4-channel H-bridge type BTL driver for CD players **BA5950FP**

The BA5950FP is a 4-channel H-bridge BTL power driver for CD players. Input is PWM, and gain and the filter constant can be changed with an attached resistor and capacitor.

Applications

CD players, CD-ROM drives and other optical disc devices

Features

- 1) 4-channel BTL driver on a HSOP 28-pin power package, allowing for application miniaturization.
- 2) Direct PWM input.

- Filter constants can be changed with an attached resistor and capacitor.
- Internal thermal shutdown circuit with hysteresis capabilities.
- 5) Internal mute circuit.

● Absolute maximum ratings (Ta = 25°C)

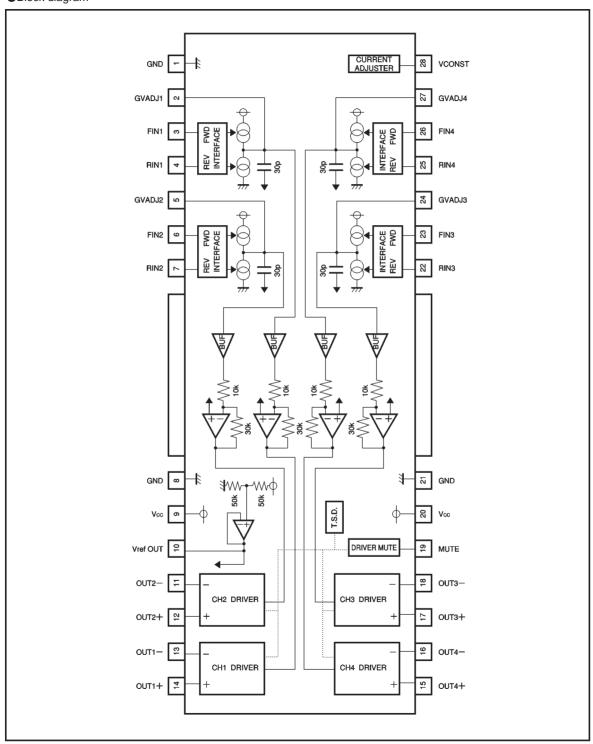
Parameter	Symbol	Limits	Unit
Power supply voltage	Vcc	18	V
Power dissipation	Pd	1800*	mW
Operating temperature	Topr	−35~+85	°C
Storage temperature	Tstg	− 55∼ + 150	°C

^{*} Reduced by 14.4 mW for ach increase in Ta of 1°C over 25°C. When mounted on a $70 \times 70 \times 1.6$ mm glass epoxy board.

● Recommended operating conditions (Ta = 25°C)

Parameter	Symbol	Min.	Тур.	Max.	Unit
Power supply voltage	Vcc	4.5	_	13.5	٧

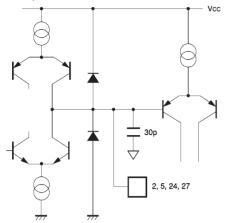
Block diagram

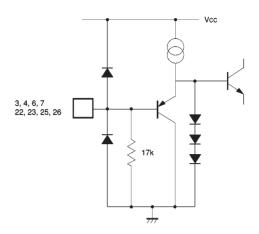


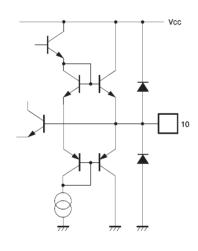
Pin descriptions

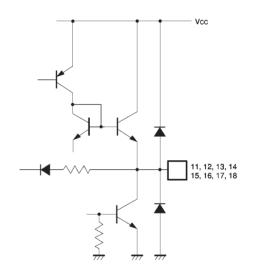
Pin No.	Pin name	Function					
1	GND	Ground					
2	GVADJ1	Channel 1 gain adjustment					
3	FIN1	Channel 1 forward input					
4	RIN1	Channel 1 reverse input					
5	GVADJ2	Channel 2 gain adjustment					
6	FIN2	Channel 2 forward input					
7	RIN2	Channel 2 reverse input					
8	GND	Substrate ground					
9	Vcc	Vcc					
10	Vref OUT	Reference voltage output					
11	OUT2-	Channel 2 negative output					
12	OUT2+	Channel 2 positive output					
13	OUT1-	Channel 1 negative output					
14	OUT1+	Channel 1 positive output					
15	OUT4+	Channel 4 positive output					
16	OUT4-	Channel 4 negative output					
17	OUT3+	Channel 3 positive output					
18	OUT3-	Channel 3 negative output					
19	MUTE	Mute					
20	Vcc	Vcc					
21	GND	Substrate ground					
22	RIN3	Channel 3 reverse input					
23	FIN3	Channel 3 forward input					
24	GVADJ3	Channel 3 gain adjustment					
25	RIN4	Channel 4 reverse input					
26	FIN4	Channel 4 forward input					
27	GVADJ4	Channel 4 gain adjustment					
28	VCONST	Output of constant voltage used to determine gain					

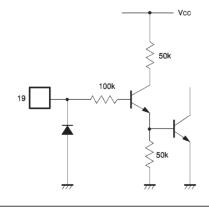
●Input / output circuits

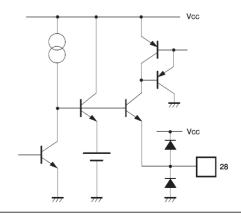












• Electrical characteristics (unless otherwise noted, Ta = 25° C, Vcc = 8V, RL = 8Ω)

Parameter	Symbol	Min.	Тур.	Max.	Unit	Conditions	Measurement Circuit
Quiescent current dissipation	la	_	14.0	18.0	mA	No load	Fig.1
Bias voltage	VBIAS	3.70	4.00	4.30	٧		Fig.1
Bias voltage variation	△VBIAS	-30	_	30	mV	1 mA, source, sink	Fig.1
Pin 28 constant voltage output	Vconst	1.10	1.25	1.40	٧	30kΩ at GND	Fig.1
Mute On voltage	VMON	2.0	_	_	٧		Fig.1
Mute Off voltage	V _{MOFF}	_	_	0.5	٧		Fig.1
⟨Interface⟩							
Input high level voltage	Vıн	2.0	_	_	٧		Fig.1
Input low level voltage	VIL	_	-	0.5	٧		Fig.1
Input high level current	lн	220	300	420	μΑ	V _{IN} =5V	Fig.1
Input low level current	lı.	0	-	-10	μA	V _{IN} =0V	Fig.1
Current pulse delay time 1	△tr	_	_	1	μs	At startup	Fig.2
Current pulse delay time 2	△tf	_	_	1	μs	At shutdown	Fig.2
Current pulse delay time differential	△tr-f	-200	_	200	ns		Fig.2
⟨Driver⟩							
Output offset voltage	Voo	-30	_	30	mV		Fig.1
Maximum output amplitude 1	V _{OMD1}	5.2	5.6	_	٧	Vcc=8V	Fig.1
Maximum output amplitude 2	V _{OMD2}	3.0	3.3	_	٧	Vcc=5V	Fig.1
Voltage gain	GVD	7.0	9.5	11.5	dB	V _{IN} =±0.5V	Fig.1
Ripple rejection	RR	_	70	_	dB	V _{IN} =100mV _{rms} , 100Hz	Fig.1

ONot designed for radiation resistance.

Measurement circuits

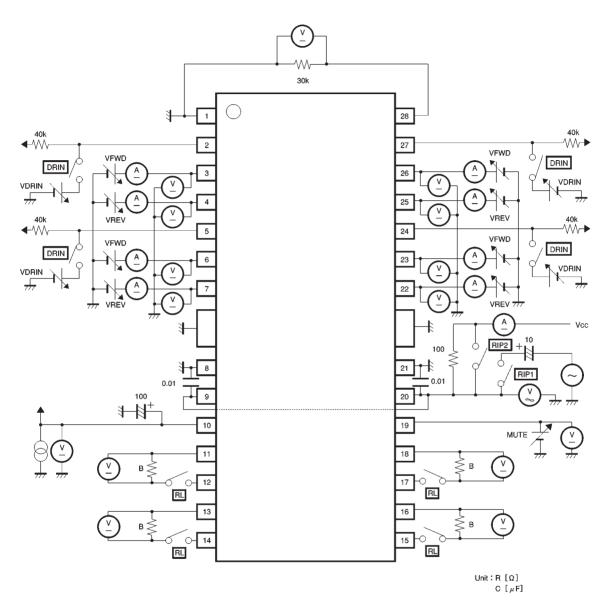


Fig.1

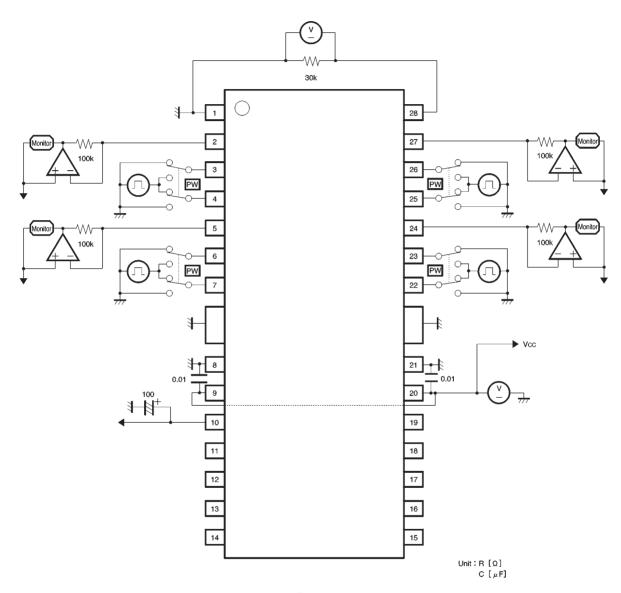


Fig.2

Switch table

Doromotor	Switch				Input				Note	
Parameter	RIP1	RIP2	RL	DRIN	MUTE	VDRIN	VFWD	VREV	Note	
Quiescent current dissipation						_				
Bias pin voltage			OFF		0.5V	_	01/	0V		
Bias voltage variation		ON.		OFF		_	0V			
Pin 28 constant voltage output	OFF	ON				_				
Mute-on voltage			ON		2.0V	_	5V	0V		
Mute-off voltage					0.5V	_	5V	0V		
⟨Interface⟩					•					
Input high level voltage F					_	_	2V	0V		
Input high level voltage R				OFF	_	_	0V	2V		
Input low level voltage F					_	_	0.5V	0V		
Input low level voltage R	٥٥٥	ON.	055		_	_	0V	0.5V		
Input high level current	OFF	FON	ON OFF		OFF	_	_	5V	5V	
Input low level current					_	_	0V	0V		
Current pulse delay 1					_	_	_	_	Measure the delay in the monitors observed	
Current pulse delay 2					_	_	_	_	waveform relative to the input waveform	
Current pulse delay differential	_	_	_	_	_	_	_	_		
(Driver)										
Output offset voltage				OFF		_	_	_		
Maximum output amplitude 1	٥	OFF ON	ON	ON	0.5V	7V, 1V	_	_		
Maximum output amplitude 2	OFF					4V, 1V	_	_		
Voltage gain						*1	_	_	(pin 10 voltage ±0.5 V) *1	
Ripple rejection	ON	OFF		OFF		_	_	_		

Circuit operation

(1) Overview

Fig. 6 shows the inputs from the digital servo IC. SW1 turns on when the forward input signal is received (HIGH level, above 2.0V). SW2 turns on when the reverse input signal is received (Fig. 3). When this happens, the constant current enters resistor R_1 and the capacitor, generating an integral waveform based on the duty of the input waveform. This is increased by a factor of 3 by the driver and output (Fig. 5).

When forward or reverse input remains at the HIGH level, the DC voltage generated at point A is:

I
$$\times$$
 R₁ [V] (reverse : -1 \times R₁ [V])

This is the voltage generated relative V_{ref} . The width setting is such that the following driver buffer output is generated :

3IR₁ [V] (reverse: -3IR₁ [V]) (1)

The time constant is:

 $R_1 (C + 30p)$

This can be changed with an attached capacitor.

Dead zone width is set by input duty ratio according to the following equation:

$$\frac{10.0 \text{ [k\Omega]} \times 1.0 \text{ [}\mu\text{A]}}{1 \times \text{R}_1} \times 100 \text{ [}\%\text{] (one side)} \text{ } \textcircled{2}$$

(2) Settings

The constant current is determined by the resistor (R_0) connected between pin 28 and pin 1 (GND).

$$I = 1.25 / R_0 [A]$$

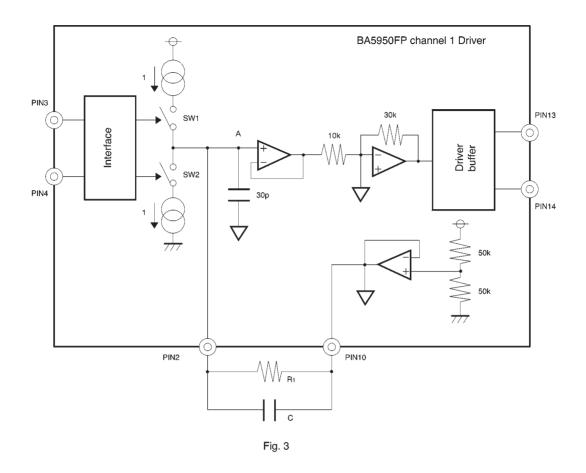
Consequently, equations (1) and (2) are as follows.

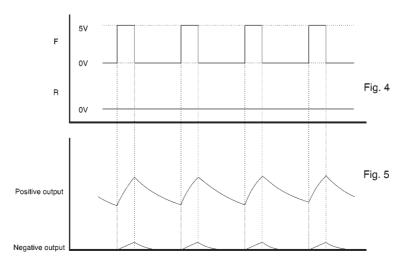
The ratio of R_0 and R_1 determine everything. R_0 must stay between $10k\Omega$ and $40k\Omega$. R_1 below $100k\Omega$.

Example : When $R_0 = 30k\Omega$, $R_1 = 40k\Omega$, input HIGH level = 5V

→5V (0dB relative to input)

 $(2)\rightarrow 0.6\%$ (input equivalent = 30mV)





Application example

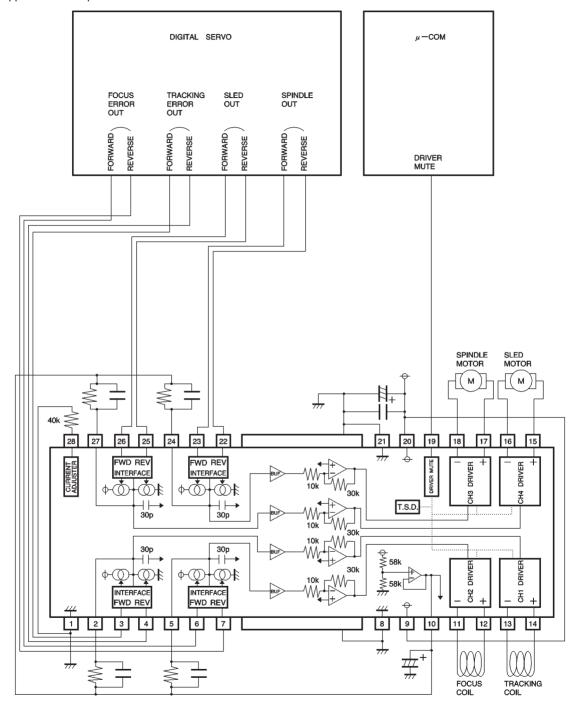


Fig. 6

Operation notes

- (1) The BA5950FP has an internal thermal shutdown circuit with hysteresis. Output current is muted when the chip temperature exceeds 175°C (typically) and restored when the chip temperature falls to 150°C (typically).
- (2) Output current can be muted by raising the mute pin (pin19) voltage above 2.0V. The mute pin must be kept below 0.5V during normal operation.
- (3) All four driver output channels are muted during thermal shutdown, muting and a drop in bias pin voltage. No other components are muted.
- (4) Connect a stabilizing capacitor (roughly $1\mu F$) to the internal reference voltage output pin (pin10).
- (5) Connect the IC to a $0.1\mu F$ bypass capacitor to the power supply, at the base of the IC.
- (6) Connect the radiating fin to an external ground.

Electrical characteristic curves

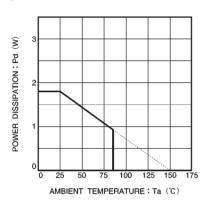


Fig. 7 Thermal derating curve

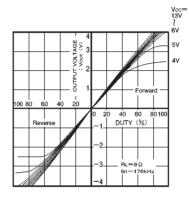


Fig. 8 Driver I/O characteristics (variable Vcc)

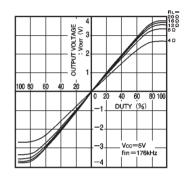


Fig. 9 Driver I/O characteristics (variable load)

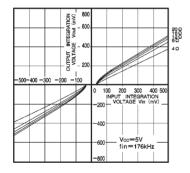


Fig. 10 Driver ultralow input I / O characteristic (variable load)

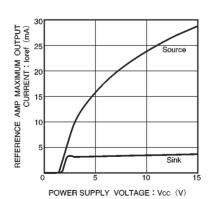


Fig. 11 Power supply voltage vs. Vref amplifier output drive current

●External dimensions (Units: mm)

