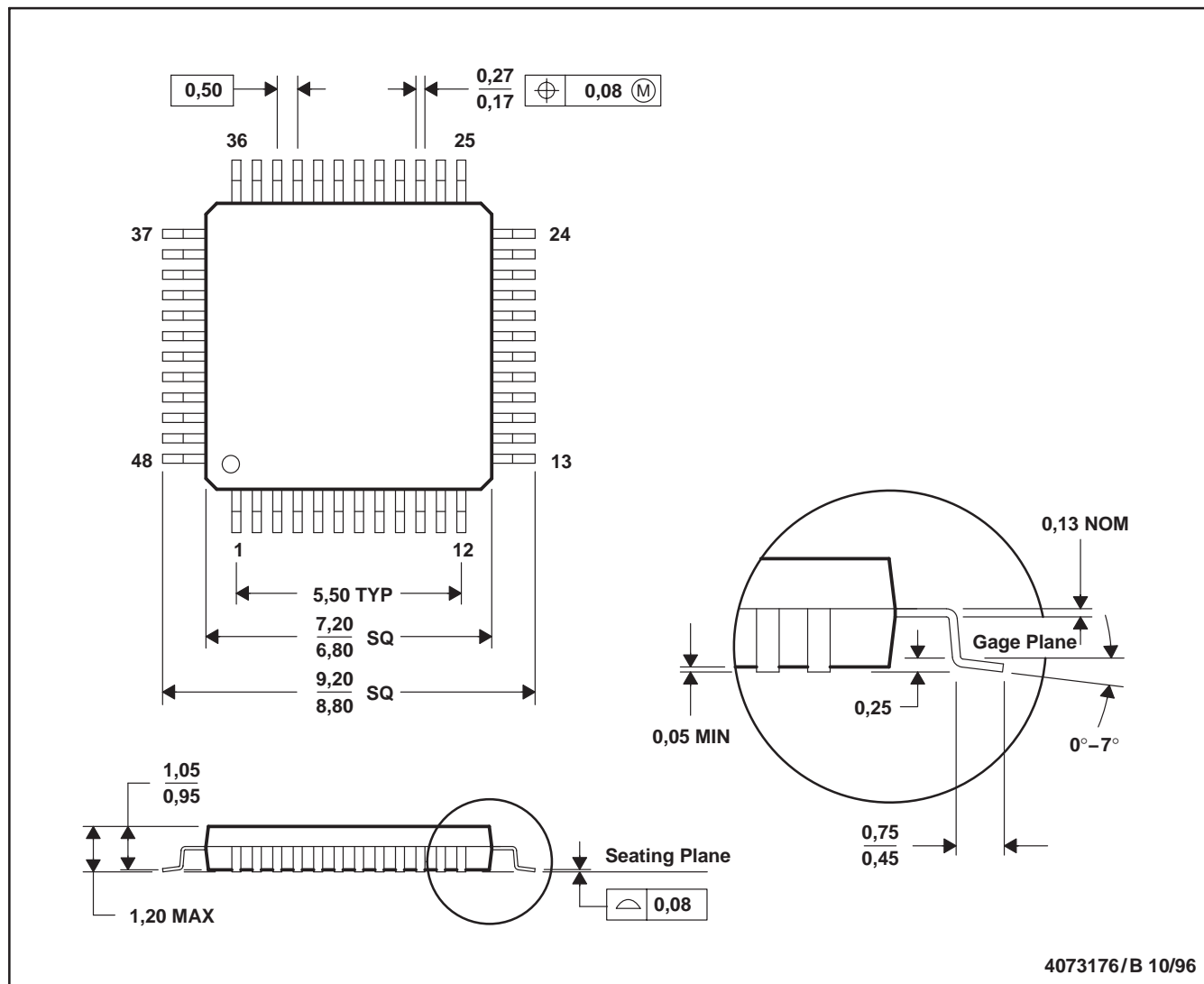


† See application note SLAA117 for values.

MECHANICAL DATA

PFB (S-PQFP-G48)

PLASTIC QUAD FLATPACK



- NOTES: A. All linear dimensions are in millimeters.
 B. This drawing is subject to change without notice.
 C. Falls within JEDEC MS-026

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