PIC16(L)F707
Data Sheet

40/44-Pin, Flash Microcontrollers with nanoWatt XLP and mTouch™ Technology

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40/44-Pin, Flash Microcontrollers with nanoWatt XLP and mTouch™ Technology

Devices included in this data sheet:

- PIC16F707
- PIC16LF707

High-Performance RISC CPU:

- Only 35 Single-Word Instructions to Learn:
  - All single-cycle instructions except branches
- Operating Speed:
  - DC – 20 MHz clock input
  - DC – 200 ns instruction cycle
- 8K x 14 Words of Flash Program Memory
- 363 Bytes of Data Memory (SRAM)
- Interrupt Capability
- 8-Level Deep Hardware Stack
- Direct, Indirect and Relative Addressing modes
- Processor Read Access to Program Memory
- Pinout Compatible to other 40-pin PIC16CXXX and PIC16FXXX Microcontrollers

Special Microcontroller Features:

- Precision Internal Oscillator:
  - 16 MHz or 500 kHz operation
  - Factory calibrated to ±1%, typical
  - Software selectable ÷1, ÷2, ÷4 or ÷8 divider
- 31 kHz Low-Power Internal Oscillator
- External Oscillator Block with:
  - 3 crystal/resonator modes up to 20 MHz
  - 3 external clock modes up to 20 MHz
- Power-on Reset (POR)
- Power-up Timer (PWRT)
- Oscillator Start-Up Timer (OST)
- Brown-out Reset (BOR):
  - Selectable between two trip points
  - Disabled in Sleep option
- Watchdog Timer (WDT)
- Programmable Code Protection
- In-Circuit Serial Programming™ (ICSP™) via two pins
- In-Circuit Debug (ICD) via Two Pins
- Multiplexed Master Clear with Pull-up/Input Pin
- Industrial and Extended Temperature Range
- High-Endurance Flash Cell:
  - 1,000 Write Flash Endurance (typical)
  - Flash Retention: >40 years
  - Power-Saving Sleep mode
- Operating Voltage Range:
  - 1.8V to 3.6V (PIC16LF707)
- 1.8V to 5.5V (PIC16F707)

Extreme Low-Power Management PIC16LF707 with nanoWatt XLP:

- Sleep mode: 20 nA @ 1.8V, typical
- Watchdog Timer: 500 nA @ 1.8V, typical
- Timer1 Oscillator: 600 nA @ 1.8V, typical @ 32 kHz

mTouch™ Technology Features:

- Up to 32 Channels
- Two Capacitive Sensing modules:
  - Acquire 2 samples simultaneously
- Multiple Power modes:
  - Operation during Sleep
  - Proximity sensing with ultra low µA current
- Adjustable Waveform Min. and Max. for Optimal Noise Performance
- 1.8V to 5.5V Operation (3.6V max. for PIC16LF707)

Analog Features:

- A/D Converter:
  - 8-bit resolution and up to 14 channels
  - Conversion available during Sleep
  - Selectable 1.024V/2.048V/4.096V voltage reference
- On-chip 3.2V Regulator (PIC16F707 device only)

Peripheral Highlights:

- Up to 35 I/O Pins and 1 Input-only Pin:
  - High current source/sink for direct LED drive
  - Interrupt-on-pin change
  - Individually programmable weak pull-ups
- Timer0/A/B: 8-Bit Timer/Counter with 8-Bit Prescaler
- Enhanced Timer1/3:
  - Dedicated low-power 32 kHz oscillator driver
  - 16-bit timer/counter with prescaler
  - External Gate Input mode with toggle and single shot modes
  - Interrupt-on-gate completion
- Timer2: 8-Bit Timer/Counter with 8-Bit Period Register, Prescaler and Postscaler
- Two Capture, Compare, PWM modules (CCP):
  - 16-bit Capture, max. resolution 12.5 ns
  - 16-bit Compare, max. resolution 200 ns
  - 10-bit PWM, max. frequency 20 kHz
- Addressable Universal Synchronous Asynchronous Receiver Transmitter (AUSART)

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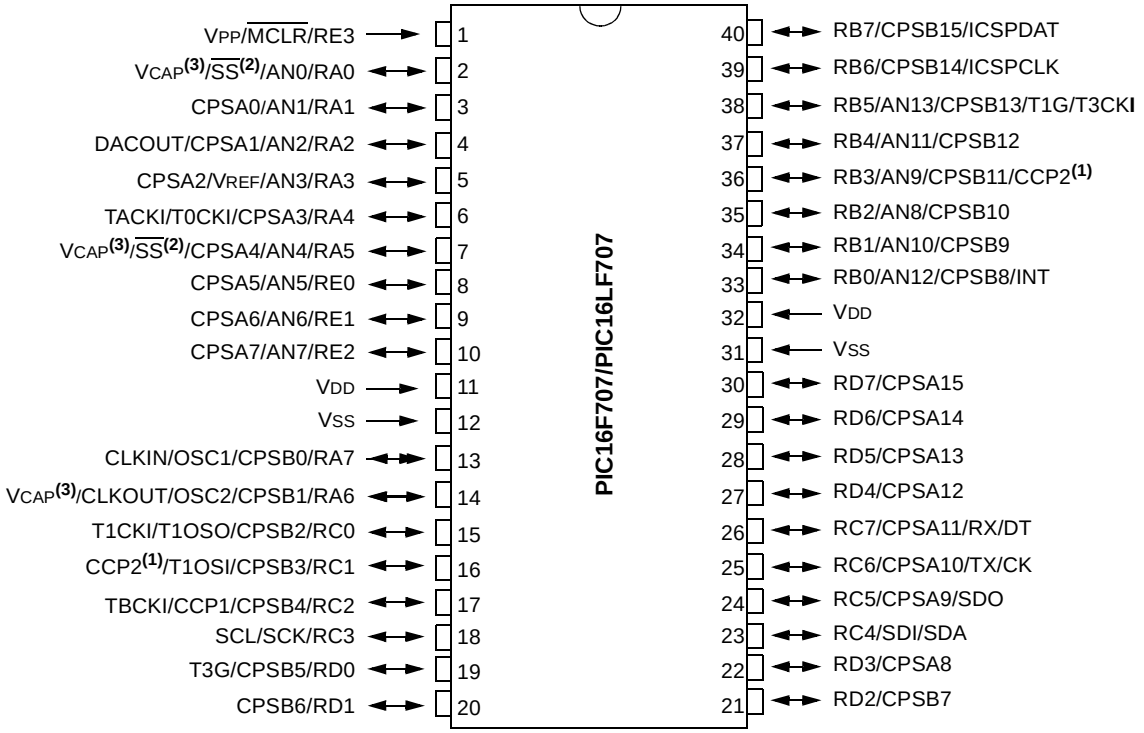
- Synchronous Serial Port (SSP):
  - SPI (Master/Slave)
  - I^2C™ (Slave) with Address Mask
- Voltage Reference module:
  - Fixed Voltage Reference (FVR) with 1.024V, 2.048V and 4.096V output levels
  - 5-bit rail-to-rail resistive DAC with positive reference selection

Device	Program Memory Flash (words)	SRAM (bytes)	I/Os	Capacitive Touch Channels	8-bit A/D (ch)	AUSART	CCP	Timers 8/16-bit
PIC16(L)F707	8192	363	36	32	14	Yes	2	4/2

Pin Diagrams

40-PIN PDIP



Note 1: CCP2 pin location may be selected as RB3 or RC1.

2: $\overline{SS}$ pin location may be selected as RA5 or RA0.

3: PIC16F707 only.

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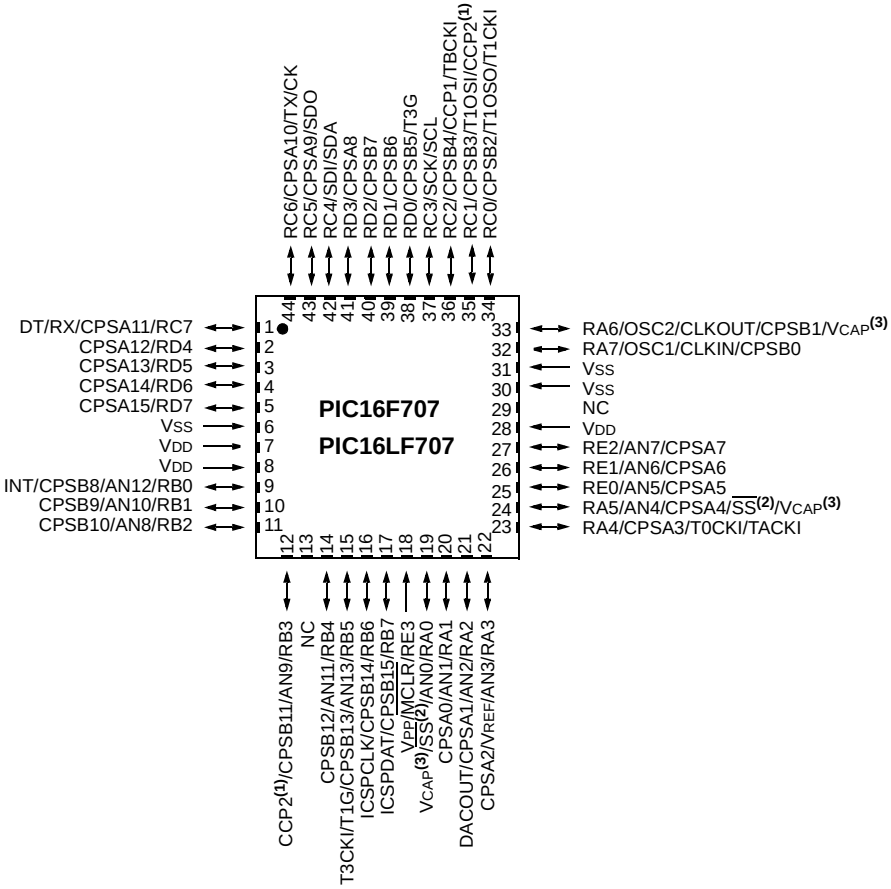
Pin Diagrams

44-PIN QFN (8x8x0.9)



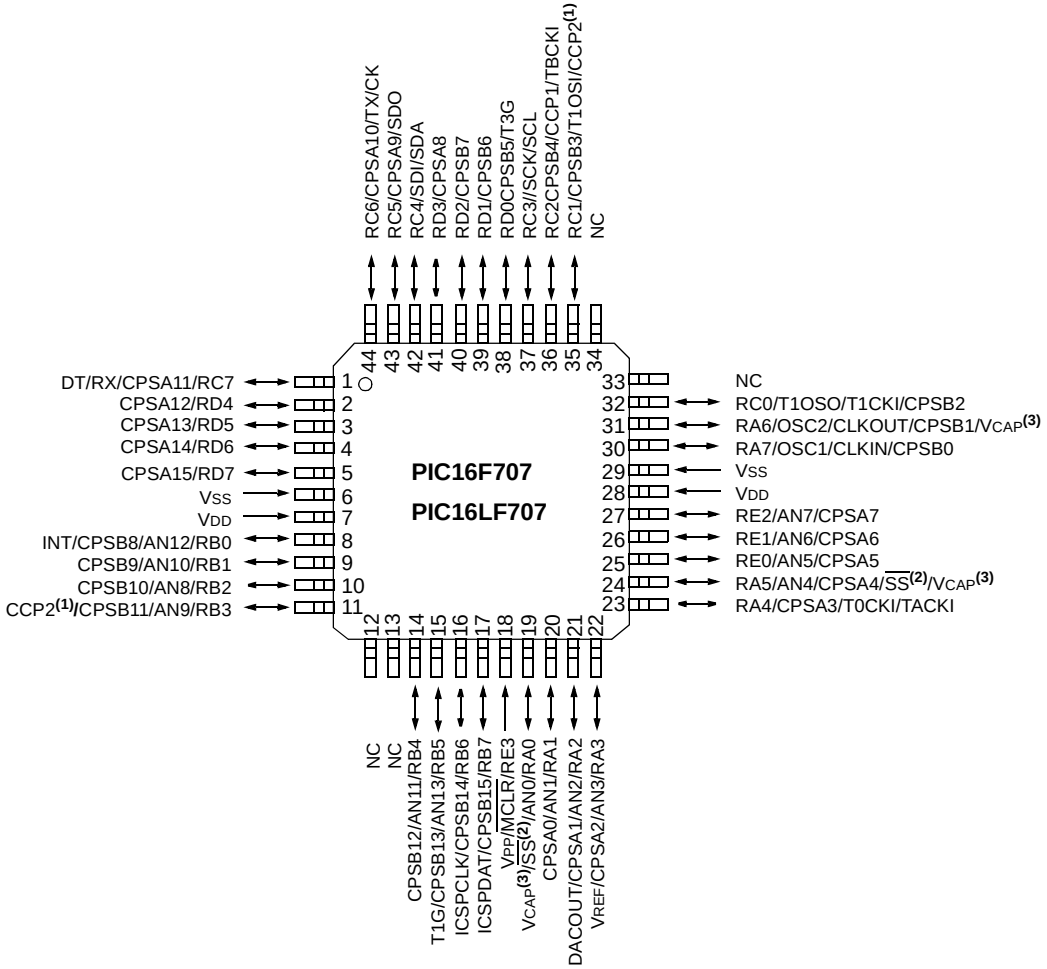
Note 1: CCP2 pin location may be selected as RB3 or RC1.

2: $\overline{SS}$ pin location may be selected as RA5 or RA0.

3: PIC16F707 only.

Pin Diagrams

44-PIN TQFP



Note 1: CCP2 pin location may be selected as RB3 or RC1.

2: $\overline{SS}$ pin location may be selected as RA5 or RA0.

3: PIC16F707 only.

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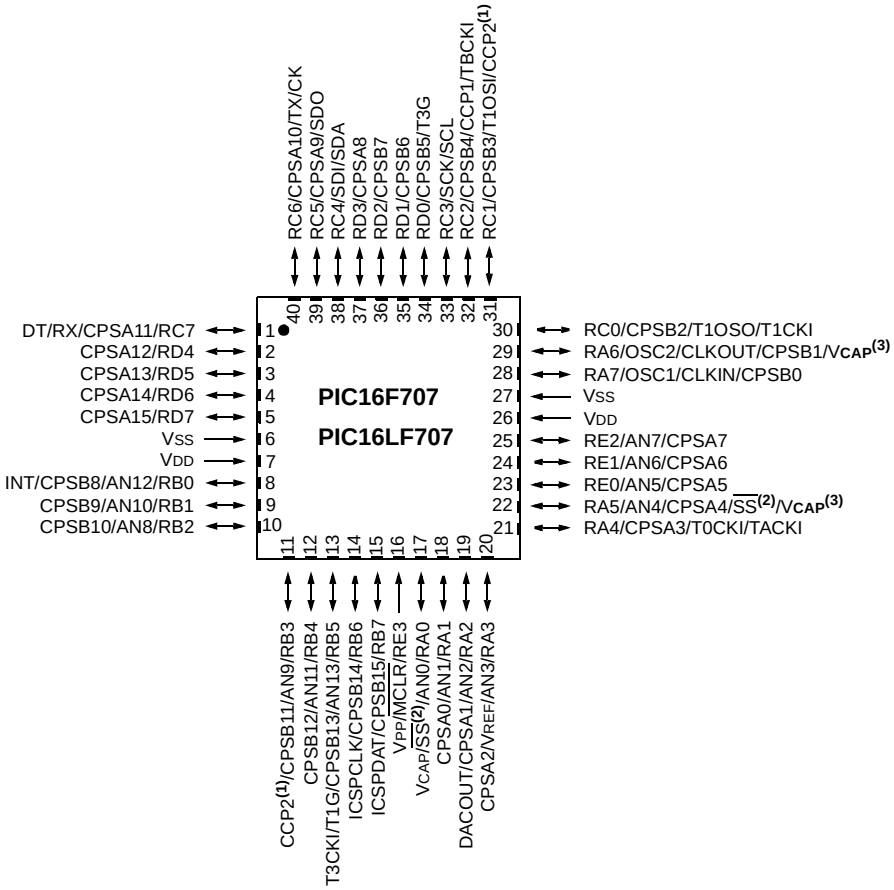
Pin Diagrams

40-PIN UQFN (5x5x0.5)



Note 1: CCP2 pin location may be selected as RB3 or RC1.
2: SS pin location may be selected as RA5 or RA0.
3: PIC16F707 only.

TABLE 1: 40/44-PIN ALLOCATION TABLE FOR PIC16F707/PIC16LF707

I/O	40-Pin PDIP	44-Pin TQFP	44-Pin QFN	40-Pin UQFN	ANSEL	A/D	DAC	Cap Sensor	Timers	CCP	AUSART	SSP	Interrupt	Pull-up	Basic
RA0	2	19	19	17	Y	AN0	—		—	—	—	$\overline{SS}$(3)	—	—	VCAP(4)
RA1	3	20	20	18	Y	AN1	—	CPSA0	—	—	—	—	—	—	—
RA2	4	21	21	19	Y	AN2	DACOUT	CPSA1	—	—	—	—	—	—	—
RA3	5	22	22	20	Y	AN3/ VREF	VREF	CPSA2	—	—	—	—	—	—	—
RA4	6	23	23	21	Y	—	—	CPSA3	T0CKI/ TACKI	—	—	—	—	—	—
RA5	7	24	24	22	Y	AN4	—	CPSA4	—	—	—	$\overline{SS}$(3)	—	—	VCAP(4)
RA6	14	31	33	29	Y	—	—	CPSB1	—	—	—	—	—	—	OSC2/ CLKOUT/ VCAP(4)
RA7	13	30	32	28	Y	—	—	CPSB0	—	—	—	—	—	—	OSC1/ CLKIN
RB0	33	8	9	8	Y	AN12	—	CPSB8	—	—	—	—	IOC/INT	Y	—
RB1	34	9	10	9	Y	AN10	—	CPSB9	—	—	—	—	IOC	Y	—
RB2	35	10	11	10	Y	AN8	—	CPSB10	—	—	—	—	IOC	Y	—
RB3	36	11	12	11	Y	AN9	—	CPSB11	—	CCP2(2)	—	—	IOC	Y	—
RB4	37	14	14	12	Y	AN11	—	CPSB12	—	—	—	—	IOC	Y	—
RB5	38	15	15	13	Y	AN13	—	CPSB13	T1G/ T3CKI	—	—	—	IOC	Y	—
RB6	39	16	16	14	Y	—	—	CPSB14	—	—	—	—	IOC	Y	ICSPCLK/ ICDCLK
RB7	40	17	17	15	Y	—	—	CPSB15	—	—	—	—	IOC	Y	ICSPDAT/ ICDDAT
RC0	15	32	34	30	Y	—	—	CPSB2	T1OSO/ T1CKI	—	—	—	—	—	—
RC1	16	35	35	31	Y	—	—	CPSB3	T1OSI	CCP2(2)	—	—	—	—	—
RC2	17	36	36	32	Y	—	—	CPSB4	TBCKI	CCP1	—	—	—	—	—
RC3	18	37	37	33	—	—	—	—	—	—	—	SCK/ SCL	—	—	—
RC4	23	42	42	38	—	—	—	—	—	—	—	SDI/ SDA	—	—	—
RC5	24	43	43	39	Y	—	—	CPSA9	—	—	—	SDO	—	—	—
RC6	25	44	44	40	Y	—	—	CPSA10	—	—	TX/CK	—	—	—	—
RC7	26	1	1	1	Y	—	—	CPSA11	—	—	RX/DT	—	—	—	—
RD0	19	38	38	34	Y	—	—	CPSB5	T3G	—	—	—	—	—	—
RD1	20	39	39	35	Y	—	—	CPSB6	—	—	—	—	—	—	—
RD2	21	40	40	36	Y	—	—	CPSB7	—	—	—	—	—	—	—
RD3	22	41	41	37	Y	—	—	CPSA8	—	—	—	—	—	—	—
RD4	27	2	2	2	Y	—	—	CPSA12	—	—	—	—	—	—	—
RD5	28	3	3	3	Y	—	—	CPSA13	—	—	—	—	—	—	—
RD6	29	4	4	4	Y	—	—	CPSA14	—	—	—	—	—	—	—
RD7	30	5	5	5	Y	—	—	CPSA15	—	—	—	—	—	—	—
RE0	8	25	25	23	Y	AN5	—	CPSA5	—	—	—	—	—	—	—
RE1	9	26	26	24	Y	AN6	—	CPSA6	—	—	—	—	—	—	—
RE2	10	27	27	25	Y	AN7	—	CPSA7	—	—	—	—	—	—	—
RE3	1	18	18	16	—	—	—	—	—	—	—	—	—	Y(1)	$\overline{MCLR}$/ VPP

Note 1: Pull-up activated only with external $\overline{MCLR}$ configuration.

2: RC1 is the default pin location for CCP2. RB3 may be selected by changing the CCP2SEL bit in the APFCON register.

3: RA5 is the default pin location for $\overline{SS}$. RA0 may be selected by changing the SSSEL bit in the APFCON register.

4: PIC16F707 only. VCAP functionality is selectable by the VCAPEN bits in Configuration Word 2.

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TABLE 1: 40/44-PIN ALLOCATION TABLE FOR PIC16F707/PIC16LF707

I/O	40-Pin PDIP	44-Pin TQFP	44-Pin QFN	40-Pin UQFN	ANSEL	A/D	DAC	Cap Sensor	Timers	CCP	AUSART	SSP	Interrupt	Pull-up	Basic
VDD	11, 32	7, 28	7, 8, 28	7, 26		—	—	—	—	—	—	—	—	—	VDD
VSS	12, 31	6, 29	6, 30, 31	6, 27		—	—	—	—	—	—	—	—	—	VSS

Note 1: Pull-up activated only with external $\overline{\text{MCLR}}$ configuration.

2: RC1 is the default pin location for CCP2. RB3 may be selected by changing the CCP2SEL bit in the APFCON register.

3: RA5 is the default pin location for $\overline{\text{SS}}$. RA0 may be selected by changing the SSSEL bit in the APFCON register.

4: PIC16F707 only. VCAP functionality is selectable by the VCAPEN bits in Configuration Word 2.

Table of Contents

1.0 Device Overview 13
2.0 Memory Organization 19
3.0 Resets 31
4.0 Interrupts 41
5.0 Low Dropout (LDO) Voltage Regulator 51
6.0 I/O Ports 53
7.0 Oscillator Module 71
8.0 Device Configuration 77
9.0 Analog-to-Digital Converter (ADC) Module 81
10.0 Fixed Voltage Reference 91
11.0 Digital-to-Analog Converter (DAC) Module 93
12.0 Timer0 Module 97
13.0 Timer1/3 Modules with Gate Control 101
14.0 TimerA/B Modules 113
15.0 Timer2 Module 117
16.0 Capacitive Sensing Module 119
17.0 Capture/Compare/PWM (CCP) Module 129
18.0 Addressable Universal Synchronous Asynchronous Receiver Transmitter (AUSART) 139
19.0 SSP Module Overview 159
20.0 Program Memory Read 181
21.0 Power-Down Mode (Sleep) 185
22.0 In-Circuit Serial Programming™ (ICSP™) 187
23.0 Instruction Set Summary 189
24.0 Development Support 199
25.0 Electrical Specifications 203
26.0 DC and AC Characteristics Graphs and Charts 233
27.0 Packaging Information 269
Appendix A: Data Sheet Revision History 279
Appendix B: Migrating From Other PIC® Devices 279
The Microchip Web Site 287
Customer Change Notification Service 287
Customer Support 287
Reader Response 288
Product Identification System 289

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1.0 DEVICE OVERVIEW

The PIC16(L)F707 devices are covered by this data sheet. They are available in 40/44-pin packages. Figure 1-1 shows a block diagram of the PIC16(L)F707 devices. Table 1-1 shows the pinout descriptions.

FIGURE 1-1: PIC16(L)F707 BLOCK DIAGRAM

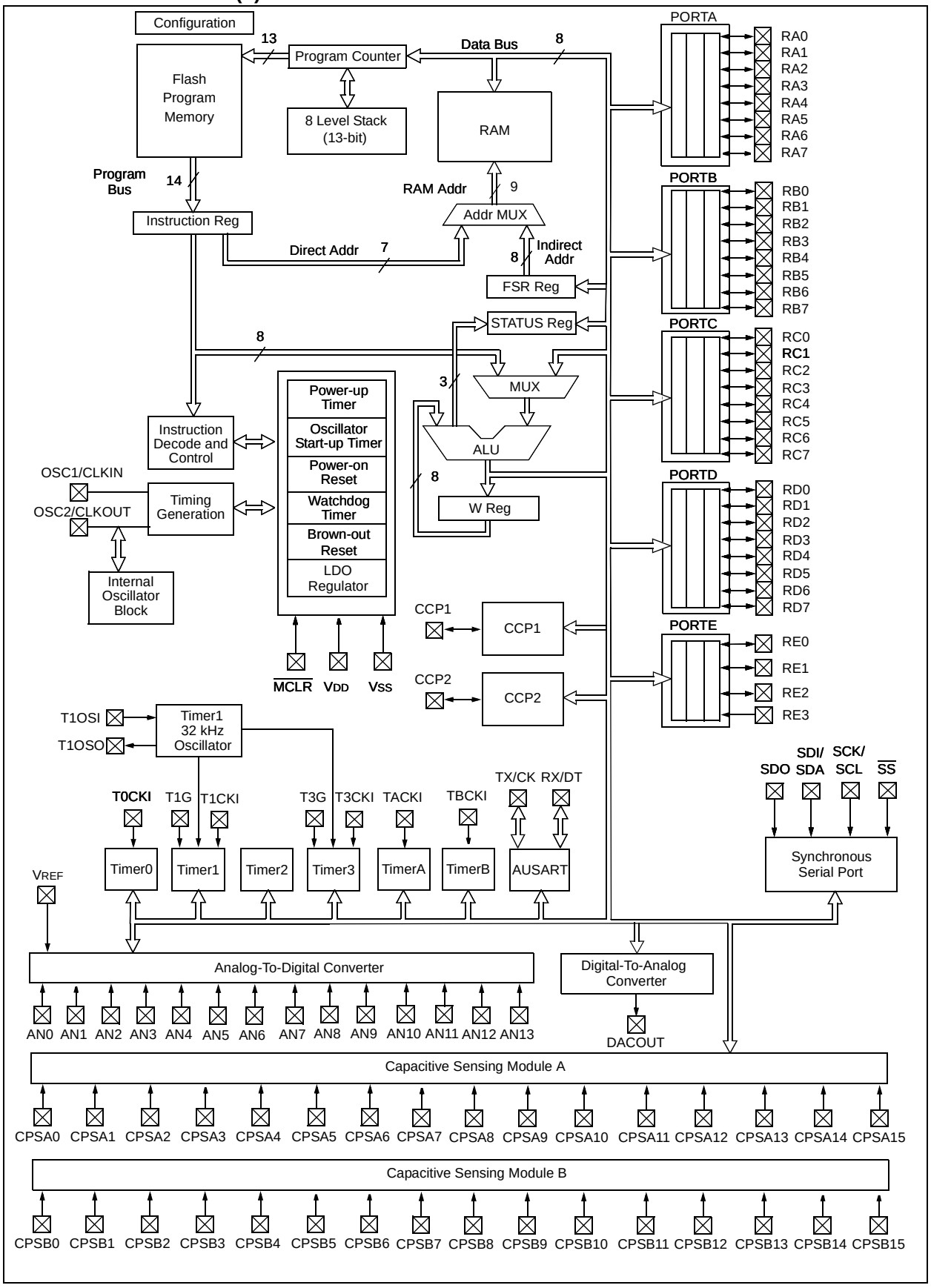

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TABLE 1-1: PIC16(L)F707 PINOUT DESCRIPTION

Name	Function	Input Type	Output Type	Description
RA0/AN0/$\overline{SS}$/V_{CAP}	RA0	TTL	CMOS	General purpose I/O.
	AN0	AN	—	A/D Channel 0 input.
	$\overline{SS}$	ST	—	Slave Select input.
	V_{CAP}	Power	Power	Filter capacitor for Voltage Regulator (PIC16F only).
RA1/AN1/CPSA0	RA1	TTL	CMOS	General purpose I/O.
	AN1	AN	—	A/D Channel 1 input.
	CPSA0	AN	—	Capacitive sensing A input 0.
RA2/AN2/CPSA1/DACOUT	RA2	TTL	CMOS	General purpose I/O.
	AN2	AN	—	A/D Channel 2 input.
	CPSA1	AN	—	Capacitive sensing A input 1.
	DACOUT	—	AN	Voltage Reference Output.
RA3/AN3/V_{REF}/CPSA2	RA3	TTL	CMOS	General purpose I/O.
	AN3	AN	—	A/D Channel 3 input.
	V_{REF}	AN	—	A/D Voltage Reference input.
	CPSA2	AN	—	Capacitive sensing A input 2.
RA4/CPSA3/T0CKI/TACKI	RA4	TTL	CMOS	General purpose I/O.
	CPSA3	AN	—	Capacitive sensing A input 3.
	T0CKI	ST	—	Timer0 clock input.
	TACKI	ST	—	TimerA clock input.
RA5/AN4/CPSA4/$\overline{SS}$/V_{CAP}	RA5	TTL	CMOS	General purpose I/O.
	AN4	AN	—	A/D Channel 4 input.
	CPSA4	AN	—	Capacitive sensing A input 4.
	$\overline{SS}$	ST	—	Slave Select input.
	V_{CAP}	Power	Power	Filter capacitor for Voltage Regulator (PIC16F only).
RA6/OSC2/CLKOUT/V_{CAP}/ CPSB1	RA6	TTL	CMOS	General purpose I/O.
	OSC2	—	XTAL	Crystal/Resonator (LP, XT, HS modes).
	CLKOUT	—	CMOS	F_{OSC}/4 output.
	V_{CAP}	Power	Power	Filter capacitor for Voltage Regulator (PIC16F only).
	CPSB1	AN	—	Capacitive sensing B input 1.
RA7/OSC1/CLKIN/CPSB0	RA7	TTL	CMOS	General purpose I/O.
	OSC1	XTAL	—	Crystal/Resonator (LP, XT, HS modes).
	CLKIN	CMOS	—	External clock input (EC mode).
	CLKIN	ST	—	RC oscillator connection (RC mode).
	CPSB0	AN	—	Capacitive sensing B input 0.
RB0/AN12/CPSB8/INT	RB0	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN12	AN	—	A/D Channel 12 input.
	CPSB8	AN	—	Capacitive sensing B input 8.
	INT	ST	—	External interrupt.
RB1/AN10/CPSB9	RB1	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN10	AN	—	A/D Channel 10 input.
	CPSB9	AN	—	Capacitive sensing B input 9.

Legend: AN = Analog input or output; CMOS = CMOS compatible input or output; OD = Open Drain
TTL = TTL compatible input; ST = Schmitt Trigger input with CMOS levels; I^2C™ = Schmitt Trigger input with I^2C levels
HV = High Voltage; XTAL = Crystal

TABLE 1-1: PIC16(L)F707 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RB2/AN8/CPSB10	RB2	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN8	AN	—	A/D Channel 8 input.
	CPSB10	AN	—	Capacitive sensing B input 10.
RB3/AN9/CPSB11/CCP2	RB3	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN9	AN	—	A/D Channel 9 input.
	CPSB11	AN	—	Capacitive sensing B input 11.
	CCP2	ST	CMOS	Capture/Compare/PWM2.
RB4/AN11/CPSB12	RB4	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN11	AN	—	A/D Channel 11 input.
	CPSB12	AN	—	Capacitive sensing B input 12.
RB5/AN13/CPSB13/T1G/T3CKI	RB5	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN13	AN	—	A/D Channel 13 input.
	CPSB13	AN	—	Capacitive sensing B input 13.
	T1G	ST	—	Timer1 gate input.
	T3CKI	ST	—	Timer3 clock input.
RB6/ICSPCLK/ICDCLK/CPSB14	RB6	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	ICSPCLK	ST	—	Serial Programming Clock.
	ICDCLK	ST	—	In-Circuit Debug Clock.
	CPSB14	AN	—	Capacitive sensing B input 14.
RB7/ICSPDAT/ICDDAT/CPSB15	RB7	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	ICSPDAT	ST	CMOS	ICSP™ Data I/O.
	ICDDAT	ST	—	In-Circuit Data I/O.
	CPSB15	AN	—	Capacitive sensing B input 15.
RC0/T1OSO/T1CKI/CPSB2	RC0	ST	CMOS	General purpose I/O.
	T1OSO	XTAL	XTAL	Timer1 oscillator connection.
	T1CKI	ST	—	Timer1 clock input.
	CPSB2	AN	—	Capacitive sensing B input 2.
RC1/T1OSI/CCP2/CPSB3	RC1	ST	CMOS	General purpose I/O.
	T1OSI	XTAL	XTAL	Timer1 oscillator connection.
	CCP2	ST	CMOS	Capture/Compare/PWM2.
	CPSB3	AN	—	Capacitive sensing B input 3.
RC2/CCP1/CPSB4/TBCKI	RC2	ST	CMOS	General purpose I/O.
	CCP1	ST	CMOS	Capture/Compare/PWM1.
	CPSB4	AN	—	Capacitive sensing B input 4.
	TBCKI	ST	—	TimerB clock input.
RC3/SCK/SCL	RC3	ST	CMOS	General purpose I/O.
	SCK	ST	CMOS	SPI clock.
	SCL	I^2C™	OD	I^2C™ clock.

Legend: AN = Analog input or output; CMOS = CMOS compatible input or output; OD = Open Drain
TTL = TTL compatible input; ST = Schmitt Trigger input with CMOS levels; I^2C™ = Schmitt Trigger input with I^2C levels
HV = High Voltage; XTAL = Crystal

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TABLE 1-1: PIC16(L)F707 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RC4/SDI/SDA	RC4	ST	CMOS	General purpose I/O.
	SDI	ST	—	SPI data input.
	SDA	I^2C™	OD	I^2C™ data input/output.
RC5/SDO/CPSA9	RC5	ST	CMOS	General purpose I/O.
	SDO	—	CMOS	SPI data output.
	CPSA9	AN	—	Capacitive sensing A input 9.
RC6/TX/CK/CPSA10	RC6	ST	CMOS	General purpose I/O.
	TX	—	CMOS	USART asynchronous transmit.
	CK	ST	CMOS	USART synchronous clock.
	CPSA10	AN	—	Capacitive sensing A input 10.
RC7/RX/DT/CPSA11	RC7	ST	CMOS	General purpose I/O.
	RX	ST	—	USART asynchronous input.
	DT	ST	CMOS	USART synchronous data.
	CPSA11	AN	—	Capacitive sensing A input 11.
RD0/CPSB5/T3G	RD0	ST	CMOS	General purpose I/O.
	CPSB5	AN	—	Capacitive sensing B input 5.
	T3G	ST	—	Timer3 Gate input.
RD1/CPSB6	RD1	ST	CMOS	General purpose I/O.
	CPSB6	AN	—	Capacitive sensing B input 6.
RD2/CPSB7	RD2	ST	CMOS	General purpose I/O.
	CPSB7	AN	—	Capacitive sensing B input 7.
RD3/CPSA8	RD3	ST	CMOS	General purpose I/O.
	CPSA8	AN	—	Capacitive sensing A input 8.
RD4/CPSA12	RD4	ST	CMOS	General purpose I/O.
	CPSA12	AN	—	Capacitive sensing A input 12.
RD5/CPSA13	RD5	ST	CMOS	General purpose I/O.
	CPSA13	AN	—	Capacitive sensing A input 13.
RD6/CPSA14	RD6	ST	CMOS	General purpose I/O.
	CPSA14	AN	—	Capacitive sensing A input 14.
RD7/CPSA15	RD7	ST	CMOS	General purpose I/O.
	CPSA15	AN	—	Capacitive sensing A input 15.
RE0/AN5/CPSA5	RE0	ST	CMOS	General purpose I/O.
	AN5	AN	—	A/D Channel 5 input.
	CPSA5	AN	—	Capacitive sensing A input 5.
RE1/AN6/CPSA6	RE1	ST	CMOS	General purpose I/O.
	AN6	AN	—	A/D Channel 6 input.
	CPSA6	AN	—	Capacitive sensing A input 6.
RE2/AN7/CPSA7	RE2	ST	CMOS	General purpose I/O.
	AN7	AN	—	A/D Channel 7 input.
	CPSA7	AN	—	Capacitive sensing A input 7.
RE3/$\overline{MCLR}$/V_{PP}	RE3	TTL	—	General purpose input.
	$\overline{MCLR}$	ST	—	Master Clear with internal pull-up.
	V_{PP}	HV	—	Programming voltage.
V_{DD}	V_{DD}	Power	—	Positive supply.

Legend: AN = Analog input or output; CMOS = CMOS compatible input or output; OD = Open Drain
TTL = TTL compatible input; ST = Schmitt Trigger input with CMOS levels; I^2C™ = Schmitt Trigger input with I^2C levels
HV = High Voltage; XTAL = Crystal

TABLE 1-1: PIC16(L)F707 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
VSS	VSS	Power	—	Ground reference.

Legend: AN = Analog input or output; CMOS = CMOS compatible input or output; OD = Open Drain
TTL = TTL compatible input; ST = Schmitt Trigger input with CMOS levels; I^2C™ = Schmitt Trigger input with I^2C levels
HV = High Voltage; XTAL = Crystal

Note: The PIC16F707 devices have an internal low dropout voltage regulator. An external capacitor must be connected to one of the available VCAP pins to stabilize the regulator. For more information, see **Section 5.0 "Low Dropout (LDO) Voltage Regulator"**. The PIC16LF707 devices do not have the voltage regulator and therefore no external capacitor is required.

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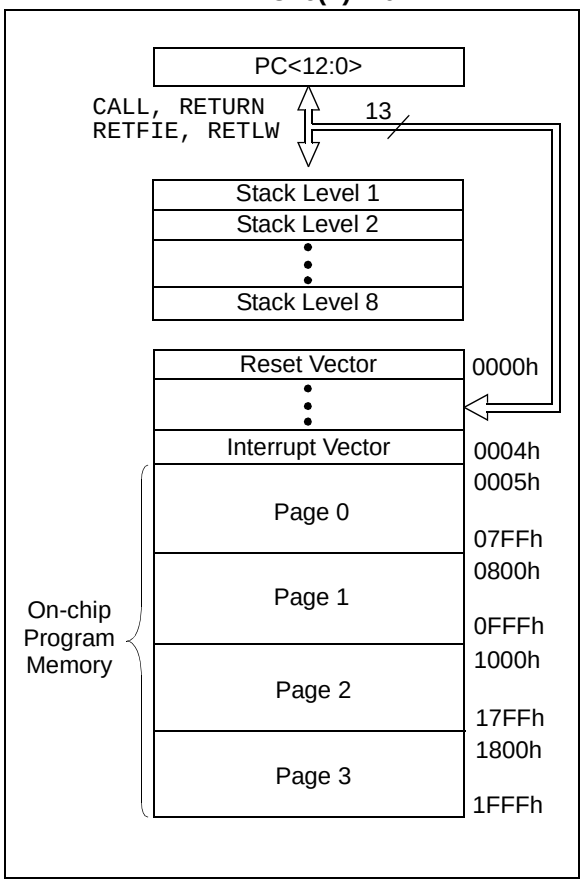
2.0 MEMORY ORGANIZATION

2.1 Program Memory Organization

The PIC16(L)F707 has a 13-bit program counter capable of addressing an 8K x 14 program memory space. The Reset vector is at 0000h and the interrupt vector is at 0004h.

FIGURE 2-1: PROGRAM MEMORY MAP AND STACK FOR THE PIC16(L)F707



2.2 Data Memory Organization

The data memory is partitioned into multiple banks which contain the General Purpose Registers (GPRs) and the Special Function Registers (SFRs). Bits RP0 and RP1 are bank select bits.

RP1	RP0		
0	0	→	Bank 0 is selected
0	1	→	Bank 1 is selected
1	0	→	Bank 2 is selected
1	1	→	Bank 3 is selected

Each bank extends up to 7Fh (128 bytes). The lower locations of each bank are reserved for the Special Function Registers. Above the Special Function Registers are the General Purpose Registers, implemented as static RAM. All implemented banks contain Special Function Registers. Some frequently used Special Function Registers from one bank are mirrored in another bank for code reduction and quicker access.

2.2.1 GENERAL PURPOSE REGISTER FILE

The register file is organized as 363 x 8 bits. Each register is accessed either directly or indirectly through the File Select Register (FSR), (Refer to **Section 2.5 "Indirect Addressing, INDF and FSR Registers"**).

2.2.2 SPECIAL FUNCTION REGISTERS

The Special Function Registers are registers used by the CPU and peripheral functions for controlling the desired operation of the device (refer to Table 2-2). These registers are static RAM.

The Special Function Registers can be classified into two sets: core and peripheral. The Special Function Registers associated with the "core" are described in this section. Those related to the operation of the peripheral features are described in the section of that peripheral feature.

TABLE 2-1: DATA MEMORY MAP FOR PIC16(L)F707

BANK 0	File Address	BANK 1	File Address	BANK 2	File Address	BANK 3	File Address
Indirect addr.(*)	00h	Indirect addr.(*)	80h	Indirect addr.(*)	100h	Indirect addr.(*)	180h
TMR0	01h	OPTION	81h	TMR0	101h	OPTION	181h
PCL	02h	PCL	82h	PCL	102h	PCL	182h
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183h
FSR	04h	FSR	84h	FSR	104h	FSR	184h
PORTA	05h	TRISA	85h	TACON	105h	ANSELA	185h
PORTB	06h	TRISB	86h	CPSBCON0	106h	ANSELB	186h
PORTC	07h	TRISC	87h	CPSBCON1	107h	ANSELC	187h
PORTD	08h	TRISD	88h	CPSACON0	108h	ANSELD	188h
PORTE	09h	TRISE	89h	CPSACON1	109h	ANSELE	189h
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18Ah
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18Bh
PIR1	0Ch	PIE1	8Ch	PMDATL	10Ch	PMCON1	18Ch
PIR2	0Dh	PIE2	8Dh	PMADRL	10Dh	Reserved	18Dh
TMR1L	0Eh	PCON	8Eh	PMDATH	10Eh	Reserved	18Eh
TMR1H	0Fh	T1GCON	8Fh	PMADRH	10Fh	Reserved	18Fh
T1CON	10h	OSCCON	90h	TMRA	110h	General Purpose Register 16 Bytes	190h
TMR2	11h	OSCTUNE	91h	TBCON	111h		191h
T2CON	12h	PR2	92h	TMRB	112h		192h
SSPBUF	13h	SSPADD/SSPMSK	93h	DACCON0	113h		193h
SSPCON	14h	SSPSTAT	94h	DACCON1	114h		194h
CCPR1L	15h	WPUB	95h	General Purpose Register 11 Bytes	115h		195h
CCPR1H	16h	IOCB	96h		116h		196h
CCP1CON	17h	T3CON	97h		117h		197h
RCSTA	18h	TXSTA	98h		118h		198h
TXREG	19h	SPBRG	99h		119h		199h
RCREG	1Ah	TMR3L	9Ah		11Ah		19Ah
CCPR2L	1Bh	TMR3H	9Bh		11Bh		19Bh
CCPR2H	1Ch	APFCON	9Ch		11Ch		19Ch
CCP2CON	1Dh	FVRCON	9Dh		11Dh		19Dh
ADRES	1Eh	T3GCON	9Eh		11Eh		19Eh
ADCON0	1Fh	ADCON1	9Fh		11Fh		19Fh
General Purpose Register 96 Bytes	20h – 7Fh	General Purpose Register 80 Bytes	A0h – EFh	General Purpose Register 80 Bytes	120h – 16Fh	General Purpose Register 80 Bytes	1A0h – 1EFh
		Accesses 70h – 7Fh	F0h – FFh	Accesses 70h – 7Fh	170h – 17Fh	Accesses 70h – 7Fh	1F0h – 1FFh

Legend: = Unimplemented data memory locations, read as '0',
* = Not a physical register

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TABLE 2-2: SPECIAL FUNCTION REGISTER SUMMARY

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets
Bank 0											
00h(2)	INDF	Addressing this location uses contents of FSR to address data memory (not a physical register)								xxxx xxxx	xxxx xxxx
01h	TMR0	Timer0 Module Register								0000 0000	0000 0000
02h(2)	PCL	Program Counter (PC) Least Significant Byte								0000 0000	0000 0000
03h(2)	STATUS	IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC	C	0001 1xxx	000q quuu
04h(2)	FSR	Indirect Data Memory Address Pointer								xxxx xxxx	uuuu uuuu
05h	PORTA	PORTA Data Latch when written: PORTA pins when read								xxxx xxxx	uuuu uuuu
06h	PORTB	PORTB Data Latch when written: PORTB pins when read								xxxx xxxx	uuuu uuuu
07h	PORTC	PORTC Data Latch when written: PORTC pins when read								xxxx xxxx	uuuu uuuu
08h	PORTD	PORTD Data Latch when written: PORTD pins when read								xxxx xxxx	uuuu uuuu
09h	PORTE	—	—	—	—	RE3	RE2	RE1	RE0	---- xxxx	---- uuuu
0Ah(1),(2)	PCLATH	—	—	—	Write Buffer for the upper 5 bits of the Program Counter					---0 0000	---0 0000
0Bh(2)	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
0Dh	PIR2	TMR3GIF	TMR3IF	TMRBIF	TMRAIF	—	—	—	CCP2IF	0000 ---0	0000 ---0
0Eh	TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
0Fh	TMR1H	Holding Register for the Most Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
10h	T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{T1SYNC}$	—	TMR1ON	0000 00-0	uuuu uu-u
11h	TMR2	Timer2 Module Register								0000 0000	0000 0000
12h	T2CON	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000
13h	SSPBUF	Synchronous Serial Port Receive Buffer/Transmit Register								xxxx xxxx	uuuu uuuu
14h	SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
15h	CCPR1L	Capture/Compare/PWM Register 1 (LSB)								xxxx xxxx	uuuu uuuu
16h	CCPR1H	Capture/Compare/PWM Register 1 (MSB)								xxxx xxxx	uuuu uuuu
17h	CCP1CON	—	—	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	--00 0000	--00 0000
18h	RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000x
19h	TXREG	USART Transmit Data Register								0000 0000	0000 0000
1Ah	RCREG	USART Receive Data Register								0000 0000	0000 0000
1Bh	CCPR2L	Capture/Compare/PWM Register 2 (LSB)								xxxx xxxx	uuuu uuuu
1Ch	CCPR2H	Capture/Compare/PWM Register 2 (MSB)								xxxx xxxx	uuuu uuuu
1Dh	CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	--00 0000	--00 0000
1Eh	ADRES	A/D Result Register								xxxx xxxx	uuuu uuuu
1Fh	ADCON0	—	—	CHS3	CHS2	CHS1	CHS0	GO/$\overline{DONE}$	ADON	--00 0000	--00 0000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved.
Shaded locations are unimplemented, read as '0'.

Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8>, whose contents are transferred to the upper byte of the program counter.

2: These registers can be addressed from any bank.

3: Accessible only when SSPM<3:0> = 1001.

TABLE 2-2: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets
Bank 1											
80h(2)	INDF	Addressing this location uses contents of FSR to address data memory (not a physical register)								xxxx xxxx	xxxx xxxx
81h	OPTION_REG	$\overline{RBPU}$	INTEDG	TMR0CS	TMR0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
82h(2)	PCL	Program Counter (PC) Least Significant Byte								0000 0000	0000 0000
83h(2)	STATUS	IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC	C	0001 1xxx	000q quuu
84h(2)	FSR	Indirect Data Memory Address Pointer								xxxx xxxx	uuuu uuuu
85h	TRISA	PORTA Data Direction Register								1111 1111	1111 1111
86h	TRISB	PORTB Data Direction Register								1111 1111	1111 1111
87h	TRISC	PORTC Data Direction Register								1111 1111	1111 1111
88h	TRISD	PORTD Data Direction Register								1111 1111	1111 1111
89h	TRISE	—	—	—	—	'1'	TRISE2	TRISE1	TRISE0	---- 1111	---- 1111
8Ah(1),(2)	PCLATH	—	—	—	Write Buffer for the upper 5 bits of the Program Counter					---0 0000	---0 0000
8Bh(2)	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
8Ch	PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
8Dh	PIE2	TMR3GIE	TMR3IE	TMRBIE	TMRAIE	—	—	—	CCP2IE	0000 ---0	0000 ---0
8Eh	PCON	—	—	—	—	—	—	$\overline{POR}$	$\overline{BOR}$	---- --qq	---- --uu
8Fh	T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/$\overline{DONE}$	T1GVAL	T1GSS1	T1GSS0	0000 0x00	uuuu uxuu
90h	OSCCON	—	—	IRCF1	IRCF0	ICSL	ICSS	—	—	--10 00--	--10 uu--
91h	OSCTUNE	—	—	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0	--00 0000	--00 0000
92h	PR2	Timer2 Period Register								1111 1111	1111 1111
93h	SSPADD	Synchronous Serial Port (I^2C mode) Address Register								0000 0000	0000 0000
93h(3)	SSPMSK	Synchronous Serial Port (I^2C mode) Address Mask Register								1111 1111	1111 1111
94h	SSPSTAT	SMP	CKE	D/$\overline{A}$	P	S	R/$\overline{W}$	UA	BF	0000 0000	0000 0000
95h	WPUB	WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0	1111 1111	1111 1111
96h	IOCB	IOCB7	IOCB6	IOCB5	IOCB4	IOCB3	IOCB2	IOCB1	IOCB0	0000 0000	0000 0000
97h	T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	—	T3SYNC	—	TMR3ON	0000 -0-0	uuuu -u-u
98h	TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
9Ah	TMR3L	Holding Register for the Least Significant Byte of the 16-bit TMR3 Register								xxxx xxxx	uuuu uuuu
9Bh	TMR3H	Holding Register for the Most Significant Byte of the 16-bit TMR3 Register								xxxx xxxx	uuuu uuuu
9Ch	APFCON	—	—	—	—	—	—	SSSEL	CCP2SEL	---- --00	---- --00
9Dh	FVRCON	FVRRDY	FVREN	—	—	CDAFVR1	CDAFVR0	ADFVR1	ADFVR0	x000 0000	x000 0000
9Eh	T3GCON	TMR3GE	T3GPOL	T3GTM	T3GSPM	T3GGO/$\overline{DONE}$	T3GVAL	T3GSS1	T3GSS0	0000 0x00	uuuu uxuu
9Fh	ADCON1	—	ADCS2	ADCS1	ADCS0	—	—	ADREF1	ADREF0	-000 --00	-000 --00

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved.
Shaded locations are unimplemented, read as '0'.

Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8>, whose contents are transferred to the upper byte of the program counter.

2: These registers can be addressed from any bank.

3: Accessible only when SSPM<3:0> = 1001.

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TABLE 2-2: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets
Bank 2											
100h(2)	INDF	Addressing this location uses contents of FSR to address data memory (not a physical register)								xxxx xxxx	xxxx xxxx
101h	TMR0	Timer0 Module Register								0000 0000	0000 0000
102h(2)	PCL	Program Counter's (PC) Least Significant Byte								0000 0000	0000 0000
103h(2)	STATUS	IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC	C	0001 1xxx	000q quuu
104h()	FSR	Indirect Data Memory Address Pointer								xxxx xxxx	uuuu uuuu
105h	TACON	TMRAON	—	TACS	TASE	TAPSA	TAPS2	TAPS1	TAPS0	0-00 0000	0-00 0000
106h	CPSBCON0	CPSBON	CPSBRM	—	—	CPSBRNG1	CPSBRNG0	CPSBOUT	TBXCS	00-- 0000	00-- 0000
107h	CPSBCON1	—	—	—	—	CPSBCH3	CPSBCH2	CPSBCH1	CPSBCH0	---- 0000	---- 0000
108h	CPSACON0	CPSAON	CPSARM	—	—	CPSARNG1	CPSARNG0	CPSAOUT	TAXCS	0--- 0000	0--- 0000
109h	CPSACON1	—	—	—	—	CPSACH3	CPSACH2	CPSACH1	CPSACH0	---- 0000	---- 0000
10Ah(1),(2)	PCLATH	—	—	—	Write Buffer for the upper 5 bits of the Program Counter					---0 0000	---0 0000
10Bh(2)	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
10Ch	PMDATL	Program Memory Read Data Register Low Byte								xxxx xxxx	uuuu uuuu
10Dh	PMADRL	Program Memory Read Address Register Low Byte								xxxx xxxx	uuuu uuuu
10Eh	PMDATH	—	—	Program Memory Read Data Register High Byte						--xx xxxx	--uu uuuu
10Fh	PMADRH	—	—	—	Program Memory Read Address Register High Byte					---x xxxx	---u uuuu
110h	TMRA	TimerA Module Register								0000 0000	0000 0000
111h	TBCON	TMRBON	—	TBCS	TBSE	TBPSA	TBPS2	TBPS1	TBPS0	0-00 0000	0-00 0000
112h	TMRB	TimerB Module Register								0000 0000	0000 0000
113h	DACCON0	DACEN	DACLPS	DACOE	—	DACPSS1	DACPSS0	—	—	000- 00--	000- 00--
114h	DACCON1	—	—	—	DACR4	DACR3	DACR2	DACR1	DACR0	---0 0000	---0 0000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved.
Shaded locations are unimplemented, read as '0'.

Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8>, whose contents are transferred to the upper byte of the program counter.
2: These registers can be addressed from any bank.
3: Accessible only when SSPM<3:0> = 1001.

TABLE 2-2: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets
Bank 3											
180h(2)	INDF	Addressing this location uses contents of FSR to address data memory (not a physical register)								xxxx xxxx	xxxx xxxx
181h	OPTION_REG	$\overline{RBPU}$	INTEDG	TMR0CS	TMR0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
182h(2)	PCL	Program Counter (PC) Least Significant Byte								0000 0000	0000 0000
183h(2)	STATUS	IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC	C	0001 1xxx	000q quuu
184h(2)	FSR	Indirect Data Memory Address Pointer								xxxx xxxx	uuuu uuuu
185h	ANSELA	ANSA7	ANSA6	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	1111 1111	1111 1111
186h	ANSELB	ANSB7	ANSB6	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	1111 1111	1111 1111
187h	ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
188h	ANSELD	ANSD7	ANSD6	ANSD5	ANSD4	ANSD3	ANSD2	ANSD1	ANSD0	1111 1111	1111 1111
189h	ANSELE	—	—	—	—	—	ANSE2	ANSE1	ANSE0	---- -111	---- -111
18Ah(1),(2)	PCLATH	—	—	—	Write Buffer for the upper 5 bits of the Program Counter					---0 0000	---0 0000
18Bh(2)	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
18Ch	PMCON1	—	—	—	—	—	—	—	RD	1--- ---0	1--- ---0
18Dh	—	Reserved								—	—
18Eh	—	Reserved								—	—
18Fh	—	Reserved								—	—

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8>, whose contents are transferred to the upper byte of the program counter.

2: These registers can be addressed from any bank.

3: Accessible only when SSPM<3:0> = 1001.

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2.2.2.1 STATUS Register

The STATUS register, shown in Register 2-1, contains:

- the arithmetic status of the ALU
- the Reset status
- the bank select bits for data memory (SRAM)

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the $\overline{TO}$ and $\overline{PD}$ bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, `CLRF STATUS` will clear the upper three bits and set the Z bit. This leaves the STATUS register as '`000u u1uu`' (where u = unchanged).

It is recommended, therefore, that only `BCF`, `BSF`, `SWAPF` and `MOVWF` instructions are used to alter the STATUS register, because these instructions do not affect any Status bits. For other instructions not affecting any Status bits (Refer to **Section 23.0 "Instruction Set Summary"**).

Note 1: The C and DC bits operate as $\overline{Borrow}$ and $\overline{Digit\ Borrow}$ out bits, respectively, in subtraction.

REGISTER 2-1: STATUS: STATUS REGISTER

R/W-0	R/W-0	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x
IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC(1)	C(1)
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **IRP:** Register Bank Select bit (used for indirect addressing)
`1` = Bank 2, 3 (100h-1FFh)
`0` = Bank 0, 1 (00h-FFh)

bit 6-5 **RP<1:0>:** Register Bank Select bits (used for direct addressing)
`00` = Bank 0 (00h-7Fh)
`01` = Bank 1 (80h-FFh)
`10` = Bank 2 (100h-17Fh)
`11` = Bank 3 (180h-1FFh)

bit 4 **$\overline{TO}$:** Time-out bit
`1` = After power-up, `CLRWDT` instruction or `SLEEP` instruction
`0` = A WDT time-out occurred

bit 3 **$\overline{PD}$:** Power-down bit
`1` = After power-up or by the `CLRWDT` instruction
`0` = By execution of the `SLEEP` instruction

bit 2 **Z:** Zero bit
`1` = The result of an arithmetic or logic operation is zero
`0` = The result of an arithmetic or logic operation is not zero

bit 1 **DC:** Digit Carry/$\overline{Digit\ Borrow}$ bit (`ADDWF`, `ADDLW`, `SUBLW`, `SUBWF` instructions)(1)
`1` = A carry-out from the 4th low-order bit of the result occurred
`0` = No carry-out from the 4th low-order bit of the result

bit 0 **C:** Carry/$\overline{Borrow}$ bit(1) (`ADDWF`, `ADDLW`, `SUBLW`, `SUBWF` instructions)(1)
`1` = A carry-out from the Most Significant bit of the result occurred
`0` = No carry-out from the Most Significant bit of the result occurred

Note 1: For $\overline{Borrow}$, the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand. For rotate (`RRF`, `RLF`) instructions, this bit is loaded with either the high-order or low-order bit of the source register.

2.2.2.2 OPTION Register

The OPTION register, shown in Register 2-2, is a readable and writable register, which contains various control bits to configure:

- Timer0/WDT prescaler
- External RB0/INT interrupt
- Timer0
- Weak pull-ups on PORTB

Note: To achieve a 1:1 prescaler assignment for Timer0, assign the prescaler to the WDT by setting PSA bit of the OPTION register to '1'. Refer to **Section 13.3 "Timer1/3 Prescaler"**.

REGISTER 2-2: OPTION_REG: OPTION REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
$\overline{\text{RBPU}}$	INTEDG	TMR0CS	TMR0SE	PSA	PS2	PS1	PS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **$\overline{\text{RBPU}}$:** PORTB Pull-up Enable bit
1 = PORTB pull-ups are disabled
0 = PORTB pull-ups are enabled by individual bits in the WPUB register

bit 6 **INTEDG:** Interrupt Edge Select bit
1 = Interrupt on rising edge of RB0/INT pin
0 = Interrupt on falling edge of RB0/INT pin

bit 5 **TMR0CS:** Timer0 Clock Source Select bit
1 = Transition on RA4/T0CKI pin
0 = Internal instruction cycle clock (FOSC/4)

bit 4 **TMR0SE:** Timer0 Source Edge Select bit
1 = Increment on high-to-low transition on RA4/T0CKI pin
0 = Increment on low-to-high transition on RA4/T0CKI pin

bit 3 **PSA:** Prescaler Assignment bit
1 = Prescaler is assigned to the WDT
0 = Prescaler is assigned to the Timer0 module

bit 2-0 **PS<2:0>:** Prescaler Rate Select bits

Bit Value	Timer0 Rate	WDT Rate
000	1 : 2	1 : 1
001	1 : 4	1 : 2
010	1 : 8	1 : 4
011	1 : 16	1 : 8
100	1 : 32	1 : 16
101	1 : 64	1 : 32
110	1 : 128	1 : 64
111	1 : 256	1 : 128

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2.2.2.3 PCON Register

The Power Control (PCON) register contains flag bits (refer to Table 3-4) to differentiate between a:

- Power-on Reset ($\overline{POR}$)
- Brown-out Reset ($\overline{BOR}$)

The PCON register bits are shown in Register 2-3.

REGISTER 2-3: PCON: POWER CONTROL REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	R/W-q	R/W-q
—	—	—	—	—	—	$\overline{POR}$	$\overline{BOR}$
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
q = Value depends on condition			

bit 7-2 **Unimplemented:** Read as '0'

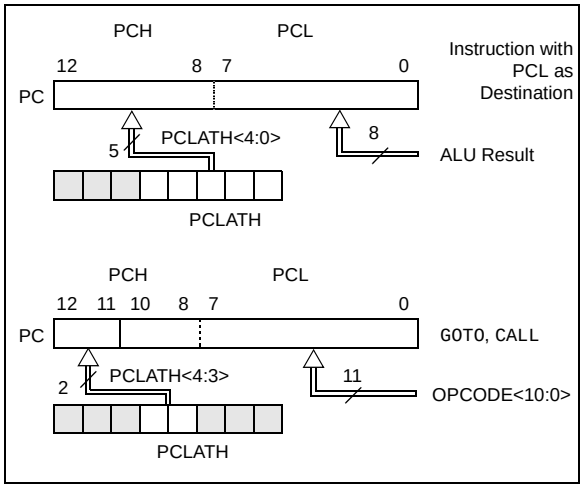
bit 1 **$\overline{POR}$:** Power-on Reset Status bit
1 = No Power-on Reset occurred
0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)

bit 0 **$\overline{BOR}$:** Brown-out Reset Status bit
1 = No Brown-out Reset occurred
0 = A Brown-out Reset occurred (must be set in software after a Power-on Reset or Brown-out Reset occurs)

2.3 PCL and PCLATH

The Program Counter (PC) is 13 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<12:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. Figure 2-2 shows the two situations for the loading of the PC. The upper example in Figure 2-2 shows how the PC is loaded on a write to PCL (PCLATH<4:0> → PCH). The lower example in Figure 2-2 shows how the PC is loaded during a CALL or GOTO instruction (PCLATH<4:3> → PCH).

FIGURE 2-2: LOADING OF PC IN DIFFERENT SITUATIONS



2.3.1 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). When performing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block). Refer to Application Note AN556, *"Implementing a Table Read"* (DS00556).

2.3.2 STACK

All devices have an 8-level x 13-bit wide hardware stack (refer to Figure 2-1). The stack space is not part of either program or data space and the Stack Pointer is not readable or writable. The PC is PUSHed onto the stack when a CALL instruction is executed or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth PUSH overwrites the value that was stored from the first PUSH. The tenth PUSH overwrites the second PUSH (and so on).

Note 1:	There are no Status bits to indicate stack overflow or stack underflow conditions.
2:	There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.

2.4 Program Memory Paging

All devices are capable of addressing a continuous 8K word block of program memory. The CALL and GOTO instructions provide only 11 bits of address to allow branching within any 2K program memory page. When doing a CALL or GOTO instruction, the upper 2 bits of the address are provided by PCLATH<4:3>. When doing a CALL or GOTO instruction, the user must ensure that the page select bits are programmed so that the desired program memory page is addressed. If a return from a CALL instruction (or interrupt) is executed, the entire 13-bit PC is POPed off the stack. Therefore, manipulation of the PCLATH<4:3> bits is not required for the RETURN instructions (which POPs the address from the stack).

Note:	The contents of the PCLATH register are unchanged after a RETURN or RETFIE instruction is executed. The user must rewrite the contents of the PCLATH register for any subsequent subroutine calls or GOTO instructions.

Example 2-1 shows the calling of a subroutine in page 1 of the program memory. This example assumes that PCLATH is saved and restored by the Interrupt Service Routine (if interrupts are used).

EXAMPLE 2-1: CALL OF A SUBROUTINE IN PAGE 1 FROM PAGE 0

```
        ORG 500h
        PAGESEL SUB_P1  ;Select page 1
                        ;(800h-FFFh)
        CALL    SUB1_P1 ;Call subroutine in
        :               ;page 1 (800h-FFFh)
        :
        ORG     900h    ;page 1 (800h-FFFh)
SUB1_P1
        :               ;called subroutine
                        ;page 1 (800h-FFFh)
        :
        RETURN          ;return to
                        ;Call subroutine
                        ;in page 0
                        ;(000h-7FFh)
```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2.5 Indirect Addressing, INDF and FSR Registers

The INDF register is not a physical register. Addressing the INDF register will cause indirect addressing.

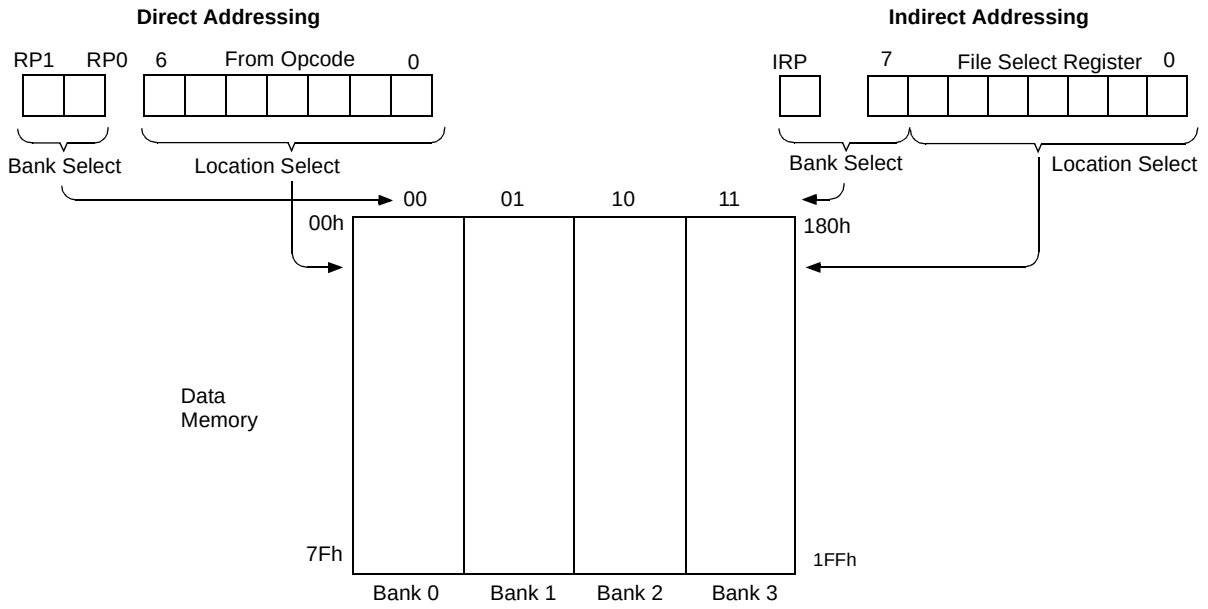
Indirect addressing is possible by using the INDF register. Any instruction using the INDF register actually accesses data pointed to by the File Select Register (FSR). Reading INDF itself indirectly will produce 00h. Writing to the INDF register indirectly results in a no operation (although Status bits may be affected). An effective 9-bit address is obtained by concatenating the 8-bit FSR register and the IRP bit of the STATUS register, as shown in Figure 2-3.

A simple program to clear RAM location 020h-02Fh using indirect addressing is shown in Example 2-2.

EXAMPLE 2-2: INDIRECT ADDRESSING

```
     MOVLW     020h   ;initialize pointer
     MOVWF     FSR    ;to RAM
     BANKISEL  020h
NEXT CLRF      INDF   ;clear INDF register
     INCF      FSR    ;inc pointer
     BTFSS     FSR,4  ;all done?
     GOTO      NEXT   ;no clear next
CONTINUE              ;yes continue
```

FIGURE 2-3: DIRECT/INDIRECT ADDRESSING



Note: For memory map detail, refer to Table 2-2.

NOTES:

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3.0 RESETS

The PIC16(L)F707 differentiates between various kinds of Reset:

a) Power-on Reset (POR)
b) WDT Reset during normal operation
c) WDT Reset during Sleep
d) $\overline{\text{MCLR}}$ Reset during normal operation
e) $\overline{\text{MCLR}}$ Reset during Sleep
f) Brown-out Reset (BOR)

Some registers are not affected in any Reset condition; their status is unknown on POR and unchanged in any other Reset. Most other registers are reset to a "Reset state" on:

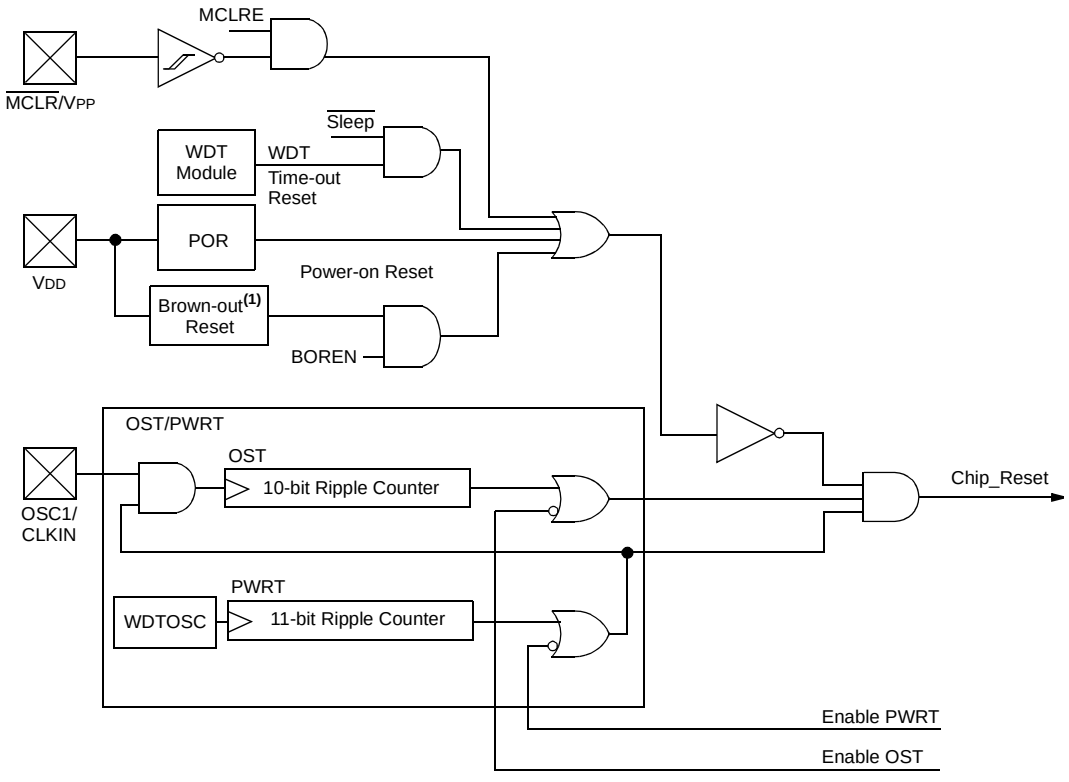
- Power-on Reset (POR)
- $\overline{\text{MCLR}}$ Reset
- $\overline{\text{MCLR}}$ Reset during Sleep
- WDT Reset
- Brown-out Reset (BOR)

Most registers are not affected by a WDT wake-up since this is viewed as the resumption of normal operation. $\overline{\text{TO}}$ and $\overline{\text{PD}}$ bits are set or cleared differently in different Reset situations, as indicated in Table 3-3. These bits are used in software to determine the nature of the Reset.

A simplified block diagram of the On-Chip Reset Circuit is shown in Figure 3-1.

The $\overline{\text{MCLR}}$ Reset path has a noise filter to detect and ignore small pulses. See **Section 25.0 "Electrical Specifications"** for pulse width specifications.

FIGURE 3-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



Note 1: Refer to the Configuration Word Register 1 (Register 8-1).

TABLE 3-1: STATUS BITS AND THEIR SIGNIFICANCE

$\overline{POR}$	$\overline{BOR}$	$\overline{TO}$	$\overline{PD}$	Condition
0	x	1	1	Power-on Reset or LDO Reset
0	x	0	x	Illegal, $\overline{TO}$ is set on $\overline{POR}$
0	x	x	0	Illegal, $\overline{PD}$ is set on $\overline{POR}$
1	0	1	1	Brown-out Reset
1	1	0	1	WDT Reset
1	1	0	0	WDT Wake-up
1	1	u	u	$\overline{MCLR}$ Reset during normal operation
1	1	1	0	$\overline{MCLR}$ Reset during Sleep or interrupt wake-up from Sleep

TABLE 3-2: RESET CONDITION FOR SPECIAL REGISTERS(2)

Condition	Program Counter	STATUS Register	PCON Register
Power-on Reset	0000h	0001 1xxx	---- --0x
$\overline{MCLR}$ Reset during normal operation	0000h	000u uuuu	---- --uu
$\overline{MCLR}$ Reset during Sleep	0000h	0001 0uuu	---- --uu
WDT Reset	0000h	0000 1uuu	---- --uu
WDT Wake-up	PC + 1	uuu0 0uuu	---- --uu
Brown-out Reset	0000h	0001 1uuu	---- --u0
Interrupt Wake-up from Sleep	PC + 1(1)	uuu1 0uuu	---- --uu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0'.

Note 1: When the wake-up is due to an interrupt and global enable bit (GIE) is set, the return address is pushed on the stack and PC is loaded with the interrupt vector (0004h) after execution of PC + 1.

2: If a Status bit is not implemented, that bit will be read as '0'.

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3.1 MCLR

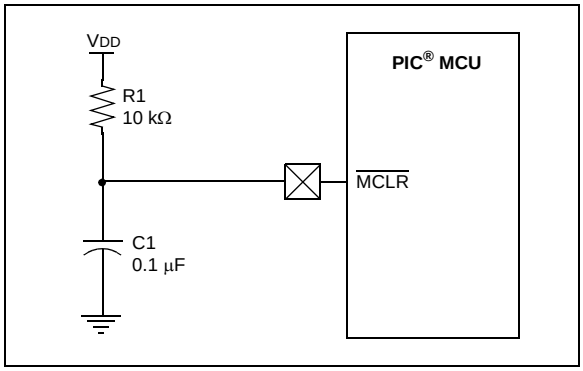
The PIC16(L)F707 has a noise filter in the MCLR Reset path. The filter will detect and ignore small pulses.

It should be noted that a Reset does not drive the MCLR pin low.

Voltages applied to the pin that exceed its specification can result in both MCLR Resets and excessive current beyond the device specification during the ESD event. For this reason, Microchip recommends that the MCLR pin no longer be tied directly to VDD. The use of an RC network, as shown in Figure 3-2, is suggested.

An internal MCLR option is enabled by clearing the MCLRE bit in the Configuration Word register. When MCLRE = 0, the Reset signal to the chip is generated internally. When the MCLRE = 1, the RE3/MCLR pin becomes an external Reset input. In this mode, the RE3/MCLR pin has a weak pull-up to VDD. In-Circuit Serial Programming is not affected by selecting the internal MCLR option.

FIGURE 3-2: RECOMMENDED MCLR CIRCUIT



3.2 Power-on Reset (POR)

The on-chip POR circuit holds the chip in Reset until VDD has reached a high enough level for proper operation. A maximum rise time for VDD is required. See **Section 25.0 "Electrical Specifications"** for details. If the BOR is enabled, the maximum rise time specification does not apply. The BOR circuitry will keep the device in Reset until VDD reaches VBOR (see **Section 3.5 "Brown-Out Reset (BOR)"**).

When the device starts normal operation (exits the Reset condition), device operating parameters (i.e., voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met.

For additional information, refer to Application Note AN607, *"Power-up Trouble Shooting"* (DS00607).

3.3 Power-up Timer (PWRT)

The Power-up Timer provides a fixed 64 ms (nominal) time-out on power-up only, from POR or Brown-out Reset. The Power-up Timer operates from the WDT oscillator. For more information, see **Section 7.3 "Internal Clock Modes"**. The chip is kept in Reset as long as PWRT is active. The PWRT delay allows the VDD to rise to an acceptable level. A Configuration bit, PWRTE, can disable (if set) or enable (if cleared or programmed) the Power-up Timer. The Power-up Timer should be enabled when Brown-out Reset is enabled, although it is not required.

The Power-up Timer delay will vary from chip-to-chip and vary due to:

- VDD variation
- Temperature variation
- Process variation

See DC parameters for details (**Section 25.0 "Electrical Specifications"**).

Note:	The Power-up Timer is enabled by the PWRTE bit in the Configuration Word 1.

3.4 Watchdog Timer (WDT)

The WDT has the following features:

- Shares an 8-bit prescaler with Timer0
- Time-out period is from 17 ms to 2.2 seconds, nominal
- Enabled by a Configuration bit

WDT is cleared under certain conditions described in Table 3-3.

3.4.1 WDT OSCILLATOR

The WDT derives its time base from 31 kHz internal oscillator.

Note:	When the Oscillator Start-up Timer (OST) is invoked, the WDT is held in Reset, because the WDT Ripple Counter is used by the OST to perform the oscillator delay count. When the OST count has expired, the WDT will begin counting (if enabled).

3.4.2 WDT CONTROL

The WDTE bit is located in the Configuration Word Register 1. When set, the WDT runs continuously.

The PSA and PS<2:0> bits of the OPTION register control the WDT period. See **Section 12.0 "Timer0 Module"** for more information.

FIGURE 3-3: WATCHDOG TIMER BLOCK DIAGRAM

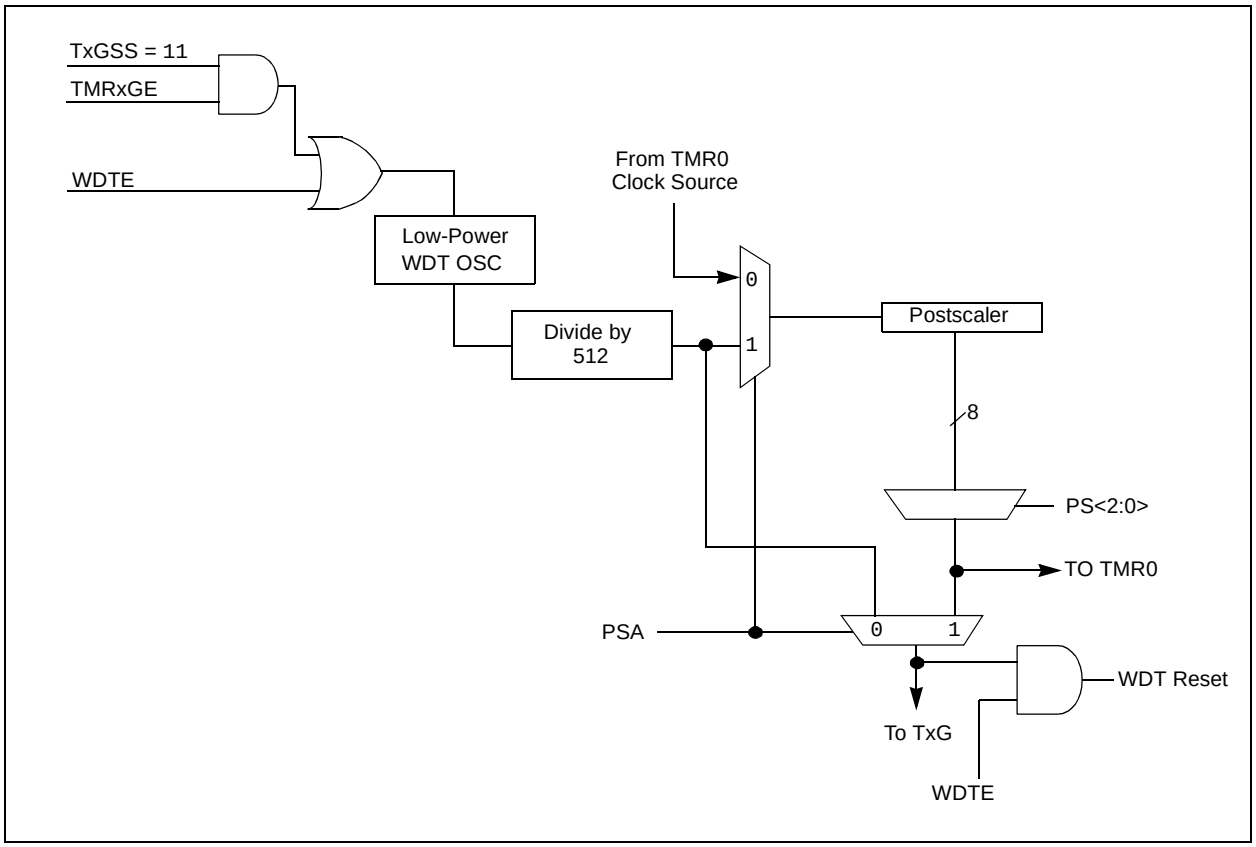


TABLE 3-3: WDT STATUS

Conditions	WDT
WDTE = 0	Cleared
CLRWDT Command	
Exit Sleep + System Clock = EXTRC, INTOSC, EXTCLK	
Exit Sleep + System Clock = XT, HS, LP	Cleared until the end of OST

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3.5 Brown-Out Reset (BOR)

Brown-out Reset is enabled by programming the BOREN<1:0> bits in the Configuration register. The brown-out trip point is selectable from two trip points via the BORV bit in the Configuration register.

Between the POR and BOR, complete voltage range coverage for execution protection can be implemented.

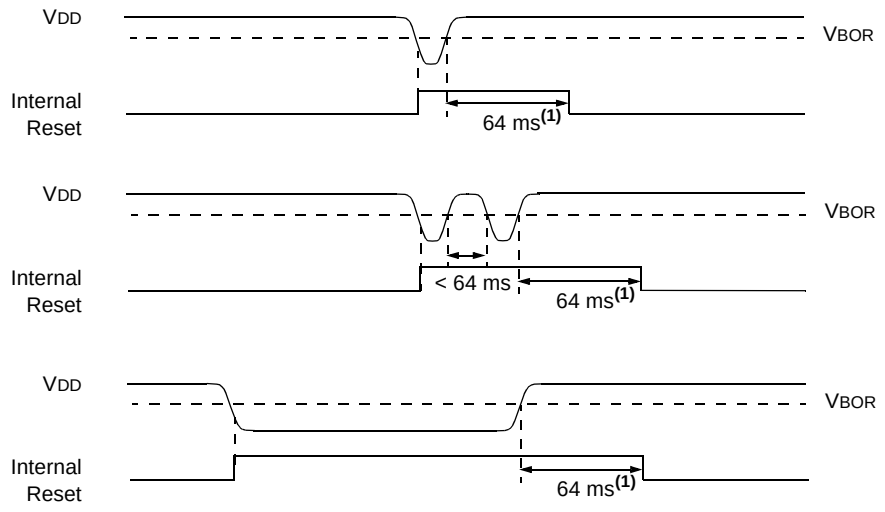
Two bits are used to enable the BOR. When BOREN = 11, the BOR is always enabled. When BOREN = 10, the BOR is enabled, but disabled during Sleep. When BOREN = 0X, the BOR is disabled.

If VDD falls below VBOR for greater than parameter (TBOR) (see **Section 25.0 "Electrical Specifications"**), the brown-out situation will reset the device. This will occur regardless of VDD slew rate. A Reset is not ensured to occur if VDD falls below VBOR for more than parameter (TBOR).

If VDD drops below VBOR while the Power-up Timer is running, the chip will go back into a Brown-out Reset and the Power-up Timer will be re-initialized. Once VDD rises above VBOR, the Power-up Timer will execute a 64 ms Reset.

Note:	When erasing Flash program memory, the BOR is forced to enabled at the minimum BOR setting to ensure that any code protection circuitry is operating properly.

FIGURE 3-4: BROWN-OUT SITUATIONS



Note 1: 64 ms delay only if $\overline{\text{PWRTE}}$ bit is programmed to '0'.

3.6 Time-out Sequence

On power-up, the time-out sequence is as follows: first, PWRT time-out is invoked after POR has expired, then OST is activated after the PWRT time-out has expired. The total time-out will vary based on oscillator configuration and $\overline{\text{PWRTE}}$ bit status. For example, in EC mode with $\overline{\text{PWRTE}}$ bit = 1 (PWRT disabled), there will be no time-out at all. Figure 3-5, Figure 3-6 and Figure 3-7 depict time-out sequences.

Since the time-outs occur from the POR pulse, if $\overline{\text{MCLR}}$ is kept low long enough, the time-outs will expire. Then, bringing $\overline{\text{MCLR}}$ high will begin execution immediately (see Figure 3-6). This is useful for testing purposes or to synchronize more than one PIC16(L)F707 device operating in parallel.

Table 3-2 shows the Reset conditions for some special registers.

3.7 Power Control (PCON) Register

The Power Control (PCON) register has two Status bits to indicate what type of Reset that last occurred.

Bit 0 is $\overline{\text{BOR}}$ (Brown-out Reset). $\overline{\text{BOR}}$ is unknown on Power-on Reset. It must then be set by the user and checked on subsequent Resets to see if $\overline{\text{BOR}}$ = 0, indicating that a brown-out has occurred. The $\overline{\text{BOR}}$ Status bit is a “don’t care” and is not necessarily predictable if the brown-out circuit is disabled (BOREN<1:0> = 00 in the Configuration Word register).

Bit 1 is $\overline{\text{POR}}$ (Power-on Reset). It is a ‘0’ on Power-on Reset and unaffected otherwise. The user must write a ‘1’ to this bit following a Power-on Reset. On a subsequent Reset, if $\overline{\text{POR}}$ is ‘0’, it will indicate that a Power-on Reset has occurred (i.e., VDD may have gone too low).

For more information, see **Section 3.5 “Brown-Out Reset (BOR)”**.

TABLE 3-4: TIME-OUT IN VARIOUS SITUATIONS

Oscillator Configuration	Power-up		Brown-out Reset		Wake-up from Sleep
	$\overline{\text{PWRTE}}$ = 0	$\overline{\text{PWRTE}}$ = 1	$\overline{\text{PWRTE}}$ = 0	$\overline{\text{PWRTE}}$ = 1	
XT, HS, LP	TPWRT + 1024 • TOSC	1024 • TOSC	TPWRT + 1024 • TOSC	1024 • TOSC	1024 • TOSC
RC, EC, INTOSC	TPWRT	—	TPWRT	—	—

FIGURE 3-5: TIME-OUT SEQUENCE ON POWER-UP (DELAYED $\overline{\text{MCLR}}$): CASE 1

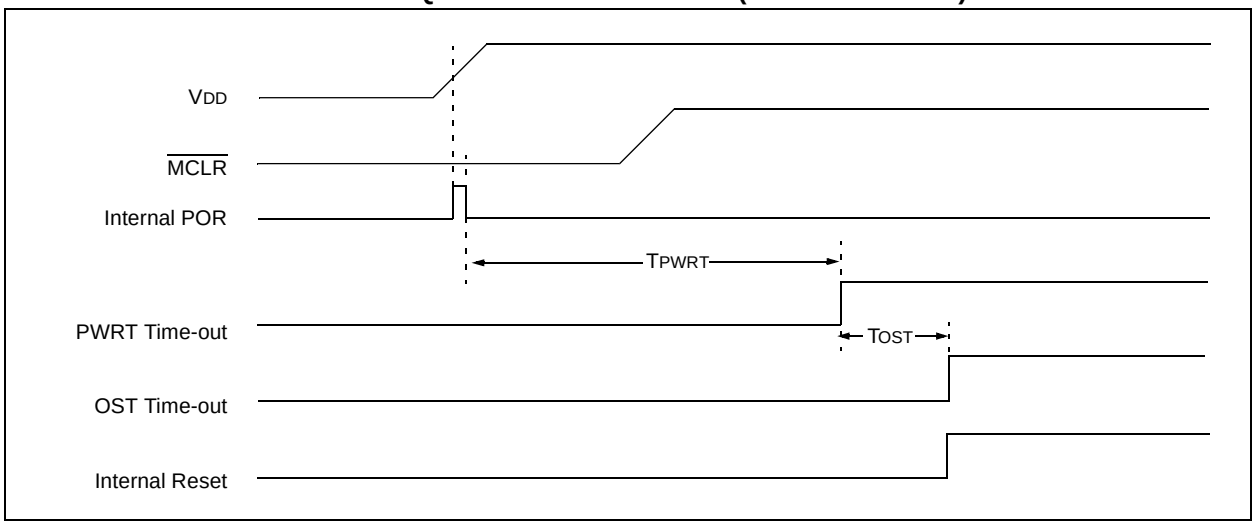


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FIGURE 3-6: TIME-OUT SEQUENCE ON POWER-UP (DELAYED $\overline{MCLR}$): CASE 2

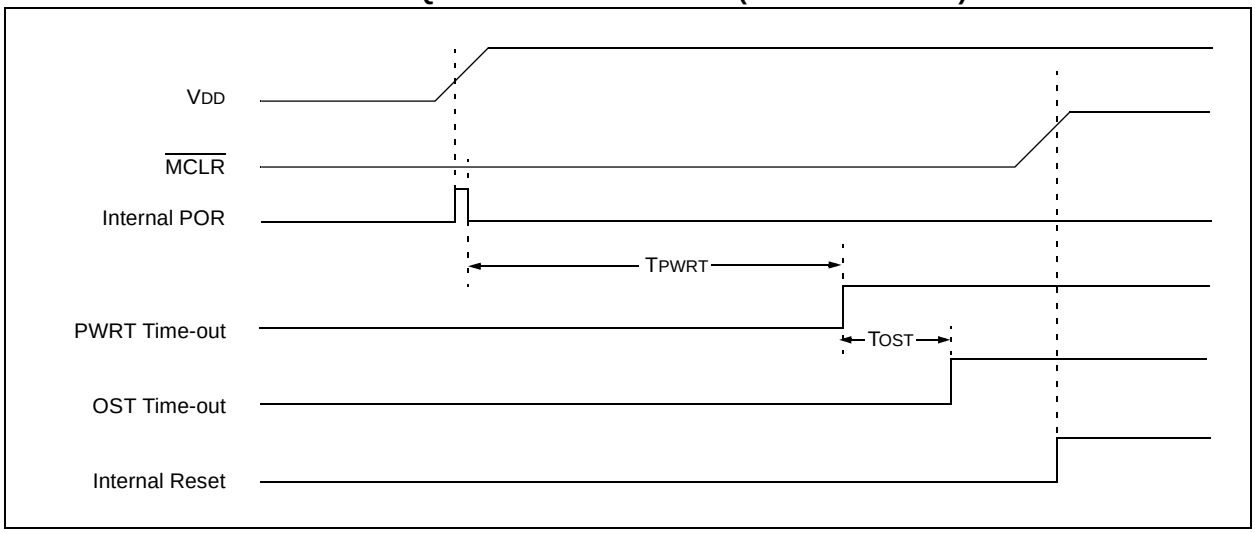

FIGURE 3-7: TIME-OUT SEQUENCE ON POWER-UP ($\overline{MCLR}$ WITH VDD): CASE 3

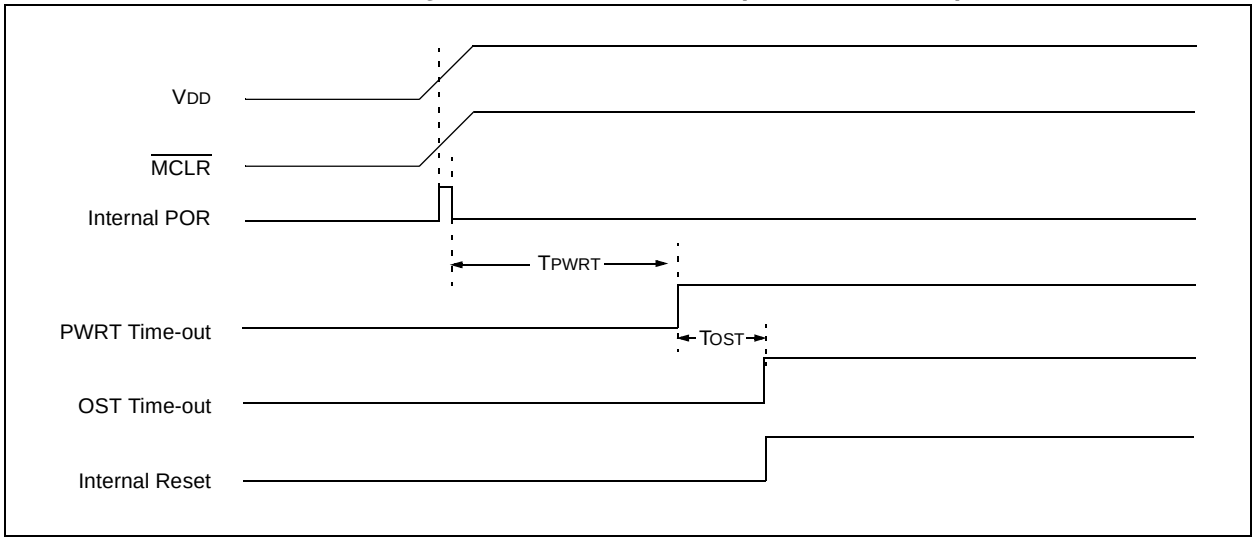

TABLE 3-5: INITIALIZATION CONDITION FOR REGISTERS

Register	Address	Power-on Reset/ Brown-out Reset(1)	$\overline{\text{MCLR}}$ Reset/ WDT Reset	Wake-up from Sleep through Interrupt/Time-out
W	—	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF	00h/80h/ 100h/180h	xxxx xxxx	xxxx xxxx	uuuu uuuu
TMR0	01h/101h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	02h/82h/ 102h/182h	0000 0000	0000 0000	PC + 1(3)
STATUS	03h/83h/ 103h/183h	0001 1xxx	000q quuu(4)	uuuq quuu(4)
FSR	04h/84h/ 104h/184h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTA	05h	xxxx xxxx	xxxx xxxx	uuuu uuuu
PORTB	06h	xxxx xxxx	xxxx xxxx	uuuu uuuu
PORTC	07h	xxxx xxxx	xxxx xxxx	uuuu uuuu
PORTD	08h	xxxx xxxx	xxxx xxxx	uuuu uuuu
PORTE	09h	---- xxxx	---- xxxx	---- uuuu
PCLATH	0Ah/8Ah/ 10Ah/18Ah	---0 0000	---0 0000	---u uuuu
INTCON	0Bh/8Bh/ 10Bh/18Bh	0000 000x	0000 000x	uuuu uuuu(2)
PIR1	0Ch	0000 0000	0000 0000	uuuu uuuu(2)
PIR2	0Dh	0000 ---0	0000 ---0	uuuu ---u(2)
TMR1L	0Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR1H	0Fh	xxxx xxxx	uuuu uuuu	uuuu uuuu
T1CON	10h	0000 00-0	uuuu uu-u	uuuu uu-u
TMR2	11h	0000 0000	0000 0000	uuuu uuuu
T2CON	12h	-000 0000	-000 0000	-uuu uuuu
SSPBUF	13h	xxxx xxxx	xxxx xxxx	uuuu uuuu
SSPCON	14h	0000 0000	0000 0000	uuuu uuuu
CCPR1L	15h	xxxx xxxx	xxxx xxxx	uuuu uuuu
CCPR1H	16h	xxxx xxxx	xxxx xxxx	uuuu uuuu
CCP1CON	17h	--00 0000	--00 0000	--uu uuuu
RCSTA	18h	0000 000x	0000 000x	uuuu uuuu
TXREG	19h	0000 0000	0000 0000	uuuu uuuu
RCREG	1Ah	0000 0000	0000 0000	uuuu uuuu
CCPR2L	1Bh	xxxx xxxx	xxxx xxxx	uuuu uuuu
CCPR2H	1Ch	xxxx xxxx	xxxx xxxx	uuuu uuuu
CCP2CON	1Dh	--00 0000	--00 0000	--uu uuuu
ADRES	1Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: If V_{DD} goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON and/or PIR1 and PIR2 will be affected (to cause wake-up).

3: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

4: See Table 3-2 for Reset value for specific condition.

5: If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.

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TABLE 3-5: INITIALIZATION CONDITION FOR REGISTERS (CONTINUED)

Register	Address	Power-on Reset/ Brown-out Reset(1)	$\overline{\text{MCLR}}$ Reset/ WDT Reset	Wake-up from Sleep through Interrupt/Time-out
ADCON0	1Fh	--00 0000	--00 0000	--uu uuuu
OPTION_REG	81h/181h	1111 1111	1111 1111	uuuu uuuu
TRISA	85h	1111 1111	1111 1111	uuuu uuuu
TRISB	86h	1111 1111	1111 1111	uuuu uuuu
TRISC	87h	1111 1111	1111 1111	uuuu uuuu
TRISD	88h	1111 1111	1111 1111	uuuu uuuu
TRISE	89h	---- 1111	---- 1111	---- uuuu
PIE1	8Ch	0000 0000	0000 0000	uuuu uuuu
PIE2	8Dh	0000 ---0	0000 ---0	uuuu ---u
PCON	8Eh	---- --qq	---- --uu(1,5)	---- --uu
T1GCON	8Fh	0000 0x00	uuuu uxuu	uuuu uxuu
OSCCON	90h	--10 qq--	--10 qq--	--uu qq--
OSCTUNE	91h	--00 0000	--uu uuuu	--uu uuuu
PR2	92h	1111 1111	1111 1111	uuuu uuuu
SSPADD	93h	0000 0000	0000 0000	uuuu uuuu
SSPMSK	93h	1111 1111	1111 1111	uuuu uuuu
SSPSTAT	94h	0000 0000	0000 0000	uuuu uuuu
WPUB	95h	1111 1111	1111 1111	uuuu uuuu
IOCB	96h	0000 0000	0000 0000	uuuu uuuu
T3CON	97h	0000 -0-0	0000 -0-0	uuuu -u-u
TXSTA	98h	0000 -010	0000 -010	uuuu -uuu
SPBRG	99h	0000 0000	0000 0000	uuuu uuuu
TMR3L	9Ah	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR3H	9Bh	xxxx xxxx	uuuu uuuu	uuuu uuuu
APFCON	9Ch	---- --00	---- --00	---- --uu
FVRCON	9Dh	q000 0000	q000 0000	q000 0000
ADCON1	9Fh	-000 --00	-000 --00	-uuu --uu
TACON	105h	0-00 0000	0-00 0000	u-uu uuuu
CPSBCON0	106h	00-- 0000	00-- 0000	uu-- uuuu
CPSBCON1	107h	---- 0000	---- 0000	---- uuuu
CPSACON0	108h	00-- 0000	00-- 0000	uu-- uuuu
CPSACON1	109h	---- 0000	---- 0000	---- uuuu
PMDATL	10Ch	xxxx xxxx	xxxx xxxx	uuuu uuuu
PMADRL	10Dh	xxxx xxxx	xxxx xxxx	uuuu uuuu
PMDATH	10Eh	--xx xxxx	--xx xxxx	--uu uuuu
PMADRH	10Fh	---x xxxx	---x xxxx	---u uuuu
TMRA	110h	0000 0000	0000 0000	uuuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: If VDD goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON and/or PIR1 and PIR2 will be affected (to cause wake-up).

3: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

4: See Table 3-2 for Reset value for specific condition.

5: If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.

TABLE 3-5: INITIALIZATION CONDITION FOR REGISTERS (CONTINUED)

Register	Address	Power-on Reset/ Brown-out Reset(1)	$\overline{\text{MCLR}}$ Reset/ WDT Reset	Wake-up from Sleep through Interrupt/Time-out
TBCON	111h	0-00 0000	0-00 0000	u-uu uuuu
TMRB	112h	0000 0000	0000 0000	uuuu uuuu
DACCON0	113h	000- 00--	000- 00--	uuu- uu--
DACCON1	114h	---0 0000	---0 0000	---u uuuu
ANSELA	185h	1111 1111	1111 1111	uuuu uuuu
ANSELB	186h	1111 1111	1111 1111	uuuu uuuu
ANSELC	187h	1111 1111	1111 1111	uuuu uuuu
ANSELD	188h	1111 1111	1111 1111	uuuu uuuu
ANSELE	189h	---- -111	---- -111	---- -uuu
PMCON1	18Ch	1--- ---0	1--- ---0	u--- ---u

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: If VDD goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON and/or PIR1 and PIR2 will be affected (to cause wake-up).

3: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

4: See Table 3-2 for Reset value for specific condition.

5: If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.

TABLE 3-6: SUMMARY OF REGISTERS ASSOCIATED WITH RESETS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets(1)
STATUS	IRP	RP1	RP0	$\overline{\text{TO}}$	$\overline{\text{PD}}$	Z	DC	C	0001 1xxx	000q quuu
PCON	—	—	—	—	—	—	$\overline{\text{POR}}$	$\overline{\text{BOR}}$	---- --qq	---- --uu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition. Shaded cells are not used by Resets.

Note 1: Other (non Power-up) Resets include $\overline{\text{MCLR}}$ Reset and Watchdog Timer Reset during normal operation.

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4.0 INTERRUPTS

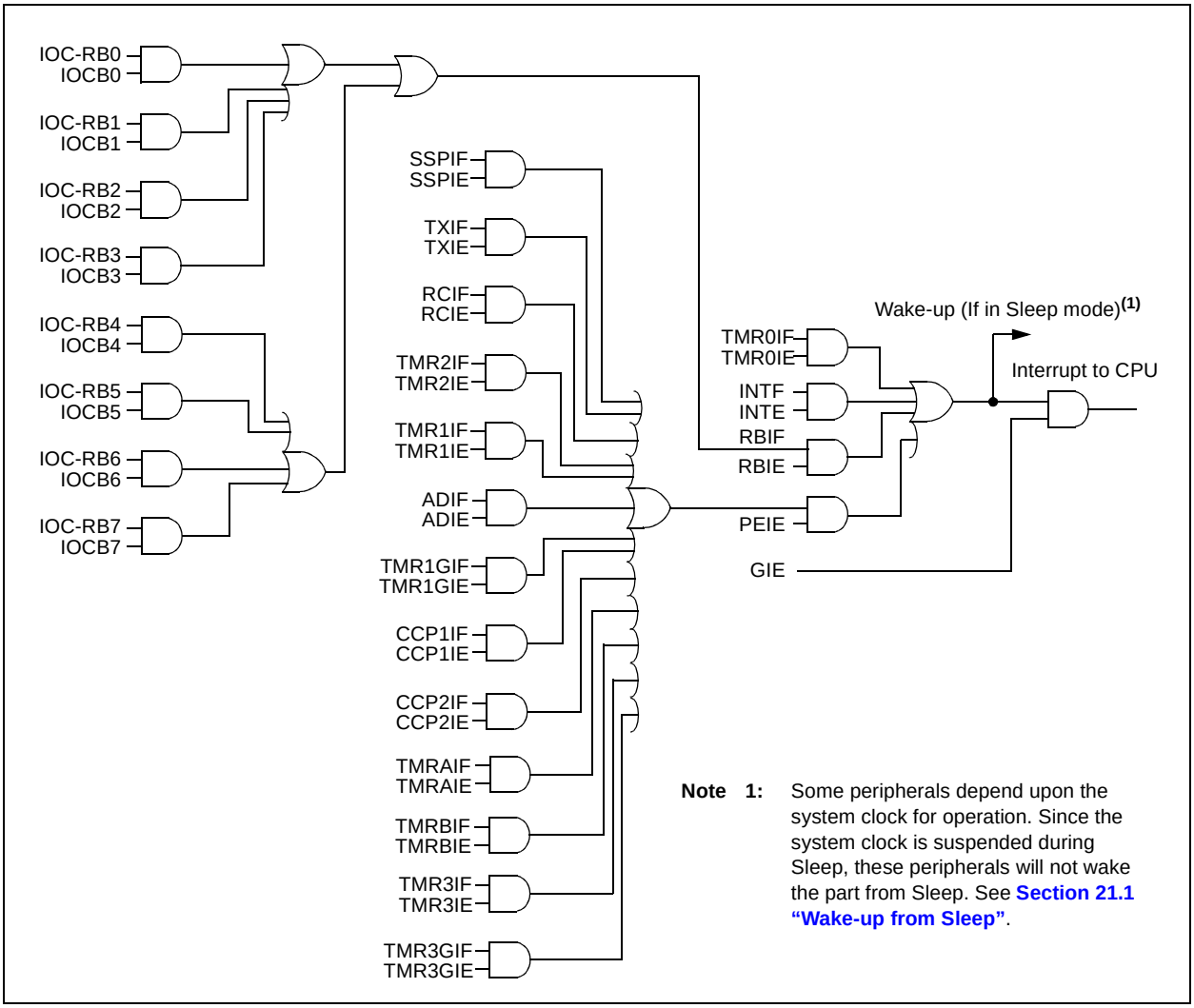
The PIC16(L)F707 device family features an interruptible core, allowing certain events to preempt normal program flow. An Interrupt Service Routine (ISR) is used to determine the source of the interrupt and act accordingly. Some interrupts can be configured to wake the MCU from Sleep mode.

The PIC16F707 family has 16 interrupt sources, differentiated by corresponding interrupt enable and flag bits:

- Timer0 Overflow Interrupt
- External Edge Detect on INT Pin Interrupt
- PORTB Change Interrupt
- Timer1 Gate Interrupt
- A/D Conversion Complete Interrupt
- AUSART Receive Interrupt
- AUSART Transmit Interrupt
- SSP Event Interrupt
- CCP1 Event Interrupt
- Timer2 Match with PR2 Interrupt
- Timer1 Overflow Interrupt
- CCP2 Event Interrupt
- TimerA Overflow Interrupt
- TimerB Overflow Interrupt
- Timer3 Overflow Interrupt
- Timer3 Gate Interrupt

A block diagram of the interrupt logic is shown in Figure 4-1.

FIGURE 4-1: INTERRUPT LOGIC



Note 1: Some peripherals depend upon the system clock for operation. Since the system clock is suspended during Sleep, these peripherals will not wake the part from Sleep. See **Section 21.1 "Wake-up from Sleep"**.

4.1 Operation

Interrupts are disabled upon any device Reset. They are enabled by setting the following bits:

- GIE bit of the INTCON register
- Interrupt enable bit(s) for the specific interrupt event(s)
- PEIE bit of the INTCON register (if the interrupt enable bit of the interrupt event is contained in the PIE1 and PIE2 registers)

The INTCON, PIR1 and PIR2 registers record individual interrupts via interrupt flag bits. Interrupt flag bits will be set, regardless of the status of the GIE, PEIE and individual Interrupt Enable bits.

The following events happen when an interrupt event occurs while the GIE bit is set:

- Current prefetched instruction is flushed
- GIE bit is cleared
- Current Program Counter (PC) is pushed onto the stack
- PC is loaded with the interrupt vector 0004h

The ISR determines the source of the interrupt by polling the interrupt flag bits. The interrupt flag bits must be cleared before exiting the ISR to avoid repeated interrupts. Because the GIE bit is cleared, any interrupt that occurs while executing the ISR will be recorded through its interrupt flag, but will not cause the processor to redirect to the interrupt vector.

The `RETFIE` instruction exits the ISR by popping the previous address from the stack and setting the GIE bit.

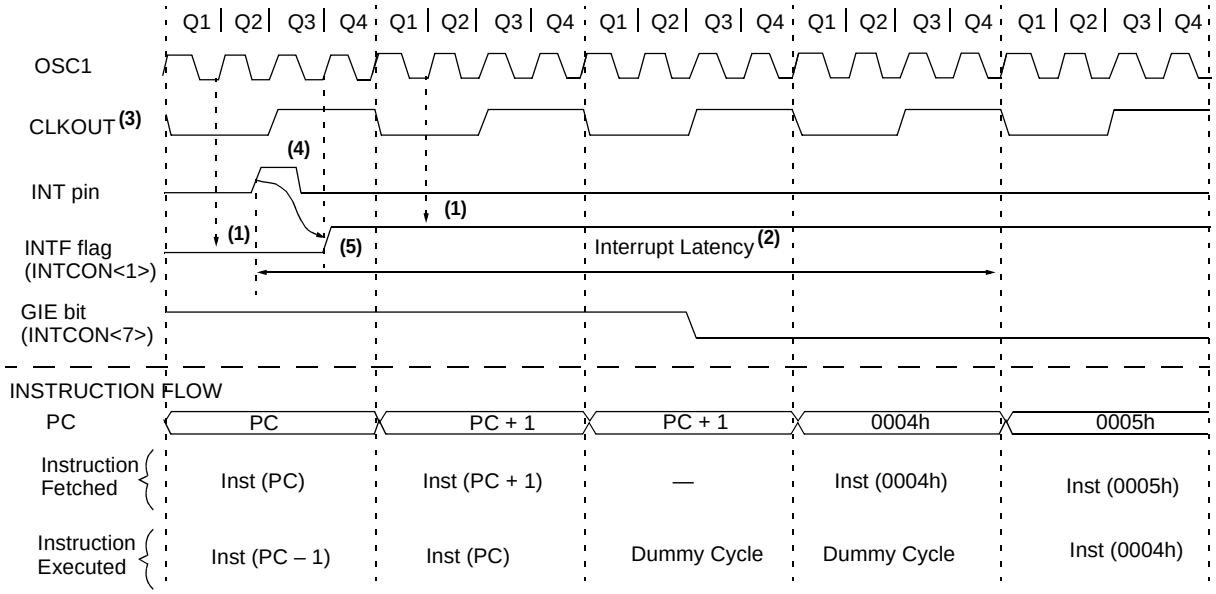
For additional information on a specific interrupt's operation, refer to its peripheral chapter.

Note 1:	Individual interrupt flag bits are set, regardless of the state of any other enable bits.
2:	All interrupts will be ignored while the GIE bit is cleared. Any interrupt occurring while the GIE bit is clear will be serviced when the GIE bit is set again.

4.2 Interrupt Latency

Interrupt latency is defined as the time from when the interrupt event occurs to the time code execution at the interrupt vector begins. The latency for synchronous interrupts is 3 instruction cycles. For asynchronous interrupts, the latency is 3 to 4 instruction cycles, depending on when the interrupt occurs. See Figure 4-2 for timing details.

FIGURE 4-2: INT PIN INTERRUPT TIMING



Note 1: INTF flag is sampled here (every Q1).

2: Asynchronous interrupt latency = 3-4 TCY. Synchronous latency = 3 TCY, where TCY = instruction cycle time. Latency is the same whether Inst (PC) is a single cycle or a 2-cycle instruction.

3: CLKOUT is available only in INTOSC and RC Oscillator modes.

4: For minimum width of INT pulse, refer to AC specifications in **Section 25.0 "Electrical Specifications"**.

5: INTF is enabled to be set any time during the Q4-Q1 cycles.

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4.3 Interrupts During Sleep

Some interrupts can be used to wake from Sleep. To wake from Sleep, the peripheral must be able to operate without the system clock. The interrupt source must have the appropriate interrupt enable bit(s) set prior to entering Sleep.

On waking from Sleep, if the GIE bit is also set, the processor will branch to the interrupt vector. Otherwise, the processor will continue executing instructions after the SLEEP instruction. The instruction directly after the SLEEP instruction will always be executed before branching to the ISR. Refer to **Section 21.0 "Power-Down Mode (Sleep)"** for more details.

4.4 INT Pin

The external interrupt, INT pin, causes an asynchronous, edge-triggered interrupt. The INTEDG bit of the OPTION register determines on which edge the interrupt will occur. When the INTEDG bit is set, the rising edge will cause the interrupt. When the INTEDG bit is clear, the falling edge will cause the interrupt. The INTF bit of the INTCON register will be set when a valid edge appears on the INT pin. If the GIE and INTE bits are also set, the processor will redirect program execution to the interrupt vector. This interrupt is disabled by clearing the INTE bit of the INTCON register.

4.5 Context Saving

When an interrupt occurs, only the return PC value is saved to the stack. If the ISR modifies or uses an instruction that modifies key registers, their values must be saved at the beginning of the ISR and restored when the ISR completes. This prevents instructions following the ISR from using invalid data. Examples of key registers include the W, STATUS, FSR and PCLATH registers.

Note:	The microcontroller does not normally require saving the PCLATH register. However, if computed GOTO's are used, the PCLATH register must be saved at the beginning of the ISR and restored when the ISR is complete to ensure correct program flow.

The code shown in Example 4-1 can be used to do the following.

- Save the W register
- Save the STATUS register
- Save the PCLATH register
- Execute the ISR program
- Restore the PCLATH register
- Restore the STATUS register
- Restore the W register

Since most instructions modify the W register, it must be saved immediately upon entering the ISR. The SWAPF instruction is used when saving and restoring the W and STATUS registers because it will not affect any bits in the STATUS register. It is useful to place W_TEMP in shared memory because the ISR cannot predict which bank will be selected when the interrupt occurs.

The processor will branch to the interrupt vector by loading the PC with 0004h. The PCLATH register will remain unchanged. This requires the ISR to ensure that the PCLATH register is set properly before using an instruction that causes PCLATH to be loaded into the PC. See **Section 2.3 "PCL and PCLATH"** for details on PC operation.

EXAMPLE 4-1: SAVING W, STATUS AND PCLATH REGISTERS IN RAM

```
MOVWF   W_TEMP              ;Copy W to W_TEMP register
SWAPF   STATUS,W            ;Swap status to be saved into W
                            ;Swaps are used because they do not affect the status bits
BANKSEL STATUS_TEMP         ;Select regardless of current bank
MOVWF   STATUS_TEMP         ;Copy status to bank zero STATUS_TEMP register
MOVF    PCLATH,W            ;Copy PCLATH to W register
MOVWF   PCLATH_TEMP         ;Copy W register to PCLATH_TEMP
:
:(ISR)                      ;Insert user code here
:
BANKSEL STATUS_TEMP         ;Select regardless of current bank
MOVF    PCLATH_TEMP,W       ;
MOVWF   PCLATH              ;Restore PCLATH
SWAPF   STATUS_TEMP,W       ;Swap STATUS_TEMP register into W
                            ;(sets bank to original state)
MOVWF   STATUS              ;Move W into STATUS register
SWAPF   W_TEMP,F            ;Swap W_TEMP
SWAPF   W_TEMP,W            ;Swap W_TEMP into W
```

4.5.1 INTCON REGISTER

The INTCON register is a readable and writable register, which contains the various enable and flag bits for TMR0 register overflow, PORTB change and external RB0/INT/SEG0 pin interrupts.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 4-1: INTCON: INTERRUPT CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x
GIE	PEIE	TMR0IE	INTE	RBIE(1)	TMR0IF(2)	INTF	RBIF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **GIE:** Global Interrupt Enable bit
1 = Enables all unmasked interrupts
0 = Disables all interrupts

bit 6 **PEIE:** Peripheral Interrupt Enable bit
1 = Enables all unmasked peripheral interrupts
0 = Disables all peripheral interrupts

bit 5 **TMR0IE:** Timer0 Overflow Interrupt Enable bit
1 = Enables the Timer0 interrupt
0 = Disables the Timer0 interrupt

bit 4 **INTE:** RB0/INT External Interrupt Enable bit
1 = Enables the RB0/INT external interrupt
0 = Disables the RB0/INT external interrupt

bit 3 **RBIE:** PORTB Change Interrupt Enable bit(1)
1 = Enables the PORTB change interrupt
0 = Disables the PORTB change interrupt

bit 2 **TMR0IF:** Timer0 Overflow Interrupt Flag bit(2)
1 = TMR0 register has overflowed (must be cleared in software)
0 = TMR0 register did not overflow

bit 1 **INTF:** RB0/INT External Interrupt Flag bit
1 = The RB0/INT external interrupt occurred (must be cleared in software)
0 = The RB0/INT external interrupt did not occur

bit 0 **RBIF:** PORTB Change Interrupt Flag bit
1 = When at least one of the PORTB general purpose I/O pins changed state (must be cleared in software)
0 = None of the PORTB general purpose I/O pins have changed state

Note 1: The appropriate bits in the IOCB register must also be set.
2: TMR0IF bit is set when Timer0 rolls over. Timer0 is unchanged on Reset and should be initialized before clearing TMR0IF bit.

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4.5.2 PIE1 REGISTER

The PIE1 register contains the interrupt enable bits, as shown in Register 4-2.

Note: Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.

REGISTER 4-2: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **TMR1GIE:** Timer1 Gate Interrupt Enable bit
1 = Enable the Timer1 gate acquisition complete interrupt
0 = Disable the Timer1 gate acquisition complete interrupt

bit 6 **ADIE:** A/D Converter (ADC) Interrupt Enable bit
1 = Enables the ADC interrupt
0 = Disables the ADC interrupt

bit 5 **RCIE:** USART Receive Interrupt Enable bit
1 = Enables the USART receive interrupt
0 = Disables the USART receive interrupt

bit 4 **TXIE:** USART Transmit Interrupt Enable bit
1 = Enables the USART transmit interrupt
0 = Disables the USART transmit interrupt

bit 3 **SSPIE:** Synchronous Serial Port (SSP) Interrupt Enable bit
1 = Enables the SSP interrupt
0 = Disables the SSP interrupt

bit 2 **CCP1IE:** CCP1 Interrupt Enable bit
1 = Enables the CCP1 interrupt
0 = Disables the CCP1 interrupt

bit 1 **TMR2IE:** TMR2 to PR2 Match Interrupt Enable bit
1 = Enables the Timer2 to PR2 match interrupt
0 = Disables the Timer2 to PR2 match interrupt

bit 0 **TMR1IE:** Timer1 Overflow Interrupt Enable bit
1 = Enables the Timer1 overflow interrupt
0 = Disables the Timer1 overflow interrupt

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4.5.3 PIE2 REGISTER

The PIE2 register contains the interrupt enable bits, as shown in Register 4-3.

Note: Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.

REGISTER 4-3: PIE2 – PERIPHERAL INTERRUPT ENABLE REGISTER 2

R/W-0	R/W-0	R/W-0	R/W-0	U-0	U-0	U-0	R/W-0
TMR3GIE	TMR3IE	TMRBIE	TMRAIE	—	—	—	CCP2IE
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7 **TMR3GIE:** Timer3 Gate Interrupt Flag bit
1 = Enable the Timer3 gate acquisition complete interrupt
0 = Disable the Timer3 gate acquisition complete interrupt

bit 6 **TMR3IE:** Timer3 Overflow Interrupt Enable bit
1 = Enables the Timer3 overflow interrupt
0 = Disables the Timer3 overflow interrupt

bit 5 **TMRBIE:** TimerB Overflow Interrupt Enable bit
1 = Enables the TimerB interrupt
0 = Disables the TimerB interrupt

bit 4 **TMRAIE:** TimerA Overflow Interrupt Enable bit
1 = Enables the TimerA interrupt
0 = Disables the TimerA interrupt

bit 3-1 **Unimplemented:** Read as '0'

bit 0 **CCP2IE:** CCP2 Interrupt Enable bit
1 = Enables the CCP2 interrupt
0 = Disables the CCP2 interrupt

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4.5.4 PIR1 REGISTER

The PIR1 register contains the interrupt flag bits, as shown in Register 4-4.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 4-4: PIR1: PERIPHERAL INTERRUPT REQUEST REGISTER 1

R/W-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0
TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **TMR1GIF:** Timer1 Gate Interrupt Flag bit
1 = Timer1 gate is inactive
0 = Timer1 gate is active

bit 6 **ADIF:** A/D Converter Interrupt Flag bit
1 = A/D conversion complete (must be cleared in software)
0 = A/D conversion has not completed or has not been started

bit 5 **RCIF:** USART Receive Interrupt Flag bit
1 = The USART receive buffer is full (cleared by reading RCREG)
0 = The USART receive buffer is not full

bit 4 **TXIF:** USART Transmit Interrupt Flag bit
1 = The USART transmit buffer is empty (cleared by writing to TXREG)
0 = The USART transmit buffer is full

bit 3 **SSPIF:** Synchronous Serial Port (SSP) Interrupt Flag bit
1 = The Transmission/Reception is complete (must be cleared in software)
0 = Waiting to Transmit/Receive

bit 2 **CCP1IF:** CCP1 Interrupt Flag bit
Capture mode:
1 = A Timer1 register capture occurred (must be cleared in software)
0 = No Timer1 register capture occurred
Compare mode:
1 = A Timer1 register compare match occurred (must be cleared in software)
0 = No Timer1 register compare match occurred
PWM mode:
Unused in this mode

bit 1 **TMR2IF:** Timer2 to PR2 Interrupt Flag bit
1 = A Timer2 to PR2 match occurred (must be cleared in software)
0 = No Timer2 to PR2 match occurred

bit 0 **TMR1IF:** Timer1 Overflow Interrupt Flag bit
1 = The Timer1 register overflowed (must be cleared in software)
0 = The Timer1 register did not overflow

4.5.5 PIR2 REGISTER

The PIR2 register contains the interrupt flag bits, as shown in Register 4-5.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 4-5: PIR2: PERIPHERAL INTERRUPT REQUEST REGISTER 2

R/W-0	R/W-0	R/W-0	R/W-0	U-0	U-0	U-0	R/W-0
TMR3GIF	TMR3IF	TMRBIF	TMRAIF	—	—	—	CCP2IF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7 **TMR3GIF:** Timer3 Gate Interrupt Flag bit
1 = Timer3 gate is inactive
0 = Timer3 gate is active

bit 6 **TMR3IF:** Timer3 Overflow Interrupt Flag bit
1 = Timer3 register overflowed (must be cleared in software)
0 = Timer3 register did not overflow

bit 5 **TMRBIF:** TimerB Overflow Interrupt Flag bit
1 = TimerB register has overflowed (must be cleared in software)
0 = TimerB register did not overflow

bit 4 **TMRAIF:** TimerA Overflow Interrupt Flag bit
1 = TimerA register has overflowed (must be cleared in software)
0 = TimerA register did not overflow

bit 3-1 **Unimplemented:** Read as ‘0’

bit 0 **CCP2IF:** CCP2 Interrupt Flag bit
Capture Mode
1 = A Timer1 register capture occurred (must be cleared in software)
0 = No Timer1 register capture occurred
Compare Mode
1 = A Timer1 register compare match occurred (must be cleared in software)
0 = No Timer1 register compare match occurred
PWM Mode
Unused in this mode

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TABLE 4-1: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPTS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
OPTION_REG	$\overline{\text{RBPU}}$	INTEDG	TMR0CS	TMR0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIE2	TMR3GIE	TMR3IE	TMRBIE	TMRAIE	—	—	—	CCP2IE	0000 ---0	0000 ---0
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIR2	TMR3GIF	TMR3IF	TMRBIF	TMRAIF	—	—	—	CCP2IF	0000 ---0	0000 ---0

Legend: - = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by interrupts.

NOTES:

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5.0 LOW DROPOUT (LDO) VOLTAGE REGULATOR

The PIC16F707 device has an internal Low Dropout Regulator (LDO) which provides operation above 3.6V. The LDO regulates a voltage for the internal device logic while permitting the VDD and I/O pins to operate at a higher voltage. There is no user enable/disable control available for the LDO, it is always active. The PIC16LF707 operates at a maximum VDD of 3.6V and does not incorporate an LDO.

A device I/O pin may be configured as the LDO voltage output, identified as the VCAP pin. Although not required, an external low-ESR capacitor may be connected to the VCAP pin for additional regulator stability.

The VCAPEN<1:0> bits of Configuration Word 2 determines which pin is assigned as the VCAP pin. Refer to Table 5-1.

TABLE 5-1: VCAPEN<1:0> SELECT BITS

VCAPEN<1:0>	Pin
00	RA0
01	RA5
10	RA6
11	No VCAP

On power-up, the external capacitor will load the LDO voltage regulator. To prevent erroneous operation, the device is held in Reset while a constant current source charges the external capacitor. After the cap is fully charged, the device is released from Reset. For more information on recommended capacitor values and the constant current rate, refer to the LDO Regulator Characteristics Table in **Section 25.0 "Electrical Specifications"**.

TABLE 5-2: SUMMARY OF CONFIGURATION WORD WITH LDO

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
CONFIG2	13:8	—	—	—	—	—	—	—	—	78
	7:0	—	—	VCAPEN1(1)	VCAPEN0(1)	—	—	—	—	

Legend: — = unimplemented locations read as '0'. Shaded cells are not used by LDO.

Note 1: PIC16F707 only.

NOTES:

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6.0 I/O PORTS

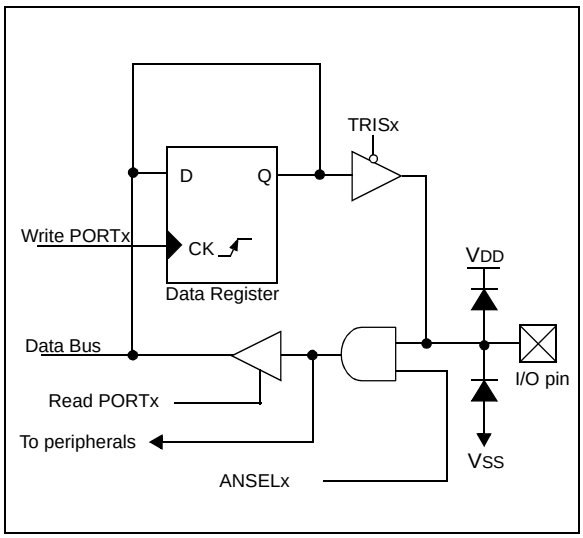
There are thirty-five general purpose I/O pins available. Depending on which peripherals are enabled, some or all of the pins may not be available as general purpose I/O. In general, when a peripheral is enabled, the associated pin may not be used as a general purpose I/O pin.

Each port has two registers for its operation. These registers are:

- TRISx registers (data direction register)
- PORTx registers (port read/write register)

Ports with analog functions also have an ANSELx register which can disable the digital input and save power. A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 6-1.

FIGURE 6-1: GENERIC I/O PORT OPERATION



6.1 Alternate Pin Function

The Alternate Pin Function Control (APFCON) register is used to steer specific peripheral input and output functions between different pins. The APFCON register is shown in Register 6-1. For this device family, the following functions can be moved between different pins.

- $\overline{SS}$ (Slave Select)
- CCP2

REGISTER 6-1: APFCON: ALTERNATE PIN FUNCTION CONTROL REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0	R/W-0
—	—	—	—	—	—	SSSEL	CCP2SEL
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-2 **Unimplemented:** Read as ‘0’.

bit 1 **SSSEL:** $\overline{SS}$ Input Pin Selection bit
0 = $\overline{SS}$ function is on RA5/AN4/CPS7/$\overline{SS}$/VCAP
1 = $\overline{SS}$ function is on RA0/AN0/$\overline{SS}$/VCAP

bit 0 **CCP2SEL:** CCP2 Input/Output Pin Selection bit
0 = CCP2 function is on RC1/T1OSI/CCP2
1 = CCP2 function is on RB3/CCP2

6.2 PORTA and TRISA Registers

PORTA is a 8-bit wide, bidirectional port. The corresponding data direction register is TRISA (Register 6-3). Setting a TRISA bit (= 1) will make the corresponding PORTA pin an input (i.e., disable the output driver). Clearing a TRISA bit (= 0) will make the corresponding PORTA pin an output (i.e., enables output driver and puts the contents of the output latch on the selected pin). Example 6-1 shows how to initialize PORTA.

Reading the PORTA register (Register 6-2) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch.

The TRISA register (Register 6-3) controls the PORTA pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISA register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

Note:	The ANSELA register must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'.

EXAMPLE 6-1: INITIALIZING PORTA

```
BANKSEL PORTA       ;
CLRF    PORTA       ;Init PORTA
BANKSEL ANSELA      ;
CLRF    ANSELA      ;digital I/O
BANKSEL TRISA       ;
MOVLW   0Ch         ;Set RA<3:2> as inputs
MOVWF   TRISA       ;and set RA<7:4,1:0>
                    ;as outputs
```

REGISTER 6-2: PORTA: PORTA REGISTER

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **RA<7:0>**: PORTA I/O Pin bits
1 = Port pin is > V_{IH}
0 = Port pin is < V_{IL}

REGISTER 6-3: TRISA: PORTA TRI-STATE REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **TRISA<7:0>:** PORTA Tri-State Control bits
1 = PORTA pin configured as an input (tri-stated)
0 = PORTA pin configured as an output

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6.2.1 ANSELA REGISTER

The ANSELA register (Register 6-4) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELA bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELA bits has no affect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

REGISTER 6-4: ANSELA: PORTA ANALOG SELECT REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANSA7	ANSA6	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	'0' = Bit is cleared
-n = Value at POR	'1' = Bit is set	x = Bit is unknown

bit 7-0 **ANSA<7:0>**: Analog Select between Analog or Digital Function on pins RA<7:0>, respectively
0 = Digital I/O. Pin is assigned to port or digital special function.
1 = Analog input. Pin is assigned as analog input(1). Digital Input buffer disabled.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

6.2.2 PIN DESCRIPTIONS

Each PORTA pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the A/D Converter (ADC), refer to the appropriate section in this data sheet.

6.2.2.1 RA0/AN0/VCAP

The RA0 pin is configurable to function as one of the following:

- General purpose I/O
- Analog input for the A/D
- Slave Select input for the SSP(1)
- Voltage Regulator Capacitor pin (PIC16F707 only)

Note 1: $\overline{SS}$ pin location may be selected as RA5 or RA0.

6.2.2.2 RA1/AN1/CPSA0

The RA1 pin is configurable to function as one of the following:

- General purpose I/O
- Analog input for the A/D
- Capacitive sensing input

6.2.2.3 RA2/AN2/CPSA1/DACOUT

The RA2 pin is configurable to function as one of the following:

- General purpose I/O
- Analog input for the A/D
- Capacitive sensing input
- DAC Output

6.2.2.4 RA3/AN3/VREF+/CPSA2

The RA3 pin is configurable to function as one of the following:

- General purpose I/O
- Analog input for the A/D
- Voltage Reference input for the A/D
- Capacitive sensing input

6.2.2.5 RA4/CPSA3/T0CKI/TACKI

The RA4 pin is configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input
- Clock input for Timer0
- Clock input for TimerA

The Timer0 clock input function works independently of any TRIS register setting. Effectively, if TRISA4 = 0, the PORTA4 register bit will output to the pad and clock Timer0 at the same time.

6.2.2.6 RA5/AN4/CPSA4/$\overline{SS}$/VCAP

The RA5 pin is configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input
- Analog input for the A/D
- Slave Select input for the SSP(1)
- Voltage Regulator Capacitor pin (PIC16F707 only)

Note 1: $\overline{SS}$ pin location may be selected as RA5 or RA0.

6.2.2.7 RA6/CPSB1/OSC2/CLKOUT/VCAP

The RA6 pin is configurable to function as one of the following:

- General purpose I/O
- Crystal/resonator connection
- Clock Output
- Voltage Regulator Capacitor pin (PIC16F707 only)
- Capacitive sensing input

6.2.2.8 RA7/CPSB0/OSC1/CLKIN

The RA7 pin is configurable to function as one of the following:

- General purpose I/O
- Crystal/resonator connection
- Clock Input
- Capacitive sensing input.

TABLE 6-1: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ADCON0	—	—	CHS3	CHS2	CHS1	CHS0	GO/$\overline{DONE}$	ADON	--00 0000	--00 0000
ADCON1	—	ADCS2	ADCS1	ADCS0	—	—	ADREF1	ADREF0	-000 --00	-000 --00
ANSELA	ANSA7	ANSA6	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	1111 1111	1111 1111
APFCON	—	—	—	—	—	—	SSSEL	CCP2SEL	---- --00	---- --00
CPSACON0	CPSAON	CPSARM	—	—	CPSARNG1	CPSARNG0	CPSAOUT	TAXCS	00-- 0000	00-- 0000
CPSACON1	—	—	—	—	CPSACH3	CPSACH2	CPSACH1	CPSACH0	---- 0000	---- 0000
CPSBCON0	CPSBON	CPSBRM	—	—	CPSBRNG1	CPSBRNG0	CPSBOUT	TBXCS	00-- 0000	00-- 0000
CPSBCON1	—	—	—	—	CPSBCH3	CPSBCH2	CPSBCH1	CPSBCH0	---- 0000	---- 0000
CONFIG2(1)	—	—	VCAPEN1	VCAPEN0	—	—	—	—	—	—
OPTION_REG	$\overline{RBPU}$	INTEDG	TMR0CS	TMR0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	xxxx xxxx	xxxx xxxx
SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111
TACON	TMRAON	—	TACS	TASE	TAPSA	TAPS2	TAPS1	TAPS0	0-00 0000	0-00 0000
DACCON0	DACEN	DACLPS	DACOE	—	DACPSS1	DACPSS0	—	—	000- 00--	000- 00--

Legend: x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by PORTA.
Note 1: PIC16F707 only.

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6.3 PORTB and TRISB Registers

PORTB is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISB (Register 6-6). Setting a TRISB bit (=1) will make the corresponding PORTB pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISB bit (= 0) will make the corresponding PORTB pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 6-2 shows how to initialize PORTB.

Reading the PORTB register (Register 6-5) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch.

The TRISB register (Register 6-6) controls the PORTB pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISB register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'. Example 6-2 shows how to initialize PORTB.

EXAMPLE 6-2: INITIALIZING PORTB

```
BANKSEL  PORTB       ;
CLRF     PORTB       ;Init PORTB
BANKSEL  ANSELB
CLRF     ANSELB      ;Make RB<7:0> digital
BANKSEL  TRISB       ;
MOVLW    B'11110000' ;Set RB<7:4> as inputs
                     ;and RB<3:0> as outputs
MOVWF    TRISB       ;
```

Note:	The ANSELB register must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'.

6.3.1 ANSELB REGISTER

The ANSELB register (Register 6-9) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELB bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELB bits has no affect on digital output functions. A pin with TRIS clear and ANSELB set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

6.3.2 WEAK PULL-UPS

Each of the PORTB pins has an individually configurable internal weak pull-up. Control bits WPUB<7:0> enable or disable each pull-up (see Register 6-7). Each weak pull-up is automatically turned off when the port pin is configured as an output. All pull-ups are disabled on a Power-on Reset by the $\overline{\text{RBPU}}$ bit of the OPTION register.

6.3.3 INTERRUPT-ON-CHANGE

All of the PORTB pins are individually configurable as an interrupt-on-change pin. Control bits IOCB<7:0> enable or disable the interrupt function for each pin. Refer to Register 6-8. The interrupt-on-change feature is disabled on a Power-on Reset.

For enabled interrupt-on-change pins, the present value is compared with the old value latched on the last read of PORTB to determine which bits have changed or mismatched the old value. The 'mismatch' outputs of the last read are OR'd together to set the PORTB Change Interrupt Flag bit (RBIF) in the INTCON register.

This interrupt can wake the device from Sleep. The user, in the Interrupt Service Routine, clears the interrupt by:

a) Any read or write of PORTB. This will end the mismatch condition.
b) Clear the flag bit RBIF.

A mismatch condition will continue to set flag bit RBIF. Reading or writing PORTB will end the mismatch condition and allow flag bit RBIF to be cleared. The latch holding the last read value is not affected by a $\overline{\text{MCLR}}$ nor Brown-out Reset. After these Resets, the RBIF flag will continue to be set if a mismatch is present.

Note:	When a pin change occurs at the same time as a read operation on PORTB, the RBIF flag will always be set. If multiple PORTB pins are configured for the interrupt-on-change, the user may not be able to identify which pin changed state.

REGISTER 6-5: PORTB: PORTB REGISTER

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **RB<7:0>**: PORTB I/O Pin bit
1 = Port pin is > V_{IH}
0 = Port pin is < V_{IL}

REGISTER 6-6: TRISB: PORTB TRI-STATE REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **TRISB<7:0>:** PORTB Tri-State Control bit
1 = PORTB pin configured as an input (tri-stated)
0 = PORTB pin configured as an output

REGISTER 6-7: WPUB: WEAK PULL-UP PORTB REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **WPUB<7:0>**: Weak Pull-up Register bits
1 = Pull-up enabled
0 = Pull-up disabled

Note 1: Global $\overline{\text{RBPU}}$ bit of the OPTION register must be cleared for individual pull-ups to be enabled.
2: The weak pull-up device is automatically disabled if the pin is in configured as an output.

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REGISTER 6-8: IOCB: INTERRUPT-ON-CHANGE PORTB REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
IOCB7	IOCB6	IOCB5	IOCB4	IOCB3	IOCB2	IOCB1	IOCB0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **IOCB<7:0>:** Interrupt-on-Change PORTB Control bits
1 = Interrupt-on-change enabled
0 = Interrupt-on-change disabled

REGISTER 6-9: ANSELB: PORTB ANALOG SELECT REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANSB7	ANSB6	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **ANSB<7:0>**: Analog Select between Analog or Digital Function on Pins RB<7:0>, respectively
0 = Digital I/O. Pin is assigned to port or digital special function.
1 = Analog input. Pin is assigned as analog input(1). Digital input buffer disabled.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

6.3.4 PIN DESCRIPTIONS

Each PORTB pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the SSP, I^2C or interrupts, refer to the appropriate section in this data sheet.

6.3.4.1 RB0/AN12/CPSB8/INT

These pins are configurable to function as one of the following:

- General purpose I/O
- Analog input for the ADC
- Capacitive sensing input
- External edge triggered interrupt

6.3.4.2 RB1/AN10/CPSB9

These pins are configurable to function as one of the following:

- General purpose I/O
- Analog input for the ADC
- Capacitive sensing input

6.3.4.3 RB2/AN8/CPSB10

These pins are configurable to function as one of the following:

- General purpose I/O
- Analog input for the ADC
- Capacitive sensing input

6.3.4.4 RB3/AN9/CPSB11/CCP2

These pins are configurable to function as one of the following:

- General purpose I/O
- Analog input for the ADC
- Capacitive sensing input
- Capture 2 input, Compare 2 output, and PWM2 output

Note: CCP2 pin location may be selected as RB3 or RC1.

6.3.4.5 RB4/AN11/CPSB12

These pins are configurable to function as one of the following:

- General purpose I/O
- Analog input for the ADC
- Capacitive sensing input

6.3.4.6 RB5/AN13/CPSB13/T1G/T3CKI

These pins are configurable to function as one of the following:

- General purpose I/O
- Analog input for the ADC
- Capacitive sensing input
- Timer1 gate input
- Timer3 clock input

6.3.4.7 RB6/ICSPCLK/CPSB14

These pins are configurable to function as one of the following:

- General purpose I/O
- In-Circuit Serial Programming clock
- Capacitive sensing input

6.3.4.8 RB7/ICSPDAT/CPSB15

These pins are configurable to function as one of the following:

- General purpose I/O
- In-Circuit Serial Programming data
- Capacitive sensing input

TABLE 6-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTB

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ADCON0	—	—	CHS3	CHS2	CHS1	CHS0	GO/$\overline{\text{DONE}}$	ADON	--00 0000	--00 0000
ANSELB	ANSB7	ANSB6	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	1111 1111	1111 1111
APFCON	—	—	—	—	—	—	SSSEL	CCP2SEL	---- --00	---- --00
CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	--00 0000	--00 0000
CPSBCON0	CPSBON	CPSBRM	—	—	CPSBRNG1	CPSBRNG0	CPSBOUT	TBXCS	00-- 0000	00-- 0000
CPSBCON1	—	—	—	—	CPSBCH3	CPSBCH2	CPSBCH1	CPSBCH0	---- 0000	---- 0000
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000X
IOCB	IOCB7	IOCB6	IOCB5	IOCB4	IOCB3	IOCB2	IOCB1	IOCB0	0000 0000	0000 0000
OPTION_REG	$\overline{\text{RBPU}}$	INTEDG	TMR0CS	TMR0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	xxxx xxxx
T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	—	T3SYNC	—	TMR3ON	0000 -0-0	0000 -0-0
T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1$\overline{\text{GGO}}$/$\overline{\text{DONE}}$	T1GVAL	T1GSS1	T1GSS0	0000 0x00	uuuu uxuu
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
WPUB	WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTB.

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6.4 PORTC and TRISC Registers

PORTC is a 8-bit wide, bidirectional port. The corresponding data direction register is TRISC (Register 6-11). Setting a TRISC bit (= 1) will make the corresponding PORTC pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISC bit (= 0) will make the corresponding PORTC pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 6-3 shows how to initialize PORTC.

Reading the PORTC register (Register 6-10) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch.

The TRISC register (Register 6-11) controls the PORTC pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISC register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

EXAMPLE 6-3: INITIALIZING PORTC

```
BANKSEL PORTC        ;
CLRF    PORTC        ;Init PORTC
BANKSEL TRISC        ;
MOVLW   B'00001100'  ;Set RC<3:2> as inputs
MOVWF   TRISC        ;and set RC<7:4,1:0>
                     ;as outputs
```

The location of the CCP2 function is controlled by the CCP2SEL bit in the APFCON register (see Register 6-1).

REGISTER 6-10: PORTC: PORTC REGISTER

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **RC<7:0>**: PORTC General Purpose I/O Pin bits
1 = Port pin is > VIH
0 = Port pin is < VIL

REGISTER 6-11: TRISC: PORTC TRI-STATE REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **TRISC<7:0>:** PORTC Tri-State Control bits
1 = PORTC pin configured as an input (tri-stated)
0 = PORTC pin configured as an output

6.4.1 ANSELC REGISTER

The ANSELC register (Register 6-12) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELC bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELC bits has no affect on digital output functions. A pin with TRIS clear and ANSELC set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note:	The ANSELC register must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'.

REGISTER 6-12: ANSELC: PORTC ANALOG SELECT REGISTER

R/W-1	R/W-1	R/W-1	U-0	U-0	R/W-1	R/W-1	R/W-1
ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5 **ANSC<7:5>**: Analog Select between Analog or Digital Function on Pins RC<7:5>, respectively
0 = Digital I/O. Pin is assigned to port or digital special function.
1 = Analog input. Pin is assigned as analog input(1). Digital input buffer disabled.

bit 4-3 **Unimplemented:** Read as '0'

bit 2-0 **ANSC<2:0>**: Analog Select between Analog or Digital Function on Pins RC<2:0>, respectively
0 = Digital I/O. Pin is assigned to port or digital special function.
1 = Analog input. Pin is assigned as analog input(1). Digital input buffer disabled.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

6.4.2 PIN DESCRIPTIONS

Each PORTC pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the SSP, I^2C or interrupts, refer to the appropriate section in this data sheet.

6.4.2.1 RC0/T1OSO/T1CKI/CPSB2

These pins are configurable to function as one of the following:

- General purpose I/O
- Timer1 oscillator output
- Timer1 clock input
- Capacitive sensing input

6.4.2.2 RC1/T1OSI/CCP2/CPSB3

These pins are configurable to function as one of the following:

- General purpose I/O
- Timer1 oscillator input
- Capture 2 input, Compare 2 output, and PWM2 output
- Capacitive sensing input

Note:	CCP2 pin location may be selected as RB3 or RC1.

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6.4.2.3 RC2/CCP1/CPSB4/TBCKI

These pins are configurable to function as one of the following:

- General purpose I/O
- Capture 1 input, Compare 1 output, and PWM1 output
- Capacitive sensing input
- TimerB Clock input

6.4.2.4 RC3/SCK/SCL

These pins are configurable to function as one of the following:

- General purpose I/O
- SPI clock
- I^2C™ clock

6.4.2.5 RC4/SDI/SDA

These pins are configurable to function as one of the following:

- General purpose I/O
- SPI data input
- I^2C data I/O

6.4.2.6 RC5/SDO/CPSA9

These pins are configurable to function as one of the following:

- General purpose I/O
- SPI data output
- Capacitive sensing input

6.4.2.7 RC6/TX/CK/CPSA10

These pins are configurable to function as one of the following:

- General purpose I/O
- Asynchronous serial output
- Synchronous clock I/O
- Capacitive sensing input

6.4.2.8 RC7/RX/DT/CPSA11

These pins are configurable to function as one of the following:

- General purpose I/O
- Asynchronous serial input
- Synchronous serial data I/O
- Capacitive sensing input

TABLE 6-3: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
APFCON	—	—	—	—	—	—	SSSEL	CCP2SEL	---- --00	---- --00
CCP1CON	—	—	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	--00 0000	--00 0000
CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	--00 0000	--00 0000
CPSACON0	CPSAON	CPSARM	—	—	CPSARNG1	CPSARNG0	CPSAOUT	TAXCS	00-- 0000	00-- 0000
CPSACON1	—	—	—	—	CPSACH3	CPSACH2	CPSACH1	CPSACH0	---- 0000	---- 0000
CPSBCON0	CPSBON	CPSBRM	—	—	CPSBRNG1	CPSBRNG0	CPSBOUT	TBXCS	00-- 0000	00-- 0000
CPSBCON1	—	—	—	—	CPSBCH3	CPSBCH2	CPSBCH1	CPSBCH0	---- 0000	---- 0000
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	xxxx xxxx	xxxx xxxx
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000x
SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
SSPSTAT	SMP	CKE	D/$\overline{A}$	P	S	R/$\overline{W}$	UA	BF	0000 0000	0000 0000
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{T1SYNC}$	—	TMR1ON	0000 00-0	uuuu uu-u
TBCON	TMRBON	—	TBCS	TBSE	TBPSA	TBPS2	TBPS1	TBPS0	0-00 0000	0-00 0000
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTC.

6.5 PORTD and TRISD Registers

PORTD is a 8-bit wide, bidirectional port. The corresponding data direction register is TRISD (Register 6-14). Setting a TRISD bit (= 1) will make the corresponding PORTD pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISD bit (= 0) will make the corresponding PORTD pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 6-4 shows how to initialize PORTD.

Reading the PORTD register (Register 6-13) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch.

The TRISD register (Register 6-14) controls the PORTD pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISD register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

EXAMPLE 6-4: INITIALIZING PORTD

```
BANKSEL PORTD        ;
CLRF    PORTD        ;Init PORTD
BANKSEL ANSELD
CLRF    ANSELD       ;Make PORTD digital
BANKSEL TRISD        ;
MOVLW   B'00001100'  ;Set RD<3:2> as inputs
MOVWF   TRISD        ;and set RD<7:4,1:0>
                     ;as outputs
```

6.5.1 ANSELD REGISTER

The ANSELD register (Register 6-15) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELD bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELD bits has no affect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note:	The ANSELD register must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'.

REGISTER 6-13: PORTD: PORTD REGISTER

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **RD<7:0>**: PORTD General Purpose I/O Pin bits
1 = Port pin is > V_{IH}
0 = Port pin is < V_{IL}

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REGISTER 6-14: TRISD: PORTD TRI-STATE REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **TRISD<7:0>:** PORTD Tri-State Control bits
1 = PORTD pin configured as an input (tri-stated)
0 = PORTD pin configured as an output

REGISTER 6-15: ANSELD: PORTD ANALOG SELECT REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANSD7	ANSD6	ANSD5	ANSD4	ANSD3	ANSD2	ANSD1	ANSD0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **ANSD<7:0>**: Analog Select between Analog or Digital Function on Pins RD<7:0>, respectively
0 = Digital I/O. Pin is assigned to port or digital special function.
1 = Analog input. Pin is assigned as analog input**(1)**. Digital input buffer disabled.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

6.5.2 PIN DESCRIPTIONS

Each PORTD pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the SSP, I^2C or interrupts, refer to the appropriate section in this data sheet.

6.5.2.1 RD0/CPSB5/T3G

These pins are configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input
- Timer3 Gate input

6.5.2.2 RD1/CPSB6

These pins are configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input

6.5.2.3 RD2/CPSB7

These pins are configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input

6.5.2.4 RD3/CPSA8

These pins are configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input

6.5.2.5 RD4/CPSA12

These pins are configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input

6.5.2.6 RD5/CPSA13

These pins are configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input

6.5.2.7 RD6/CPSA14

These pins are configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input

6.5.2.8 RD7/CPSA15

These pins are configurable to function as one of the following:

- General purpose I/O
- Capacitive sensing input

TABLE 6-4: SUMMARY OF REGISTERS ASSOCIATED WITH PORTD

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELD	ANSD7	ANSD6	ANSD5	ANSD4	ANSD3	ANSD2	ANSD1	ANSD0	1111 1111	1111 1111
CPSACON0	CPSAON	CPSARM	—	—	CPSARNG1	CPSARNG0	CPSAOUT	TAXCS	00-- 0000	00-- 0000
CPSACON1	—	—	—	—	CPSACH3	CPSACH2	CPSACH1	CPSACH0	---- 0000	---- 0000
CPSBCON0	CPSBON	CPSBRM	—	—	CPSBRNG1	CPSBRNG0	CPSBOUT	TBXCS	00-- 0000	00-- 0000
CPSBCON1	—	—	—	—	CPSBCH3	CPSBCH2	CPSBCH1	CPSBCH0	---- 0000	---- 0000
T3GCON	TMR3GE	T3GPOL	T3GTM	T3GSPM	T3GGO/ DONE	T3GVAL	T3GSS1	T3GSS0	0000 0x00	uuuu uxuu
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	xxxx xxxx	xxxx xxxx
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by PORTD.

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6.6 PORTE and TRISE Registers

PORTE is a 4-bit wide, bidirectional port. The corresponding data direction register is TRISE. Setting a TRISE bit (= 1) will make the corresponding PORTE pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISE bit (= 0) will make the corresponding PORTE pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). The exception is RE3, which is input only and its TRIS bit will always read as '1'. Example 6-5 shows how to initialize PORTE.

Reading the PORTE register (Register 6-16) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch. RE3 reads '0' when MCLRE = 1.

The TRISE register (Register 6-17) controls the PORTE pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISE register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

Note:	The ANSELE register must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'.

EXAMPLE 6-5: INITIALIZING PORTE

```
BANKSEL PORTE        ;
CLRF    PORTE        ;Init PORTE
BANKSEL ANSELE       ;
CLRF    ANSELE       ;digital I/O
BANKSEL TRISE        ;
MOVLW   B'00001100'  ;Set RE<2> as an input
MOVWF   TRISE        ;and set RE<1:0>
                     ;as outputs
```

6.6.1 ANSELE REGISTER

The ANSELE register (Register 6-18) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELE bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELE bits has no affect on digital output functions. A pin with TRIS clear and ANSELE set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

REGISTER 6-16: PORTE: PORTE REGISTER

U-0	U-0	U-0	U-0	R-x	R/W-x	R/W-x	R/W-x
—	—	—	—	RE3	RE2	RE1	RE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-4 **Unimplemented**: Read as '0'

bit 3-0 **RE<3:0>**: PORTE I/O Pin bits
1 = Port pin is > V_{IH}
0 = Port pin is < V_{IL}

REGISTER 6-17: TRISE: PORTE TRI-STATE REGISTER

U-0	U-0	U-0	U-0	R-1	R/W-1	R/W-1	R/W-1
—	—	—	—	TRISE3	TRISE2	TRISE1	TRISE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-4 **Unimplemented**: Read as '0'

bit 3 **TRISE3:** RE3 Port Tri-state Control bit
This bit is always '1' as RE3 is an input only

bit 2-0 **TRISE<2:0>:** RE<2:0> Tri-State Control bits(1)
1 = PORTE pin configured as an input (tri-stated)
0 = PORTE pin configured as an output

REGISTER 6-18: ANSELE: PORTE ANALOG SELECT REGISTER

U-0	U-0	U-0	U-0	U-0	R/W-1	R/W-1	R/W-1
—	—	—	—	—	ANSE2	ANSE1	ANSE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-3 **Unimplemented:** Read as '0'

bit 2-0 **ANSE<2:0>**: Analog Select between Analog or Digital Function on Pins RE<2:0>, respectively
0 = Digital I/O. Pin is assigned to port or digital special function.
1 = Analog input. Pin is assigned as analog input(1). Digital input buffer disabled.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

6.6.2 PIN DESCRIPTIONS

Each PORTE pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the SSP, I^2C or interrupts, refer to the appropriate section in this data sheet.

6.6.2.1 RE0/AN5/CPSA5

These pins are configurable to function as one of the following:

- General purpose I/O
- Analog input for the ADC
- Capacitive sensing input

6.6.2.2 RE1/AN6/CPSA6

These pins are configurable to function as one of the following:

- General purpose I/O
- Analog input for the ADC
- Capacitive sensing input

6.6.2.3 RE2/AN7/CPSA7

These pins are configurable to function as one of the following:

- General purpose I/O
- Analog input for the ADC
- Capacitive sensing input

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6.6.2.4 RE3/$\overline{MCLR}$/VPP

These pins are configurable to function as one of the following:

- General purpose input
- Master Clear Reset with weak pull-up
- Programming voltage reference input

TABLE 6-5: SUMMARY OF REGISTERS ASSOCIATED WITH PORTE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ADCON0	—	—	CHS3	CHS2	CHS1	CHS0	GO/$\overline{DONE}$	ADON	--00 0000	--00 0000
ANSELE	—	—	—	—	—	ANSE2	ANSE1	ANSE0	---- -111	---- -111
CPSACON0	CPSAON	CPSARM	—	—	CPSARNG1	CPSARNG0	CPSAOUT	TAXCS	00-- 0000	00-- 0000
CPSACON1	—	—	—	—	CPSACH3	CPSACH2	CPSACH1	CPSACH0	---- 0000	---- 0000
PORTE	—	—	—	—	RE3	RE2	RE1	RE0	---- xxxx	---- xxxx
TRISE	—	—	—	—	TRISE3(1)	TRISE2	TRISE1	TRISE0	---- 1111	---- 1111

Legend: x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by PORTE.
Note 1: This bit is always '1' as RE3 is input only.

NOTES:

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7.0 OSCILLATOR MODULE

7.1 Overview

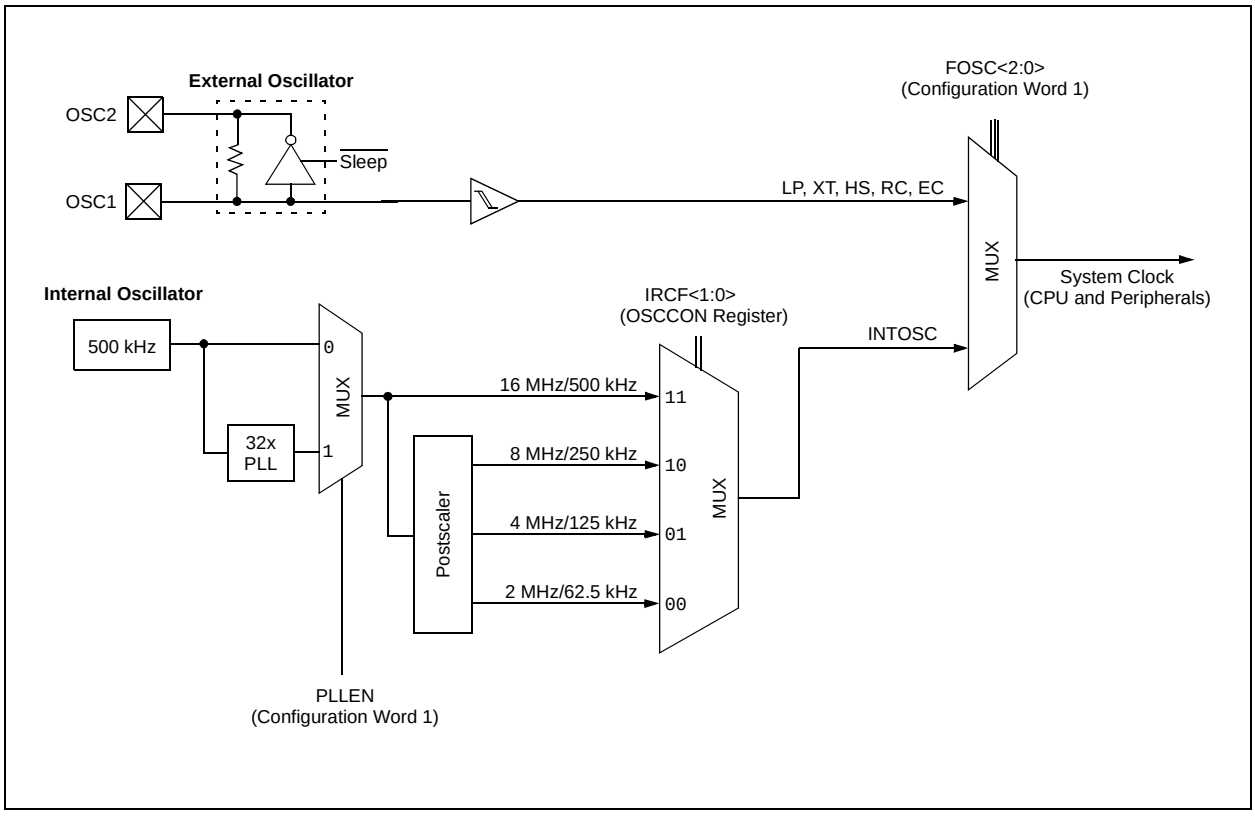
The oscillator module has a wide variety of clock sources and selection features that allow it to be used in a wide range of applications while maximizing performance and minimizing power consumption. Figure 7-1 illustrates a block diagram of the oscillator module.

Clock sources can be configured from external oscillators, quartz crystal resonators, ceramic resonators and Resistor-Capacitor (RC) circuits. In addition, the system can be configured to use an internal calibrated high-frequency oscillator as clock source, with a choice of selectable speeds via software.

Clock source modes are configured by the FOSC bits in Configuration Word 1 (CONFIG1). The oscillator module can be configured for one of eight modes of operation.

1. RC – External Resistor-Capacitor (RC) with FOSC/4 output on OSC2/CLKOUT.
2. RCIO – External Resistor-Capacitor (RC) with I/O on OSC2/CLKOUT.
3. INTOSC – Internal oscillator with FOSC/4 output on OSC2 and I/O on OSC1/CLKIN.
4. INTOSCIO – Internal oscillator with I/O on OSC1/CLKIN and OSC2/CLKOUT.
5. EC – External clock with I/O on OSC2/CLKOUT.
6. HS – High Gain Crystal or Ceramic Resonator mode.
7. XT – Medium Gain Crystal or Ceramic Resonator Oscillator mode.
8. LP – Low-Power Crystal mode.

FIGURE 7-1: SIMPLIFIED PIC® MCU CLOCK SOURCE BLOCK DIAGRAM



7.2 Clock Source Modes

Clock source modes can be classified as external or internal.

- Internal clock source (INTOSC) is contained within the oscillator module and derived from a 500 kHz high precision oscillator. The oscillator module has eight selectable output frequencies, with a maximum internal frequency of 16 MHz.
- External clock modes rely on external circuitry for the clock source. Examples are: oscillator modules (EC mode), quartz crystal resonators or ceramic resonators (LP, XT and HS modes) and Resistor-Capacitor (RC) mode circuits.

The system clock can be selected between external or internal clock sources via the FOSC bits of the Configuration Word 1.

7.3 Internal Clock Modes

The oscillator module has eight output frequencies derived from a 500 kHz high precision oscillator. The IRCF bits of the OSCCON register select the postscaler applied to the clock source dividing the frequency by 1, 2, 4 or 8. Setting the PLLEN bit of the Configuration Word 1 locks the internal clock source to 16 MHz before the postscaler is selected by the IRCF bits. The PLLEN bit must be set or cleared at the time of programming; therefore, only the upper or low four clock source frequencies are selectable in software.

7.3.1 INTOSC AND INTOSCIO MODES

The INTOSC and INTOSCIO modes configure the internal oscillators as the system clock source when the device is programmed using the oscillator selection or the FOSC<2:0> bits in the CONFIG1 register. See **Section 8.0 "Device Configuration"** for more information.

In INTOSC mode, OSC1/CLKIN is available for general purpose I/O. OSC2/CLKOUT outputs the selected internal oscillator frequency divided by 4. The CLKOUT signal may be used to provide a clock for external circuitry, synchronization, calibration, test or other application requirements.

In INTOSCIO mode, OSC1/CLKIN and OSC2/CLKOUT are available for general purpose I/O.

7.3.2 FREQUENCY SELECT BITS (IRCF)

The output of the 500 kHz INTOSC and 16 MHz INTOSC, with Phase Locked Loop enabled, connect to a postscaler and multiplexer (see Figure 7-1). The Internal Oscillator Frequency Select bits (IRCF) of the OSCCON register select the frequency output of the internal oscillator. Depending upon the PLLEN bit, one of four frequencies of two frequency sets can be selected via software:

If PLLEN = 1, frequency selection is as follows:

- 16 MHz
- 8 MHz (Default after Reset)
- 4 MHz
- 2 MHz

If PLLEN = 0, frequency selection is as follows:

- 500 kHz
- 250 kHz (Default after Reset)
- 125 kHz
- 62.5 kHz

Note:	Following any Reset, the IRCF<1:0> bits of the OSCCON register are set to '10' and the frequency selection is set to 8 MHz or 250 kHz. The user can modify the IRCF bits to select a different frequency.

There is no start-up delay before a new frequency selected in the IRCF bits takes effect. This is because the old and new frequencies are derived from INTOSC via the postscaler and multiplexer.

Start-up delay specifications are located in the Table 25-4 in **Section 25.0 "Electrical Specifications"**.

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7.4 Oscillator Control

The Oscillator Control (OSCCON) register (Figure 7-1) displays the status and allows frequency selection of the internal oscillator (INTOSC) system clock. The OSCCON register contains the following bits:

- Frequency selection bits (IRCF)
- Status Locked bits (ICSL)
- Status Stable bits (ICSS)

REGISTER 7-1: OSCCON: OSCILLATOR CONTROL REGISTER

U-0	U-0	R/W-1	R/W-0	R-q	R-q	U-0	U-0
—	—	IRCF1	IRCF0	ICSL	ICSS	—	—
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
q = Value depends on condition			

bit 7-6 **Unimplemented:** Read as '0'

bit 5-4 **IRCF<1:0>:** Internal Oscillator Frequency Select bits

When PLLEN = 1 (16 MHz INTOSC)
11 = 16 MHz
10 = 8 MHz (POR value)
01 = 4 MHz
00 = 2 MHz
When PLLEN = 0 (500 kHz INTOSC)
11 = 500 kHz
10 = 250 kHz (POR value)
01 = 125 kHz
00 = 62.5 kHz

bit 3 **ICSL:** Internal Clock Oscillator Status Locked bit (2% Stable)

1 = 16 MHz/500 kHz Internal Oscillator (HFIOSC) is in lock.
0 = 16 MHz/500 kHz Internal Oscillator (HFIOSC) has not yet locked.

bit 2 **ICSS:** Internal Clock Oscillator Status Stable bit (0.5% Stable)

1 = 16 MHz/500 kHz Internal Oscillator (HFIOSC) has stabilized to its maximum accuracy
0 = 16 MHz/500 kHz Internal Oscillator (HFIOSC) has not yet reached its maximum accuracy

bit 1-0 **Unimplemented:** Read as '0'

7.5 Oscillator Tuning

The INTOSC is factory calibrated but can be adjusted in software by writing to the OSCTUNE register (Register 7-2).

The default value of the OSCTUNE register is '0'. The value is a 6-bit two's complement number.

When the OSCTUNE register is modified, the INTOSC frequency will begin shifting to the new frequency. Code execution continues during this shift. There is no indication that the shift has occurred.

REGISTER 7-2: OSCTUNE: OSCILLATOR TUNING REGISTER

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Read as '0'

bit 5-0 **TUN<5:0>:** Frequency Tuning bits

01 1111 = Maximum frequency
01 1110 =
•
•
•
00 0001 =
00 0000 = Oscillator module is running at the factory-calibrated frequency.
11 1111 =
•
•
•
10 0000 = Minimum frequency

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7.6 External Clock Modes

7.6.1 OSCILLATOR START-UP TIMER (OST)

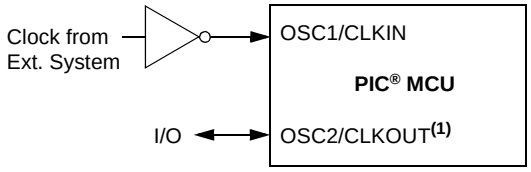
If the oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations on the OSC1 pin before the device is released from Reset. This occurs following a Power-on Reset (POR) and when the Power-up Timer (PWRT) has expired (if configured), or a wake-up from Sleep. During this time, the program counter does not increment and program execution is suspended. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the oscillator module.

7.6.2 EC MODE

The External Clock (EC) mode allows an externally generated logic level as the system clock source. When operating in this mode, an external clock source is connected to the OSC1 input and the OSC2 is available for general purpose I/O. Figure 7-2 shows the pin connections for EC mode.

The Oscillator Start-up Timer (OST) is disabled when EC mode is selected. Therefore, there is no delay in operation after a Power-on Reset (POR) or wake-up from Sleep. Because the PIC® MCU design is fully static, stopping the external clock input will have the effect of halting the device while leaving all data intact. Upon restarting the external clock, the device will resume operation as if no time had elapsed.

FIGURE 7-2: EXTERNAL CLOCK (EC) MODE OPERATION



Note 1: Alternate pin functions are described in **Section 6.1 "Alternate Pin Function"**.

7.6.3 LP, XT, HS MODES

The LP, XT and HS modes support the use of quartz crystal resonators or ceramic resonators connected to OSC1 and OSC2 (Figure 7-3). The mode selects a low, medium or high gain setting of the internal inverter-amplifier to support various resonator types and speed.

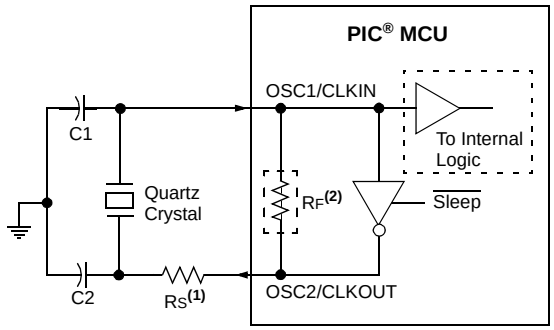
LP Oscillator mode selects the lowest gain setting of the internal inverter-amplifier. LP mode current consumption is the least of the three modes. This mode is best suited to drive resonators with a low drive level specification, for example, tuning fork type crystals.

XT Oscillator mode selects the intermediate gain setting of the internal inverter-amplifier. XT mode current consumption is the medium of the three modes. This mode is best suited to drive resonators with a medium drive level specification.

HS Oscillator mode selects the highest gain setting of the internal inverter-amplifier. HS mode current consumption is the highest of the three modes. This mode is best suited for resonators that require a high drive setting.

Figure 7-3 and Figure 7-4 show typical circuits for quartz crystal and ceramic resonators, respectively.

FIGURE 7-3: QUARTZ CRYSTAL OPERATION (LP, XT OR HS MODE)



Note 1: A series resistor (RS) may be required for quartz crystals with low drive level.

2: The value of RF varies with the Oscillator mode selected.

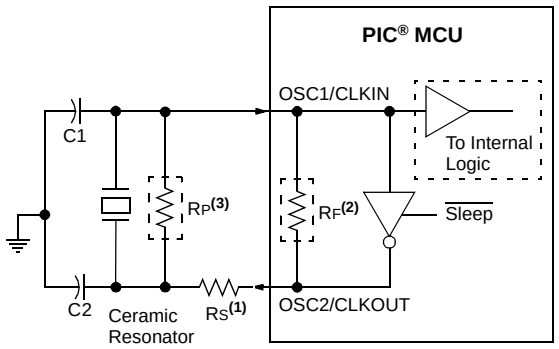
Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.

2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.

3: For oscillator design assistance, reference the following Microchip Applications Notes:

- AN826, "*Crystal Oscillator Basics and Crystal Selection for rfPIC® and PIC® Devices*" (DS00826)
- AN849, "*Basic PIC® Oscillator Design*" (DS00849)
- AN943, "*Practical PIC® Oscillator Analysis and Design*" (DS00943)
- AN949, "*Making Your Oscillator Work*" (DS00949)

FIGURE 7-4: CERAMIC RESONATOR OPERATION (XT OR HS MODE)



Note 1: A series resistor (RS) may be required for ceramic resonators with low drive level.

2: The value of RF varies with the Oscillator mode selected.

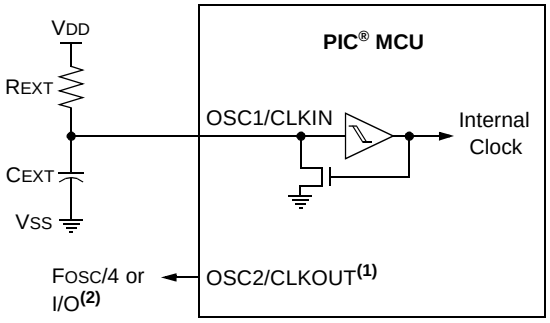
3: An additional parallel feedback resistor (RP) may be required for proper ceramic resonator operation.

7.6.4 EXTERNAL RC MODES

The external Resistor-Capacitor (RC) modes support the use of an external RC circuit. This allows the designer maximum flexibility in frequency choice while keeping costs to a minimum when clock accuracy is not required. There are two modes: RC and RCIO.

In RC mode, the RC circuit connects to OSC1. OSC2/CLKOUT outputs the RC oscillator frequency divided by 4. This signal may be used to provide a clock for external circuitry, synchronization, calibration, test or other application requirements. Figure 7-5 shows the external RC mode connections.

FIGURE 7-5: EXTERNAL RC MODES



Recommended values: 10 kΩ ≤ REXT ≤ 100 kΩ, <3V
3 kΩ ≤ REXT ≤ 100 kΩ, 3-5V
CEXT > 20 pF, 2-5V

Note 1: Alternate pin functions are described in **Section 6.1 "Alternate Pin Function"**.

2: Output depends upon RC or RCIO clock mode.

In RCIO mode, the RC circuit is connected to OSC1. OSC2 becomes an additional general purpose I/O pin.

The RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values and the operating temperature. Other factors affecting the oscillator frequency are:

- threshold voltage variation
- component tolerances
- packaging variations in capacitance

The user also needs to take into account variation due to tolerance of external RC components used.

TABLE 7-1: SUMMARY OF REGISTERS ASSOCIATED WITH CLOCK SOURCES

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets(1)
CONFIG1(1)	—	$\overline{CP}$	MCLRE	$\overline{PWRTE}$	WDTE	FOSC2	FOSC1	FOSC0	—	—
OSCCON	—	—	IRCF1	IRCF0	ICSL	ICSS	—	—	--10 qq--	--10 qq--
OSCTUNE	—	—	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0	--00 0000	--uu uuuu

Legend: x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by oscillators.

Note 1: See Configuration Word 1 (Register 8-1) for operation of all bits.

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8.0 DEVICE CONFIGURATION

Device Configuration consists of Configuration Word 1 and Configuration Word 2 registers, Code Protection and Device ID.

8.1 Configuration Words

There are several Configuration Word bits that allow different oscillator and memory protection options. These are implemented as Configuration Word 1 register at 2007h and Configuration Word 2 register at 2008h. These registers are only accessible during programming.

REGISTER 8-1: CONFIG1: CONFIGURATION WORD REGISTER 1

R/P-1	R/P-1	U-1(4)	R/P-1	R/P-1	R/P-1
$\overline{\text{DEBUG}}$	PLLEN	—	BORV	BOREN1	BOREN0
bit 13					bit 8

U-1(4)	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1
—	$\overline{\text{CP}}$	MCLRE	$\overline{\text{PWRTE}}$	WDTE	FOSC2	FOSC1	FOSC0
bit 7							bit 0

Legend:	P = Programmable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 13 **$\overline{\text{DEBUG}}$:** In-Circuit Debugger Mode bit
1 = In-Circuit Debugger disabled, RB6/ICSPCLK and RB7/ICSPDAT are general purpose I/O pins
0 = In-Circuit Debugger enabled, RB6/ICSPCLK and RB7/ICSPDAT are dedicated to the debugger

bit 12 **PLLEN:** INTOSC PLL Enable bit
0 = INTOSC frequency is 500 kHz
1 = INTOSC frequency is 16 MHz (32x)

bit 11 **Unimplemented:** Read as ‘1’

bit 10 **BORV:** Brown-out Reset Voltage Selection bit
0 = Brown-out Reset Voltage (V_{BOR}) set to 2.5 V nominal
1 = Brown-out Reset Voltage (V_{BOR}) set to 1.9 V nominal

bit 9-8 **BOREN<1:0>:** Brown-out Reset Selection bits(1)
0x = BOR disabled (Preconditioned State)
10 = BOR enabled during operation and disabled in Sleep
11 = BOR enabled

bit 7 **Unimplemented:** Read as ‘1’

bit 6 **$\overline{\text{CP}}$:** Code Protection bit(2)
1 = Program memory code protection is disabled
0 = Program memory code protection is enabled

bit 5 **MCLRE:** RE3/$\overline{\text{MCLR}}$ Pin Function Select bit(3)
1 = RE3/$\overline{\text{MCLR}}$ pin function is $\overline{\text{MCLR}}$
0 = RE3/$\overline{\text{MCLR}}$ pin function is digital input, $\overline{\text{MCLR}}$ internally tied to V_{DD}

bit 4 **$\overline{\text{PWRTE}}$:** Power-up Timer Enable bit
1 = PWRT disabled
0 = PWRT enabled

bit 3 **WDTE:** Watchdog Timer Enable bit
1 = WDT enabled
0 = WDT disabled

Note 1: Enabling Brown-out Reset does not automatically enable Power-up Timer.
2: The entire program memory will be erased when the code protection is turned off.
3: When $\overline{\text{MCLR}}$ is asserted in INTOSC or RC mode, the internal clock oscillator is disabled.
4: MPLAB® IDE masks unimplemented Configuration bits to ‘0’.

REGISTER 8-1: CONFIG1: CONFIGURATION WORD REGISTER 1 (CONTINUED)

bit 2-0 **FOSC<2:0>:** Oscillator Selection bits
- 111 = RC oscillator: CLKOUT function on RA6/OSC2/CLKOUT pin, RC on RA7/OSC1/CLKIN
- 110 = RCIO oscillator: I/O function on RA6/OSC2/CLKOUT pin, RC on RA7/OSC1/CLKIN
- 101 = INTOSC oscillator: CLKOUT function on RA6/OSC2/CLKOUT pin, I/O function on RA7/OSC1/CLKIN
- 100 = INTOSCIO oscillator: I/O function on RA6/OSC2/CLKOUT pin, I/O function on RA7/OSC1/CLKIN
- 011 = EC: I/O function on RA6/OSC2/CLKOUT pin, CLKIN on RA7/OSC1/CLKIN
- 010 = HS oscillator: High-speed crystal/resonator on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN
- 001 = XT oscillator: Crystal/resonator on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN
- 000 = LP oscillator: Low-power crystal on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN

Note 1: Enabling Brown-out Reset does not automatically enable Power-up Timer.
2: The entire program memory will be erased when the code protection is turned off.
3: When $\overline{MCLR}$ is asserted in INTOSC or RC mode, the internal clock oscillator is disabled.
4: MPLAB® IDE masks unimplemented Configuration bits to '0'.

REGISTER 8-2: CONFIG2: CONFIGURATION WORD REGISTER 2

U-1(1)	U-1(1)	U-1(1)	U-1(1)	U-1(1)	U-1(1)	U-1(1)	U-1(1)
—	—	—	—	—	—	—	—
bit 15							bit 8

U-1(1)	U-1(1)	R/P-1	R/P-1	U-1(1)	U-1(1)	U-1(1)	U-1(1)
—	—	VCAPEN1	VCAPEN0	—	—	—	—
bit 7							bit 0

Legend:	P = Programmable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 15-6 **Unimplemented**: Read as '1'

bit 5-4 **VCAPEN<1:0>:** Voltage Regulator Capacitor Enable bits
For the PIC16LF707:
These bits are ignored. All VCAP pin functions are disabled.
For the PIC16F707:
- 00 = VCAP functionality is enabled on RA0
- 01 = VCAP functionality is enabled on RA5
- 10 = VCAP functionality is enabled on RA6
- 11 = All VCAP functions are disabled (not recommended)

bit 3-0 **Unimplemented**: Read as '1'

Note 1: MPLAB® IDE masks unimplemented Configuration bits to '0'.

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8.2 Code Protection

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out using ICSP™ for verification purposes.

Note:	The entire Flash program memory will be erased when the code protection is turned off. See the "*PIC16F707/PIC16LF707 Memory Programming Specification*" (DS41332) for more information.

8.3 User ID

Four memory locations (2000h-2003h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are not accessible during normal execution, but are readable and writable during Program/Verify mode. Only the Least Significant 7 bits of the ID locations are reported when using MPLAB IDE. See the "*PIC16F707/PIC16LF707 Memory Programming Specification*" (DS41332) for more information.

NOTES:

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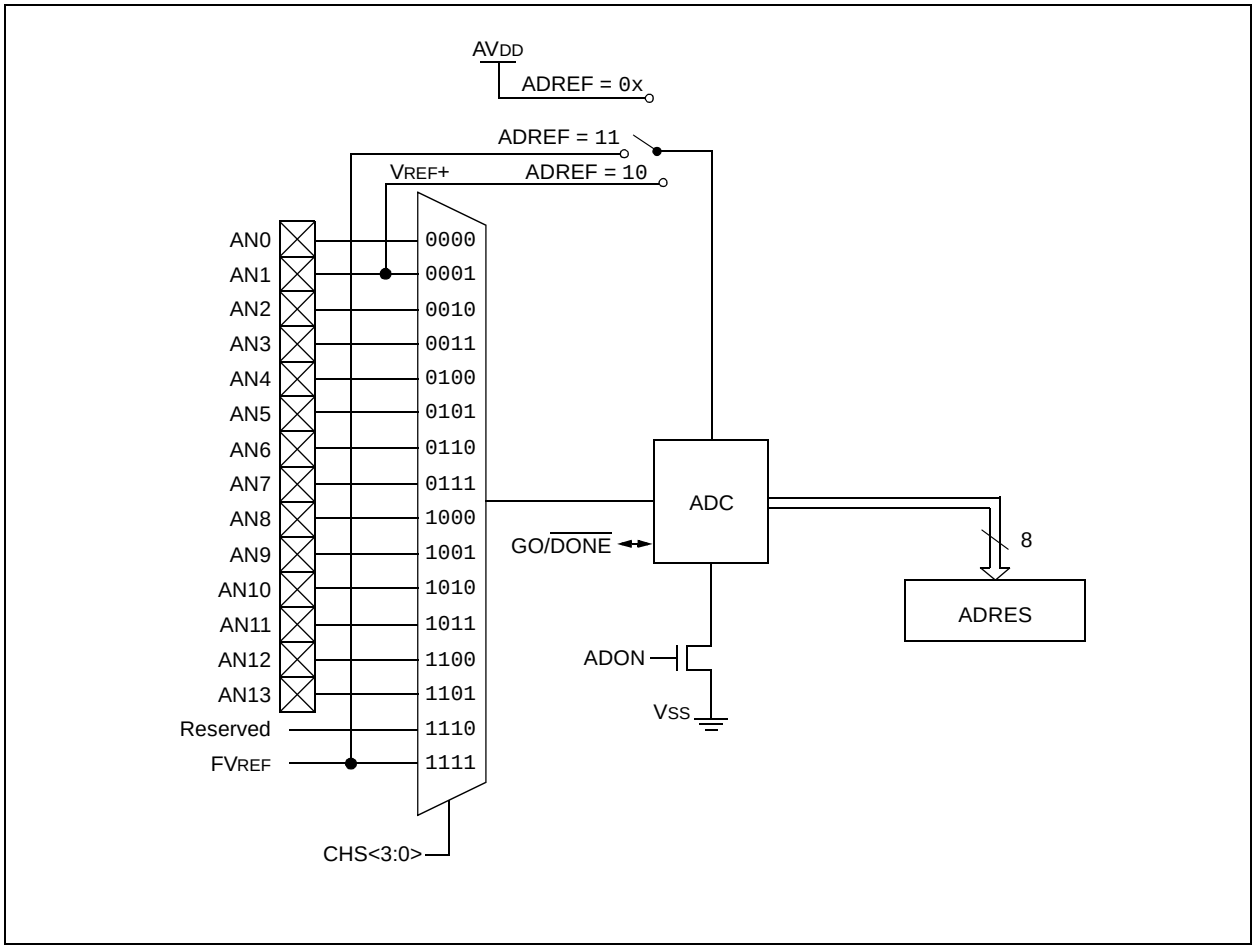
9.0 ANALOG-TO-DIGITAL CONVERTER (ADC) MODULE

The Analog-to-Digital Converter (ADC) allows conversion of an analog input signal to a 8-bit binary representation of that signal. This device uses analog inputs, which are multiplexed into a single sample and hold circuit. The output of the sample and hold is connected to the input of the converter. The converter generates a 8-bit binary result via successive approximation and stores the conversion result into the ADC result register (ADRES). Figure 9-1 shows the block diagram of the ADC.

The ADC voltage reference is software selectable to be either internally generated or externally supplied.

The ADC can generate an interrupt upon completion of a conversion. This interrupt can be used to wake-up the device from Sleep.

FIGURE 9-1: ADC BLOCK DIAGRAM



9.1 ADC Configuration

When configuring and using the ADC the following functions must be considered:

- Port configuration
- Channel selection
- ADC voltage reference selection
- ADC conversion clock source
- Interrupt control
- Results formatting

9.1.1 PORT CONFIGURATION

The ADC can be used to convert both analog and digital signals. When converting analog signals, the I/O pin should be configured for analog by setting the associated TRIS and ANSEL bits. Refer to **Section 6.0 "I/O Ports"** for more information.

Note:	Analog voltages on any pin that is defined as a digital input may cause the input buffer to conduct excess current.

9.1.2 CHANNEL SELECTION

The CHS bits of the ADCON0 register determine which channel is connected to the sample and hold circuit.

When changing channels, a delay is required before starting the next conversion. Refer to **Section 9.2 "ADC Operation"** for more information.

9.1.3 ADC VOLTAGE REFERENCE

The ADREF bits of the ADCON1 register provides control of the positive voltage reference. The positive voltage reference can be either VDD, an external voltage source or the internal Fixed Voltage Reference. The negative voltage reference is always connected to the ground reference. See **Section 10.0 "Fixed Voltage Reference"** for more details on the Fixed Voltage Reference.

9.1.4 CONVERSION CLOCK

The source of the conversion clock is software selectable via the ADCS bits of the ADCON1 register. There are seven possible clock options:

- FOSC/2
- FOSC/4
- FOSC/8
- FOSC/16
- FOSC/32
- FOSC/64
- FRC (dedicated internal oscillator)

The time to complete one bit conversion is defined as TAD. One full 8-bit conversion requires 10 TAD periods as shown in Figure 9-2.

For correct conversion, the appropriate TAD specification must be met. Refer to the A/D conversion requirements in **Section 25.0 "Electrical Specifications"** for more information. Table 9-1 gives examples of appropriate ADC clock selections.

Note:	Unless using the FRC, any changes in the system clock frequency will change the ADC clock frequency, which may adversely affect the ADC result.

TABLE 9-1: ADC CLOCK PERIOD (TAD) VS. DEVICE OPERATING FREQUENCIES

ADC Clock Period (TAD)		Device Frequency (FOSC)				
ADC Clock Source	**ADCS<2:0>**	**20 MHz**	**16 MHz**	**8 MHz**	**4 MHz**	**1 MHz**
Fosc/2	000	100 ns(2)	125 ns(2)	250 ns(2)	500 ns(2)	2.0 µs
Fosc/4	100	200 ns(2)	250 ns(2)	500 ns(2)	1.0 µs	4.0 µs
Fosc/8	001	400 ns(2)	0.5 µs(2)	1.0 µs	2.0 µs	8.0 µs(3)
Fosc/16	101	800 ns	1.0 µs	2.0 µs	4.0 µs	16.0 µs(3)
Fosc/32	010	1.6 µs	2.0 µs	4.0 µs	8.0 µs(3)	32.0 µs(3)
Fosc/64	110	3.2 µs	4.0 µs	8.0 µs(3)	16.0 µs(3)	64.0 µs(3)
FRC	x11	1.0-6.0 µs(1,4)	1.0-6.0 µs(1,4)	1.0-6.0 µs(1,4)	1.0-6.0 µs(1,4)	1.0-6.0 µs(1,4)

Legend: Shaded cells are outside of recommended range.

Note 1: The FRC source has a typical TAD time of 1.6 µs for VDD.

2: These values violate the minimum required TAD time.

3: For faster conversion times, the selection of another clock source is recommended.

4: When the device frequency is greater than 1 MHz, the FRC clock source is only recommended if the conversion will be performed during Sleep.

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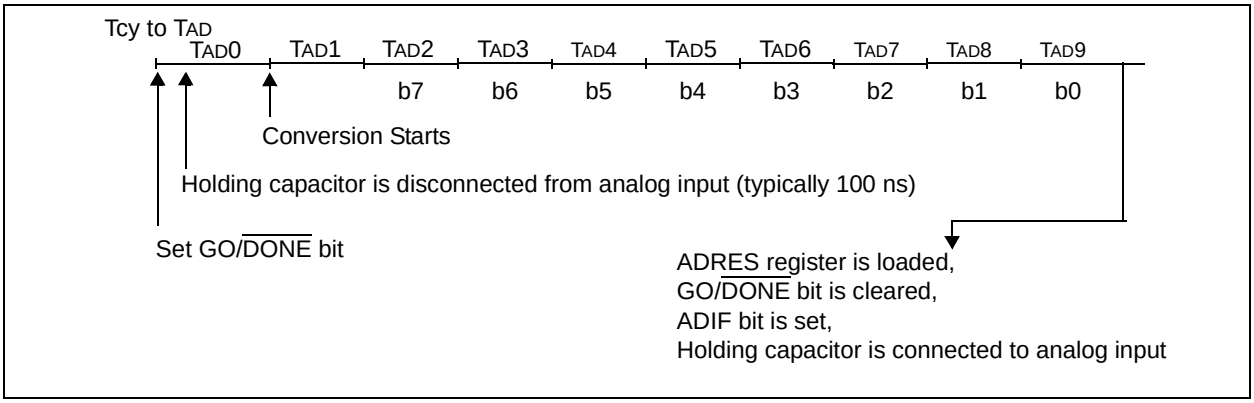
FIGURE 9-2: ANALOG-TO-DIGITAL CONVERSION TAD CYCLES



9.1.5 INTERRUPTS

The ADC module allows for the ability to generate an interrupt upon completion of an Analog-to-Digital conversion. The ADC interrupt flag is the ADIF bit in the PIR1 register. The ADC Interrupt Enable is the ADIE bit in the PIE1 register. The ADIF bit must be cleared in software.

Note 1:	The ADIF bit is set at the completion of every conversion, regardless of whether or not the ADC interrupt is enabled.
2:	The ADC operates during Sleep only when the FRC oscillator is selected.

This interrupt can be generated while the device is operating or while in Sleep. If the device is in Sleep, the interrupt will wake-up the device. Upon waking from Sleep, the next instruction following the SLEEP instruction is always executed. If the user is attempting to wake-up from Sleep and resume in-line code execution, the GIE and PEIE bits of the INTCON register must be disabled. If the GIE and PEIE bits of the INTCON register are enabled, execution will switch to the Interrupt Service Routine.

Please refer to **Section 9.1.5 "Interrupts"** for more information.

9.2 ADC Operation

9.2.1 STARTING A CONVERSION

To enable the ADC module, the ADON bit of the ADCON0 register must be set to a '1'. Setting the GO/DONE bit of the ADCON0 register to a '1' will start the Analog-to-Digital conversion.

Note:	The GO/DONE bit should not be set in the same instruction that turns on the ADC. Refer to **Section 9.2.6 "A/D Conversion Procedure"**.

9.2.2 COMPLETION OF A CONVERSION

When the conversion is complete, the ADC module will:

- Clear the GO/DONE bit
- Set the ADIF interrupt flag bit
- Update the ADRES register with new conversion result

9.2.3 TERMINATING A CONVERSION

If a conversion must be terminated before completion, the GO/DONE bit can be cleared in software. The ADRES register will be updated with the partially complete Analog-to-Digital conversion sample. Incomplete bits will match the last bit converted.

Note:	A device Reset forces all registers to their Reset state. Thus, the ADC module is turned off and any pending conversion is terminated.

9.2.4 ADC OPERATION DURING SLEEP

The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. When the FRC clock source is selected, the ADC waits one additional instruction before starting the conversion. This allows the SLEEP instruction to be executed, which can reduce system noise during the conversion. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

9.2.5 SPECIAL EVENT TRIGGER

The Special Event Trigger of the CCP module allows periodic ADC measurements without software intervention. When this trigger occurs, the GO/DONE bit is set by hardware and the Timer1 counter resets to zero.

Using the Special Event Trigger does not assure proper ADC timing. It is the user's responsibility to ensure that the ADC timing requirements are met.

Refer to **Section 17.0 "Capture/Compare/PWM (CCP) Module"** for more information.

9.2.6 A/D CONVERSION PROCEDURE

This is an example procedure for using the ADC to perform an Analog-to-Digital conversion:

1. Configure Port:
   - Disable pin output driver (Refer to the TRIS register)
   - Configure pin as analog (Refer to the ANSEL register)
2. Configure the ADC module:
   - Select ADC conversion clock
   - Configure voltage reference
   - Select ADC input channel
   - Turn on ADC module
3. Configure ADC interrupt (optional):
   - Clear ADC interrupt flag
   - Enable ADC interrupt
   - Enable peripheral interrupt
   - Enable global interrupt**(1)**
4. Wait the required acquisition time**(2)**.
5. Start conversion by setting the GO/DONE bit.
6. Wait for ADC conversion to complete by one of the following:
   - Polling the GO/DONE bit
   - Waiting for the ADC interrupt (interrupts enabled)
7. Read ADC Result.
8. Clear the ADC interrupt flag (required if interrupt is enabled).

Note 1:	The global interrupt can be disabled if the user is attempting to wake-up from Sleep and resume in-line code execution.
2:	Refer to **Section 9.3 "A/D Acquisition Requirements"**.

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EXAMPLE 9-1: A/D CONVERSION

```
;This code block configures the ADC
;for polling, Vdd reference, Frc clock
;and AN0 input.
;
;Conversion start & polling for completion
; are included.
;
BANKSEL    ADCON1     ;
MOVLW      B'01110000';ADC Frc clock,
                      ;VDD reference
MOVWF      ADCON1     ;
BANKSEL    TRISA      ;
BSF        TRISA,0    ;Set RA0 to input
BANKSEL    ANSELA     ;
BSF        ANSELA,0   ;Set RA0 to analog
BANKSEL    ADCON0     ;
MOVLW      B'00000001';AN0, On
MOVWF      ADCON0     ;
CALL       SampleTime ;Acquisiton delay
BSF        ADCON0,GO  ;Start conversion
BTFSC      ADCON0,GO  ;Is conversion done?
GOTO       $-1        ;No, test again
BANKSEL    ADRES      ;
MOVF       ADRES,W    ;Read result
MOVWF      RESULT     ;store in GPR space
```

9.2.7 ADC REGISTER DEFINITIONS

The following registers are used to control the operation of the ADC.

REGISTER 9-1: ADCON0: A/D CONTROL REGISTER 0

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	CHS3	CHS2	CHS1	CHS0	GO/$\overline{\text{DONE}}$	ADON
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Read as '0'

bit 5-2 **CHS<3:0>:** Analog Channel Select bits

0000 = AN0
0001 = AN1
0010 = AN2
0011 = AN3
0100 = AN4
0101 = AN5
0110 = AN6
0111 = AN7
1000 = AN8
1001 = AN9
1010 = AN10
1011 = AN11
1100 = AN12
1101 = AN13
1110 = Reserved
1111 = Fixed Voltage Reference (FV_{REF})

bit 1 **GO/$\overline{\text{DONE}}$:** A/D Conversion Status bit

1 = A/D conversion cycle in progress. Setting this bit starts an A/D conversion cycle.
This bit is automatically cleared by hardware when the A/D conversion has completed.
0 = A/D conversion completed/not in progress

bit 0 **ADON:** ADC Enable bit

1 = ADC is enabled
0 = ADC is disabled and consumes no operating current

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REGISTER 9-2: ADCON1: A/D CONTROL REGISTER 1

U-0	R/W-0	R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0
—	ADCS2	ADCS1	ADCS0	—	—	ADREF1	ADREF0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **Unimplemented:** Read as '0'

bit 6-4 **ADCS<2:0>:** A/D Conversion Clock Select bits
000 = FOSC/2
001 = FOSC/8
010 = FOSC/32
011 = FRC (clock supplied from a dedicated RC oscillator)
100 = FOSC/4
101 = FOSC/16
110 = FOSC/64
111 = FRC (clock supplied from a dedicated RC oscillator)

bit 3-2 **Unimplemented:** Read as '0'

bit 1-0 **ADREF<1:0>:** Voltage Reference Configuration bits
0x = VREF is connected to VDD
10 = VREF is connected to external VREF (RA3/AN3)
11 = VREF is connected to internal Fixed Voltage Reference

REGISTER 9-3: ADRES: ADC RESULT REGISTER

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
ADRES7	ADRES6	ADRES5	ADRES4	ADRES3	ADRES2	ADRES1	ADRES0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **ADRES<7:0>**: ADC Result Register bits
8-bit conversion result.

9.3 A/D Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 9-3. The source impedance (RS) and the internal sampling switch (RSS) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (RSS) impedance varies over the device voltage (VDD), refer to Figure 9-3. **The maximum recommended impedance for analog sources is 10 kΩ.** As the source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an A/D acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 9-1 may be used. This equation assumes that 1/2 LSb error is used (256 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

EQUATION 9-1: ACQUISITION TIME EXAMPLE

Assumptions: $Temperature = 50°C$ *and external impedance of* $10k\Omega$ *5.0V* V_{DD}

$$T_{ACQ} = \text{Amplifier Settling Time} + \text{Hold Capacitor Charging Time} + \text{Temperature Coefficient}$$
$$= T_{AMP} + T_C + T_{COFF}$$
$$= 2\mu s + T_C + [(Temperature - 25°C)(0.05\mu s/°C)]$$

The value for T_C *can be approximated with the following equations:*

$$V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) = V_{CHOLD} \quad ;[1]\ V_{CHOLD}\ \text{charged to within 1/2 lsb}$$

$$V_{APPLIED}\left(1 - e^{\frac{-T_C}{RC}}\right) = V_{CHOLD} \quad ;[2]\ V_{CHOLD}\ \text{charge response to}\ V_{APPLIED}$$

$$V_{APPLIED}\left(1 - e^{\frac{-T_C}{RC}}\right) = V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) \quad ;\text{combining [1] and [2]}$$

Note: *Where n = number of bits of the ADC.*

Solving for T_C*:*

$$T_C = -C_{HOLD}(R_{IC} + R_{SS} + R_S)\ ln(1/511)$$
$$= -10pF(1k\Omega + 7k\Omega + 10k\Omega)\ ln(0.001957)$$
$$= 1.12\mu s$$

Therefore:

$$T_{ACQ} = 2\mu s + 1.12\mu s + [(50°C - 25°C)(0.05\mu s/°C)]$$
$$= 4.42\mu s$$

Note 1: The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

3: The maximum recommended impedance for analog sources is 10 kΩ. This is required to meet the pin leakage specification.

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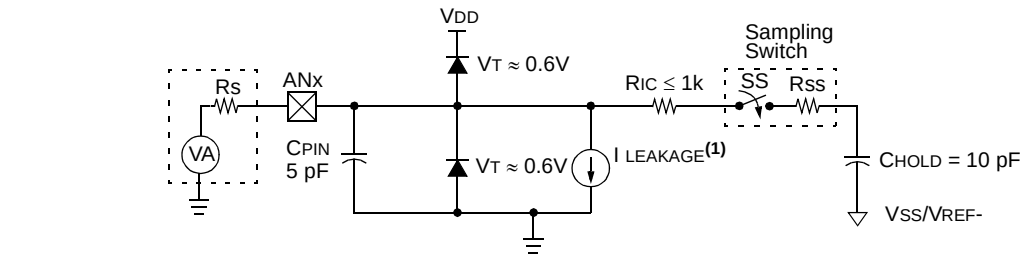
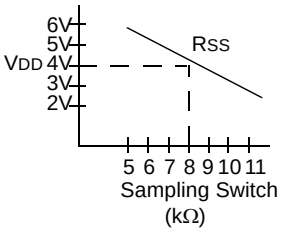
FIGURE 9-3: ANALOG INPUT MODEL





Legend:

CHOLD	=	Sample/Hold Capacitance
CPIN	=	Input Capacitance
I LEAKAGE	=	Leakage current at the pin due to various junctions
RIC	=	Interconnect Resistance
RSS	=	Resistance of Sampling Switch
SS	=	Sampling Switch
VT	=	Threshold Voltage

Note 1: Refer to **Section 25.0 "Electrical Specifications"**.

FIGURE 9-4: ADC TRANSFER FUNCTION

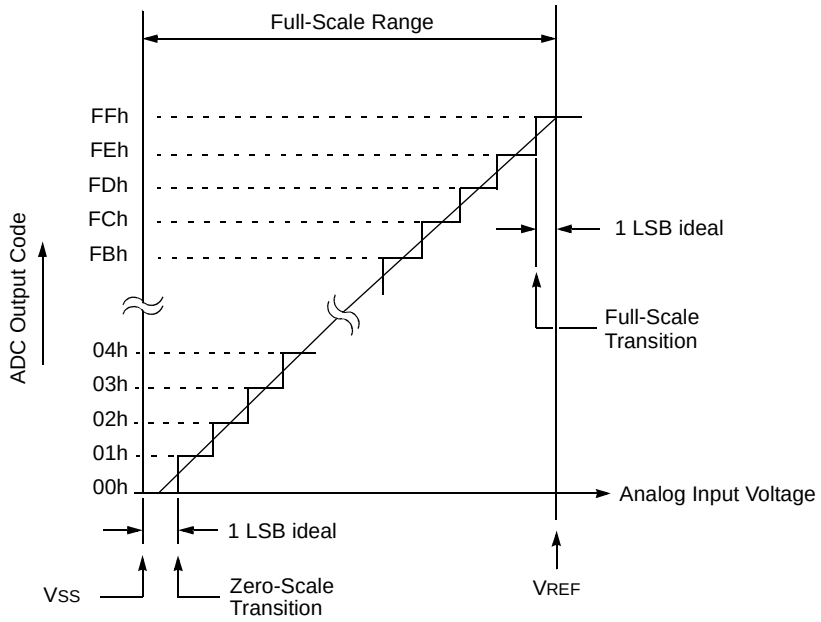


TABLE 9-2: SUMMARY OF ASSOCIATED ADC REGISTERS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ADCON0	—	—	CHS3	CHS2	CHS1	CHS0	GO/$\overline{\text{DONE}}$	ADON	--00 0000	--00 0000
ADCON1	—	ADCS2	ADCS1	ADCS0	—	—	ADREF1	ADREF0	-000 --00	-000 --00
ANSELA	ANSA7	ANSA6	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	1111 1111	1111 1111
ANSELB	ANSB7	ANSB6	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	1111 1111	1111 1111
ANSELE	—	—	—	—	—	ANSE2	ANSE1	ANSE0	---- -111	---- -111
ADRES	A/D Result Register Byte								xxxx xxxx	uuuu uuuu
CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	--00 0000	--00 0000
FVRCON	FVRRDY	FVREN	—	—	CDAFVR1	CDAFVR0	ADFVR1	ADFVR0	q000 0000	q000 0000
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
TRISE	—	—	—	—	TRISE3	TRISE2	TRISE1	TRISE0	---- 1111	---- 1111

Legend: x = unknown, u = unchanged, – = unimplemented read as '0', q = value depends on condition. Shaded cells are not used for ADC module.

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10.0 FIXED VOLTAGE REFERENCE

The Fixed Voltage Reference, or FVR, is a stable voltage reference independent of VDD with 1.024V, 2.048V or 4.096V selectable output levels. The output of the FVR can be configured to supply a reference voltage to the following:

- ADC input channel
- ADC positive reference
- Digital-to-Analog Converter (DAC)
- Capacitive Sensing Modules (CSM)

The FVR can be enabled by setting the FVREN bit of the FVRCON register.

10.1 Independent Gain Amplifiers

The output of the FVR supplied to the ADC and CSM/DAC modules is routed through the two independent programmable gain amplifiers. Each amplifier can be configured to amplify the reference voltage by 1x, 2x or 4x.

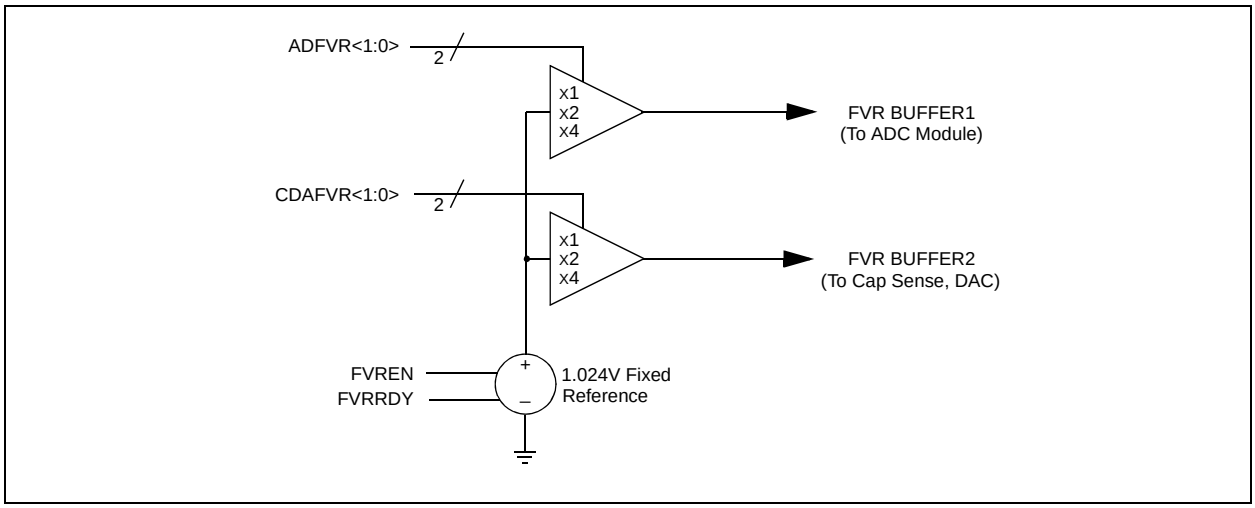
The ADFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the ADC module. Reference **Section 9.0 "Analog-to-Digital Converter (ADC) Module"** for additional information on selecting the appropriate input channel.

The CDAFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the capacitive sensing and digital-to-analog converter modules. Reference **Section 16.0 "Capacitive Sensing Module"** and **Section 11.0 "Digital-to-Analog Converter (DAC) Module"** for additional information.

10.2 FVR Stabilization Period

When the Fixed Voltage Reference module is enabled, it requires time for the reference and amplifier circuits to stabilize. Once the circuits stabilize and are ready for use, the FVRRDY bit of the FVRCON register will be set. See **Section 25.0 "Electrical Specifications"** for the minimum delay requirement.

FIGURE 10-1: VOLTAGE REFERENCE BLOCK DIAGRAM



REGISTER 10-1: FVRCON: FIXED VOLTAGE REFERENCE REGISTER

R-q	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
FVRRDY(1)	FVREN	—	—	CDAFVR1(2)	CDAFVR0(2)	ADFVR1(2)	ADFVR0(2)
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
q = Value depends on condition			

bit 7 **FVRRDY:** Fixed Voltage Reference Ready Flag bit(1)
- 0 = Fixed Voltage Reference output is not active or stable
- 1 = Fixed Voltage Reference output is ready for use

bit 6 **FVREN:** Fixed Voltage Reference Enable bit
- 0 = Fixed Voltage Reference is disabled
- 1 = Fixed Voltage Reference is enabled

bit 5-4 **Reserved:** Read as '0'. Maintain these bits clear

bit 3-2 **CDAFVR<1:0>:** Cap Sense and D/A Converter Fixed Voltage Reference Selection bit(2)
- 00 = CSM and D/A Converter Fixed Voltage Reference Peripheral output is off.
- 01 = CSM and D/A Converter Fixed Voltage Reference Peripheral output is 1x (1.024V)
- 10 = CSM and D/A Converter Fixed Voltage Reference Peripheral output is 2x (2.048V)
- 11 = CSM and D/A Converter Fixed Voltage Reference Peripheral output is 4x (4.096V)

bit 1-0 **ADFVR<1:0>:** A/D Converter Fixed Voltage Reference Selection bit(2)
- 00 = A/D Converter Fixed Voltage Reference Peripheral output is off.
- 01 = A/D Converter Fixed Voltage Reference Peripheral output is 1x (1.024V)
- 10 = A/D Converter Fixed Voltage Reference Peripheral output is 2x (2.048V)
- 11 = A/D Converter Fixed Voltage Reference Peripheral output is 4x (4.096V)

Note 1: FVRRDY is always '1' on PIC16F707 devices.
2: Fixed Voltage Reference output cannot exceed V_{DD}.

TABLE 10-1: REGISTERS ASSOCIATED WITH VOLTAGE REFERENCE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
FVRCON	FVRRDY	FVREN	Reserved	Reserved	CDAFVR1	CDAFVR0	ADFVR1	ADFVR0	q000 0000	q000 0000

Legend: Shaded cells are not used by the voltage reference module.

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11.0 DIGITAL-TO-ANALOG CONVERTER (DAC) MODULE

The Digital-to-Analog Converter supplies a variable voltage reference, ratiometric with VDD, with 32 selectable output levels. The output of the DAC can be configured to supply a reference voltage to the following:

- DACOUT device pin
- Capacitive sensing modules

The Digital-to-Analog Converter (DAC) can be enabled by setting the DACEN bit of the DACCON0 register.

11.1 Output Voltage Selection

The DAC has 32 voltage level ranges. The 32 levels are set with the DACR<4:0> bits of the DACCON1 register.

The DAC output voltage is determined by the following equation:

EQUATION 11-1:

IF DACEN = 1

$$V_{OUT} = \left((V_{SOURCE+} - V_{SOURCE-}) \times \frac{DACR[4:0]}{2^5} \right) + V_{SOURCE-}$$

IF DACEN = 0 & DACLPS = 1 & DACR[4:0] = 11111

$$V_{OUT} = V_{SOURCE+}$$

IF DACEN = 0 & DACLPS = 0 & DACR[4:0] = 00000

$$V_{OUT} = V_{SOURCE-}$$

$V_{SOURCE+}$ = V_{DD}, V_{REF}, or FVR BUFFER 2

$V_{SOURCE-}$ = V_{SS}

11.2 Output Clamped to Vss

The DAC output voltage can be set to VSS with no power consumption by setting the DACEN bit of the DACCON0 register to '0'.

11.3 Output Ratiometric to VDD

The DAC is VDD derived and therefore, the DAC output changes with fluctuations in VDD. The tested absolute accuracy of the DAC can be found in **Section 25.0 "Electrical Specifications"**.

11.4 Voltage Reference Output

The DAC can be output to the device DACOUT pin by setting the DACOE bit of the DACCON0 register to '1'. Selecting the reference voltage for output on the DACOUT pin automatically overrides the digital output buffer and digital input threshold detector functions of that pin. Reading the DACOUT pin when it has been configured for reference voltage output will always return a '0'.

Due to the limited current drive capability, a buffer must be used on the voltage reference output for external connections to DACOUT. Example 11-1 shows an example buffering technique.

11.5 Operation During Sleep

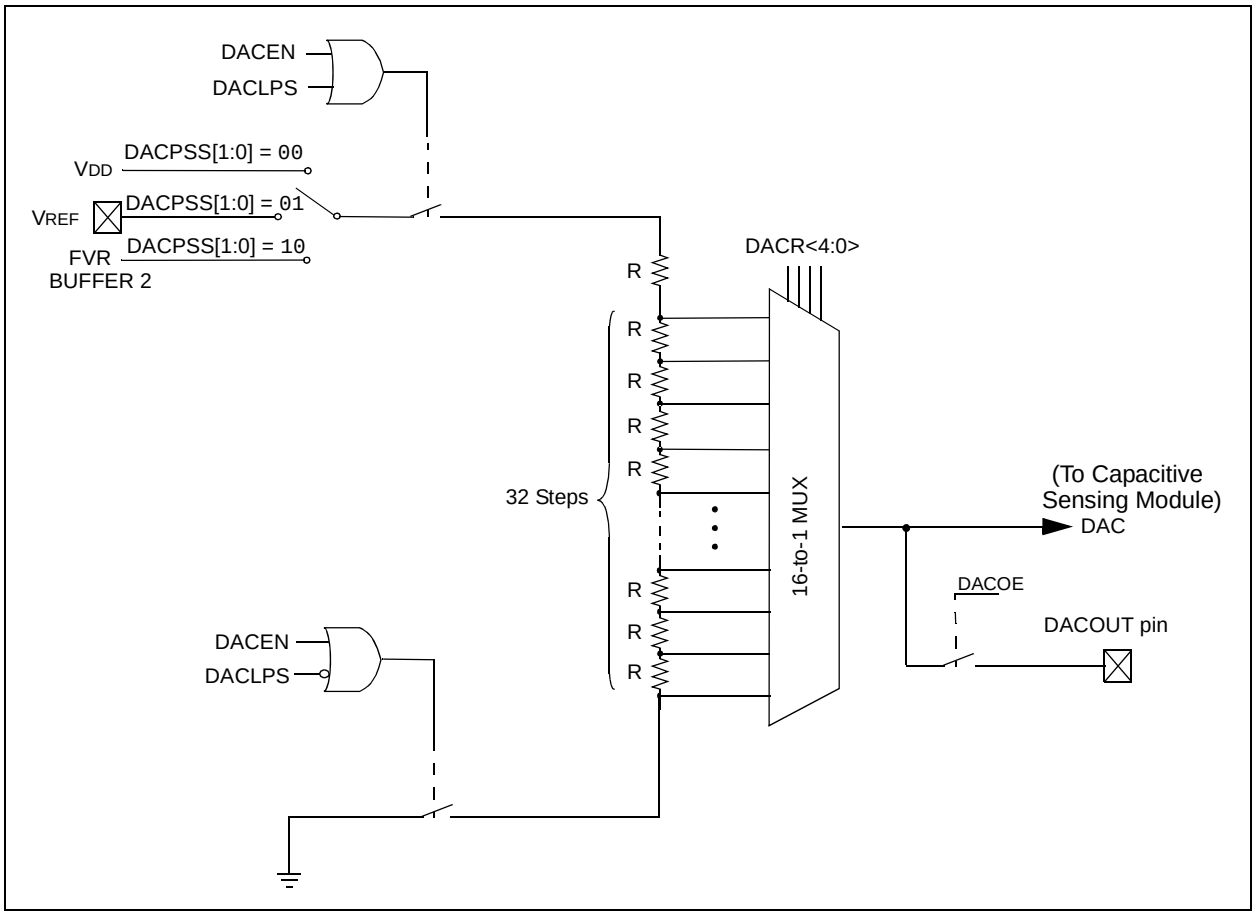
When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the DACCON0 register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

11.6 Effects of a Reset

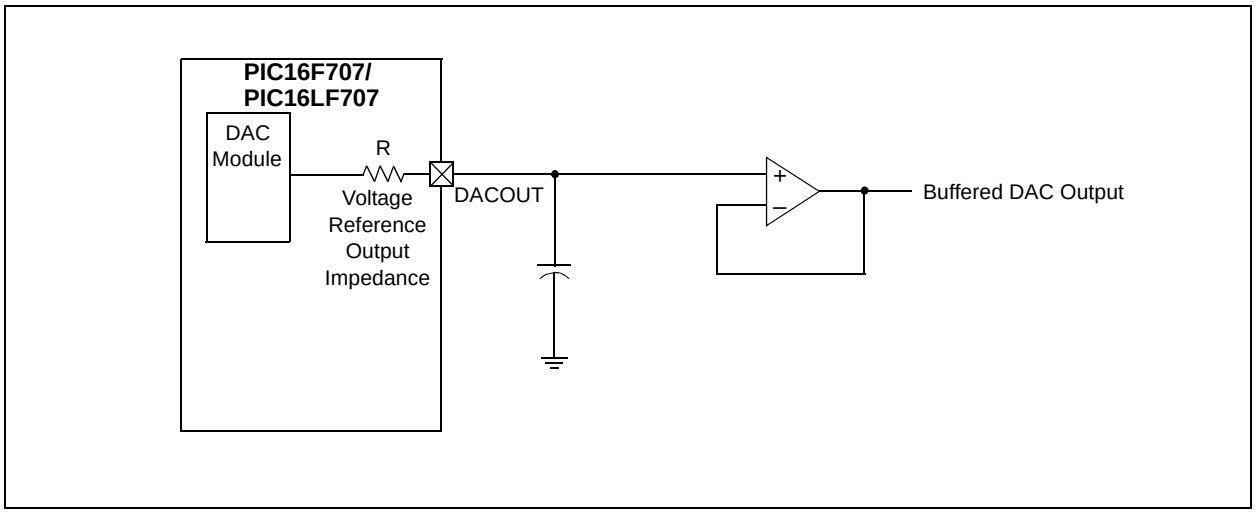
A device Reset affects the following:

- Voltage reference is disabled
- Fixed voltage reference is disabled
- DAC is removed from the DACOUT pin
- The DACR<4:0> range select bits are cleared

FIGURE 11-1: DIGITAL-TO-ANALOG CONVERTER BLOCK DIAGRAM



EXAMPLE 11-1: VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE



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REGISTER 11-1: DACCON0: VOLTAGE REFERENCE CONTROL REGISTER 0

R/W-0/0	R/W-0/0	R/W-0/0	U-0	R/W-0/0	R/W-0/0	U-0	U-0
DACEN	DACLPS	DACOE	—	DACPSS1	DACPSS0	—	—
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	q = Value depends on condition

bit 7 **DACEN:** Digital-to-Analog Converter Enable bit
0 = Digital-to-Analog Converter is disabled
1 = Digital-to-Analog Converter is enabled

bit 6 **DACLPS:** DAC Low-Power Voltage State Select bit
0 = V_{DAC} = DAC negative reference source selected
1 = V_{DAC} = DAC positive reference source selected

bit 5 **DACOE:** DAC Voltage Output Enable bit
0 = DAC voltage level is output on the DACOUT pin
1 = DAC voltage level is disconnected from the DACOUT pin

bit 4 **Unimplemented:** Read as ‘0’

bit 3-2 **DACPSS<1:0>:** DAC Positive Source Select bits
00 = V_{DD}
01 = V_{REF}
10 = FVR Buffer 2 output
11 = Reserved, do not use

bit 1-0 **Unimplemented:** Read as ‘0’

REGISTER 11-2: DACCON1: VOLTAGE REFERENCE CONTROL REGISTER 1

U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	DACR4	DACR3	DACR2	DACR1	DACR0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-5 **Unimplemented:** Read as ‘0’

bit 4-0 **DACR<4:0>:** DAC Voltage Output Select bits

TABLE 11-1: REGISTERS ASSOCIATED WITH THE DIGITAL-TO-ANALOG CONVERTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
FVRCON	FVRRDY	FVREN	Reserved	Reserved	CDAFVR1	CDAFVR0	ADFVR1	ADFVR0	q000 0000	q000 0000
DACCON0	DACEN	DACLPS	DACOE	—	DACPSS1	DACPSS0	—	—	000- 00--	000- 00--
DACCON1	—	—	—	DACR4	DACR3	DACR2	DACR1	DACR0	---0 0000	---0 0000

Legend: — = Unimplemented locations, read as '0'. Shaded cells are not used by the DAC module.

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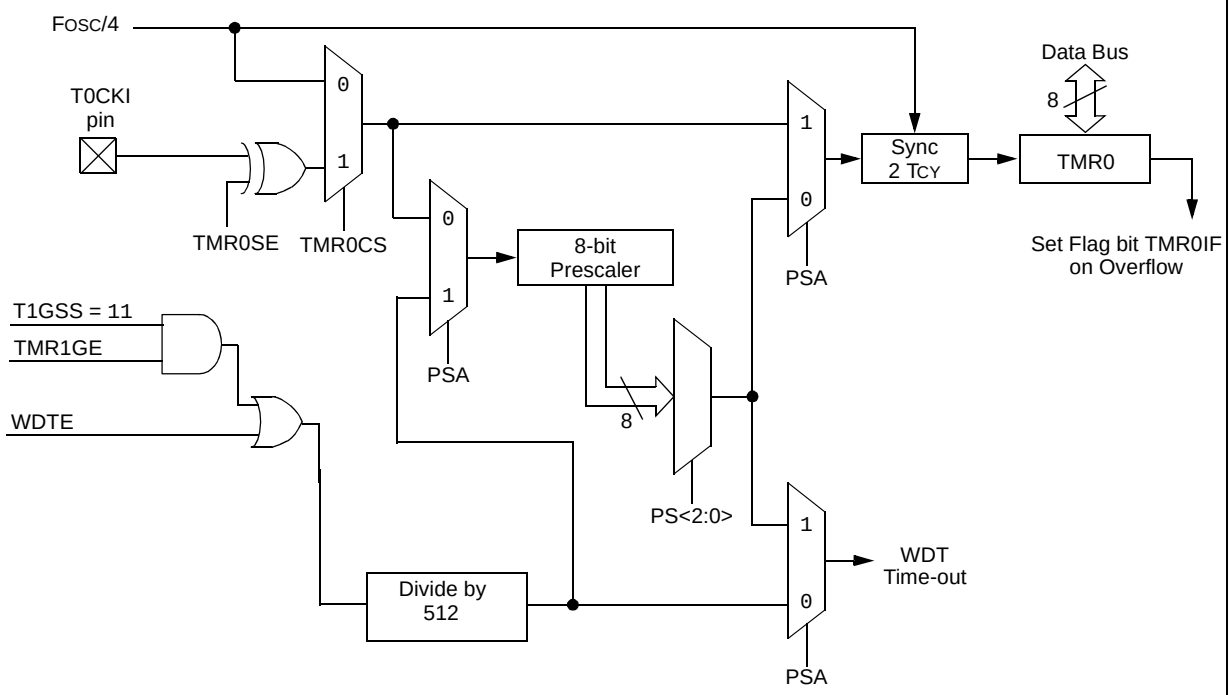
12.0 TIMER0 MODULE

The Timer0 module is an 8-bit timer/counter with the following features:

- 8-bit timer/counter register (TMR0)
- 8-bit prescaler (shared with Watchdog Timer)
- Programmable internal or external clock source
- Programmable external clock edge selection
- Interrupt on overflow

Figure 12-1 is a block diagram of the Timer0 module.

FIGURE 12-1: BLOCK DIAGRAM OF THE TIMER0/WDT PRESCALER



Note 1: TMR0SE, TMR0CS, PSA, PS<2:0> are bits in the OPTION register.

2: WDTE bit is in Configuration Word 1.

3: T1GSS and TMR1GE are in the T1GCON register.

12.1 Timer0 Operation

The Timer0 module can be used as either an 8-bit timer or an 8-bit counter.

12.1.1 8-BIT TIMER MODE

The Timer0 module will increment every instruction cycle, if used without a prescaler. 8-bit Timer mode is selected by clearing the TMR0CS bit of the OPTION register.

When TMR0 is written, the increment is inhibited for two instruction cycles immediately following the write.

Note:	The value written to the TMR0 register can be adjusted, in order to account for the two instruction cycle delay when TMR0 is written.

12.1.2 8-BIT COUNTER MODE

In 8-bit Counter mode, the Timer0 module will increment on every rising or falling edge of the T0CKI pin. 8-bit Counter mode using the T0CKI pin is selected by setting the TMR0CS bit of the OPTION register to '1'.

The rising or falling transition of the incrementing edge for either input source is determined by the TMR0SE bit in the OPTION register.

12.1.3 SOFTWARE PROGRAMMABLE PRESCALER

A single software programmable prescaler is available for use with either Timer0 or the Watchdog Timer (WDT), but not both simultaneously. The prescaler assignment is controlled by the PSA bit of the OPTION register. To assign the prescaler to Timer0, the PSA bit must be cleared to a '0'.

There are 8 prescaler options for the Timer0 module ranging from 1:2 to 1:256. The prescale values are selectable via the PS<2:0> bits of the OPTION register. In order to have a 1:1 prescaler value for the Timer0 module, the prescaler must be assigned to the WDT module.

The prescaler is not readable or writable. When the prescaler is enabled or assigned to the Timer0 module, all instructions writing to the TMR0 register will clear the prescaler.

Note:	When the prescaler is assigned to WDT, a CLRWDT instruction will clear the prescaler along with the WDT.

12.1.4 TIMER0 INTERRUPT

Timer0 will generate an interrupt when the TMR0 register overflows from FFh to 00h. The TMR0IF interrupt flag bit of the INTCON register is set every time the TMR0 register overflows, regardless of whether or not the Timer0 interrupt is enabled. The TMR0IF bit can only be cleared in software. The Timer0 interrupt enable is the TMR0IE bit of the INTCON register.

Note:	The Timer0 interrupt cannot wake the processor from Sleep since the timer is frozen during Sleep.

12.1.5 USING TIMER0 WITH AN EXTERNAL CLOCK

When Timer0 is in Counter mode, the synchronization of the T0CKI input and the Timer0 register is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks. Therefore, the high and low periods of the external clock source must meet the timing requirements as shown in **Section 25.0 "Electrical Specifications"**.

12.1.6 TIMER ENABLE

Operation of Timer0 is always enabled and the module will operate according to the settings of the OPTION register.

12.1.7 OPERATION DURING SLEEP

Timer0 cannot operate while the processor is in Sleep mode. The contents of the TMR0 register will remain unchanged while the processor is in Sleep mode.

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REGISTER 12-1: OPTION_REG: OPTION REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
$\overline{\text{RBPU}}$	INTEDG	TMR0CS	TMR0SE	PSA	PS2	PS1	PS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **$\overline{\text{RBPU}}$:** PORTB Pull-up Enable bit
1 = PORTB pull-ups are disabled
0 = PORTB pull-ups are enabled by individual PORT latch values

bit 6 **INTEDG:** Interrupt Edge Select bit
1 = Interrupt on rising edge of INT pin
0 = Interrupt on falling edge of INT pin

bit 5 **TMR0CS:** TMR0 Clock Source Select bit
1 = Transition on T0CKI pin
0 = Internal instruction cycle clock (FOSC/4)

bit 4 **TMR0SE:** TMR0 Source Edge Select bit
1 = Increment on high-to-low transition on T0CKI pin
0 = Increment on low-to-high transition on T0CKI pin

bit 3 **PSA:** Prescaler Assignment bit
1 = Prescaler is assigned to the WDT
0 = Prescaler is assigned to the Timer0 module

bit 2-0 **PS<2:0>:** Prescaler Rate Select bits

BIT VALUE	TMR0 RATE	WDT RATE
000	1 : 2	1 : 1
001	1 : 4	1 : 2
010	1 : 8	1 : 4
011	1 : 16	1 : 8
100	1 : 32	1 : 16
101	1 : 64	1 : 32
110	1 : 128	1 : 64
111	1 : 256	1 : 128

TABLE 12-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER0

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
OPTION_REG	$\overline{\text{RBPU}}$	INTEDG	TMR0CS	TMR0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
TMR0	Timer0 Module Register								xxxx xxxx	uuuu uuuu
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111

Legend: – = Unimplemented locations, read as '0'. Shaded cells are not used by the Timer0 module.

NOTES:

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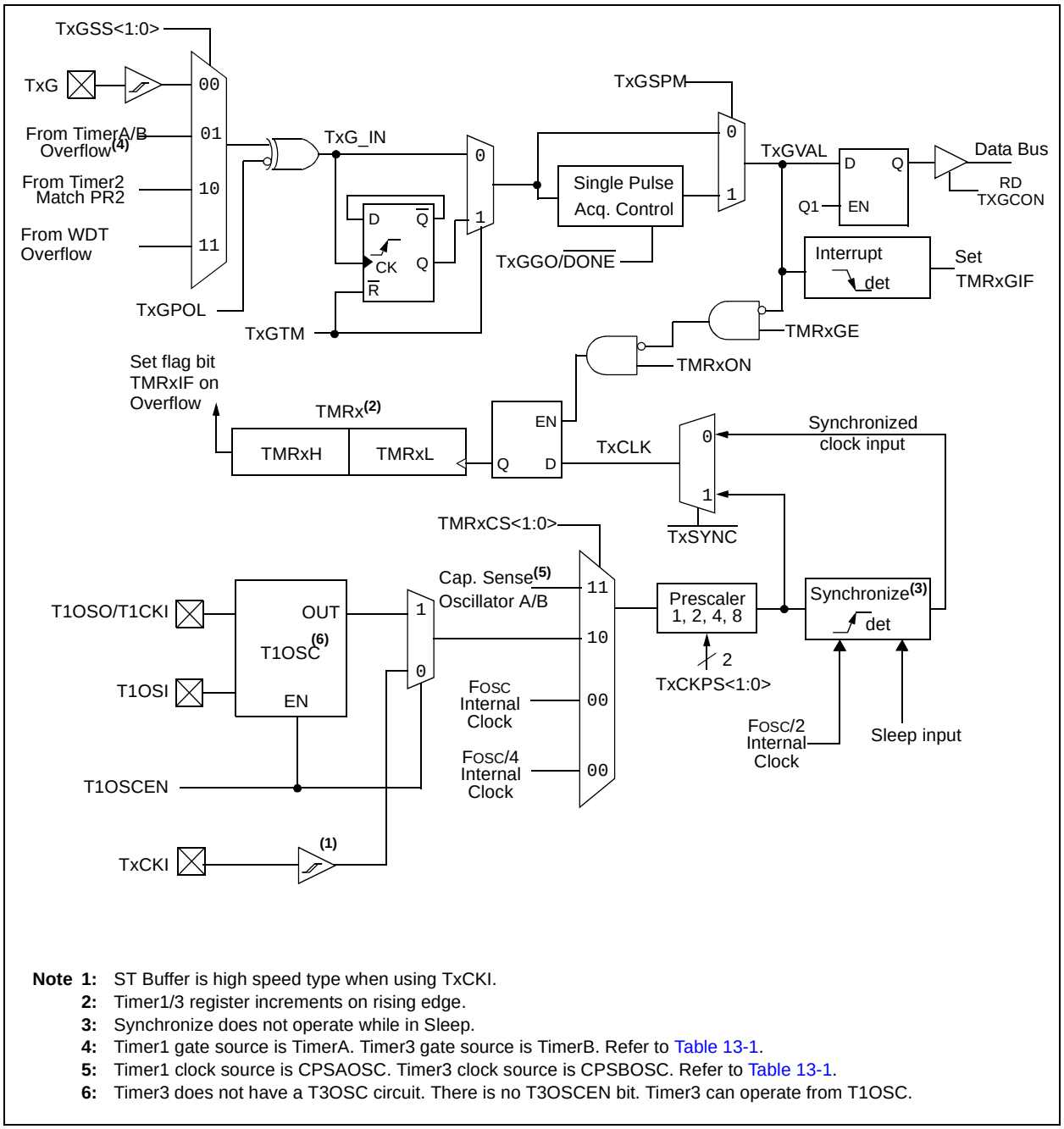
13.0 TIMER1/3 MODULES WITH GATE CONTROL

The Timer1 and Timer3 modules are 16-bit timers/counters with the following features:

- 16-bit timer/counter register pair (TMRxH:TMRxL)
- Programmable internal or external clock source
- 3-bit prescaler
- Dedicated LP oscillator circuit (Timer1 only)
- Synchronous or asynchronous operation
- Multiple Timer1/3 gate (count enable) sources
- Interrupt on overflow
- Wake-up on overflow (external clock, Asynchronous mode only)
- Time base for the Capture/Compare function (Timer1 only)
- Special Event Trigger with CCP (Timer1 only)
- Selectable Gate Source Polarity
- Gate Toggle mode
- Gate Single-pulse mode
- Gate Value Status
- Gate Event Interrupt

Figure 13-1 is a block diagram of the Timer1/3 modules.

FIGURE 13-1: TIMER1/TIMER3 BLOCK DIAGRAM



Note 1: ST Buffer is high speed type when using TxCKI.
2: Timer1/3 register increments on rising edge.
3: Synchronize does not operate while in Sleep.
4: Timer1 gate source is TimerA. Timer3 gate source is TimerB. Refer to Table 13-1.
5: Timer1 clock source is CPSAOSC. Timer3 clock source is CPSBOSC. Refer to Table 13-1.
6: Timer3 does not have a T3OSC circuit. There is no T3OSCEN bit. Timer3 can operate from T1OSC.

TABLE 13-1: CPSOSC/TIMER ASSOCIATION

Period Measurement	Cap Sense Oscillator	Divider Timer (Gate Source)
Timer1	CPS A	TimerA
Timer3	CPS B	TimerB

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13.1 Timer1/3 Operation

The Timer1 and Timer3 modules are 16-bit incrementing counters which are accessed through the TMRxH:TMRxL register pair. Writes to TMRxH or TMRxL directly update the counter.

When used with an internal clock source, the module is a timer and increments on every instruction cycle. When used with an external clock source, the module can be used as either a timer or counter and increments on every selected edge of the external source.

Timer1/3 is enabled by configuring the TMRxON and TMRxGE bits in the TxCON and TxGCON registers, respectively. Table 13-2 displays the Timer1/3 enable selections.

TABLE 13-2: TIMER1/3 ENABLE SELECTIONS

TMRxON	TMRxGE	Timer1/3 Operation
0	0	Off
0	1	Off
1	0	Always On
1	1	Count Enabled

13.2 Clock Source Selection

The TMRxCS<1:0> bits of the TxCON register and the T1OSCEN bit of the T1CON register are used to select the clock source for Timer1/3. Table 13-3 displays the clock source selections.

13.2.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected, the TMRxH:TMRxL register pair will increment on multiples of FOSC as determined by the Timer1/3 prescaler.

13.2.2 EXTERNAL CLOCK SOURCE

When the external clock source is selected, the Timer1/3 modules may work as a timer or a counter.

When enabled to count, Timer1/3 is incremented on the rising edge of the external clock input TxCKI or a capacitive sensing oscillator signal. Either of these external clock sources can be synchronized to the microcontroller system clock or they can be run asynchronously. If set for the capacitive sensing oscillator signal, Timer1 will use the CPS A signal and Timer3 will use the CPS B signal (see Table 13-1).

When used as a timer with a clock oscillator, an external 32.768 kHz crystal can be used in conjunction with the dedicated internal oscillator circuit. Only one dedicated internal oscillator circuit is available. See **Section 13.4 "Timer1/3 Oscillator"** for more information.

Note: In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge after any one or more of the following conditions:

- Timer1/3 enabled after POR reset
- Write to TMRxH or TMRxL
- Timer1/3 is disabled
- Timer1/3 is disabled (TMRxON = 0) when TxCKI is high, then Timer1/3 is enabled (TMRxON=1) when TxCKI is low.

TABLE 13-3: CLOCK SOURCE SELECTIONS

TMRxCS1	TMRxCS0	T1OSCEN	Timer1 Clock Source	Timer3 Clock Source
0	1	x	System Clock (FOSC)	System Clock (FOSC)
0	0	x	Instruction Clock (FOSC/4)	Instruction Clock (FOSC/4)
1	1	x	Capacitive Sensing A Oscillator	Capacitive Sensing B Oscillator
1	0	0	External Clocking on T1CKI Pin	External Clocking on T3CKI Pin
1	0	1	Oscillator Circuit on T1OSI/ T1OSO Pins	Oscillator Circuit on T1OSI/ T1OSO Pins

13.3 Timer1/3 Prescaler

Timer1 and Timer3 have four prescaler options allowing 1, 2, 4 or 8 divisions of the clock input. The TxCKPS bits of the TxCON register control the prescale counter. The prescale counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMRxH or TMRxL.

13.4 Timer1/3 Oscillator

A dedicated low-power 32.768 kHz oscillator circuit is built-in between pins T1OSI (input) and T1OSO (amplifier output). This internal circuit is to be used in conjunction with an external 32.768 kHz crystal.

The oscillator circuit is enabled by setting the T1OSCEN bit of the T1CON register. The oscillator can provide a clock source to Timer1 and/or Timer3. The oscillator will continue to run during Sleep.

Note:	The oscillator requires a start-up and stabilization time before use. Thus, T1OSCEN should be set and a suitable delay observed prior to enabling Timer1/3.

13.5 Timer1/3 Operation in Asynchronous Counter Mode

If control bit $\overline{\text{TxSYNC}}$ of the TxCON register is set, the external clock input is not synchronized. The timer increments asynchronously to the internal phase clocks. If external clock source is selected, then the timer will continue to run during Sleep and can generate an interrupt on overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (see **Section 13.5.1 "Reading and Writing Timer1/3 in Asynchronous Counter Mode"**).

13.5.1 READING AND WRITING TIMER1/3 IN ASYNCHRONOUS COUNTER MODE

Reading TMRxH or TMRxL while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the TMRxH:TMRxL register pair.

13.6 Timer1/3 Gate

Timer1/3 can be configured to count freely or the count can be enabled and disabled using Timer1/3 gate circuitry. This is also referred to as Timer1/3 gate count enable.

Timer1/3 gate can also be driven by multiple selectable sources.

13.6.1 TIMER1/3 GATE COUNT ENABLE

The Timer1/3 gate is enabled by setting the TMRxGE bit of the TxGCON register. The polarity of the Timer1/3 gate is configured using the TxGPOL bit of the TxGCON register.

When Timer1/3 gate (TxG) input is active, Timer1/3 will increment on the rising edge of the Timer1/3 clock source. When Timer1/3 gate input is inactive, no incrementing will occur and Timer1/3 will hold the current count. See Figure 13-3 for timing details.

TABLE 13-4: TIMER1/3 GATE ENABLE SELECTIONS

TxCLK	TxGPOL	TxG	Timer1/3 Operation
↑	0	0	Counts
↑	0	1	Holds Count
↑	1	0	Holds Count
↑	1	1	Counts

13.6.2 TIMER1/3 GATE SOURCE SELECTION

The Timer1/3 gate source can be selected from one of four different sources. Source selection is controlled by the TxGSS bits of the TxGCON register. The polarity for each available source is also selectable. Polarity selection is controlled by the TxGPOL bit of the TxGCON register.

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TABLE 13-5: TIMER1/3 GATE SOURCES

TxGSS	Timer1 Gate Source	Timer3 Gate Source
00	Timer1 Gate Pin	Timer3 Gate Pin
01	Overflow of TimerA (TMRA increments from FFh to 00h)	Overflow of TimerB (TMRB increments from FFh to 00h)
10	Timer2 match PR2 (TMR2 increments to match PR2)	Timer2 match PR2 (TMR2 increments to match PR2)
11	Count Enabled by WDT Overflow (Watchdog Time-out interval expired)	Count Enabled by WDT Overflow (Watchdog Time-out interval expired)

13.6.3 TxG PIN GATE OPERATION

The TxG pin is one source for Timer1/3 gate control. It can be used to supply an external source to the Timer1/3 gate circuitry. Timer1 gate can be configured for the T1G pin and Timer3 gate can be configured for the T3G pin.

13.6.4 TIMERA/B OVERFLOW GATE OPERATION

When TimerA/B increments from FFh to 00h a low-to-high pulse will automatically be generated and internally supplied to the Timer1/3 gate circuitry. Timer1 gate can be configured for TimerA overflow and Timer3 gate can be configured for TimerB overflow.

13.6.5 TIMER2 MATCH GATE OPERATION

The TMR2 register will increment until it matches the value in the PR2 register. On the very next increment cycle, TMR2 will be reset to 00h. When this Reset occurs, a low-to-high pulse will automatically be generated and internally supplied to the Timer1/3 gate circuitry. Both Timer1 gate and Timer3 gate can be configured for the Timer2 match.

13.6.6 WATCHDOG OVERFLOW GATE OPERATION

The Watchdog Timer oscillator, prescaler and counter will be automatically turned on when TMRxGE = 1 and TxGSS selects the WDT as a gate source for Timer1/3 (TxGSS = 11). TMRxON does not factor into the oscillator, prescaler and counter enable. See Table 13-6. Both Timer1 gate and Timer3 gate can be configured for Watchdog overflow.

The PSA and PS bits of the OPTION register still control what time-out interval is selected. Changing the prescaler during operation may result in a spurious capture.

Enabling the Watchdog Timer oscillator does not automatically enable a Watchdog Reset or wake-up from Sleep upon counter overflow.

Note:	When using the WDT as a gate source for Timer1/3, operations that clear the Watchdog Timer (CLRWDT, SLEEP instructions) will affect the time interval being measured for capacitive sensing. This includes waking from Sleep. All other interrupts that might wake the device from Sleep should be disabled to prevent them from disturbing the measurement period.

As the gate signal coming from the WDT counter will generate different pulse widths, depending on if the WDT is enabled, when the CLRWDT instruction is executed, and so on, Toggle mode must be used. A specific sequence is required to put the device into the correct state to capture the next WDT counter interval.

TABLE 13-6: WDT/TIMER1/3 GATE INTERRACTION

WDTE	TMRxGE = 1 and TxGSS = 11	WDT Oscillator Enable	WDT Reset	Wake-up	WDT Available for TxG Source
1	N	Y	Y	Y	N
1	Y	Y	Y	Y	Y
0	Y	Y	N	N	Y
0	N	N	N	N	N

13.6.7 TIMER1/3 GATE TOGGLE MODE

When Timer1/3 Gate Toggle mode is enabled, it is possible to measure the full-cycle length of a Timer1/3 gate signal, as opposed to the duration of a single level pulse.

The Timer1/3 gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. See Figure 13-4 for timing details.

Timer1/3 Gate Toggle mode is enabled by setting the TxGTM bit of the TxGCON register. When the TxGTM bit is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.

Note:	Enabling Toggle mode at the same time as changing the gate polarity may result in indeterminate operation.

13.6.8 TIMER1/3 GATE SINGLE-PULSE MODE

When Timer1/3 Gate Single-Pulse mode is enabled, it is possible to capture a single pulse gate event. Timer1/3 Gate Single-Pulse mode is first enabled by setting the TxGSPM bit in the TxGCON register. Next, the TxGGO/DONE bit in the TxGCON register must be set. The Timer1/3 will be fully enabled on the next incrementing edge. On the next trailing edge of the pulse, the TxGGO/DONE bit will automatically be cleared. No other gate events will be allowed to increment Timer1/3 until the TxGGO/DONE bit is once again set in software.

Clearing the TxGSPM bit of the TxGCON register will also clear the TxGGO/DONE bit. See Figure 13-5 for timing details.

Enabling the Toggle mode and the Single-Pulse mode simultaneously will permit both sections to work together. This allows the cycle times on the Timer1/3 gate source to be measured. See Figure 13-6 for timing details.

13.6.9 TIMER1/3 GATE VALUE STATUS

When Timer1/3 gate value status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the TxGVAL bit in the TxGCON register. The TxGVAL bit is valid even when the Timer1/3 gate is not enabled (TMRxGE bit is cleared).

13.6.10 TIMER1/3 GATE EVENT INTERRUPT

When Timer1/3 gate event interrupt is enabled, it is possible to generate an interrupt upon the completion of a gate event. When the falling edge of TxGVAL occurs, the TMRxGIF flag bit in the PIRx register will be set. If the TMRxGIE bit in the PIEx register is set, then an interrupt will be recognized. See Table 13-7 for interrupt bit locations.

The TMRxGIF flag bit operates even when the Timer1/3 gate is not enabled (TMRxGE bit is cleared).

TABLE 13-7: TIMER1/3 INTERRUPT BIT LOCATIONS

	Timer1	Timer3
Interrupt Flag	TMR1IF bit in PIR1 register	TMR3IF bit in PIR2 register
Interrupt Enable	TMR1IE bit in PIE1 register	TMR3IE bit in PIE2 register
Gate Interrupt Flag	TMR1GIF bit in PIR1 register	TMR3GIF bit in PIR2 register
Gate Interrupt Enable	TMR1GIE bit in PIE1 register	TMR3GIE bit in PIE2 register

13.7 Timer1/3 Interrupt

The Timer1/3 register pair (TMRxH:TMRxL) increments to FFFFh and rolls over to 0000h. When Timer1/3 rolls over, the Timer1/3 interrupt flag bit of the PIRx register is set. See Table 13-7 for interrupt bit locations.

To enable the interrupt on rollover, you must set these bits:

- TMRxON bit of the TxCON register
- TMRxIE bit of the PIEx register
- PEIE bit of the INTCON register
- GIE bit of the INTCON register

The interrupt is cleared by clearing the TMRxIF bit in the Interrupt Service Routine.

Note:	The TMRxH:TMRxL register pair and the TMRxIF bit should be cleared before enabling interrupts.

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13.8 Timer1/3 Operation During Sleep

Timer1/3 can only operate during Sleep when setup in Asynchronous Counter mode. In this mode, an external crystal or clock source can be used to increment the counter. To set up the timer to wake the device:

- TMRxON bit of the TxCON register must be set
- TMRxIE bit of the PIEx register must be set
- PEIE bit of the INTCON register must be set
- $\overline{\text{TxSYNC}}$ bit of the TxCON register must be set
- TMRxCS bits of the TxCON register must be configured
- T1OSCEN bit of the T1CON register must be configured
- TMRxGIE bit of the TxGCON register must be configured

The device will wake-up on an overflow and execute the next instructions. If the GIE bit of the INTCON register is set, the device will call the Interrupt Service Routine (0004h).

13.9 CCP Capture/Compare Time Base (Timer1 Only)

The CCP module uses the TMR1H:TMR1L register pair as the time base when operating in Capture or Compare mode.

In Capture mode, the value in the TMR1H:TMR1L register pair is copied into the CCPR1H:CCPR1L register pair on a configured event.

In Compare mode, an event is triggered when the value CCPR1H:CCPR1L register pair matches the value in the TMR1H:TMR1L register pair. This event can be a Special Event Trigger.

For more information, see **Section 17.0 "Capture/Compare/PWM (CCP) Module"**.

13.10 CCP Special Event Trigger (Timer1 only)

When the CCP is configured to trigger a special event, the trigger will clear the TMR1H:TMR1L register pair. This special event does not cause a Timer1 interrupt. The CCP module may still be configured to generate a CCP interrupt.

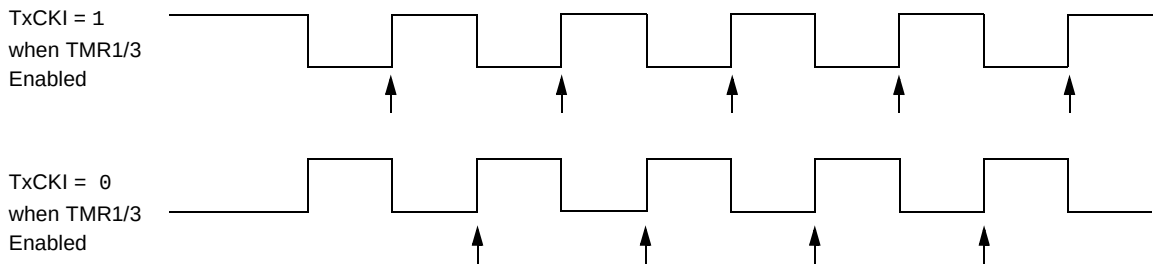
In this mode of operation, the CCPR1H:CCPR1L register pair becomes the period register for Timer1.

Timer1 should be synchronized to the FOSC/4 to utilize the Special Event Trigger. Asynchronous operation of Timer1 can cause a Special Event Trigger to be missed.

In the event that a write to TMR1H or TMR1L coincides with a Special Event Trigger from the CCP, the write will take precedence.

For more information, see **Section 17.2.4 "Special Event Trigger"**.

FIGURE 13-2: TIMER1/TIMER3 INCREMENTING EDGE



Note 1: Arrows indicate counter increments.

2: In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge of the clock.

FIGURE 13-3: TIMER1/TIMER3 GATE COUNT ENABLE MODE

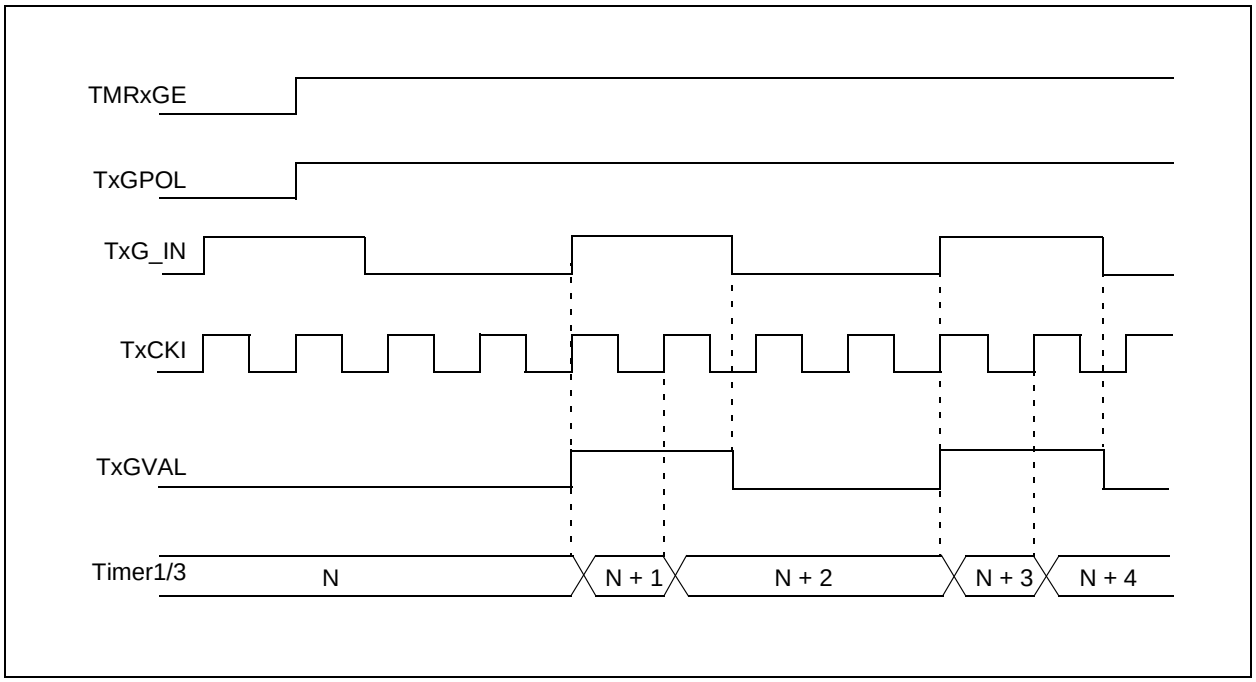


FIGURE 13-4: TIMER1/TIMER3 GATE TOGGLE MODE

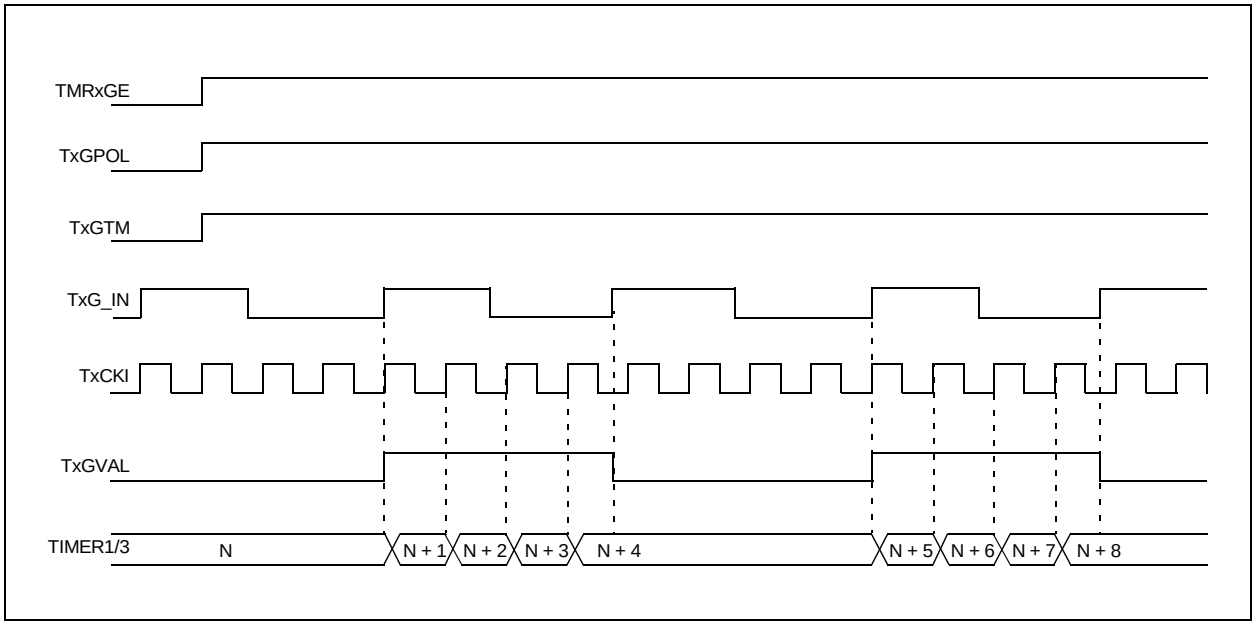


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FIGURE 13-5: TIMER1/TIMER3 GATE SINGLE-PULSE MODE

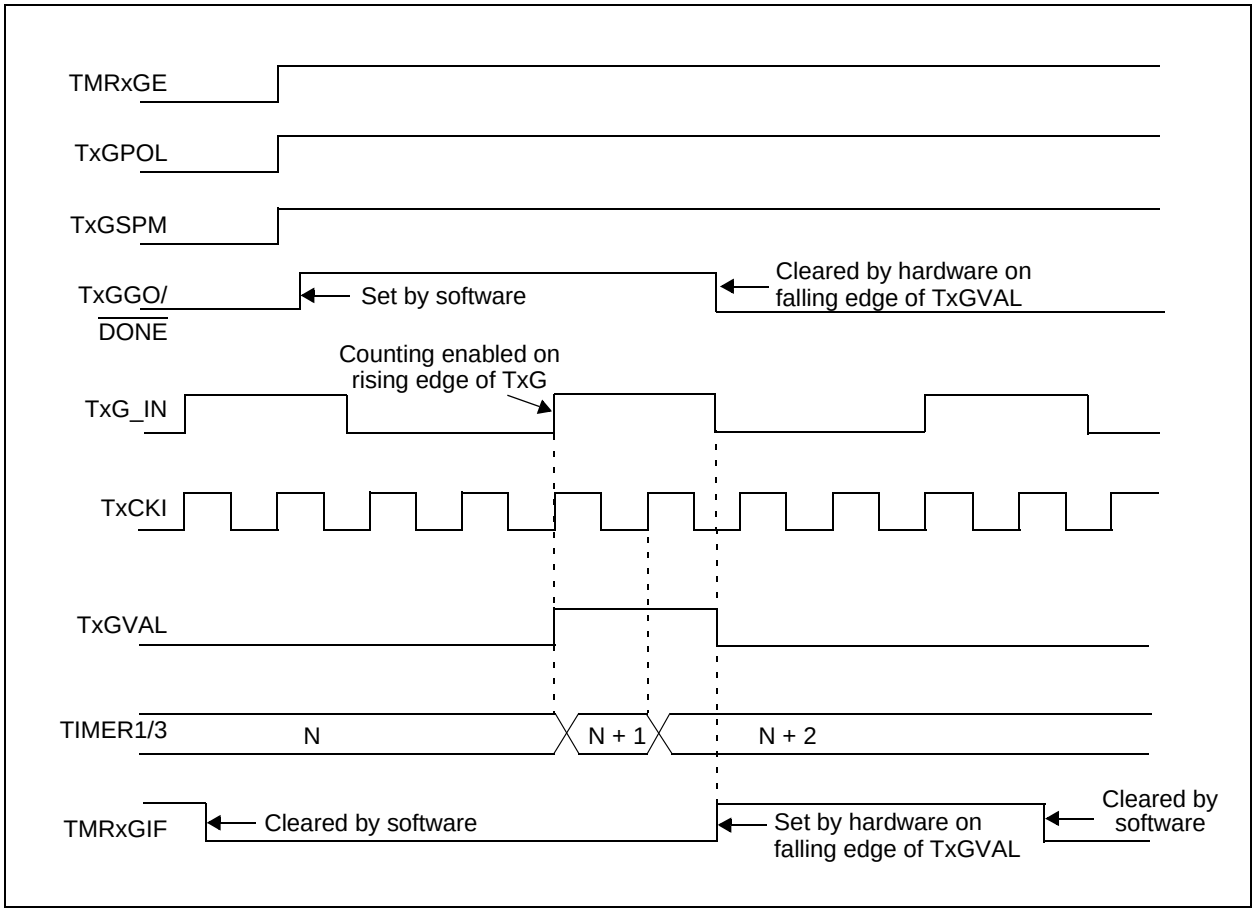
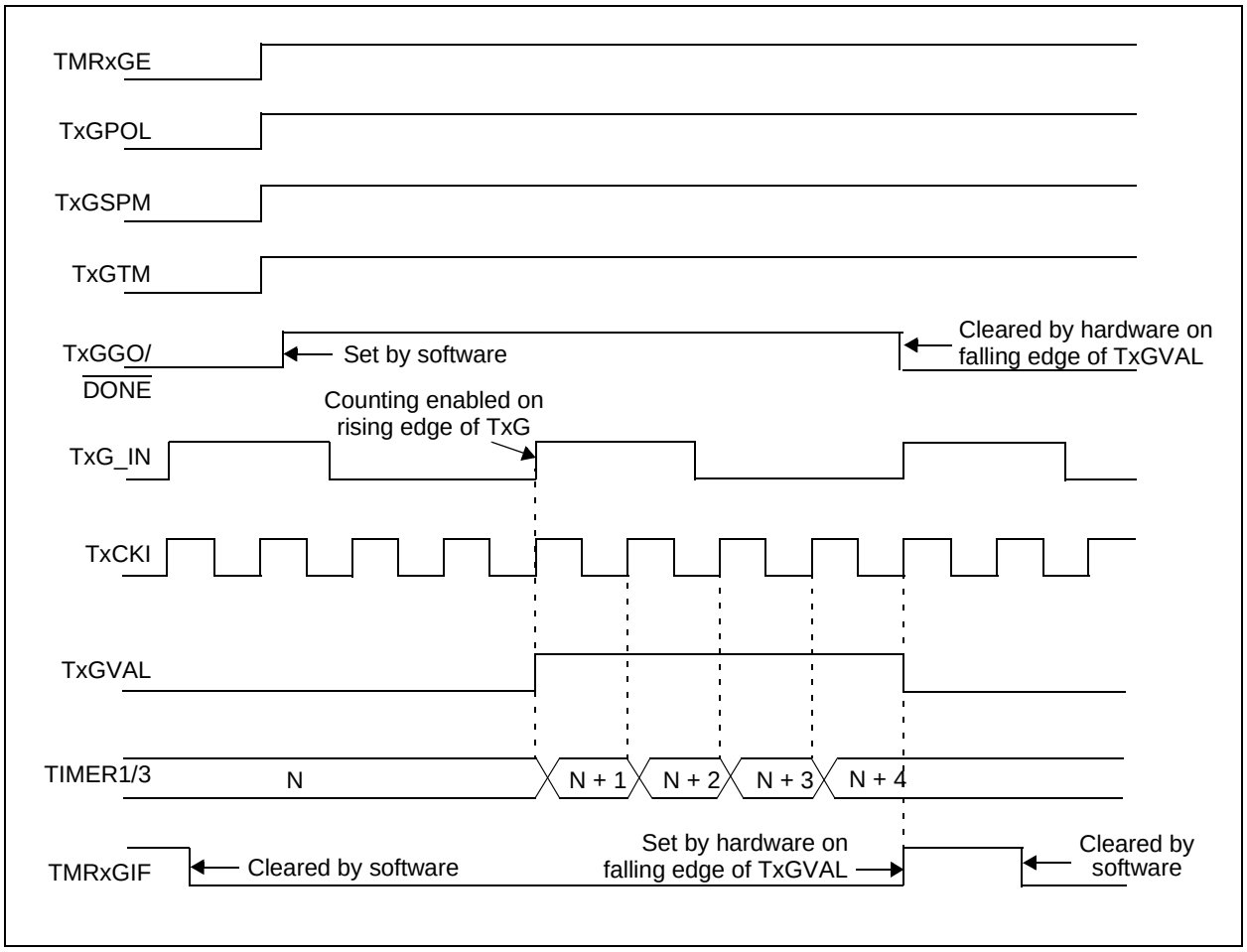


FIGURE 13-6: TIMER1/TIMER3 GATE SINGLE-PULSE AND TOGGLE COMBINED MODE



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13.11 Timer1/3 Control Register

The Timer1/3 Control register (TxCON), shown in Register 13-1, is used to control Timer1/3 and select the various features of the Timer1/3 module.

REGISTER 13-1: TxCON: TIMER1/TIMER3 CONTROL REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	U-0	R/W-0/0
TMRxCS1	TMRxCS0	TxCKPS1	TxCKPS0	T1OSCEN(1)	$\overline{\text{TxSYNC}}$	—	TMRxON
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 **TMRxCS<1:0>:** Timerx Clock Source Select bits
- 11 = Timerx clock source is Capacitive Sensing Oscillator (CPSxOSC)
- 10 = Timerx clock source is pin or oscillator:
  - If T1OSCEN = 0:
  - External clock from TxCKI pin (on the rising edge)
  - If T1OSCEN = 1:
  - Crystal oscillator on T1OSI/T1OSO pins
- 01 = Timerx clock source is system clock (FOSC)
- 00 = Timerx clock source is instruction clock (FOSC/4)

bit 5-4 **TxCKPS<1:0>:** Timerx Input Clock Prescale Select bits
- 11 = 1:8 Prescale value
- 10 = 1:4 Prescale value
- 01 = 1:2 Prescale value
- 00 = 1:1 Prescale value

bit 3 **T1OSCEN:** LP Oscillator Enable Control bit(1)
- 1 = Dedicated Timer1/3 oscillator circuit enabled
- 0 = Dedicated Timer1/3 oscillator circuit disabled

bit 2 **$\overline{\text{TxSYNC}}$:** Timerx External Clock Input Synchronization Control bit

If TMRxCS<1:0> = 1X
- 1 = Do not synchronize external clock input
- 0 = Synchronize external clock input with system clock (FOSC)

If TMRxCS<1:0> = 0X

This bit is ignored. Timerx uses the internal clock when TMR1CS<1:0> = 0X.

bit 1 **Unimplemented:** Read as '0'

bit 0 **TMRxON:** Timerx on bit
- 1 = Enables Timerx
- 0 = Stops Timerx
  - Clears Timerx gate flip-flop

REGISTER 13-2: TxGCON: TIMER1/TIMER3 GATE CONTROL REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R-0/0	R/W-0/0	R/W-0/0
TMRxGE	TxGPOL	TxGTM	TxGSPM	TxGGO/$\overline{\text{DONE}}$	TxGVAL	TxGSS1	TxGSS0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	HC = Bit is cleared by hardware

bit 7 **TMRxGE:** Timerx Gate Enable bit
If TMRxON = 0:
This bit is ignored.
If TMRxON = 1:
1 = Timerx counting is controlled by the Timerx gate function
0 = Timerx counts regardless of Timerx gate function

bit 6 **TxGPOL:** Timerx Gate Polarity bit
1 = Timerx gate is active-high (Timerx counts when gate is high)
0 = Timerx gate is active-low (Timerx counts when gate is low)

bit 5 **TxGTM:** Timerx Gate Toggle Mode bit
1 = Timerx Gate Toggle mode is enabled
0 = Timerx Gate Toggle mode is disabled and toggle flip-flop is cleared
Timerx gate flip-flop toggles on every rising edge.

bit 4 **TxGSPM:** Timerx Gate Single-Pulse Mode bit
1 = Timerx gate Single-Pulse mode is enabled and is controlling Timerx gate
0 = Timerx gate Single-Pulse mode is disabled

bit 3 **TxGGO/$\overline{\text{DONE}}$:** Timerx Gate Single-Pulse Acquisition Status bit
1 = Timerx gate single-pulse acquisition is ready, waiting for an edge
0 = Timerx gate single-pulse acquisition has completed or has not been started
This bit is automatically cleared when T1GSPM is cleared.

bit 2 **TxGVAL:** Timerx Gate Current State bit
Indicates the current state of the Timerx gate that could be provided to TMRxH:TMRxL.
Unaffected by Timerx Gate Enable (TMRxGE).

bit 1-0 **TxGSS<1:0>:** Timerx Gate Source Select bits
00 = Timerx gate pin
01 = TimerA/B overflow output
10 = TMR2 Match PR2 output
11 = Watchdog Timer scaler overflow
Watchdog Timer oscillator is turned on if TMRxGE = 1, regardless of the state of TMR1ON.

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14.0 TIMERA/B MODULES

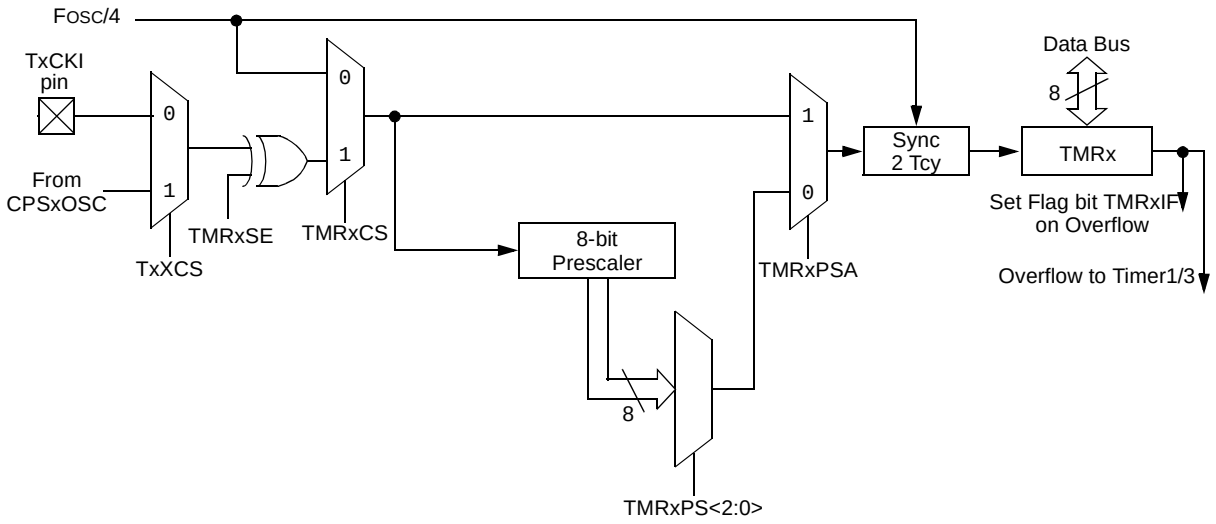
TimerA and TimerB are two more Timer0-type modules. Timers A and B are available as general-purpose timers/counters, and are closely integrated with the capacitive sensing modules.

The TimerA/B modules incorporate the following features:

- 8-bit timer/counter register (TMRx)
- 8-bit prescaler
- Programmable internal or external clock source
- Programmable external clock edge selection
- Interrupt on overflow
- TMRA can be used to gate Timer1
- TMRB can be used to gate Timer3

Figure 14-1 is a block diagram of the TimerA/TimerB modules.

FIGURE 14-1: BLOCK DIAGRAM OF THE TIMERA/TIMERB PRESCALER



Note 1: TxXCS is in the CPSxCON0 register.

14.1 TimerA/B Operation

The TimerA/B modules can be used as either 8-bit timers or 8-bit counters. Additionally, the modules can also be used to set Timer1's/Timer3's period of measurement for the capacitive sensing modules via Timer1's or Timer3's gate feature.

TABLE 14-1: CPSOSC/TIMER ASSOCIATION

Cap Sense Oscillator	Divider Timer	Period Measurement
CPS A	TimerA	Timer1
CPS B	TimerB	Timer3

14.1.1 8-BIT TIMER MODE

The TimerA/B modules will increment every instruction cycle, if used without a prescaler. 8-bit Timer mode is selected by clearing the TMRxCS bit of the TxCON registers.

When TMRx is written, the increment is inhibited for two instruction cycles immediately following the write.

Note:	The value written to the TMRx register can be adjusted, in order to account for the two instruction cycle delay when TMRx is written.

14.1.2 8-BIT COUNTER MODE

In 8-bit Counter mode, the TimerA/B modules will increment on every rising or falling edge of the TxCKI pin or the Capacitive Sensing Oscillator (CPSxOSC) signal. 8-bit Counter mode using the TxCKI pin is selected by setting the TMRxCS bit of the TxCON register to '1' and resetting the TxXCS bit in the CPSxCON0 register to '0'. 8-bit Counter mode using the Capacitive Sensing Oscillator (CPSxOSC) signal is selected by setting the TMRxCS bit in the TxCON register to '1' and setting the TxXCS bit in the CPSxCON0 register to '1'.

The rising or falling transition of the incrementing edge for either input source is determined by the TMRxSE bit in the TxCON register.

14.1.3 SOFTWARE PROGRAMMABLE PRESCALER

For TimerA/B modules, the software programmable prescaler is exclusive to the Timer. The prescaler is enabled by clearing the TMRxPSA bit of the TxCON register.

There are 8 prescaler options for TimerA/B modules ranging from 1:2 to 1:256. The prescale values are selectable via the TMRxPS<2:0> bits of the TxCON register for TimerA/B. In order to have a 1:1 prescaler value for the TimerA/B modules, the prescaler must be disabled.

The prescaler is not readable or writable. When the prescaler is enabled or assigned to the Timer module, all instructions writing to the TMRx register will clear the prescaler. Enabling the TimerA/B modules also clears the prescaler.

14.1.4 TIMERA/B INTERRUPT

TimerA/B will generate an interrupt when the corresponding TMR register overflows from FFh to 00h. The TMRxIF interrupt flag bit of the PIR2 register is set every time the TMRx register overflows. These interrupt flag bits are set regardless of whether or not the relative Timer interrupt is enabled. The interrupt flag bits can only be cleared in software. The TimerA/B interrupt enable bits are the TMRxIE in the PIE2 register.

Note:	TimerA/B interrupts cannot wake the processor from Sleep since the timer is frozen during Sleep.

14.1.5 USING TIMERA/B WITH AN EXTERNAL CLOCK

When TimerA/B is in Counter mode, the synchronization of the TxCKI input and the TMRx register is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks. Therefore, the high and low periods of the external clock source must meet the timing requirements as shown in **Section 25.0 "Electrical Specifications"**.

14.1.6 TIMER ENABLE

Operation of TimerA/B is enabled by setting the TMRxON bit of the TxCON register. When the module is disabled, the value in the TMRx register is maintained. Enabling the TMRx module will reset the prescaler used by the counter.

14.1.7 OPERATION DURING SLEEP

TimerA and TimerB cannot operate while the processor is in Sleep mode. The contents of the TMRx registers will remain unchanged while the processor is in Sleep mode.

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REGISTER 14-1: TxCON: TIMERA/TIMERB CONTROL REGISTER

R/W-0/0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
TMRxON	—	TMRxCS	TMRxSE	$\overline{\text{TMRxPSA}}$	TMRxPS2	TMRxPS1	TMRxPS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **TMRxON:** TimerA/TimerB On/Off Control bit
1 = Timerx is enabled
0 = Timerx is disabled

bit 6 **Unimplemented:** Read as '0'

bit 5 **TMRxCS:** TMRx Clock Source Select bit
1 = Transition on TxCKI pin or CPSxOSC signal
0 = Internal instruction cycle clock (FOSC/4)

bit 4 **TMRxSE:** TMRx Source Edge Select bit
1 = Increment on high-to-low transition on TxCKI pin
0 = Increment on low-to-high transition on TxCKI pin

bit 3 **$\overline{\text{TMRxPSA}}$:** Prescaler Assignment bit
1 = Prescaler is disabled. Timer clock input bypasses prescaler.
0 = Prescaler is enabled. Timer clock input comes from the prescaler output.

bit 2-0 **TMRxPS<2:0>:** Prescaler Rate Select bits

BIT VALUE	TMRx RATE
000	1 : 2
001	1 : 4
010	1 : 8
011	1 : 16
100	1 : 32
101	1 : 64
110	1 : 128
111	1 : 256

TABLE 14-2: SUMMARY OF REGISTERS ASSOCIATED WITH TIMERA/B

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CPSACON0	CPSAON	CPSARM	—	—	CPSARNG1	CPSARNG0	CPSAOUT	TAXCS	00-- 0000	00-- 0000
CPSBCON0	CPSBON	CPSBRM	—	—	CPSBRNG1	CPSBRNG0	CPSBOUT	TBXCS	00-- 0000	00-- 0000
PIE2	TMR3GIE	TMR3IE	TMRBIE	TMRAIE	—	—	—	CCP2IE	0000 ---0	0000 ---0
PIR2	TMR3GIF	TMR3IF	TMRBIF	TMRAIF	—	—	—	CCP2IF	0000 ---0	0000 ---0
TACON	TMRAON	—	TACS	TASE	TAPSA	TAPS2	TAPS1	TAPS0	0-00 0000	0-00 0000
TBCON	TMRBON	—	TBCS	TBSE	TBPSA	TBPS2	TBPS1	TBPS0	0-00 0000	0-00 0000
TMRA	TimerA Module Register								0000 0000	0000 0000
TMRB	TimerB Module Register								0000 0000	0000 0000
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

Legend: – = Unimplemented locations, read as '0'. Shaded cells are not used by the TimerA/B modules.

NOTES:

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15.0 TIMER2 MODULE

The Timer2 module is an 8-bit timer with the following features:

- 8-bit timer register (TMR2)
- 8-bit period register (PR2)
- Interrupt on TMR2 match with PR2
- Software programmable prescaler (1:1, 1:4, 1:16)
- Software programmable postscaler (1:1 to 1:16)

See Figure 15-1 for a block diagram of Timer2.

15.1 Timer2 Operation

The clock input to the Timer2 module is the system instruction clock (FOSC/4). The clock is fed into the Timer2 prescaler, which has prescale options of 1:1, 1:4 or 1:16. The output of the prescaler is then used to increment the TMR2 register.

The values of TMR2 and PR2 are constantly compared to determine when they match. TMR2 will increment from 00h until it matches the value in PR2. When a match occurs, two things happen:

- TMR2 is reset to 00h on the next increment cycle.
- The Timer2 postscaler is incremented.

The match output of the Timer2/PR2 comparator is then fed into the Timer2 postscaler. The postscaler has postscale options of 1:1 to 1:16 inclusive. The output of the Timer2 postscaler is used to set the TMR2IF interrupt flag bit in the PIR1 register.

The TMR2 and PR2 registers are both fully readable and writable. On any Reset, the TMR2 register is set to 00h and the PR2 register is set to FFh.

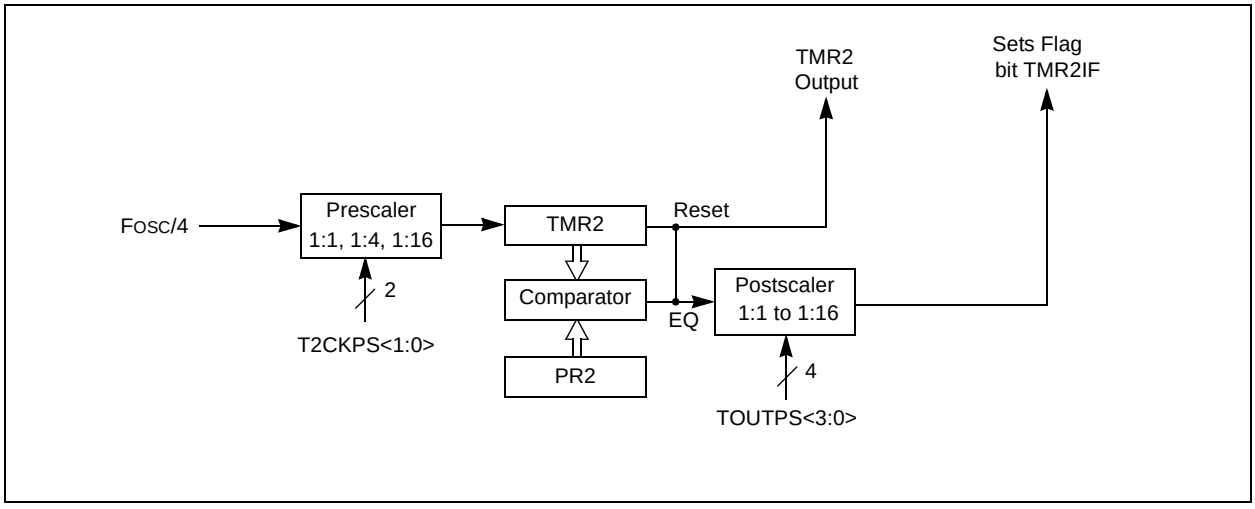
Timer2 is turned on by setting the TMR2ON bit in the T2CON register to a '1'. Timer2 is turned off by clearing the TMR2ON bit to a '0'.

The Timer2 prescaler is controlled by the T2CKPS bits in the T2CON register. The Timer2 postscaler is controlled by the TOUTPS bits in the T2CON register. The prescaler and postscaler counters are cleared when:

- A write to TMR2 occurs.
- A write to T2CON occurs.
- Any device Reset occurs (Power-on Reset, $\overline{\text{MCLR}}$ Reset, Watchdog Timer Reset, or Brown-out Reset).

Note: TMR2 is not cleared when T2CON is written.

FIGURE 15-1: TIMER2 BLOCK DIAGRAM



REGISTER 15-1: T2CON: TIMER2 CONTROL REGISTER

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **Unimplemented:** Read as '0'

bit 6-3 **TOUTPS<3:0>:** Timer2 Output Postscaler Select bits

0000 = 1:1 Postscaler
0001 = 1:2 Postscaler
0010 = 1:3 Postscaler
0011 = 1:4 Postscaler
0100 = 1:5 Postscaler
0101 = 1:6 Postscaler
0110 = 1:7 Postscaler
0111 = 1:8 Postscaler
1000 = 1:9 Postscaler
1001 = 1:10 Postscaler
1010 = 1:11 Postscaler
1011 = 1:12 Postscaler
1100 = 1:13 Postscaler
1101 = 1:14 Postscaler
1110 = 1:15 Postscaler
1111 = 1:16 Postscaler

bit 2 **TMR2ON:** Timer2 On bit

1 = Timer2 is on
0 = Timer2 is off

bit 1-0 **T2CKPS<1:0>:** Timer2 Clock Prescale Select bits

00 = Prescaler is 1
01 = Prescaler is 4
1x = Prescaler is 16

TABLE 15-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER2

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PR2	Timer2 Module Period Register								1111 1111	1111 1111
TMR2	Holding Register for the 8-bit TMR2 Register								0000 0000	0000 0000
T2CON	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000

Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used for Timer2 module.

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16.0 CAPACITIVE SENSING MODULE

The capacitive sensing modules (CSM) allow for an interaction with an end user without a mechanical interface. In a typical application, the capacitive sensing module is attached to a pad on a Printed Circuit Board (PCB), which is electrically isolated from the end user. When the end user places their finger over the PCB pad, a capacitive load is added, causing a frequency shift in the capacitive sensing module. The capacitive sensing module requires software and at least one timer resource to determine the change in frequency. Key features of this module include:

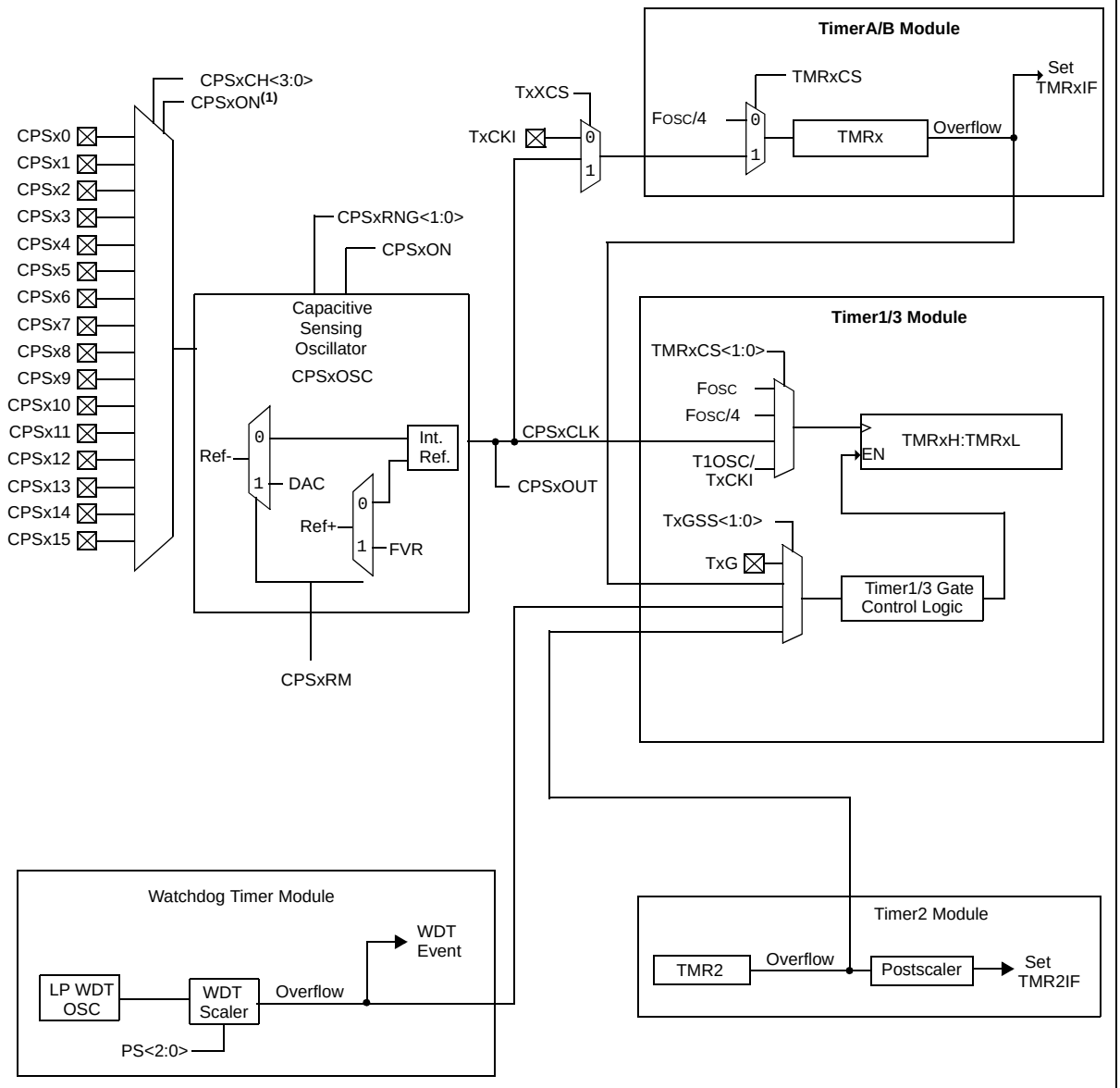
- Analog MUX for monitoring multiple inputs
- Capacitive sensing oscillator
- Multiple Power modes
- High power range with variable voltage references
- Multiple timer resources
- Software control
- Operation during sleep
- Acquire two samples simultaneously (when using both CSM modules)

Two identical capacitive sensing modules are implemented on the PIC16F707/PIC16LF707. The modules are named CPSA and CPSB. The timer module integration for both capacitive sensing modules is shown in Table 16-1. A block diagram of the capacitive sensing module is shown in Figure 16-1 and Figure 16-2.

TABLE 16-1: CPSOSC TIMER USAGE

Cap Sense Oscillator	Mode	Frequency Measurement	Duration Control
Cap Sense Oscillator A	TimerA/Software	TimerA	Software
	Timer1/Software	Timer1	Software
	Timer1/TimerA	Timer1	TimerA
Cap Sense Oscillator B	TimerB/Software	TimerB	Software
	Timer3/Software	Timer3	Software
	Timer3/TimerB	Timer3	TimerB

FIGURE 16-1: CAPACITIVE SENSING BLOCK DIAGRAM



Note 1: If CPSxON = 0, disabling capacitive sensing, no channel is selected.

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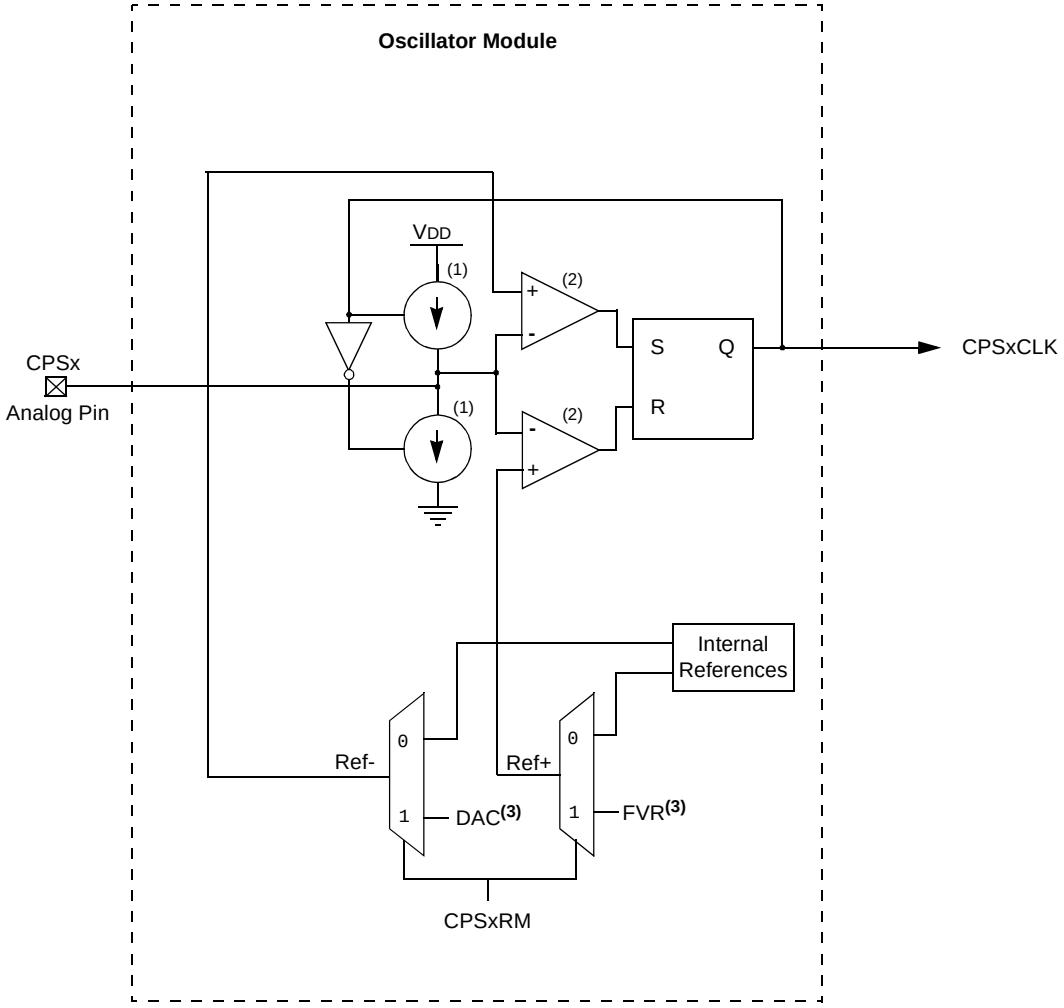
FIGURE 16-2: CAPACITIVE SENSING OSCILLATOR BLOCK DIAGRAM



Note 1: Module Enable and Power mode selections are not shown.
2: Comparators remain active in Noise Detection mode.

16.1 Analog MUX

Each capacitive sensing module can monitor up to 16 inputs, providing 32 capacitive sensing inputs in total. The capacitive sensing inputs are defined as CPSA<15:0> for capacitive sensing module A, and CPSB<15:0> for capacitive sensing module B. To determine if a frequency change has occurred the use must:

- Select the appropriate CPS pin by setting the CPSxCH<3:0> bits of the CPSxCON1 register.
- Set the corresponding ANSEL bit.
- Set the corresponding TRIS bit.
- Run the software algorithm.

Selection of the CPSx pin while the module is enabled will cause the capacitive sensing oscillator to be on the CPSx pin. Failure to set the corresponding ANSEL and TRIS bits can cause the capacitive sensing oscillator to stop, leading to false frequency readings.

16.2 Capacitive Sensing Oscillator

The capacitive sensing oscillator consists of a constant current source and a constant current sink, to produce a triangle waveform. The CPSxOUT bit of the CPSxCON0 register shows the status of the capacitive sensing oscillator, whether it is sinking or sourcing current. The oscillator is designed to drive a capacitive load (single PCB pad) and at the same time, be a clock source to either TimerA/B or Timer1/3. The oscillator has three different current settings as defined by CPSxRNG<1:0> of the CPSxCON0 register. The different current settings for the oscillator serve two purposes:

- Maximize the number of counts in a timer for a fixed time base.
- Maximize the count differential in the timer during a change in frequency.

16.3 Voltage References

The capacitive sensing oscillator uses voltage references to provide two voltage thresholds for oscillation. The upper voltage threshold is referred to as Ref+ and the lower voltage threshold is referred to as Ref-.

The user can elect to use fixed voltage references, which are internal to the capacitive sensing oscillator, or variable voltage references, which are supplied by the Fixed Voltage Reference (FVR) module and the Digital-to-Analog Converter (DAC) module.

When the fixed voltage references are used, the V_{SS} voltage determines the lower threshold level (Ref-) and the V_{DD} voltage determines the upper threshold level (Ref+).

When the variable voltage references are used, the DAC voltage determines the lower threshold level (Ref-) and the FVR voltage determines the upper threshold level (Ref+). An advantage of using these reference sources is that oscillation frequency remains constant with changes in V_{DD}.

Different oscillation frequencies can be obtained through the use of these variable voltage references. The more the upper voltage reference level is lowered and the more the lower voltage reference level is raised, the higher the capacitive sensing oscillator frequency becomes.

Selection between the voltage references is controlled by the CPSxRM bit of the CPSxCON0 register. Setting this bit selects the variable voltage references and clearing this bit selects the fixed voltage references.

Please see **Section 10.0 “Fixed Voltage Reference”** and **Section 11.0 “Digital-to-Analog Converter (DAC) Module”** for more information on configuring the variable voltage levels.

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16.4 Power Modes

The capacitive sensing oscillator can operate in one of seven different power modes. The power modes are separated into two ranges; the low range and the high range.

When the oscillator's low range is selected, the fixed internal voltage references of the capacitive sensing oscillator are being used. When the oscillator's high range is selected, the variable voltage references supplied by the FVR and DAC modules are being used. Selection between the voltage references is controlled by the CPSxRM bit of the CPSxCON0 register. See **Section 16.3 "Voltage References"** for more information.

Within each range there are three distinct power modes; Low, Medium and High. Current consumption is dependent upon the range and mode selected. Selecting power modes within each range is accomplished by configuring the CPSxRNG <1:0> bits in the CPSxCON0 register. See Table 16-2 for proper power mode selection.

The remaining mode is a Noise Detection mode that resides within the high range. The Noise Detection mode is unique in that it disables the sinking and sourcing of current on the analog pin but leaves the rest of the oscillator circuitry active. This reduces the oscillation frequency on the analog pin to zero and also greatly reduces the current consumed by the oscillator module.

When noise is introduced onto the pin, the oscillator is driven at the frequency determined by the noise. This produces a detectable signal at the comparator output, indicating the presence of activity on the pin.

Figure 16-2 shows a more detailed drawing of the current sources and comparators associated with the oscillator.

TABLE 16-2: POWER MODE SELECTION

CPSxRM	Range	CPSxRNG<1:0>	Mode	Nominal Current (1)
0	Low	00	Off	0.0 µA
		01	Low	0.1 µA
		10	Medium	1.2 µA
		11	High	18 µA
1	High	00	Noise Detection	0.0 µA
		01	Low	9 µA
		10	Medium	30 µA
		11	High	100 µA

Note: See **Section 25.0 "Electrical Specifications"** for more information.

16.5 Timer Resources

To measure the change in frequency of the capacitive sensing oscillator, a fixed time base is required. For the period of the fixed time base, the capacitive sensing oscillator is used to clock either TimerA/B or Timer1/3 (for CPSA/B, respectively). The frequency of the capacitive sensing oscillator is equal to the number of counts in the timer divided by the period of the fixed time base.

16.6 Fixed Time Base

To measure the frequency of the capacitive sensing oscillator, a fixed time base is required. Any timer resource or software loop can be used to establish the fixed time base. It is up to the end user to determine the method in which the fixed time base is generated.

Note:	The fixed time base can not be generated by the timer resource that the capacitive sensing oscillator is clocking.

16.6.1 TIMERA/B

To select TimerA/B as the timer resource for the capacitive sensing module:

- Set the TAXCS/TBXCS bit of the CPSACON0/ CPSBCON0 register.
- Clear the TMRACS/TMRBCS bit of the TACON/ TBCON register.

When TimerA/B is chosen as the timer resource, the capacitive sensing oscillator will be the clock source for TimerA/B. Refer to **Section 14.0 "TimerA/B Modules"** for additional information.

16.6.2 TIMER1/3

To select Timer1/3 as the timer resource for the capacitive sensing module, set the TMRxCS<1:0> of the TxCON register to '11'. When Timer1/3 is chosen as the timer resource, the capacitive sensing oscillator will be the clock source for Timer1/3. Because the Timer1/3 module has a gate control, developing a time base for the frequency measurement can be simplified by using the TimerA/B overflow flag.

It is recommend that the TimerA/B overflow flag, in conjunction with the Toggle mode of the Timer1/3 gate, be used to develop the fixed time base required by the software portion of the capacitive sensing module. Refer to **Section 13.11 "Timer1/3 Control Register "** for additional information.

TABLE 16-3: TIMER1/3 ENABLE FUNCTION

TMRxON	TMRxGE	Timerx Operation
0	0	Off
0	1	Off
1	0	On
1	1	Count Enabled by Input

16.7 Software Control

The software portion of the capacitive sensing module is required to determine the change in frequency of the capacitive sensing oscillator. This is accomplished by the following:

- Setting a fixed time base to acquire counts on TimerA/B or Timer1/3.
- Establishing the nominal frequency for the capacitive sensing oscillator.
- Establishing the reduced frequency for the capacitive sensing oscillator due to an additional capacitive load.
- Set the frequency threshold.

16.7.1 NOMINAL FREQUENCY (NO CAPACITIVE LOAD)

To determine the nominal frequency of the capacitive sensing oscillator:

- Remove any extra capacitive load on the selected CPSx pin.
- At the start of the fixed time base, clear the timer resource.
- At the end of the fixed time base, save the value in the timer resource.

The value of the timer resource is the number of oscillations of the capacitive sensing oscillator for the given time base. The frequency of the capacitive sensing oscillator is equal to the number of counts on the timer divided by the period of the fixed time base.

16.7.2 REDUCED FREQUENCY (ADDITIONAL CAPACITIVE LOAD)

The extra capacitive load will cause the frequency of the capacitive sensing oscillator to decrease. To determine the reduced frequency of the capacitive sensing oscillator:

- Add a typical capacitive load on the selected CPSx pin.
- Use the same fixed time base as the nominal frequency measurement.
- At the start of the fixed time base, clear the timer resource.
- At the end of the fixed time base, save the value in the timer resource.

The value of the timer resource is the number of oscillations of the capacitive sensing oscillator with an additional capacitive load. The frequency of the capacitive sensing oscillator is equal to the number of counts on the timer divided by the period of the fixed time base. This frequency should be less than the value obtained during the nominal frequency measurement.

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16.7.3 FREQUENCY THRESHOLD

The frequency threshold should be placed midway between the value of nominal frequency and the reduced frequency of the capacitive sensing oscillator. Refer to Application Note AN1103, "*Software Handling for Capacitive Sensing*" (DS01103) for more detailed information on the software required for capacitive sensing module.

Note:	For more information on general capacitive sensing refer to Application Notes: • AN1101, *"Introduction to Capacitive Sensing"* (DS01101) • AN1102, *"Layout and Physical Design Guidelines for Capacitive Sensing"* (DS01102).

16.8 Operation during Sleep

The capacitive sensing oscillator will continue to run as long as the module is enabled, independent of the part being in Sleep. In order for the software to determine if a frequency change has occurred, the part must be awake. However, the part does not have to be awake when the timer resource is acquiring counts.

Note:	TimerA/B does not operate when in Sleep, and therefore cannot be used for capacitive sense measurements in Sleep.

REGISTER 16-1: CPSxCON0: CAPACITIVE SENSING CONTROL REGISTER 0

R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0	R-0/0	R/W-0/0
CPSxON	CPSxRM	—	—	CPSxRNG1	CPSxRNG0	CPSxOUT	TxXCS
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7 **CPSxON:** Capacitive Sensing Module Enable bit
1 = Capacitive sensing module is enabled
0 = Capacitive sensing module is disabled

bit 6 **CPSxRM:** Capacitive Sensing Reference Mode bit
1 = Capacitive sensing module is in high range. DAC and FVR provide oscillator voltage references.
0 = Capacitive sensing module is in low range. Internal oscillator voltage references are used.

bit 5-4 **Unimplemented:** Read as '0'

bit 3-2 **CPSxRNG<1:0>:** Capacitive Sensing Current Range bits
If CPSxRM = 0 (low range):
11 = Oscillator is in high range: Charge/discharge current is nominally 18 µA.
10 = Oscillator is in medium range. Charge/discharge current is nominally 1.2 µA.
01 = Oscillator is in low range. Charge/discharge current is nominally 0.1 µA.
00 = Oscillator is off.
If CPSxRM = 1 (high range):
11 = Oscillator is in high range: Charge/discharge current is nominally 100 µA.
10 = Oscillator is in medium range. Charge/discharge current is nominally 30 µA.
01 = Oscillator is in low range. Charge/discharge current is nominally 9 µA.
00 =Oscillator is on; Noise Detection mode; No charge/discharge current is supplied.

bit 1 **CPSxOUT:** Capacitive Sensing Oscillator Status bit
1 = Oscillator is sourcing current (Current flowing out of the pin)
0 = Oscillator is sinking current (Current flowing into the pin)

bit 0 **TxXCS:** TimerA/B External Clock Source Select bit
If TMRxCS = 1:
The TxXCS bit controls which clock external to the core/TimerA/B module supplies TimerA/B:
1 = TimerA/B clock source is the capacitive sensing oscillator
0 = TimerA/B clock source is the TxCKI pin
If TMRxCS = 0:
TimerA/B clock source is controlled by the core/TimerA/B module and is FOSC/4.

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REGISTER 16-2: CPSxCON1: CAPACITIVE SENSING CONTROL REGISTER 1

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	—	CPSxCH3	CPSxCH2	CPSxCH1	CPSxCH0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-4 **Unimplemented:** Read as ‘0’

bit 3-0 **CPSxCH<3:0>:** Capacitive Sensing Channel Select bits

If CPSxON = 0:

These bits are ignored. No channel is selected.

If CPSxON = 1:

0000 = channel 0, (CPSx0)
0001 = channel 1, (CPSx1)
0010 = channel 2, (CPSx2)
0011 = channel 3, (CPSx3)
0100 = channel 4, (CPSx4)
0101 = channel 5, (CPSx5)
0110 = channel 6, (CPSx6)
0111 = channel 7, (CPSx7)
1000 = channel 8, (CPSx8)
1001 = channel 9, (CPSx9)
1010 = channel 10, (CPSx10)
1011 = channel 11, (CPSx11)
1100 = channel 12, (CPSx12)
1101 = channel 13, (CPSx13)
1110 = channel 14, (CPSx14)
1111 = channel 15, (CPSx15)

TABLE 16-4: SUMMARY OF REGISTERS ASSOCIATED WITH CAPACITIVE SENSING

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELA	ANSA7	ANSA6	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	1111 1111	1111 1111
ANSELB	ANSB7	ANSB6	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	1111 1111	1111 1111
ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
ANSELD	ANSD7	ANSD6	ANSD5	ANSD4	ANSD3	ANSD2	ANSD1	ANSD0	1111 1111	1111 1111
ANSELE	—	—	—	—	—	ANSE2	ANSE1	ANSE0	---- -111	---- -111
CPSACON0	CPSAON	CPSARM	—	—	CPSARNG1	CPSARNG0	CPSAOUT	TAXCS	00-- 0000	00-- 0000
CPSACON1	—	—	—	—	CPSACH3	CPSACH2	CPSACH1	CPSACH0	---- 0000	---- 0000
CPSBCON0	CPSBON	CPSBRM	—	—	CPSBRNG1	CPSBRNG0	CPSBOUT	TBXCS	00-- 0000	00-- 0000
CPSBCON1	—	—	—	—	CPSBCH3	CPSBCH2	CPSBCH1	CPSBCH0	---- 0000	---- 0000
TACON	TMRAON	—	TACS	TASE	TAPSA	TAPS2	TAPS1	TAPS0	0-00 0000	0-00 0000
TBCON	TMRBON	—	TBCS	TBSE	TBPSA	TBPS2	TBPS1	TBPS0	0-00 0000	0-00 0000
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{\text{T1SYNC}}$	—	TMR1ON	0000 00-0	0000 00-0
T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	—	T3SYNC	—	TMR3ON	0000 -0-0	0000 -0-0
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	1111 1111	1111 1111
TRISE	—	—	—	—	TRISE3	TRISE2	TRISE1	TRISE0	---- 1111	---- 1111

Legend: — = Unimplemented locations, read as '0'. Shaded cells are not used by the capacitive sensing modules.

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17.0 CAPTURE/COMPARE/PWM (CCP) MODULE

The Capture/Compare/PWM module is a peripheral which allows the user to time and control different events. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external event when a predetermined amount of time has expired. The PWM mode can generate a pulse-width modulated signal of varying frequency and duty cycle.

The timer resources used by the module are shown in Table 17-2.

Additional information on CCP modules is available in Application Note AN594, *"Using the CCP Modules"* (DS00594).

TABLE 17-1: CCP MODE – TIMER RESOURCES REQUIRED

CCP Mode	Timer Resource
Capture	Timer1
Compare	Timer1
PWM	Timer2

Note: Timer3 has no connection to either CCP.

TABLE 17-2: INTERACTION OF TWO CCP MODULES

CCP1 Mode	CCP2 Mode	Interaction
Capture	Capture	Same TMR1 time base
Capture	Compare	Same TMR1 time base(1, 2)
Compare	Compare	Same TMR1 time base(1, 2)
PWM	PWM	The PWMs will have the same frequency and update rate (TMR2 interrupt). The rising edges will be aligned.
PWM	Capture	None
PWM	Compare	None

Note 1: If CCP2 is configured as a Special Event Trigger, CCP1 will clear Timer1, affecting the value captured on the CCP2 pin.

2: If CCP1 is in Capture mode and CCP2 is configured as a Special Event Trigger, CCP2 will clear Timer1, affecting the value captured on the CCP1 pin.

Note: CCPRx and CCPx throughout this document refer to CCPR1 or CCPR2 and CCP1 or CCP2, respectively.

REGISTER 17-1: CCPxCON: CCPx CONTROL REGISTER

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	DCxB1	DCxB0	CCPxM3	CCPxM2	CCPxM1	CCPxM0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Read as '0'

bit 5-4 **DCxB<1:0>**: PWM Duty Cycle Least Significant bits
Capture mode:
Unused
Compare mode:
Unused
PWM mode:
These bits are the two LSbs of the PWM duty cycle. The eight MSbs are found in CCPRxL.

bit 3-0 **CCPxM<3:0>:** CCP Mode Select bits
0000 = Capture/Compare/PWM off (resets CCP module)
0001 = Unused (reserved)
0010 = Compare mode, toggle output on match (CCPxIF bit of the PIRx register is set)
0011 = Unused (reserved)
0100 = Capture mode, every falling edge
0101 = Capture mode, every rising edge
0110 = Capture mode, every 4th rising edge
0111 = Capture mode, every 16th rising edge
1000 = Compare mode, set output on match (CCPxIF bit of the PIRx register is set)
1001 = Compare mode, clear output on match (CCPxIF bit of the PIRx register is set)
1010 = Compare mode, generate software interrupt on match (CCPxIF bit is set of the PIRx register, CCPx pin is unaffected)
1011 = Compare mode, trigger special event (CCPxIF bit of the PIRx register is set, TMR1 is reset and A/D conversion(1) is started if the ADC module is enabled. CCPx pin is unaffected.)
11xx = PWM mode.

Note 1: A/D conversion start feature is available only on CCP2.

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17.1 Capture Mode

In Capture mode, CCPRxH:CCPRxL captures the 16-bit value of the TMR1 register when an event occurs on pin CCPx. An event is defined as one of the following and is configured by the CCPxM<3:0> bits of the CCPxCON register:

- Every falling edge
- Every rising edge
- Every 4th rising edge
- Every 16th rising edge

When a capture is made, the interrupt request flag bit CCPxIF of the PIRx register is set. The interrupt flag must be cleared in software. If another capture occurs before the value in the CCPRxH, CCPRxL register pair is read, the old captured value is overwritten by the new captured value (refer to Figure 17-1).

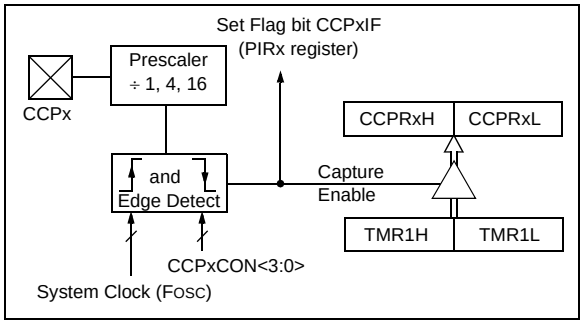
17.1.1 CCPx PIN CONFIGURATION

In Capture mode, the CCPx pin should be configured as an input by setting the associated TRIS control bit.

Either RC1 or RB3 can be selected as the CCP2 pin. Refer to **Section 6.1 "Alternate Pin Function"** for more information.

Note:	If the CCPx pin is configured as an output, a write to the port can cause a capture condition.

FIGURE 17-1: CAPTURE MODE OPERATION BLOCK DIAGRAM



17.1.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode or Synchronized Counter mode for the CCP module to use the capture feature. In Asynchronous Counter mode or when Timer1 is clocked at FOSC, the capture operation may not work.

17.1.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCPxIE interrupt enable bit of the PIEx register clear to avoid false interrupts. Additionally, the user should clear the CCPxIF interrupt flag bit of the PIRx register following any change in operating mode.

Note:	Clocking Timer1 from the system clock (FOSC) should not be used in Capture mode. In order for Capture mode to recognize the trigger event on the CCPx pin, Timer1 must be clocked from the instruction clock (FOSC/4) or from an external clock source.

17.1.4 CCP PRESCALER

There are four prescaler settings specified by the CCPxM<3:0> bits of the CCPxCON register. Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. Any Reset will clear the prescaler counter.

Switching from one capture prescaler to another does not clear the prescaler and may generate a false interrupt. To avoid this unexpected operation, turn the module off by clearing the CCPxCON register before changing the prescaler (refer to Example 17-1).

EXAMPLE 17-1: CHANGING BETWEEN CAPTURE PRESCALERS

```
BANKSEL CCP1CON     ;Set Bank bits to point
                    ;to CCP1CON
CLRF    CCP1CON     ;Turn CCP module off
MOVLW   NEW_CAPT_PS ;Load the W reg with
                    ; the new prescaler
                    ; move value and CCP ON
MOVWF   CCP1CON     ;Load CCP1CON with this
                    ; value
```

17.1.5 CAPTURE DURING SLEEP

Capture mode depends upon the Timer1 module for proper operation. There are two options for driving the Timer1 module in Capture mode. It can be driven by the instruction clock (FOSC/4), or by an external clock source.

If Timer1 is clocked by FOSC/4, then Timer1 will not increment during Sleep. When the device wakes from Sleep, Timer1 will continue from its previous state.

If Timer1 is clocked by an external clock source, then Capture mode will operate as defined in **Section 17.1 "Capture Mode"**.

TABLE 17-3: SUMMARY OF REGISTERS ASSOCIATED WITH CAPTURE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELB	ANSB7	ANSB6	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	1111 1111	1111 1111
ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
APFCON	—	—	—	—	—	—	SSSEL	CCP2SEL	---- --00	---- --00
CCP1CON	—	—	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	--00 0000	--00 0000
CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	--00 0000	--00 0000
CCPRxL	Capture/Compare/PWM Register X Low Byte								xxxx xxxx	uuuu uuuu
CCPRxH	Capture/Compare/PWM Register X High Byte								xxxx xxxx	uuuu uuuu
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIE2	TMR3GIE	TMR3IE	TMRBIE	TMRAIE	—	—	—	CCP2IE	0000 ---0	0000 ---0
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIR2	TMR3GIF	TMR3IF	TMRBIF	TMRAIF	—	—	—	CCP2IF	0000 ---0	0000 ---0
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{\text{T1SYNC}}$	—	TMR1ON	0000 00-0	uuuu uu-u
T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T$\overline{\text{1GGO}}$/ $\overline{\text{DONE}}$	T1GVAL	T1GSS1	T1GSS0	0000 0x00	0000 0x00
TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
TMR1H	Holding Register for the Most Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

Legend: - = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the capture.

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17.2 Compare Mode

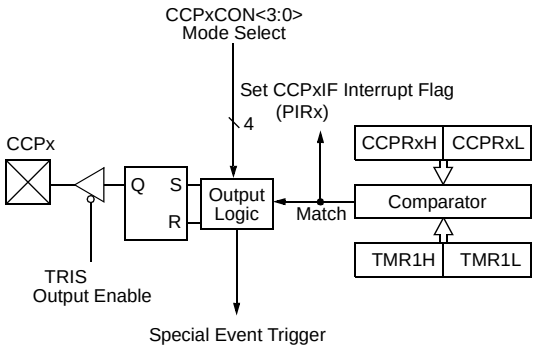
In Compare mode, the 16-bit CCPRx register value is constantly compared against the TMR1 register pair value. When a match occurs, the CCPx module may:

- Toggle the CCPx output
- Set the CCPx output
- Clear the CCPx output
- Generate a Special Event Trigger
- Generate a Software Interrupt

The action on the pin is based on the value of the CCPxM<3:0> control bits of the CCPxCON register.

All Compare modes can generate an interrupt.

FIGURE 17-2: COMPARE MODE OPERATION BLOCK DIAGRAM



Special Event Trigger will:

- Clear TMR1H and TMR1L registers.
- NOT set interrupt flag bit TMR1IF of the PIR1 register.
- Set the GO/$\overline{\text{DONE}}$ bit to start the ADC conversion (CCP2 only).

17.2.1 CCPx PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the associated TRIS bit.

Either RC1 or RB3 can be selected as the CCP2 pin. Refer to **Section 6.1 "Alternate Pin Function"** for more information.

Note:	Clearing the CCPxCON register will force the CCPx compare output latch to the default low level. This is not the PORT I/O data latch.

17.2.2 TIMER1 MODE SELECTION

In Compare mode, Timer1 must be running in either Timer mode or Synchronized Counter mode. The compare operation may not work in Asynchronous Counter mode.

Note:	Clocking Timer1 from the system clock (FOSC) should not be used in Compare mode. For the Compare operation of the TMR1 register to the CCPRx register to occur, Timer1 must be clocked from the instruction clock (FOSC/4) or from an external clock source.

17.2.3 SOFTWARE INTERRUPT MODE

When Software Interrupt mode is chosen (CCPxM<3:0> = 1010), the CCPxIF bit in the PIRx register is set and the CCPx module does not assert control of the CCPx pin (refer to the CCPxCON register).

17.2.4 SPECIAL EVENT TRIGGER

When Special Event Trigger mode is chosen (CCPxM<3:0> = 1011), the CCPx module does the following:

- Resets Timer1
- Starts an ADC conversion if ADC is enabled (CCP2 only)

The CCPx module does not assert control of the CCPx pin in this mode (refer to the CCPxCON register).

The Special Event Trigger output of the CCP occurs immediately upon a match between the TMR1H, TMR1L register pair and the CCPRxH, CCPRxL register pair. The TMR1H, TMR1L register pair is not reset until the next rising edge of the Timer1 clock. This allows the CCPRxH, CCPRxL register pair to effectively provide a 16-bit programmable period register for Timer1.

Note 1:	The Special Event Trigger from the CCP module does not set interrupt flag bit TMR1IF of the PIR1 register.
2:	Removing the match condition by changing the contents of the CCPRxH and CCPRxL register pair, between the clock edge that generates the Special Event Trigger and the clock edge that generates the Timer1 Reset, will preclude the Reset from occurring.

17.2.5 COMPARE DURING SLEEP

The Compare mode is dependent upon the system clock (FOSC) for proper operation. Since FOSC is shut down during Sleep mode, the Compare mode will not function properly during Sleep.

TABLE 17-4: SUMMARY OF REGISTERS ASSOCIATED WITH COMPARE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ADCON0	—	—	CHS3	CHS2	CHS1	CHS0	GO/$\overline{\text{DONE}}$	ADON	--00 0000	--00 0000
ANSELB	ANSB7	ANSB6	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	1111 1111	1111 1111
ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
APFCON	—	—	—	—	—	—	SSSEL	CCP2SEL	---- --00	---- --00
CCP1CON	—	—	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	--00 0000	--00 0000
CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	--00 0000	--00 0000
CCPRxL	Capture/Compare/PWM Register X Low Byte								xxxx xxxx	uuuu uuuu
CCPRxH	Capture/Compare/PWM Register X High Byte								xxxx xxxx	uuuu uuuu
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIE2	TMR3GIE	TMR3IE	TMRBIE	TMRAIE	—	—	—	CCP2IE	0000 ---0	0000 ---0
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIR2	TMR3GIF	TMR3IF	TMRBIF	TMRAIF	—	—	—	CCP2IF	0000 ---0	0000 ---0
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{\text{T1SYNC}}$	—	TMR1ON	0000 00-0	uuuu uu-u
T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/$\overline{\text{DONE}}$	T1GVAL	T1GSS1	T1GSS0	0000 0x00	0000 0x00
TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
TMR1H	Holding Register for the Most Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

Legend: - = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the compare.

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17.3 PWM Mode

The PWM mode generates a pulse-width modulated signal on the CCPx pin. The duty cycle, period and resolution are determined by the following registers:

- PR2
- T2CON
- CCPRxL
- CCPxCON

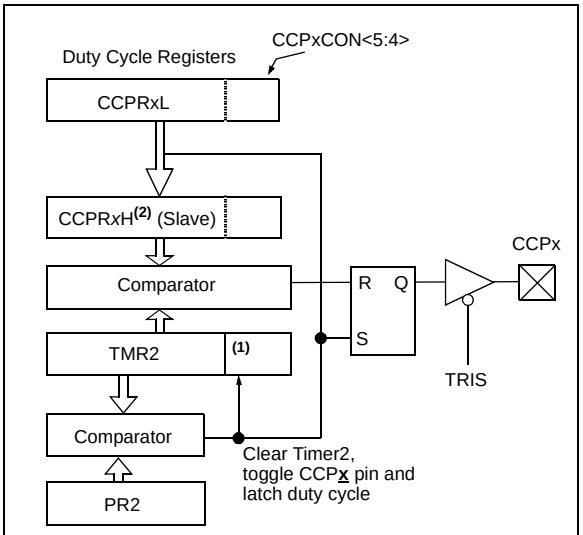
In Pulse-Width Modulation (PWM) mode, the CCP module produces up to a 10-bit resolution PWM output on the CCPx pin.

Figure 17-3 shows a simplified block diagram of PWM operation.

Figure 17-4 shows a typical waveform of the PWM signal.

For a step-by-step procedure on how to set up the CCP module for PWM operation, refer to **Section 17.3.8 "Setup for PWM Operation"**.

FIGURE 17-3: SIMPLIFIED PWM BLOCK DIAGRAM

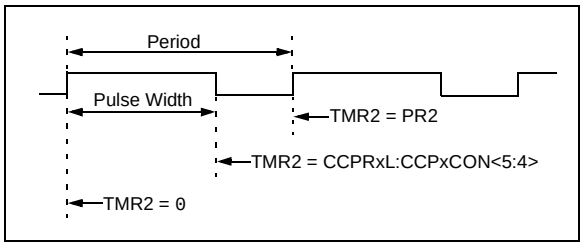


Note 1: The 8-bit timer TMR2 register is concatenated with the 2-bit internal system clock (FOSC), or 2 bits of the prescaler, to create the 10-bit time base.

2: In PWM mode, CCPRxH is a read-only register.

The PWM output (Figure 17-4) has a time base (period) and a time that the output stays high (duty cycle).

FIGURE 17-4: CCP PWM OUTPUT



17.3.1 CCPX PIN CONFIGURATION

In PWM mode, the CCPx pin is multiplexed with the PORT data latch. The user must configure the CCPx pin as an output by clearing the associated TRIS bit.

Either RC1 or RB3 can be selected as the CCP2 pin. Refer to **Section 6.1 "Alternate Pin Function"** for more information.

Note:	Clearing the CCPxCON register will relinquish CCPx control of the CCPx pin.

17.3.2 PWM PERIOD

The PWM period is specified by the PR2 register of Timer2. The PWM period can be calculated using the formula of Equation 17-1.

EQUATION 17-1: PWM PERIOD

$$PWM\ Period = [(PR2) + 1] \bullet 4 \bullet TOSC \bullet (TMR2\ Prescale\ Value)$$

Note: TOSC = 1/FOSC

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCPx pin is set. (Exception: If the PWM duty cycle = 0%, the pin will not be set.)
- The PWM duty cycle is latched from CCPRxL into CCPRxH.

Note: The Timer2 postscaler (refer to **Section 15.1 "Timer2 Operation"**) is not used in the determination of the PWM frequency.

17.3.3 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to multiple registers: CCPRxL register and DCxB<1:0> bits of the CCPxCON register. The CCPRxL contains the eight MSbs and the DCxB<1:0> bits of the CCPxCON register contain the two LSbs. CCPRxL and DCxB<1:0> bits of the CCPxCON register can be written to at any time. The duty cycle value is not latched into CCPRxH until after the period completes (i.e., a match between PR2 and TMR2 registers occurs). While using the PWM, the CCPRxH register is read-only.

Equation 17-2 is used to calculate the PWM pulse width.

Equation 17-3 is used to calculate the PWM duty cycle ratio.

EQUATION 17-2: PULSE WIDTH

$$Pulse\ Width = (CCPRxL:CCPxCON<5:4>) \bullet TOSC \bullet (TMR2\ Prescale\ Value)$$

Note: TOSC = 1/FOSC

EQUATION 17-3: DUTY CYCLE RATIO

$$Duty\ Cycle\ Ratio = \frac{(CCPRxL:CCPxCON<5:4>)}{4(PR2 + 1)}$$

The CCPRxH register and a 2-bit internal latch are used to double buffer the PWM duty cycle. This double buffering is essential for glitchless PWM operation.

The 8-bit timer TMR2 register is concatenated with either the 2-bit internal system clock (FOSC), or 2 bits of the prescaler, to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.

When the 10-bit time base matches the CCPRxH and 2-bit latch, then the CCPx pin is cleared (refer to Figure 17-3).

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17.3.4 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is 10 bits when PR2 is 255. The resolution is a function of the PR2 register value as shown by Equation 17-4.

EQUATION 17-4: PWM RESOLUTION

$$Resolution = \frac{log[4(PR2 + 1)]}{log(2)} \ bits$$

Note: If the pulse width value is greater than the period the assigned PWM pin(s) will remain unchanged.

TABLE 17-5: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (FOSC = 20 MHz)

PWM Frequency	1.22 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescale (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	6.6

TABLE 17-6: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (FOSC = 8 MHz)

PWM Frequency	1.22 kHz	4.90 kHz	19.61 kHz	76.92 kHz	153.85 kHz	200.0 kHz
Timer Prescale (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0x65	0x65	0x65	0x19	0x0C	0x09
Maximum Resolution (bits)	8	8	8	6	5	5

17.3.5 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the CCPx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

17.3.6 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency (FOSC). Any changes in the system clock frequency will result in changes to the PWM frequency. Refer to **Section 7.0 "Oscillator Module"** for additional details.

17.3.7 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

17.3.8 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

1. Disable the PWM pin (CCPx) output driver(s) by setting the associated TRIS bit(s).
2. Load the PR2 register with the PWM period value.
3. Configure the CCP module for the PWM mode by loading the CCPxCON register with the appropriate values.
4. Load the CCPRxL register and the DCxBx bits of the CCPxCON register, with the PWM duty cycle value.
5. Configure and start Timer2:
   - Clear the TMR2IF interrupt flag bit of the PIR1 register. See Note below.
   - Configure the T2CKPS bits of the T2CON register with the Timer2 prescale value.
   - Enable Timer2 by setting the TMR2ON bit of the T2CON register.

6. Enable PWM output pin:
   - Wait until Timer2 overflows, TMR2IF bit of the PIR1 register is set. See the Note below.
   - Enable the PWM pin (CCPx) output driver(s) by clearing the associated TRIS bit(s).

Note: In order to send a complete duty cycle and period on the first PWM output, the above steps must be included in the setup sequence. If it is not critical to start with a complete PWM signal on the first output, then step 6 may be ignored.

TABLE 17-7: SUMMARY OF REGISTERS ASSOCIATED WITH PWM

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELB	ANSB7	ANSB6	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	1111 1111	1111 1111
ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
APFCON	—	—	—	—	—	—	SSSEL	CCP2SEL	---- --00	---- --00
CCP1CON	—	—	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	--00 0000	--00 0000
CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	--00 0000	--00 0000
CCPRxL	Capture/Compare/PWM Register X Low Byte								xxxx xxxx	uuuu uuuu
CCPRxH	Capture/Compare/PWM Register X High Byte								xxxx xxxx	uuuu uuuu
PR2	Timer2 Period Register								1111 1111	1111 1111
T2CON	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000
TMR2	Timer2 Module Register								0000 0000	0000 0000
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

Legend: - = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the PWM.

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18.0 ADDRESSABLE UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (AUSART)

The Addressable Universal Synchronous Asynchronous Receiver Transmitter (AUSART) module is a serial I/O communications peripheral. It contains all the clock generators, shift registers and data buffers necessary to perform an input or output serial data transfer independent of device program execution. The AUSART, also known as a Serial Communications Interface (SCI), can be configured as a full-duplex asynchronous system or half-duplex synchronous system. Full-Duplex mode is useful for communications with peripheral systems, such as CRT terminals and personal computers. Half-Duplex Synchronous mode is intended for communications with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs or other microcontrollers. These devices typically do not have internal clocks for baud rate generation and require the external clock signal provided by a master synchronous device.

The AUSART module includes the following capabilities:

- Full-duplex asynchronous transmit and receive
- Two-character input buffer
- One-character output buffer
- Programmable 8-bit or 9-bit character length
- Address detection in 9-bit mode
- Input buffer overrun error detection
- Received character framing error detection
- Half-duplex synchronous master
- Half-duplex synchronous slave
- Sleep operation

Block diagrams of the AUSART transmitter and receiver are shown in Figure 18-1 and Figure 18-2.

FIGURE 18-1: AUSART TRANSMIT BLOCK DIAGRAM

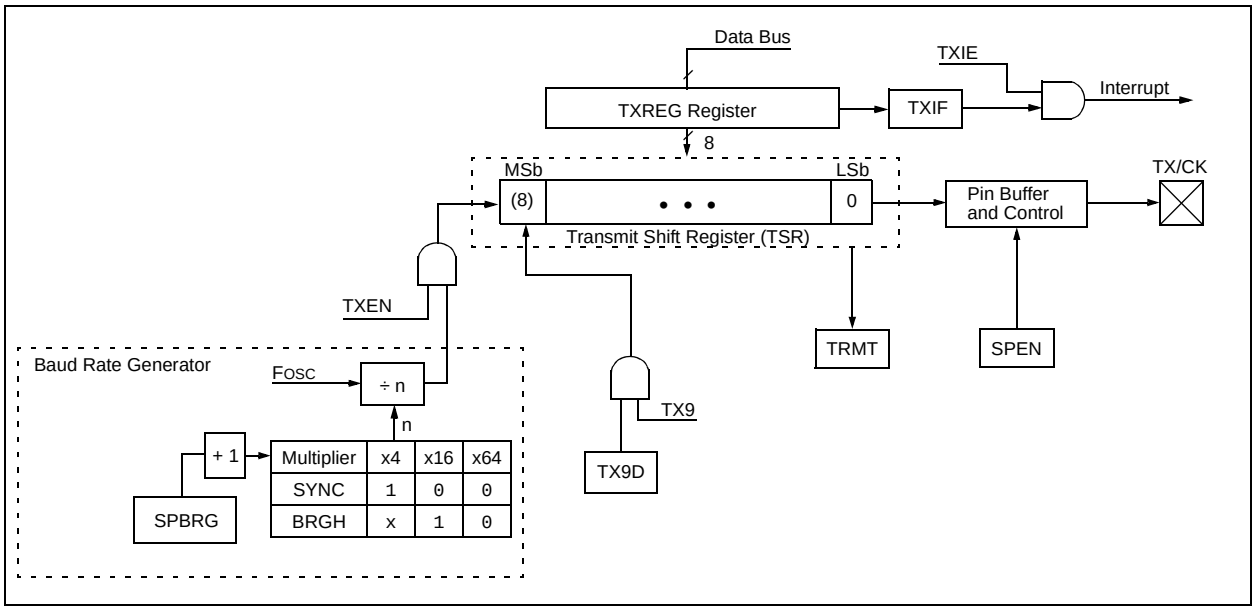
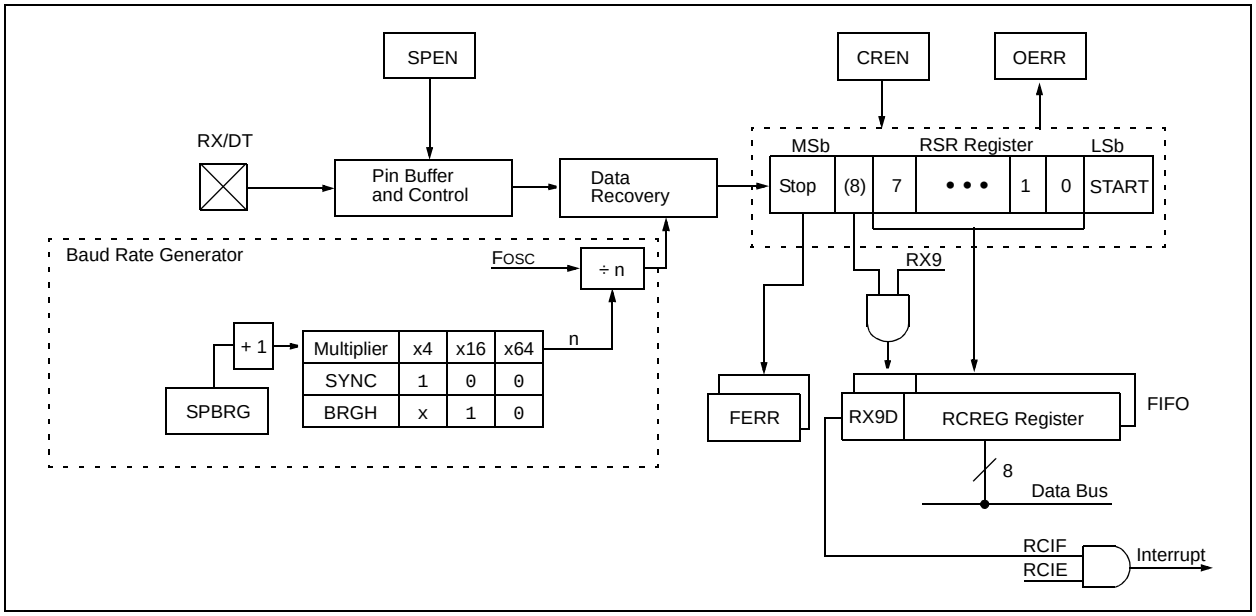


FIGURE 18-2: AUSART RECEIVE BLOCK DIAGRAM



The operation of the AUSART module is controlled through two registers:

- Transmit Status and Control (TXSTA)
- Receive Status and Control (RCSTA)

These registers are detailed in Register 18-1 and Register 18-2, respectively.

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18.1 AUSART Asynchronous Mode

The AUSART transmits and receives data using the standard non-return-to-zero (NRZ) format. NRZ is implemented with two levels: a VOH mark state which represents a '1' data bit, and a VOL space state which represents a '0' data bit. NRZ refers to the fact that consecutively transmitted data bits of the same value stay at the output level of that bit without returning to a neutral level between each bit transmission. An NRZ transmission port idles in the mark state. Each character transmission consists of one Start bit followed by eight or nine data bits and is always terminated by one or more Stop bits. The Start bit is always a space and the Stop bits are always marks. The most common data format is 8 bits. Each transmitted bit persists for a period of 1/(Baud Rate). An on-chip dedicated 8-bit Baud Rate Generator is used to derive standard baud rate frequencies from the system oscillator. Refer to Table 18-5 for examples of baud rate configurations.

The AUSART transmits and receives the LSb first. The AUSART's transmitter and receiver are functionally independent, but share the same data format and baud rate. Parity is not supported by the hardware, but can be implemented in software and stored as the ninth data bit.

18.1.1 AUSART ASYNCHRONOUS TRANSMITTER

The AUSART transmitter block diagram is shown in Figure 18-1. The heart of the transmitter is the serial Transmit Shift Register (TSR), which is not directly accessible by software. The TSR obtains its data from the transmit buffer, which is the TXREG register.

18.1.1.1 Enabling the Transmitter

The AUSART transmitter is enabled for asynchronous operations by configuring the following three control bits:

- TXEN = 1
- SYNC = 0
- SPEN = 1

All other AUSART control bits are assumed to be in their default state.

Setting the TXEN bit of the TXSTA register enables the transmitter circuitry of the AUSART. Clearing the SYNC bit of the TXSTA register configures the AUSART for asynchronous operation. Setting the SPEN bit of the RCSTA register enables the AUSART and automatically configures the TX/CK I/O pin as an output.

Note 1: When the SPEN bit is set, the RX/DT I/O pin is automatically configured as an input, regardless of the state of the corresponding TRIS bit and whether or not the AUSART receiver is enabled. The RX/DT pin data can be read via a normal PORT read but PORT latch data output is precluded.

2: The corresponding ANSEL bit must be cleared for the RX/DT port pin to ensure proper AUSART functionality.

3: The TXIF transmitter interrupt flag is set when the TXEN enable bit is set.

18.1.1.2 Transmitting Data

A transmission is initiated by writing a character to the TXREG register. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXREG is immediately transferred to the TSR register. If the TSR still contains all or part of a previous character, the new character data is held in the TXREG until the Stop bit of the previous character has been transmitted. The pending character in the TXREG is then transferred to the TSR in one TCY immediately following the Stop bit transmission. The transmission of the Start bit, data bits and Stop bit sequence commences immediately following the transfer of the data to the TSR from the TXREG.

18.1.1.3 Transmit Interrupt Flag

The TXIF interrupt flag bit of the PIR1 register is set whenever the AUSART transmitter is enabled and no character is being held for transmission in the TXREG. In other words, the TXIF bit is only clear when the TSR is busy with a character and a new character has been queued for transmission in the TXREG. The TXIF flag bit is not cleared immediately upon writing TXREG. TXIF becomes valid in the second instruction cycle following the write execution. Polling TXIF immediately following the TXREG write will return invalid results. The TXIF bit is read-only, it cannot be set or cleared by software.

The TXIF interrupt can be enabled by setting the TXIE interrupt enable bit of the PIE1 register. However, the TXIF flag bit will be set whenever the TXREG is empty, regardless of the state of TXIE enable bit.

To use interrupts when transmitting data, set the TXIE bit only when there is more data to send. Clear the TXIE interrupt enable bit upon writing the last character of the transmission to the TXREG.

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18.1.1.4 TSR Status

The TRMT bit of the TXSTA register indicates the status of the TSR register. This is a read-only bit. The TRMT bit is set when the TSR register is empty and is cleared when a character is transferred to the TSR register from the TXREG. The TRMT bit remains clear until all bits have been shifted out of the TSR register. No interrupt logic is tied to this bit, so the user has to poll this bit to determine the TSR status.

Note:	The TSR register is not mapped in data memory, so it is not available to the user.

18.1.1.5 Transmitting 9-Bit Characters

The AUSART supports 9-bit character transmissions. When the TX9 bit of the TXSTA register is set the AUSART will shift 9 bits out for each character transmitted. The TX9D bit of the TXSTA register is the ninth, and Most Significant, data bit. When transmitting 9-bit data, the TX9D data bit must be written before writing the 8 Least Significant bits into the TXREG. All nine bits of data will be transferred to the TSR shift register immediately after the TXREG is written.

A special 9-bit Address mode is available for use with multiple receivers. Refer to **Section 18.1.2.7 "Address Detection"** for more information on the Address mode.

18.1.1.6 Asynchronous Transmission Set-up:

1. Initialize the SPBRG register and the BRGH bit to achieve the desired baud rate (Refer to **Section 18.2 "AUSART Baud Rate Generator (BRG)"**).
2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
3. If 9-bit transmission is desired, set the TX9 control bit. A set ninth data bit will indicate that the 8 Least Significant data bits are an address when the receiver is set for address detection.
4. Enable the transmission by setting the TXEN control bit. This will cause the TXIF interrupt bit to be set.
5. If interrupts are desired, set the TXIE interrupt enable bit of the PIE1 register. An interrupt will occur immediately provided that the GIE and PEIE bits of the INTCON register are also set.
6. If 9-bit transmission is selected, the ninth bit should be loaded into the TX9D data bit.
7. Load 8-bit data into the TXREG register. This will start the transmission.

FIGURE 18-3: ASYNCHRONOUS TRANSMISSION

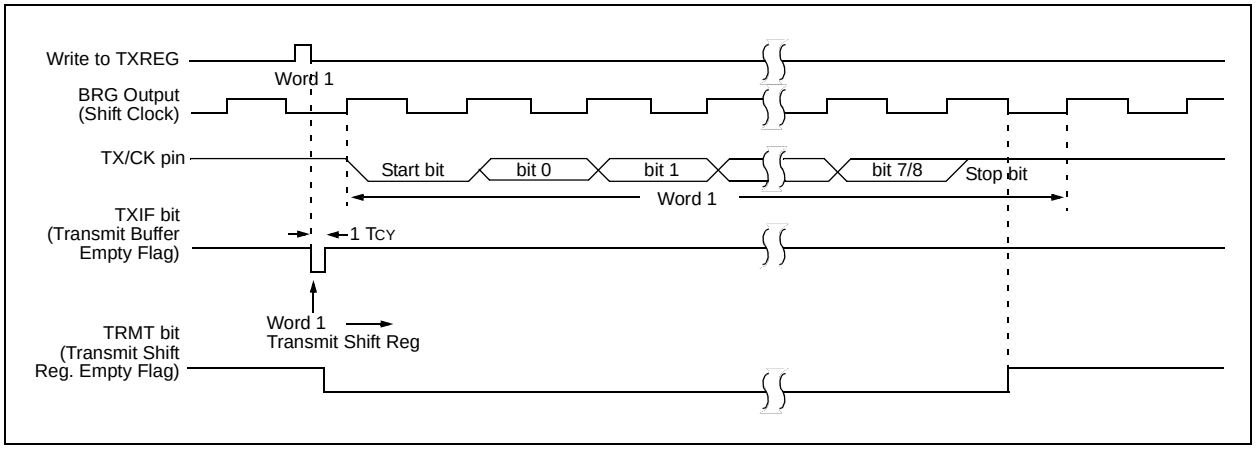

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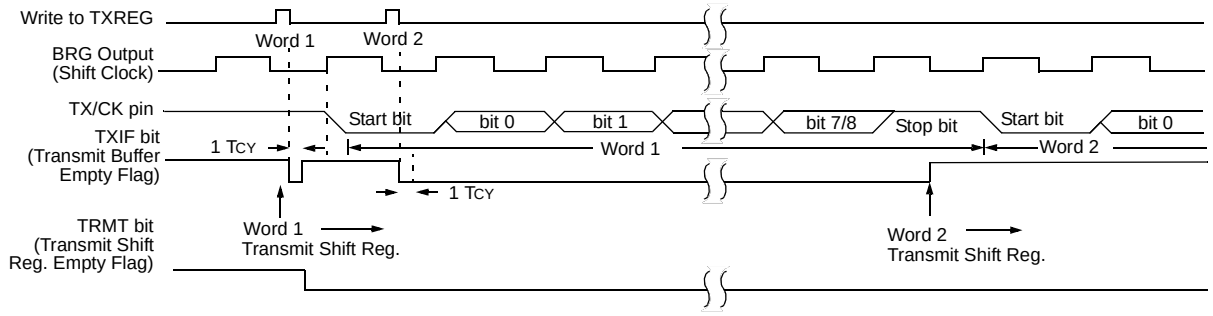
FIGURE 18-4: ASYNCHRONOUS TRANSMISSION (BACK-TO-BACK)



Note: This timing diagram shows two consecutive transmissions.

TABLE 18-1: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000x
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXREG	AUSART Transmit Data Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for asynchronous transmission.

18.1.2 AUSART ASYNCHRONOUS RECEIVER

The Asynchronous mode is typically used in RS-232 systems. The receiver block diagram is shown in Figure 18-2. The data is received on the RX/DT pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at 16 times the baud rate, whereas the serial Receive Shift Register (RSR) operates at the bit rate. When all 8 or 9 bits of the character have been shifted in, they are immediately transferred to a two character First-In First-Out (FIFO) memory. The FIFO buffering allows reception of two complete characters and the start of a third character before software must start servicing the AUSART receiver. The FIFO and RSR registers are not directly accessible by software. Access to the received data is via the RCREG register.

18.1.2.1 Enabling the Receiver

The AUSART receiver is enabled for asynchronous operation by configuring the following three control bits:

- CREN = 1
- SYNC = 0
- SPEN = 1

All other AUSART control bits are assumed to be in their default state.

Setting the CREN bit of the RCSTA register enables the receiver circuitry of the AUSART. Clearing the SYNC bit of the TXSTA register configures the AUSART for asynchronous operation. Setting the SPEN bit of the RCSTA register enables the AUSART and automatically configures the RX/DT I/O pin as an input.

Note 1: When the SPEN bit is set, the TX/CK I/O pin is automatically configured as an output, regardless of the state of the corresponding TRIS bit and whether or not the AUSART transmitter is enabled. The PORT latch is disconnected from the output driver so it is not possible to use the TX/CK pin as a general purpose output.

2: The corresponding ANSEL bit must be cleared for the RX/DT port pin to ensure proper AUSART functionality.

18.1.2.2 Receiving Data

The receiver data recovery circuit initiates character reception on the falling edge of the first bit. The first bit, also known as the Start bit, is always a zero. The data recovery circuit counts one-half bit time to the center of the Start bit and verifies that the bit is still a zero. If it is not a zero then the data recovery circuit aborts character reception, without generating an error, and resumes looking for the falling edge of the Start bit. If the Start bit zero verification succeeds then the data recovery circuit counts a full bit time to the center of the next bit. The bit is then sampled by a majority detect circuit and the resulting '0' or '1' is shifted into the RSR. This repeats until all data bits have been sampled and shifted into the RSR. One final bit time is measured and the level sampled. This is the Stop bit, which is always a '1'. If the data recovery circuit samples a '0' in the Stop bit position then a framing error is set for this character, otherwise the framing error is cleared for this character. Refer to **Section 18.1.2.4 "Receive Framing Error"** for more information on framing errors.

Immediately after all data bits and the Stop bit have been received, the character in the RSR is transferred to the AUSART receive FIFO and the RCIF interrupt flag bit of the PIR1 register is set. The top character in the FIFO is transferred out of the FIFO by reading the RCREG register.

Note: If the receive FIFO is overrun, no additional characters will be received until the overrun condition is cleared. Refer to ***Section 18.1.2.5 "Receive Overrun Error"*** for more information on overrun errors.

18.1.2.3 Receive Interrupts

The RCIF interrupt flag bit of the PIR1 register is set whenever the AUSART receiver is enabled and there is an unread character in the receive FIFO. The RCIF interrupt flag bit is read-only, it cannot be set or cleared by software.

RCIF interrupts are enabled by setting all of the following bits:

- RCIE, Receive Interrupt Enable bit of the PIE1 register
- PEIE, Peripheral Interrupt Enable bit of the INTCON register
- GIE, Global Interrupt Enable bit of the INTCON register

The RCIF interrupt flag bit of the PIR1 register will be set when there is an unread character in the FIFO, regardless of the state of interrupt enable bits.

18.1.2.4 Receive Framing Error

Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RCSTA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RCREG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error (FERR = 1) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RCSTA register which resets the AUSART. Clearing the CREN bit of the RCSTA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

Note: If all receive characters in the receive FIFO have framing errors, repeated reads of the RCREG will not clear the FERR bit.

18.1.2.5 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RCSTA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RCSTA register or by setting the AUSART by clearing the SPEN bit of the RCSTA register.

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18.1.2.6 Receiving 9-bit Characters

The AUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set the AUSART will shift 9 bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the 8 Least Significant bits from the RCREG.

18.1.2.7 Address Detection

A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RCSTA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit of the PIR1 register. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.

18.1.2.8 Asynchronous Reception Set-up:

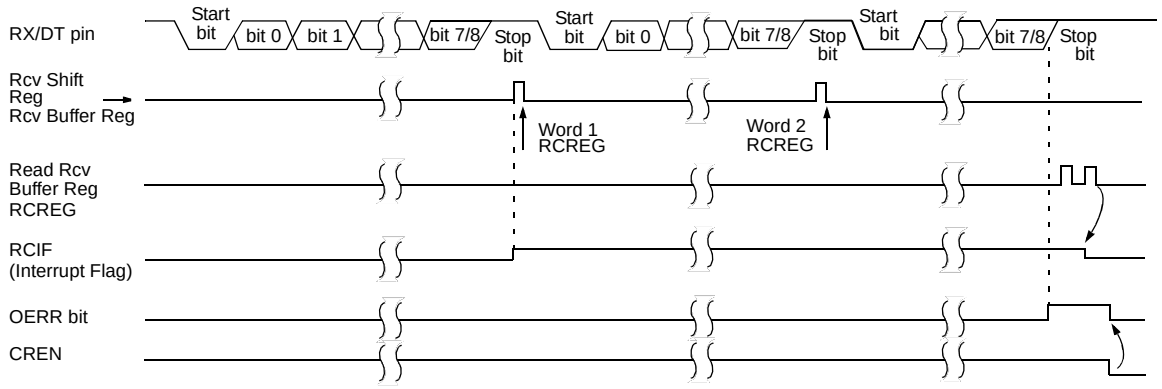
1. Initialize the SPBRG register and the BRGH bit to achieve the desired baud rate (refer to **Section 18.2 "AUSART Baud Rate Generator (BRG)"**).
2. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
3. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
4. If 9-bit reception is desired, set the RX9 bit.
5. Enable reception by setting the CREN bit.
6. The RCIF interrupt flag bit of the PIR1 register will be set when a character is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE bit of the PIE1 register was also set.
7. Read the RCSTA register to get the error flags and, if 9-bit data reception is enabled, the ninth data bit.
8. Get the received 8 Least Significant data bits from the receive buffer by reading the RCREG register.
9. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.

18.1.2.9 9-bit Address Detection Mode Set-up

This mode would typically be used in RS-485 systems. To set up an asynchronous reception with address detect enable:

1. Initialize the SPBRG register and the BRGH bit to achieve the desired baud rate (refer to **Section 18.2 "AUSART Baud Rate Generator (BRG)"**).
2. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
3. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
4. Enable 9-bit reception by setting the RX9 bit.
5. Enable address detection by setting the ADDEN bit.
6. Enable reception by setting the CREN bit.
7. The RCIF interrupt flag bit of the PIR1 register will be set when a character with the ninth bit set is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit of the PIE1 register was also set.
8. Read the RCSTA register to get the error flags. The ninth data bit will always be set.
9. Get the received 8 Least Significant data bits from the receive buffer by reading the RCREG register. Software determines if this is the device's address.
10. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.
11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and generate interrupts.

FIGURE 18-5: ASYNCHRONOUS RECEPTION



Note: This timing diagram shows three words appearing on the RX input. The RCREG (receive buffer) is read after the third word, causing the OERR (overrun) bit to be set.

TABLE 18-2: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
RCREG	AUSART Receive Data Register								0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000x
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for asynchronous reception.

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REGISTER 18-1: TXSTA: TRANSMIT STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R-1	R/W-0
CSRC	TX9	TXEN(1)	SYNC	—	BRGH	TRMT	TX9D
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7 **CSRC:** Clock Source Select bit
Asynchronous mode:
Don’t care
Synchronous mode:
1 = Master mode (clock generated internally from BRG)
0 = Slave mode (clock from external source)

bit 6 **TX9:** 9-bit Transmit Enable bit
1 = Selects 9-bit transmission
0 = Selects 8-bit transmission

bit 5 **TXEN:** Transmit Enable bit(1)
1 = Transmit enabled
0 = Transmit disabled

bit 4 **SYNC:** AUSART Mode Select bit
1 = Synchronous mode
0 = Asynchronous mode

bit 3 **Unimplemented:** Read as ‘0’

bit 2 **BRGH:** High Baud Rate Select bit
Asynchronous mode:
1 = High speed
0 = Low speed
Synchronous mode:
Unused in this mode

bit 1 **TRMT:** Transmit Shift Register Status bit
1 = TSR empty
0 = TSR full

bit 0 **TX9D:** Ninth bit of Transmit Data
Can be address/data bit or a parity bit.

Note 1: SREN/CREN overrides TXEN in Synchronous mode.

REGISTER 18-2: RCSTA: RECEIVE STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7 **SPEN:** Serial Port Enable bit(1)
1 = Serial port enabled (configures RX/DT and TX/CK pins as serial port pins)
0 = Serial port disabled (held in Reset)

bit 6 **RX9:** 9-bit Receive Enable bit
1 = Selects 9-bit reception
0 = Selects 8-bit reception

bit 5 **SREN:** Single Receive Enable bit
Asynchronous mode:
Don’t care
Synchronous mode – Master:
1 = Enables single receive
0 = Disables single receive
This bit is cleared after reception is complete.
Synchronous mode – Slave:
Don’t care

bit 4 **CREN:** Continuous Receive Enable bit
Asynchronous mode:
1 = Enables receiver
0 = Disables receiver
Synchronous mode:
1 = Enables continuous receive until enable bit CREN is cleared (CREN overrides SREN)
0 = Disables continuous receive

bit 3 **ADDEN:** Address Detect Enable bit
Asynchronous mode 9-bit (RX9 = 1):
1 = Enables address detection, enable interrupt and load the receive buffer when RSR<8> is set
0 = Disables address detection, all bytes are received and ninth bit can be used as parity bit
Asynchronous mode 8-bit (RX9 = 0):
Don’t care
Synchronous mode:
Must be set to ‘0’

bit 2 **FERR:** Framing Error bit
1 = Framing error (can be updated by reading RCREG register and receive next valid byte)
0 = No framing error

bit 1 **OERR:** Overrun Error bit
1 = Overrun error (can be cleared by clearing bit CREN)
0 = No overrun error

bit 0 **RX9D:** Ninth bit of Received Data
This can be address/data bit or a parity bit and must be calculated by user firmware.

Note 1: The AUSART module automatically changes the pin from tri-state to drive as needed. Configure TRISx = 1.

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18.2 AUSART Baud Rate Generator (BRG)

The Baud Rate Generator (BRG) is an 8-bit timer that is dedicated to the support of both the asynchronous and synchronous AUSART operation.

The SPBRG register determines the period of the free running baud rate timer. In Asynchronous mode the multiplier of the baud rate period is determined by the BRGH bit of the TXSTA register. In Synchronous mode, the BRGH bit is ignored.

Table 18-3 contains the formulas for determining the baud rate. Example 18-1 provides a sample calculation for determining the baud rate and baud rate error.

Typical baud rates and error values for various Asynchronous modes have been computed for your convenience and are shown in Table 18-5. It may be advantageous to use the high baud rate (BRGH = 1), to reduce the baud rate error.

Writing a new value to the SPBRG register causes the BRG timer to be reset (or cleared). This ensures that the BRG does not wait for a timer overflow before outputting the new baud rate.

EXAMPLE 18-1: CALCULATING BAUD RATE ERROR

For a device with FOSC of 16 MHz, desired baud rate of 9600, and Asynchronous mode with SYNC = 0 and BRGH = 0 (as seen in Table 18-5):

$$Desired\ Baud\ Rate = \frac{FOSC}{64(SPBRG + 1)}$$

Solving for SPBRG:

$$SPBRG = \left(\frac{FOSC}{64(Desired\ Baud\ Rate)}\right) - 1$$

$$= \left(\frac{16000000}{64(9600)}\right) - 1$$

$$= [25.042] = 25$$

$$Actual\ Baud\ Rate = \frac{16000000}{64(25 + 1)}$$

$$= 9615$$

$$\%\ Error = \left(\frac{Actual\ Baud\ Rate - Desired\ Baud\ Rate}{Desired\ Baud\ Rate}\right) 100$$

$$= \left(\frac{9615 - 9600}{9600}\right) 100 = 0.16\%$$

TABLE 18-3: BAUD RATE FORMULAS

Configuration Bits		AUSART Mode	Baud Rate Formula
SYNC	BRGH		
0	0	Asynchronous	FOSC/[64 (n+1)]
0	1	Asynchronous	FOSC/[16 (n+1)]
1	x	Synchronous	FOSC/[4 (n+1)]

Legend: x = Don't care, n = value of SPBRG register

TABLE 18-4: REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000x
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for the Baud Rate Generator.

TABLE 18-5: BAUD RATES FOR ASYNCHRONOUS MODES

BAUD RATE	SYNC = 0, BRGH = 0											
	Fosc = 20.000 MHz			Fosc = 18.432 MHz			Fosc = 16.0000 MHz			Fosc = 11.0592 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	—	—	—	—	—	—	—	—	—	—	—	—
1200	1221	1.73	255	1200	0.00	239	1201	0.08	207	1200	0.00	143
2400	2404	0.16	129	2400	0.00	119	2403	0.16	103	2400	0.00	71
9600	9470	-1.36	32	9600	0.00	29	9615	0.16	25	9600	0.00	17
10417	10417	0.00	29	10286	-1.26	27	10416	-0.01	23	10165	-2.42	16
19.2k	19.53k	1.73	15	19.20k	0.00	14	19.23k	0.16	12	19.20k	0.00	8
57.6k	—	—	—	57.60k	0.00	7	—	—	—	57.60k	0.00	2
115.2k	—	—	—	—	—	—	—	—	—	—	—	—

BAUD RATE	SYNC = 0, BRGH = 0											
	Fosc = 8.000 MHz			Fosc = 4.000 MHz			Fosc = 3.6864 MHz			Fosc = 1.000 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	—	—	—	300	0.16	207	300	0.00	191	300	0.16	51
1200	1202	0.16	103	1202	0.16	51	1200	0.00	47	1202	0.16	12
2400	2404	0.16	51	2404	0.16	25	2400	0.00	23	—	—	—
9600	9615	0.16	12	—	—	—	9600	0.00	5	—	—	—
10417	10417	0.00	11	10417	0.00	5	—	—	—	—	—	—
19.2k	—	—	—	—	—	—	19.20k	0.00	2	—	—	—
57.6k	—	—	—	—	—	—	57.60k	0.00	0	—	—	—
115.2k	—	—	—	—	—	—	—	—	—	—	—	—

BAUD RATE	SYNC = 0, BRGH = 1											
	Fosc = 20.000 MHz			Fosc = 18.432 MHz			Fosc = 16.0000 MHz			Fosc = 11.0592 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	—	—	—	—	—	—	—	—	—	—	—	—
1200	—	—	—	—	—	—	—	—	—	—	—	—
2400	—	—	—	—	—	—	—	—	—	—	—	—
9600	9615	0.16	129	9600	0.00	119	9615	0.16	103	9600	0.00	71
10417	10417	0.00	119	10378	-0.37	110	10417	0.00	95	10473	0.53	65
19.2k	19.23k	0.16	64	19.20k	0.00	59	19.23k	0.16	51	19.20k	0.00	35
57.6k	56.82k	-1.36	21	57.60k	0.00	19	58.8k	2.12	16	57.60k	0.00	11
115.2k	113.64k	-1.36	10	115.2k	0.00	9	—	—	—	115.2k	0.00	5

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TABLE 18-5: BAUD RATES FOR ASYNCHRONOUS MODES

BAUD RATE	SYNC = 0, BRGH = 1											
	FOSC = 8.000 MHz			FOSC = 4.000 MHz			FOSC = 3.6864 MHz			FOSC = 1.000 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	—	—	—	—	—	—	—	—	—	300	0.16	207
1200	—	—	—	1202	0.16	207	1200	0.00	191	1202	0.16	51
2400	2404	0.16	207	2404	0.16	103	2400	0.00	95	2404	0.16	25
9600	9615	0.16	51	9615	0.16	25	9600	0.00	23	—	—	—
10417	10417	0.00	47	10417	0.00	23	10473	0.53	21	10417	0.00	5
19.2k	19231	0.16	25	19.23k	0.16	12	19.2k	0.00	11	—	—	—
57.6k	55556	-3.55	8	—	—	—	57.60k	0.00	3	—	—	—
115.2k	—	—	—	—	—	—	115.2k	0.00	1	—	—	—

18.3 AUSART Synchronous Mode

Synchronous serial communications are typically used in systems with a single master and one or more slaves. The master device contains the necessary circuitry for baud rate generation and supplies the clock for all devices in the system. Slave devices can take advantage of the master clock by eliminating the internal clock generation circuitry.

There are two signal lines in Synchronous mode: a bidirectional data line and a clock line. Slaves use the external clock supplied by the master to shift the serial data into and out of their respective receive and transmit shift registers. Since the data line is bidirectional, synchronous operation is half-duplex only. Half-duplex refers to the fact that master and slave devices can receive and transmit data but not both simultaneously. The AUSART can operate as either a master or slave device.

Start and Stop bits are not used in synchronous transmissions.

18.3.1 SYNCHRONOUS MASTER MODE

The following bits are used to configure the AUSART for Synchronous Master operation:

- SYNC = 1
- CSRC = 1
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXSTA register configures the device for synchronous operation. Setting the CSRC bit of the TXSTA register configures the device as a master. Clearing the SREN and CREN bits of the RCSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCSTA register enables the AUSART.

18.3.1.1 Master Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a master transmits the clock on the TX/CK line. The TX/CK pin output driver is automatically enabled when the AUSART is configured for synchronous transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One clock cycle is generated for each data bit. Only as many clock cycles are generated as there are data bits.

18.3.1.2 Synchronous Master Transmission

Data is transferred out of the device on the RX/DT pin. The RX/DT and TX/CK pin output drivers are automatically enabled when the AUSART is configured for synchronous master transmit operation.

A transmission is initiated by writing a character to the TXREG register. If the TSR still contains all or part of a previous character, the new character data is held in the TXREG until the last bit of the previous character has been transmitted. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXREG is immediately transferred to the TSR. The transmission of the character commences immediately following the transfer of the data to the TSR from the TXREG.

Each data bit changes on the leading edge of the master clock and remains valid until the subsequent leading clock edge.

Note:	The TSR register is not mapped in data memory, so it is not available to the user.

18.3.1.3 Synchronous Master Transmission Set-up:

1. Initialize the SPBRG register and the BRGH bit to achieve the desired baud rate (refer to **Section 18.2 “AUSART Baud Rate Generator (BRG)”**).
2. Enable the synchronous master serial port by setting bits SYNC, SPEN, and CSRC.
3. Disable Receive mode by clearing bits SREN and CREN.
4. Enable Transmit mode by setting the TXEN bit.
5. If 9-bit transmission is desired, set the TX9 bit.
6. If interrupts are desired, set the TXIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
7. If 9-bit transmission is selected, the ninth bit should be loaded in the TX9D bit.
8. Start transmission by loading data to the TXREG register.

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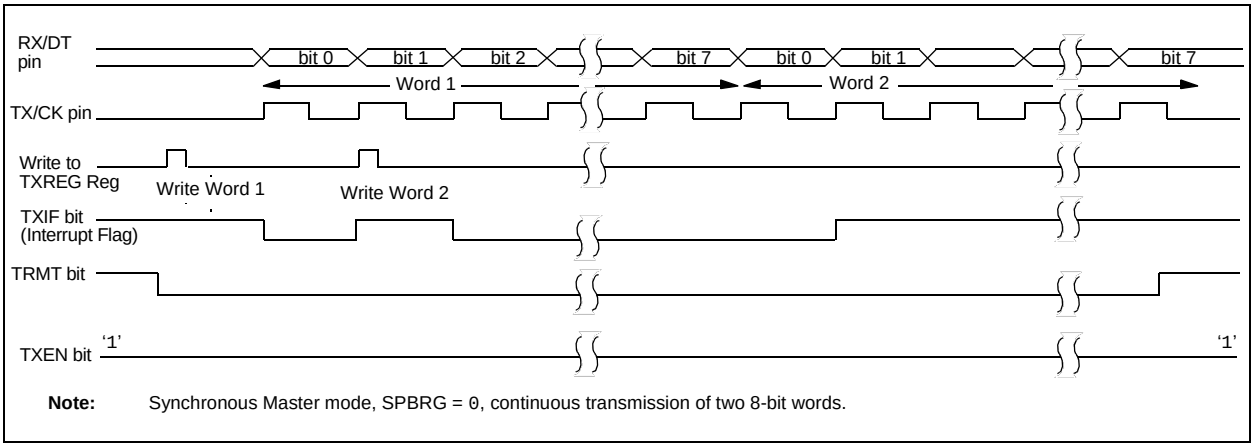
FIGURE 18-6: SYNCHRONOUS TRANSMISSION



Note: Synchronous Master mode, SPBRG = 0, continuous transmission of two 8-bit words.

FIGURE 18-7: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)

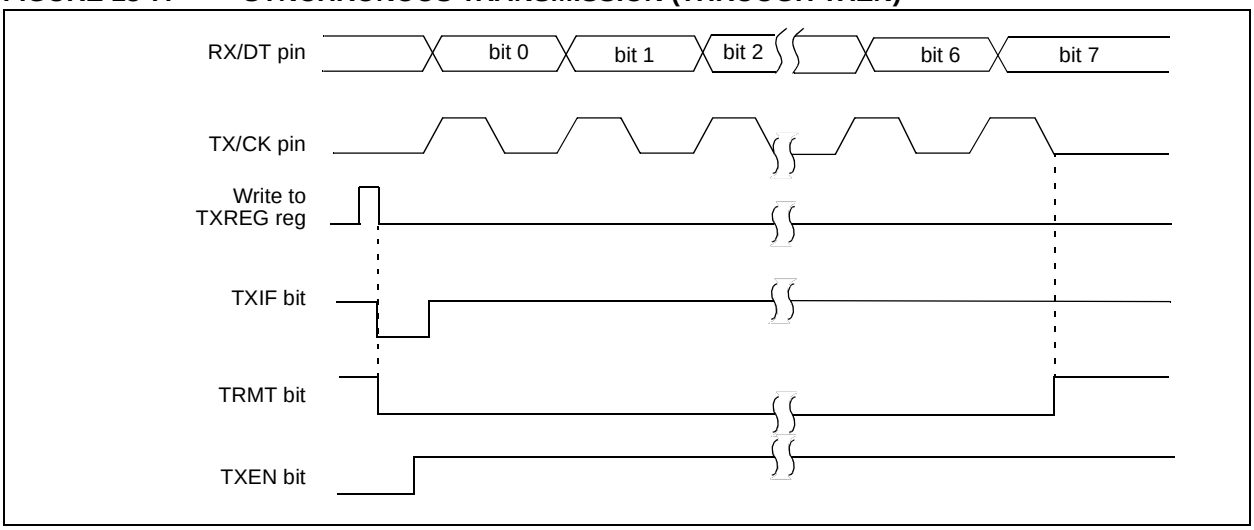


TABLE 18-6: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000x
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXREG	AUSART Transmit Data Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for synchronous master transmission.

18.3.1.4 Synchronous Master Reception

Data is received at the RX/DT pin. The RX/DT pin output driver is automatically disabled when the AUSART is configured for synchronous master receive operation.

In Synchronous mode, reception is enabled by setting either the Single Receive Enable bit (SREN of the RCSTA register) or the Continuous Receive Enable bit (CREN of the RCSTA register).

When SREN is set and CREN is clear, only as many clock cycles are generated as there are data bits in a single character. The SREN bit is automatically cleared at the completion of one character. When CREN is set, clocks are continuously generated until CREN is cleared. If CREN is cleared in the middle of a character the CK clock stops immediately and the partial character is discarded. If SREN and CREN are both set, then SREN is cleared at the completion of the first character and CREN takes precedence.

To initiate reception, set either SREN or CREN. Data is sampled at the RX/DT pin on the trailing edge of the TX/CK clock pin and is shifted into the Receive Shift Register (RSR). When a complete character is received into the RSR, the RCIF bit of the PIR1 register is set and the character is automatically transferred to the two character receive FIFO. The Least Significant eight bits of the top character in the receive FIFO are available in RCREG. The RCIF bit remains set as long as there are un-read characters in the receive FIFO.

18.3.1.5 Slave Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a slave receives the clock on the TX/CK line. The TX/CK pin output driver is automatically disabled when the device is configured for synchronous slave transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One data bit is transferred for each clock cycle. Only as many clock cycles should be received as there are data bits.

18.3.1.6 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before RCREG is read to access the FIFO. When this happens the OERR bit of the RCSTA register is set. Previous data in the FIFO will not be overwritten. The two characters in the FIFO buffer can be read, however, no additional characters will be received until the error is cleared. The OERR bit can only be cleared by clearing the overrun condition. If the overrun error occurred when the SREN bit is set and CREN is clear then the error is cleared by reading RCREG. If the overrun occurred when the CREN bit is set then the error condition is cleared by either clearing the CREN bit of the RCSTA register.

18.3.1.7 Receiving 9-bit Characters

The AUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set, the AUSART will shift 9-bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth, and Most Significant, data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the 8 Least Significant bits from the RCREG.

Address detection in Synchronous modes is not supported, therefore the ADDEN bit of the RCSTA register must be cleared.

18.3.1.8 Synchronous Master Reception Set-up:

1. Initialize the SPBRG register for the appropriate baud rate. Set or clear the BRGH bit, as required, to achieve the desired baud rate.
2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
3. Ensure bits CREN and SREN are clear.
4. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
5. If 9-bit reception is desired, set bit RX9.
6. Verify address detection is disabled by clearing the ADDEN bit of the RCSTA register.
7. Start reception by setting the SREN bit or for continuous reception, set the CREN bit.
8. Interrupt flag bit RCIF of the PIR1 register will be set when reception of a character is complete. An interrupt will be generated if the RCIE interrupt enable bit of the PIE1 register was set.
9. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
10. Read the 8-bit received data by reading the RCREG register.
11. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit, which resets the AUSART.

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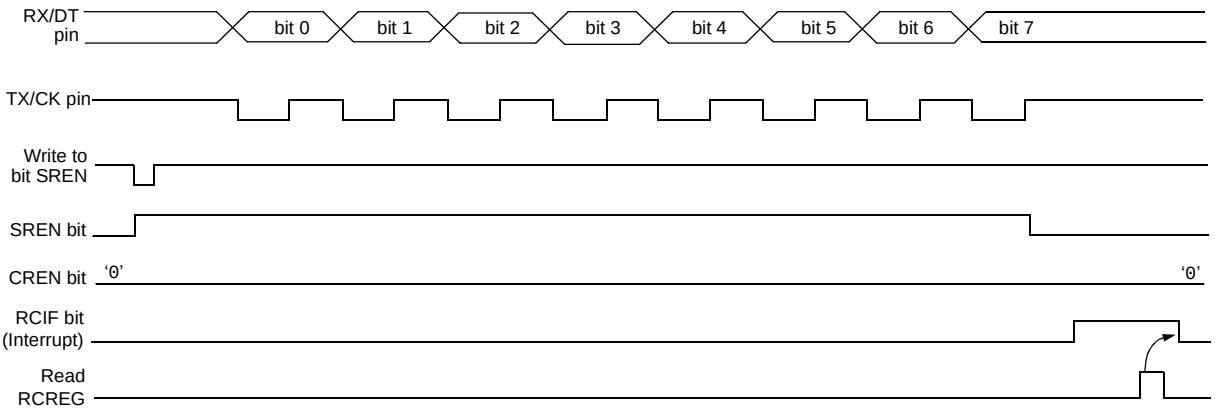
FIGURE 18-8: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)



Note: Timing diagram demonstrates Synchronous Master mode with bit SREN = 1 and bit BRGH = 0.

TABLE 18-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
RCREG	AUSART Receive Data Register								0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000X	0000 000X
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for synchronous master reception.

18.3.2 SYNCHRONOUS SLAVE MODE

The following bits are used to configure the AUSART for synchronous slave operation:

- SYNC = 1
- CSRC = 0
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXSTA register configures the device for synchronous operation. Clearing the CSRC bit of the TXSTA register configures the device as a slave. Clearing the SREN and CREN bits of the RCSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCSTA register enables the AUSART.

18.3.2.1 AUSART Synchronous Slave Transmit

The operation of the Synchronous Master and Slave modes are identical (refer to **Section 18.3.1.2 "Synchronous Master Transmission"**), except in the case of the Sleep mode.

If two words are written to the TXREG and then the SLEEP instruction is executed, the following will occur:

1. The first character will immediately transfer to the TSR register and transmit.
2. The second word will remain in TXREG register.
3. The TXIF bit will not be set.
4. After the first character has been shifted out of TSR, the TXREG register will transfer the second character to the TSR and the TXIF bit will now be set.
5. If the PEIE and TXIE bits are set, the interrupt will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will call the Interrupt Service Routine.

18.3.2.2 Synchronous Slave Transmission Set-up:

1. Set the SYNC and SPEN bits and clear the CSRC bit.
2. Clear the CREN and SREN bits.
3. If using interrupts, ensure that the GIE and PEIE bits of the INTCON register are set and set the TXIE bit.
4. If 9-bit transmission is desired, set the TX9 bit.
5. Enable transmission by setting the TXEN bit.
6. Verify address detection is disabled by clearing the ADDEN bit of the RCSTA register.
7. If 9-bit transmission is selected, insert the Most Significant bit into the TX9D bit.
8. Start transmission by writing the Least Significant 8 bits to the TXREG register.

TABLE 18-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000X	0000 000X
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXREG	AUSART Transmit Data Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for synchronous slave transmission.

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18.3.2.3 AUSART Synchronous Slave Reception

The operation of the Synchronous Master and Slave modes is identical (**Section 18.3.1.4 "Synchronous Master Reception"**), with the following exceptions:

- Sleep
- CREN bit is always set, therefore the receiver is never Idle
- SREN bit, which is a "don't care" in Slave mode

A character may be received while in Sleep mode by setting the CREN bit prior to entering Sleep. Once the word is received, the RSR register will transfer the data to the RCREG register. If the RCIE interrupt enable bit of the PIE1 register is set, the interrupt generated will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will branch to the interrupt vector.

18.3.2.4 Synchronous Slave Reception Set-up:

1. Set the SYNC and SPEN bits and clear the CSRC bit.
2. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
3. If 9-bit reception is desired, set the RX9 bit.
4. Verify address detection is disabled by clearing the ADDEN bit of the RCSTA register.
5. Set the CREN bit to enable reception.
6. The RCIF bit of the PIR1 register will be set when reception is complete. An interrupt will be generated if the RCIE bit of the PIE1 register was set.
7. If 9-bit mode is enabled, retrieve the Most Significant bit from the RX9D bit of the RCSTA register.
8. Retrieve the 8 Least Significant bits from the receive FIFO by reading the RCREG register.
9. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCSTA register.

TABLE 18-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELC	ANSC7	ANSC6	ANSC5	—	—	ANSC2	ANSC1	ANSC0	111- -111	111- -111
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
RCREG	AUSART Receive Data Register								0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000X	0000 000X
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for synchronous slave reception.

18.4 AUSART Operation During Sleep

The AUSART will remain active during Sleep only in the Synchronous Slave mode. All other modes require the system clock and therefore cannot generate the necessary signals to run the transmit or receive shift registers during Sleep.

Synchronous Slave mode uses an externally generated clock to run the transmit and receive shift registers.

18.4.1 SYNCHRONOUS RECEIVE DURING SLEEP

To receive during Sleep, all the following conditions must be met before entering Sleep mode:

- RCSTA and TXSTA control registers must be configured for synchronous slave reception (refer to **Section 18.3.2.4 "Synchronous Slave Reception Set-up:"**).
- If interrupts are desired, set the RCIE bit of the PIE1 register and the PEIE bit of the INTCON register.
- The RCIF interrupt flag must be cleared by reading RCREG to unload any pending characters in the receive buffer.

Upon entering Sleep mode, the device will be ready to accept data and clocks on the RX/DT and TX/CK pins, respectively. When the data word has been completely clocked in by the external device, the RCIF interrupt flag bit of the PIR1 register will be set. Thereby, waking the processor from Sleep.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit of the INTCON register is also set, then the Interrupt Service Routine at address 0004h will be called.

18.4.2 SYNCHRONOUS TRANSMIT DURING SLEEP

To transmit during Sleep, all the following conditions must be met before entering Sleep mode:

- RCSTA and TXSTA control registers must be configured for synchronous slave transmission (refer to **Section 18.3.2.2 "Synchronous Slave Transmission Set-up:"**).
- The TXIF interrupt flag must be cleared by writing the output data to the TXREG, thereby filling the TSR and transmit buffer.
- If interrupts are desired, set the TXIE bit of the PIE1 register and the PEIE bit of the INTCON register.

Upon entering Sleep mode, the device will be ready to accept clocks on TX/CK pin and transmit data on the RX/DT pin. When the data word in the TSR has been completely clocked out by the external device, the pending byte in the TXREG will transfer to the TSR and the TXIF flag will be set. Thereby, waking the processor from Sleep. At this point, the TXREG is available to accept another character for transmission, which will clear the TXIF flag.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit is also set then the Interrupt Service Routine at address 0004h will be called.

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19.0 SSP MODULE OVERVIEW

The Synchronous Serial Port (SSP) module is a serial interface useful for communicating with other peripherals or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D converters, etc. The SSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit (I^2C™)

19.1 SPI Mode

The SPI mode allows 8 bits of data to be synchronously transmitted and received, simultaneously. The SSP module can be operated in one of two SPI modes:

- Master mode
- Slave mode

SPI is a full-duplex protocol, with all communication being bidirectional and initiated by a master device. All clocking is provided by the master device and all bits are transmitted, MSb first. Care must be taken to ensure that all devices on the SPI bus are setup to allow all controllers to send and receive data at the same time.

A typical SPI connection between microcontroller devices is shown in Figure 19-1. Addressing of more than one slave device is accomplished via multiple hardware slave select lines. External hardware and additional I/O pins must be used to support multiple slave select addressing. This prevents extra overhead in software for communication.

For SPI communication, typically three pins are used:

- Serial Data Out (SDO)
- Serial Data In (SDI)
- Serial Clock (SCK)

Additionally, a fourth pin may be used when in a Slave mode of operation:

- Slave Select ($\overline{SS}$)

FIGURE 19-1: TYPICAL SPI MASTER/SLAVE CONNECTION

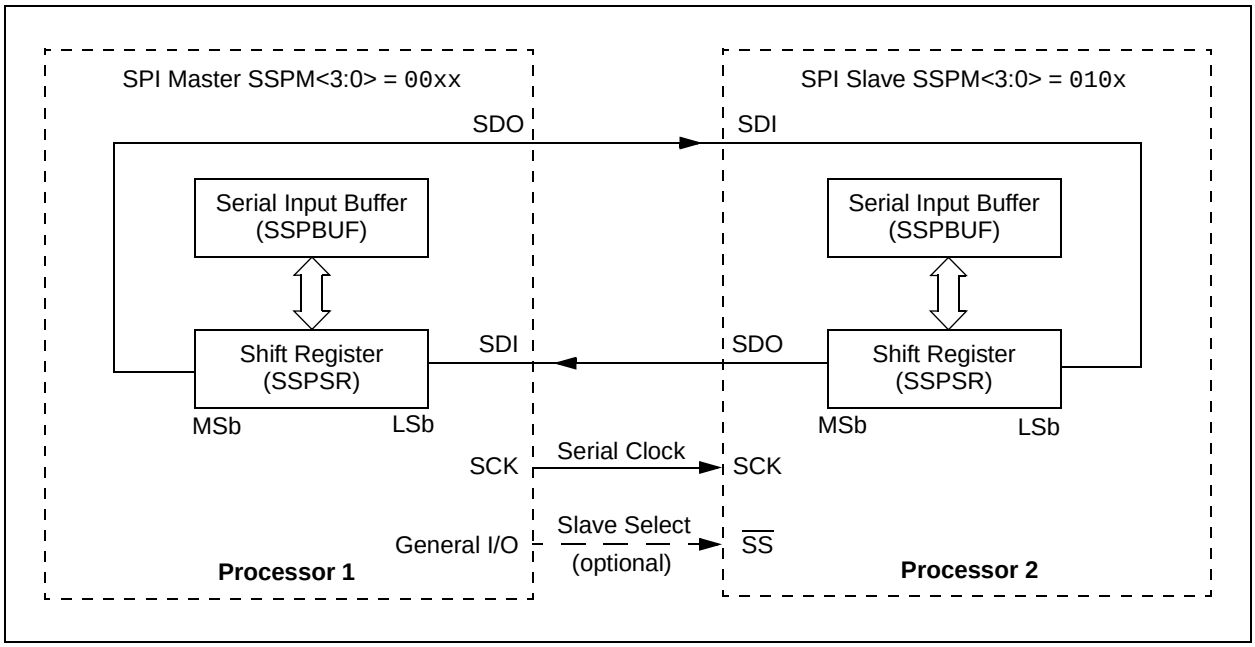
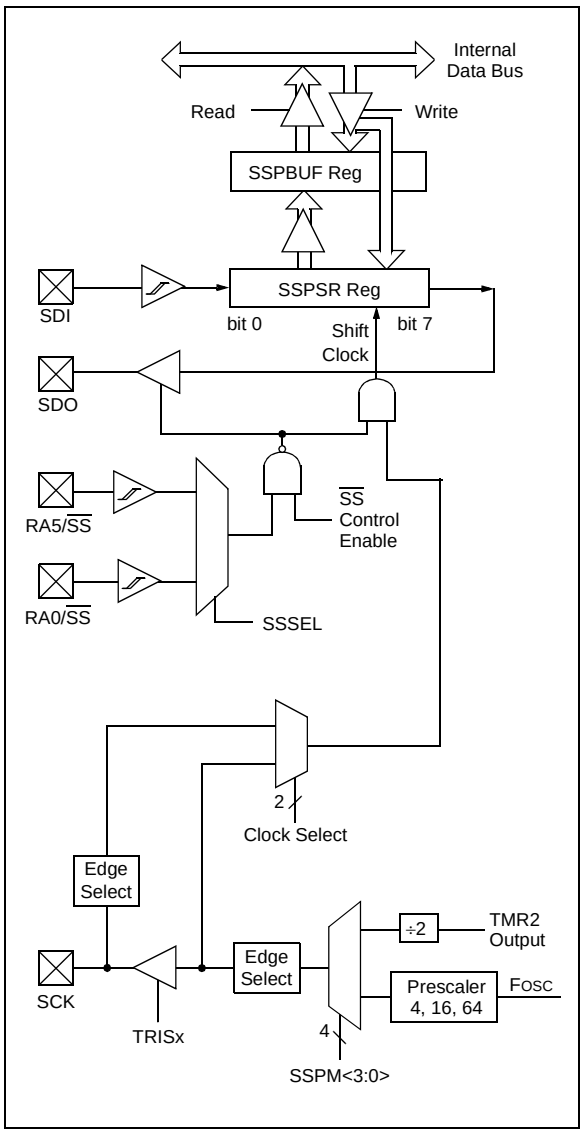


FIGURE 19-2: SPI MODE BLOCK DIAGRAM



19.1.1 MASTER MODE

In Master mode, data transfer can be initiated at any time because the master controls the SCK line. Master mode determines when the slave (Figure 19-1, Processor 2) transmits data via control of the SCK line.

19.1.1.1 Master Mode Operation

The SSP consists of a transmit/receive shift register (SSPSR) and a buffer register (SSPBUF). The SSPSR register shifts the data in and out of the device, MSb first. The SSPBUF register holds the data that is written out of the master until the received data is ready. Once the eight bits of data have been received, the byte is moved to the SSPBUF register. The Buffer Full Status bit, BF of the SSPSTAT register, and the SSP Interrupt Flag bit, SSPIF of the PIR1 register, are then set.

Any write to the SSPBUF register during transmission/reception of data will be ignored and the Write Collision Detect bit, WCOL of the SSPCON register, will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPBUF register completed successfully.

When the application software is expecting to receive valid data, the SSPBUF should be read before the next byte of data is written to the SSPBUF. The BF bit of the SSPSTAT register is set when SSPBUF has been loaded with the received data (transmission is complete). When the SSPBUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. The SSP interrupt may be used to determine when the transmission/reception is complete and the SSPBUF must be read and/or written. If interrupts are not used, then software polling can be done to ensure that a write collision does not occur. Example 19-1 shows the loading of the SSPBUF (SSPSR) for data transmission.

Note:	The SSPSR is not directly readable or writable and can only be accessed by addressing the SSPBUF register.

19.1.1.2 Enabling Master I/O

To enable the serial port, the SSPEN bit of the SSPCON register, must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, re-initialize the SSPCON register and then set the SSPEN bit. If a Master mode of operation is selected in the SSPM bits of the SSPCON register, the SDI, SDO and SCK pins will be assigned as serial port pins.

For these pins to function as serial port pins, they must have their corresponding data direction bits set or cleared in the associated TRIS register as follows:

- SDI configured as input
- SDO configured as output
- SCK configured as output

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19.1.1.3 Master Mode Setup

In Master mode, the data is transmitted/received as soon as the SSPBUF register is loaded with a byte value. If the master is only going to receive, SDO output could be disabled (programmed and used as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate.

When initializing SPI Master mode operation, several options need to be specified. This is accomplished by programming the appropriate control bits in the SSPCON and SSPSTAT registers. These control bits allow the following to be specified:

- SCK as clock output
- Idle state of SCK (CKP bit)
- Data input sample phase (SMP bit)
- Output data on rising/falling edge of SCK (CKE bit)
- Clock bit rate

In Master mode, the SPI clock rate (bit rate) is user selectable to be one of the following:

- FOSC/4 (or TCY)
- FOSC/16 (or 4 • TCY)
- FOSC/64 (or 16 • TCY)
- (Timer2 output)/2

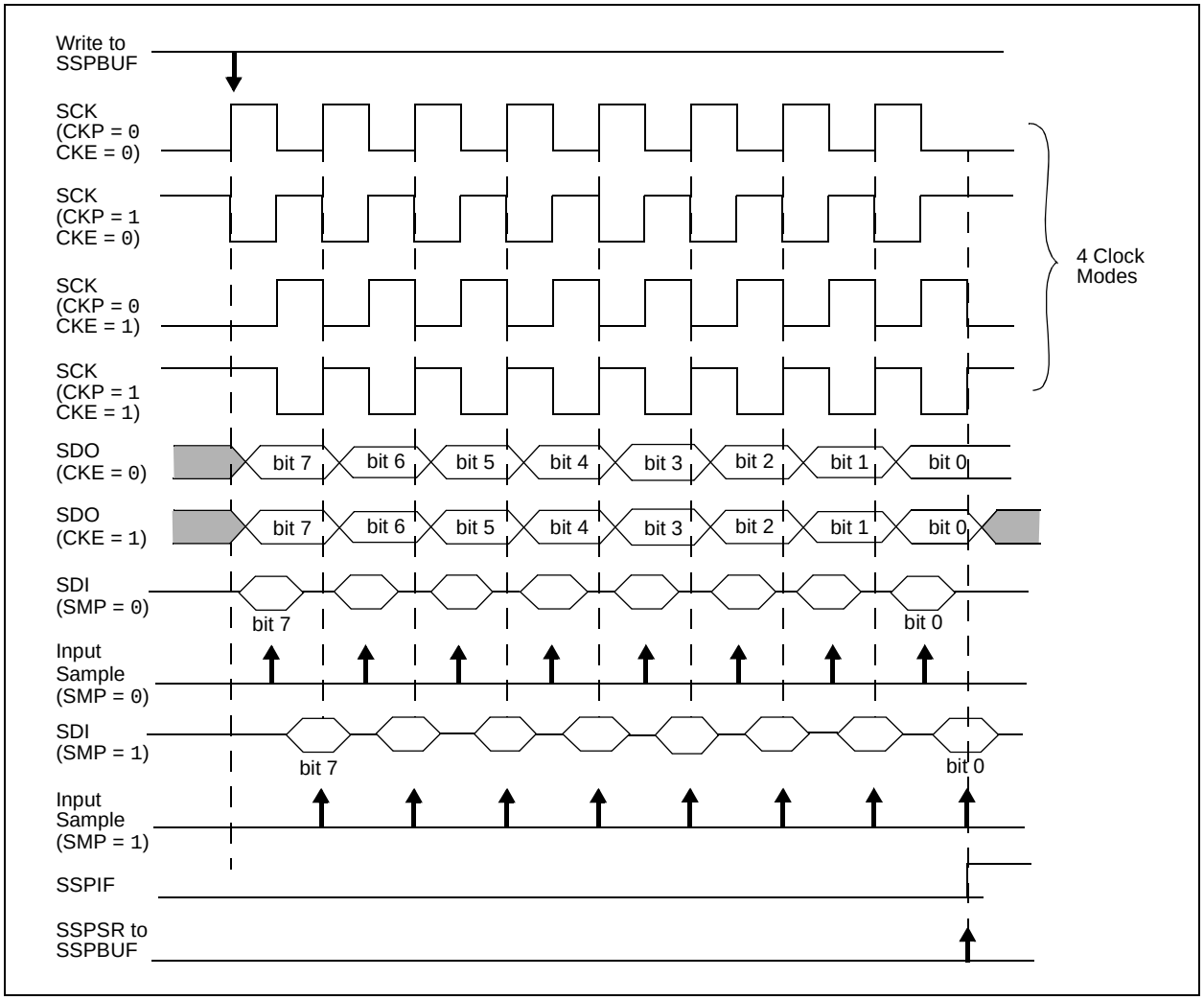
This allows a maximum data rate of 5 Mbps (at FOSC = 20 MHz).

Figure 19-3 shows the waveforms for Master mode. The clock polarity is selected by appropriately programming the CKP bit of the SSPCON register. When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The sample time of the input data is shown based on the state of the SMP bit and can occur at the middle or end of the data output time. The time when the SSPBUF is loaded with the received data is shown.

19.1.1.4 Sleep in Master Mode

In Master mode, all module clocks are halted and the transmission/reception will remain in their current state, paused, until the device wakes from Sleep. After the device wakes up from Sleep, the module will continue to transmit/receive data.

FIGURE 19-3: SPI MASTER MODE WAVEFORM



EXAMPLE 19-1: LOADING THE SSPBUF (SSPSR) REGISTER

```
        BANKSEL   SSPSTAT      ;
LOOP    BTFSS     SSPSTAT, BF  ;Has data been received(transmit complete)?
        GOTO      LOOP         ;No
        BANKSEL   SSPBUF       ;
        MOVF      SSPBUF, W    ;WREG reg = contents of SSPBUF
        MOVWF     RXDATA       ;Save in user RAM, if data is meaningful
        MOVF      TXDATA, W    ;W reg = contents of TXDATA
        MOVWF     SSPBUF       ;New data to xmit
```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19.1.2 SLAVE MODE

For any SPI device acting as a slave, the data is transmitted and received as external clock pulses appear on the SCK pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

19.1.2.1 Slave Mode Operation

The SSP consists of a transmit/receive shift register (SSPSR) and a buffer register (SSPBUF). The SSPSR shifts the data in and out of the device, MSb first. The SSPBUF holds the data that was written to the SSPSR until the received data is ready.

The slave has no control as to when data will be clocked in or out of the device. All data that is to be transmitted, to a master or another slave, must be loaded into the SSPBUF register before the first clock pulse is received.

Once eight bits of data have been received:

- Received byte is moved to the SSPBUF register
- BF bit of the SSPSTAT register is set
- SSPIF bit of the PIR1 register is set

Any write to the SSPBUF register during transmission/ reception of data will be ignored and the Write Collision Detect bit, WCOL of the SSPCON register, will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPBUF register completed successfully.

The user's firmware must read SSPBUF, clearing the BF flag, or the SSPOV bit of the SSPCON register will be set with the reception of the next byte and communication will be disabled.

A SPI module transmits and receives at the same time, occasionally causing dummy data to be transmitted/ received. It is up to the user to determine which data is to be used and what can be discarded.

19.1.2.2 Enabling Slave I/O

To enable the serial port, the SSPEN bit of the SSPCON register must be set. If a Slave mode of operation is selected in the SSPM bits of the SSPCON register, the SDI, SDO and SCK pins will be assigned as serial port pins.

For these pins to function as serial port pins, they must have their corresponding data direction bits set or cleared in the associated TRIS register as follows:

- SDI configured as input
- SDO configured as output
- SCK configured as input

Optionally, a fourth pin, Slave Select ($\overline{SS}$) may be used in Slave mode. Slave Select may be configured to operate on one of the following pins via the SSSEL bit in the APFCON register.

- RA5/AN4/$\overline{SS}$
- RA0/AN0/$\overline{SS}$

Upon selection of a Slave Select pin, the appropriate bits must be set in the ANSELA and TRISA registers. Slave Select must be set as an input by setting the corresponding bit in TRISA, and digital I/O must be enabled on the $\overline{SS}$ pin by clearing the corresponding bit of the ANSELA register.

19.1.2.3 Slave Mode Setup

When initializing the SSP module to SPI Slave mode, compatibility must be ensured with the master device. This is done by programming the appropriate control bits of the SSPCON and SSPSTAT registers. These control bits allow the following to be specified:

- SCK as clock input
- Idle state of SCK (CKP bit)
- Data input sample phase (SMP bit)
- Output data on rising/falling edge of SCK (CKE bit)

Figure 19-4 and Figure 19-5 show example waveforms of Slave mode operation.

FIGURE 19-4: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 0)

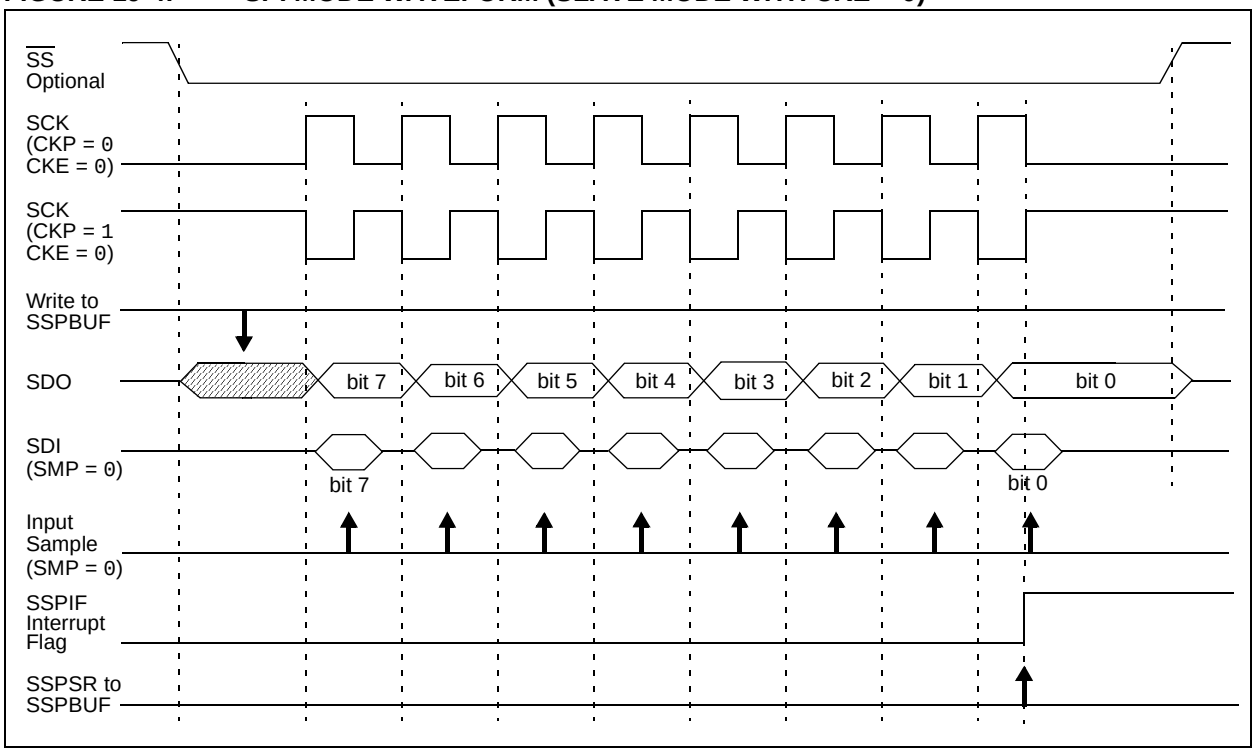
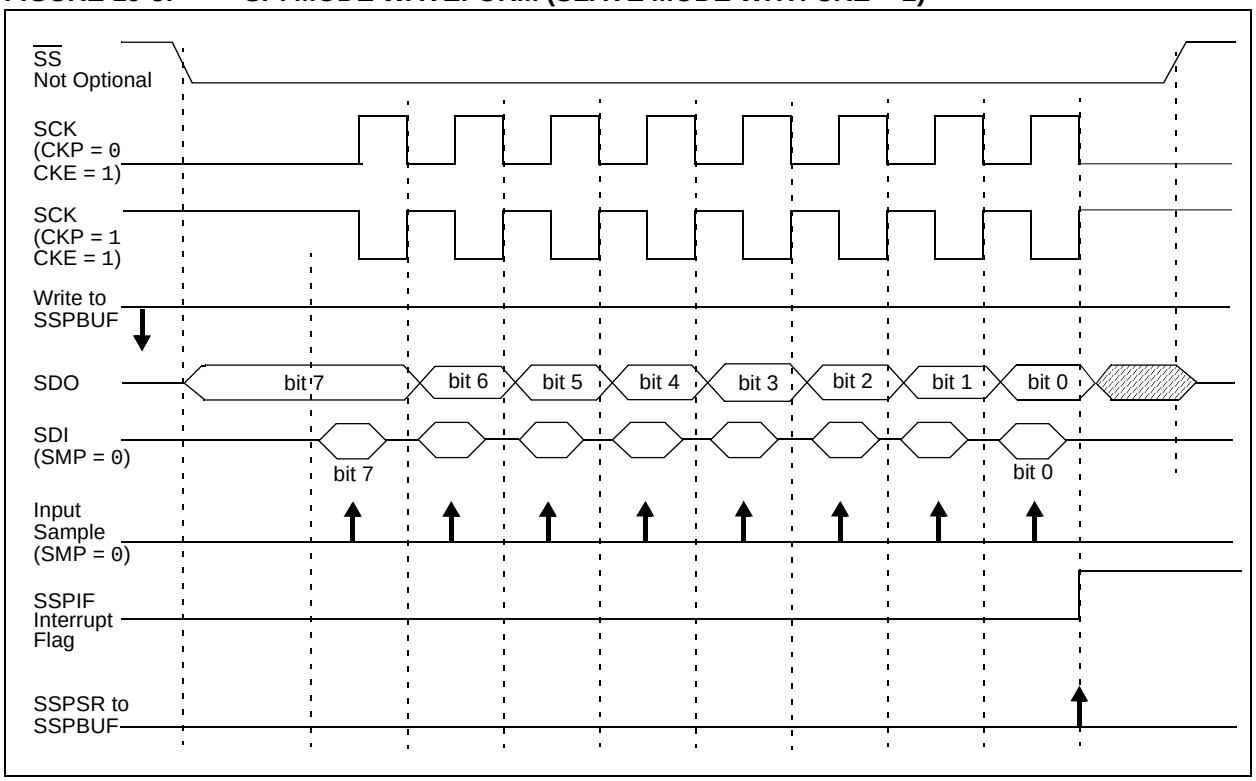


FIGURE 19-5: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)



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19.1.2.4 Slave Select Operation

The $\overline{SS}$ pin allows Synchronous Slave mode operation. The SPI must be in Slave mode with $\overline{SS}$ pin control enabled (SSPM<3:0> = 0100). The associated TRIS bit for the $\overline{SS}$ pin must be set, making $\overline{SS}$ an input.

In Slave Select mode, when:

- $\overline{SS}$ = 0, The device operates as specified in **Section 19.1.2 "Slave Mode"**.
- $\overline{SS}$ = 1, The SPI module is held in Reset and the SDO pin will be tri-stated.

Note 1:	When the SPI is in Slave mode with $\overline{SS}$ pin control enabled (SSPM<3:0> = 0100), the SPI module will reset if the $\overline{SS}$ pin is driven high.
2:	If the SPI is used in Slave mode with CKE set, the $\overline{SS}$ pin control must be enabled.

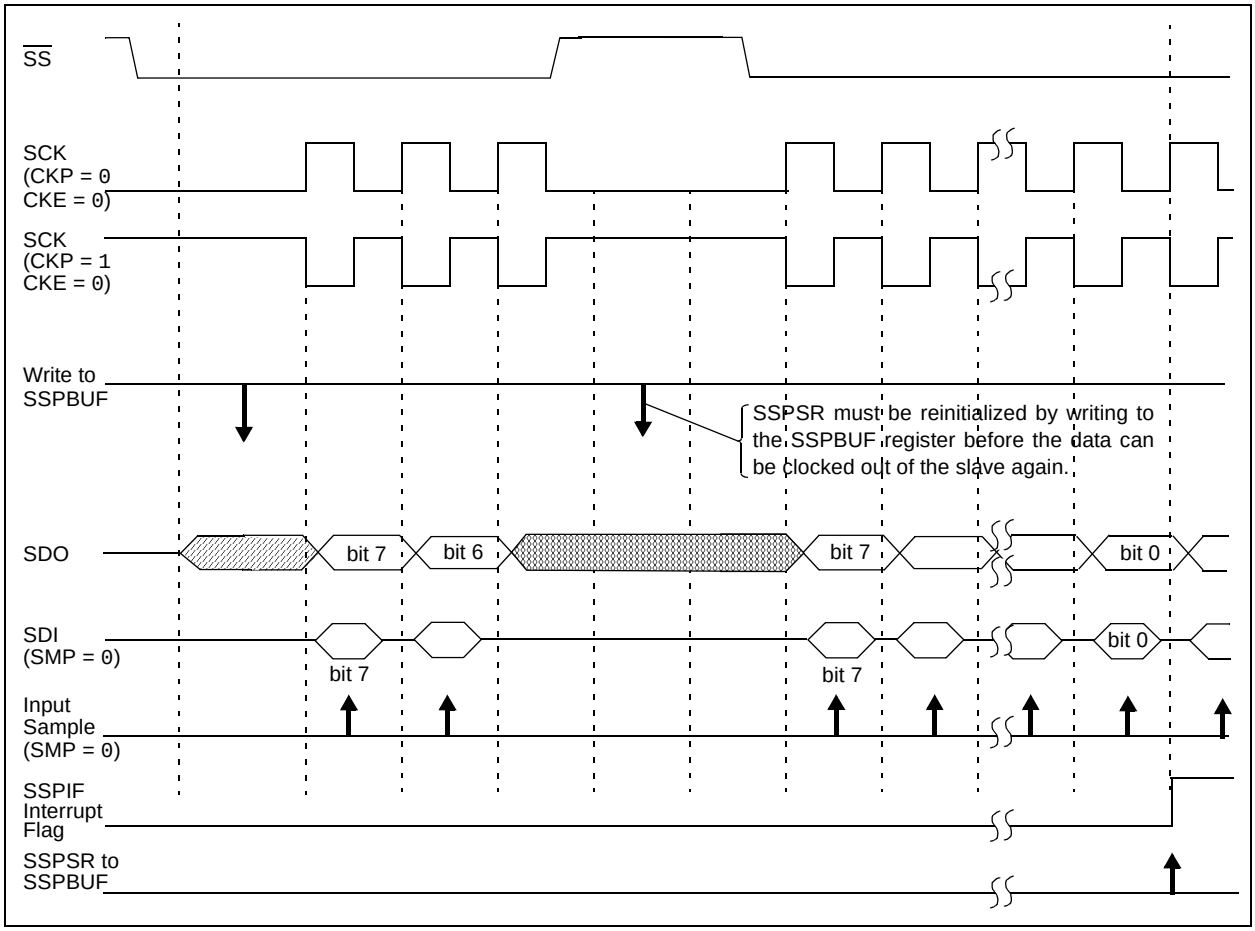
When the SPI module resets, the bit counter is cleared to '0'. This can be done by either forcing the $\overline{SS}$ pin to a high level or clearing the SSPEN bit. Figure 19-6 shows the timing waveform for such a synchronization event.

Note:	SSPSR must be reinitialized by writing to the SSPBUF register before the data can be clocked out of the slave again.

19.1.2.5 Sleep in Slave Mode

While in Sleep mode, the slave can transmit/receive data. The SPI Transmit/Receive Shift register operates asynchronously to the device on the externally supplied clock source. This allows the device to be placed in Sleep mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the SSP interrupt flag bit will be set and if enabled, will wake the device from Sleep.

FIGURE 19-6: SLAVE SELECT SYNCHRONIZATION WAVEFORM



REGISTER 19-1: SSPCON: SYNC SERIAL PORT CONTROL REGISTER (SPI MODE)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **WCOL**: Write Collision Detect bit
- 1 = The SSPBUF register is written while it is still transmitting the previous word (must be cleared in software)
- 0 = No collision

bit 6 **SSPOV**: Receive Overflow Indicator bit
- 1 = A new byte is received while the SSPBUF register is still holding the previous data. In case of overflow, the data in SSPSR is lost. Overflow can only occur in Slave mode. The user must read the SSPBUF, even if only transmitting data, to avoid setting overflow. In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSPBUF register.
- 0 = No overflow

bit 5 **SSPEN**: Synchronous Serial Port Enable bit
- 1 = Enables serial port and configures SCK, SDO and SDI as serial port pins(1)
- 0 = Disables serial port and configures these pins as I/O port pins

bit 4 **CKP**: Clock Polarity Select bit
- 1 = Idle state for clock is a high level
- 0 = Idle state for clock is a low level

bit 3-0 **SSPM<3:0>**: Synchronous Serial Port Mode Select bits
- 0000 = SPI Master mode, clock = FOSC/4
- 0001 = SPI Master mode, clock = FOSC/16
- 0010 = SPI Master mode, clock = FOSC/64
- 0011 = SPI Master mode, clock = TMR2 output/2
- 0100 = SPI Slave mode, clock = SCK pin. $\overline{SS}$ pin control enabled.
- 0101 = SPI Slave mode, clock = SCK pin. $\overline{SS}$ pin control disabled. $\overline{SS}$ can be used as I/O pin.

Note 1: When enabled, these pins must be properly configured as input or output.

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REGISTER 19-2: SSPSTAT: SYNC SERIAL PORT STATUS REGISTER (SPI MODE)

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0
SMP	CKE	D/$\overline{A}$	P	S	R/$\overline{W}$	UA	BF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7 **SMP**: SPI Data Input Sample Phase bit
SPI Master mode:
1 = Input data sampled at end of data output time
0 = Input data sampled at middle of data output time
SPI Slave mode:
SMP must be cleared when SPI is used in Slave mode

bit 6 **CKE**: SPI Clock Edge Select bit
SPI mode, CKP = 0:
1 = Data stable on rising edge of SCK
0 = Data stable on falling edge of SCK
SPI mode, CKP = 1:
1 = Data stable on falling edge of SCK
0 = Data stable on rising edge of SCK

bit 5 **D/$\overline{A}$:** Data/$\overline{\text{Address}}$ bit
Used in I^2C mode only.

bit 4 **P:** Stop bit
Used in I^2C mode only.

bit 3 **S:** Start bit
Used in I^2C mode only.

bit 2 **R/$\overline{W}$:** Read/$\overline{\text{Write}}$ Information bit
Used in I^2C mode only.

bit 1 **UA:** Update Address bit
Used in I^2C mode only.

bit 0 **BF**: Buffer Full Status bit
1 = Receive complete, SSPBUF is full
0 = Receive not complete, SSPBUF is empty

TABLE 19-1: SUMMARY OF REGISTERS ASSOCIATED WITH SPI OPERATION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELA	ANSA7	ANSA6	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	1111 1111	1111 1111
APFCON	—	—	—	—	—	—	SSSEL	CCP2SEL	---- --00	---- --00
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PR2	Timer2 Period Register								1111 1111	1111 1111
SSPBUF	Synchronous Serial Port Receive Buffer/Transmit Register								xxxx xxxx	uuuu uuuu
SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
SSPSTAT	SMP	CKE	D/$\overline{A}$	P	S	R/$\overline{W}$	UA	BF	0000 0000	0000 0000
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
T2CON	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000

Legend: x = unknown, u = unchanged, – = unimplemented, read as '0'. Shaded cells are not used by the SSP in SPI mode.

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19.2 I²C Mode

The SSP module, in I²C mode, implements all slave functions, except general call support. It provides interrupts on Start and Stop bits in hardware to facilitate firmware implementations of the master functions. The SSP module implements the I²C Standard mode specifications:

- I²C Slave mode (7-bit address)
- I²C Slave mode (10-bit address)
- Start and Stop bit interrupts enabled to support firmware Master mode
- Address masking

Two pins are used for data transfer; the SCL pin (clock line) and the SDA pin (data line). The user must configure the two pin's data direction bits as inputs in the appropriate TRIS register. Upon enabling I²C mode, the I²C slew rate limiters in the I/O pads are controlled by the SMP bit of the SSPSTAT register. The SSP module functions are enabled by setting the SSPEN bit of the SSPCON register.

Data is sampled on the rising edge and shifted out on the falling edge of the clock. This ensures that the SDA signal is valid during the SCL high time. The SCL clock input must have minimum high and low times for proper operation. Refer to **Section 25.0 "Electrical Specifications"**.

FIGURE 19-7: I²C™ MODE BLOCK DIAGRAM

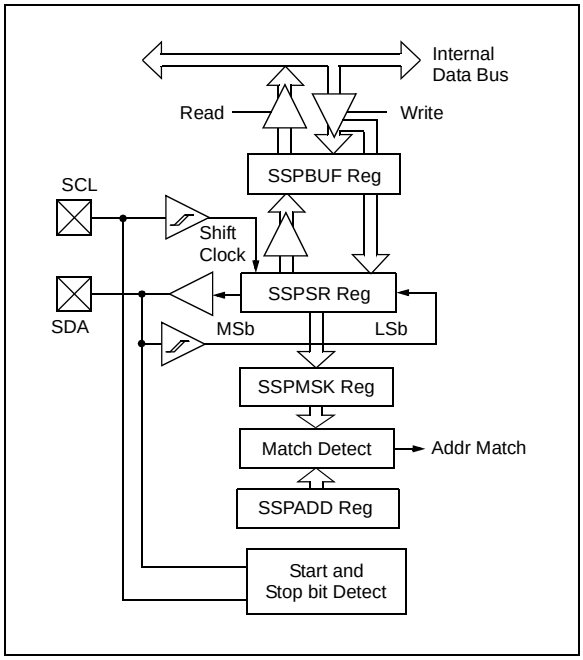
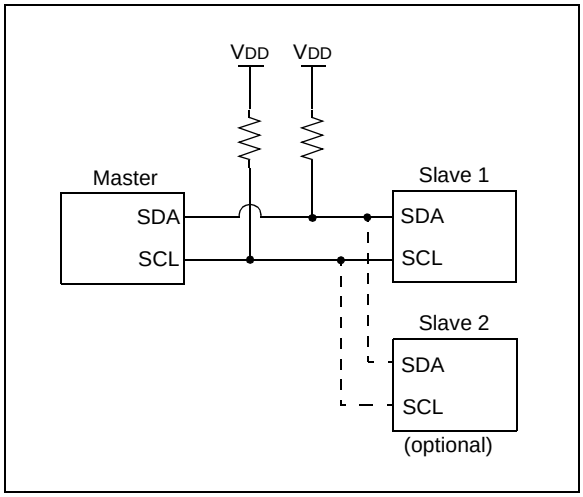


FIGURE 19-8: TYPICAL I²C™ CONNECTIONS



The SSP module has six registers for I²C operation. They are:

- SSP Control (SSPCON) register
- SSP Status (SSPSTAT) register
- Serial Receive/Transmit Buffer (SSPBUF) register
- SSP Shift Register (SSPSR), not directly accessible
- SSP Address (SSPADD) register
- SSP Address Mask (SSPMSK) register

19.2.1 HARDWARE SETUP

Selection of I²C mode, with the SSPEN bit of the SSPCON register set, forces the SCL and SDA pins to be open drain, provided these pins are programmed as inputs by setting the appropriate TRISC bits. The SSP module will override the input state with the output data, when required, such as for Acknowledge and slave-transmitter sequences.

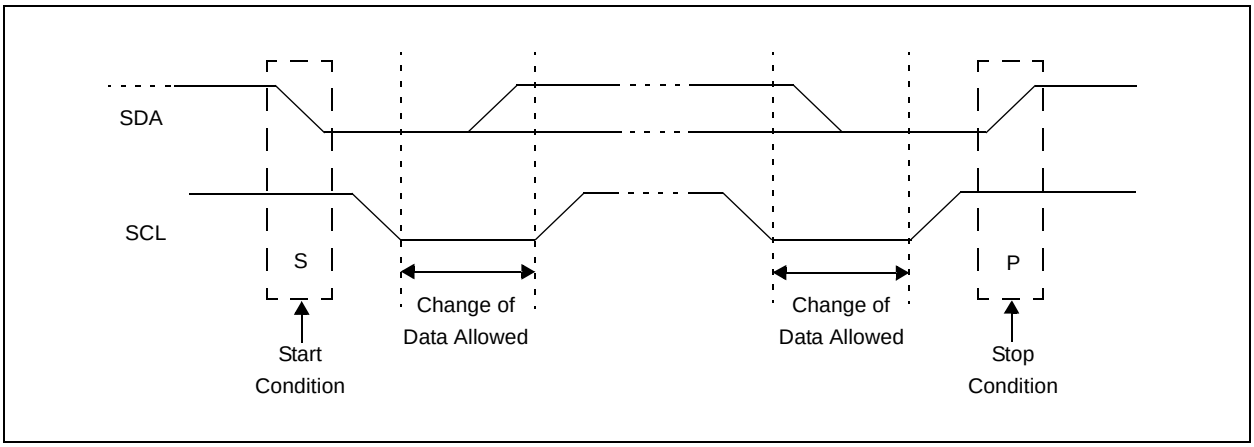
Note:	Pull-up resistors must be provided externally to the SCL and SDA pins for proper operation of the I²C module.

19.2.2 START AND STOP CONDITIONS

During times of no data transfer (Idle time), both the clock line (SCL) and the data line (SDA) are pulled high through external pull-up resistors. The Start and Stop conditions determine the start and stop of data transmission. The Start condition is defined as a high-to-low transition of the SDA line while SCL is high. The Stop condition is defined as a low-to-high transition of the SDA line while SCL is high.

Figure 19-9 shows the Start and Stop conditions. A master device generates these conditions for starting and terminating data transfer. Due to the definition of the Start and Stop conditions, when data is being transmitted, the SDA line can only change state when the SCL line is low.

FIGURE 19-9: START AND STOP CONDITIONS



19.2.3 ACKNOWLEDGE

After the valid reception of an address or data byte, the hardware automatically will generate the Acknowledge ($\overline{ACK}$) pulse and load the SSPBUF register with the received value currently in the SSPSR register. There are certain conditions that will cause the SSP module not to generate this $\overline{ACK}$ pulse. They include any or all of the following:

- The Buffer Full bit, BF of the SSPSTAT register, was set before the transfer was received.
- The SSP Overflow bit, SSPOV of the SSPCON register, was set before the transfer was received.
- The SSP Module is being operated in Firmware Master mode.

In such a case, the SSPSR register value is not loaded into the SSPBUF, but bit SSPIF of the PIR1 register is set. Table 19-2 shows the results of when a data transfer byte is received, given the status of bits BF and SSPOV. Flag bit BF is cleared by reading the SSPBUF register, while bit SSPOV is cleared through software.

TABLE 19-2: DATA TRANSFER RECEIVED BYTE ACTIONS

Status Bits as Data Transfer is Received		SSPSR → SSPBUF	Generate $\overline{ACK}$ Pulse	Set bit SSPIF (SSP Interrupt occurs if enabled)
BF	SSPOV			
0	0	Yes	Yes	Yes
1	0	No	No	Yes
1	1	No	No	Yes
0	1	No	No	Yes

Note: Shaded cells show the conditions where the user software did not properly clear the overflow condition.

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19.2.4 ADDRESSING

Once the SSP module has been enabled, it waits for a Start condition to occur. Following the Start condition, the 8 bits are shifted into the SSPSR register. All incoming bits are sampled with the rising edge of the clock line (SCL).

19.2.4.1 7-bit Addressing

In 7-bit Addressing mode (Figure 19-10), the value of register SSPSR<7:1> is compared to the value of register SSPADD<7:1>. The address is compared on the falling edge of the eighth clock (SCL) pulse. If the addresses match, and the BF and SSPOV bits are clear, the following events occur:

- The SSPSR register value is loaded into the SSPBUF register.
- The BF bit is set.
- An $\overline{\text{ACK}}$ pulse is generated.
- SSP interrupt flag bit, SSPIF of the PIR1 register, is set (interrupt is generated if enabled) on the falling edge of the ninth SCL pulse.

19.2.4.2 10-bit Addressing

In 10-bit Address mode, two address bytes need to be received by the slave (Figure 19-11). The five Most Significant bits (MSbs) of the first address byte specify if it is a 10-bit address. The R/$\overline{\text{W}}$ bit of the SSPSTAT register must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal '1111 0 A9 A8 0', where A9 and A8 are the two MSbs of the address.

The sequence of events for 10-bit address is as follows for reception:

1. Load SSPADD register with high byte of address.
2. Receive first (high) byte of address (bits SSPIF, BF and UA of the SSPSTAT register are set).
3. Read the SSPBUF register (clears bit BF).
4. Clear the SSPIF flag bit.
5. Update the SSPADD register with second (low) byte of address (clears UA bit and releases the SCL line).
6. Receive low byte of address (bits SSPIF, BF and UA are set).
7. Update the SSPADD register with the high byte of address. If match releases SCL line, this will clear bit UA.
8. Read the SSPBUF register (clears bit BF).
9. Clear flag bit SSPIF.

If data is requested by the master, once the slave has been addressed:

1. Receive repeated Start condition.
2. Receive repeat of high byte address with R/$\overline{\text{W}}$ = 1, indicating a read.
3. BF bit is set and the CKP bit is cleared, stopping SCL and indicating a read request.
4. SSPBUF is written, setting BF, with the data to send to the master device.
5. CKP is set in software, releasing the SCL line.

19.2.4.3 Address Masking

The Address Masking register (SSPMSK) is only accessible while the SSPM bits of the SSPCON register are set to '1001'. In this register, the user can select which bits of a received address the hardware will compare when determining an address match. Any bit that is set to a zero in the SSPMSK register, the corresponding bit in the received address byte and SSPADD register are ignored when determining an address match. By default, the register is set to all ones, requiring a complete match of a 7-bit address or the lower eight bits of a 10-bit address.

19.2.5 RECEPTION

When the R/$\overline{W}$ bit of the received address byte is clear, the master will write data to the slave. If an address match occurs, the received address is loaded into the SSPBUF register. An address byte overflow will occur if that loaded address is not read from the SSPBUF before the next complete byte is received.

An SSP interrupt is generated for each data transfer byte. The BF, R/$\overline{W}$ and D/$\overline{A}$ bits of the SSPSTAT register are used to determine the status of the last received byte.

FIGURE 19-10: I²C™ WAVEFORMS FOR RECEPTION (7-BIT ADDRESS)

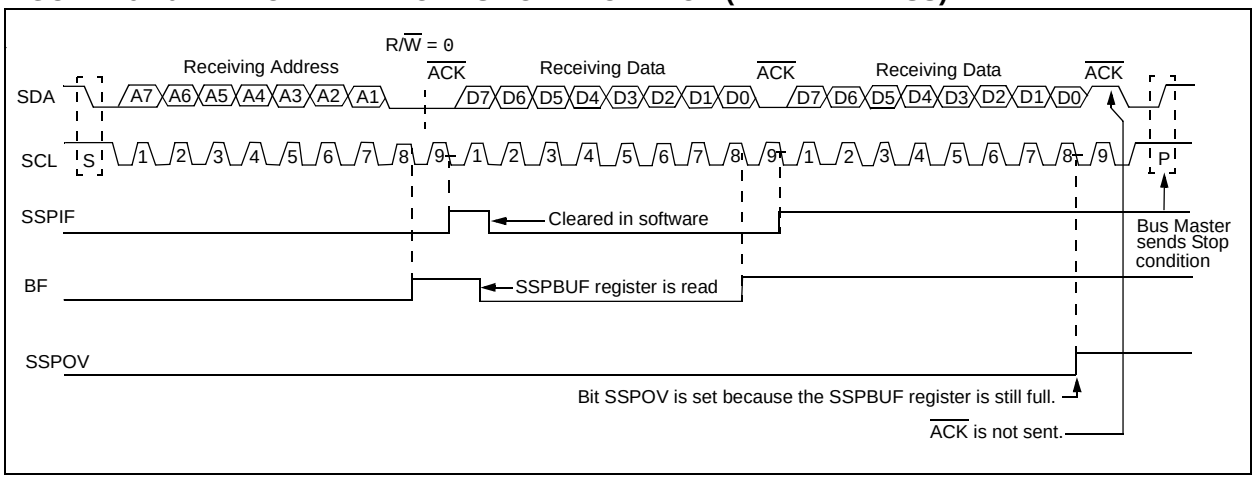

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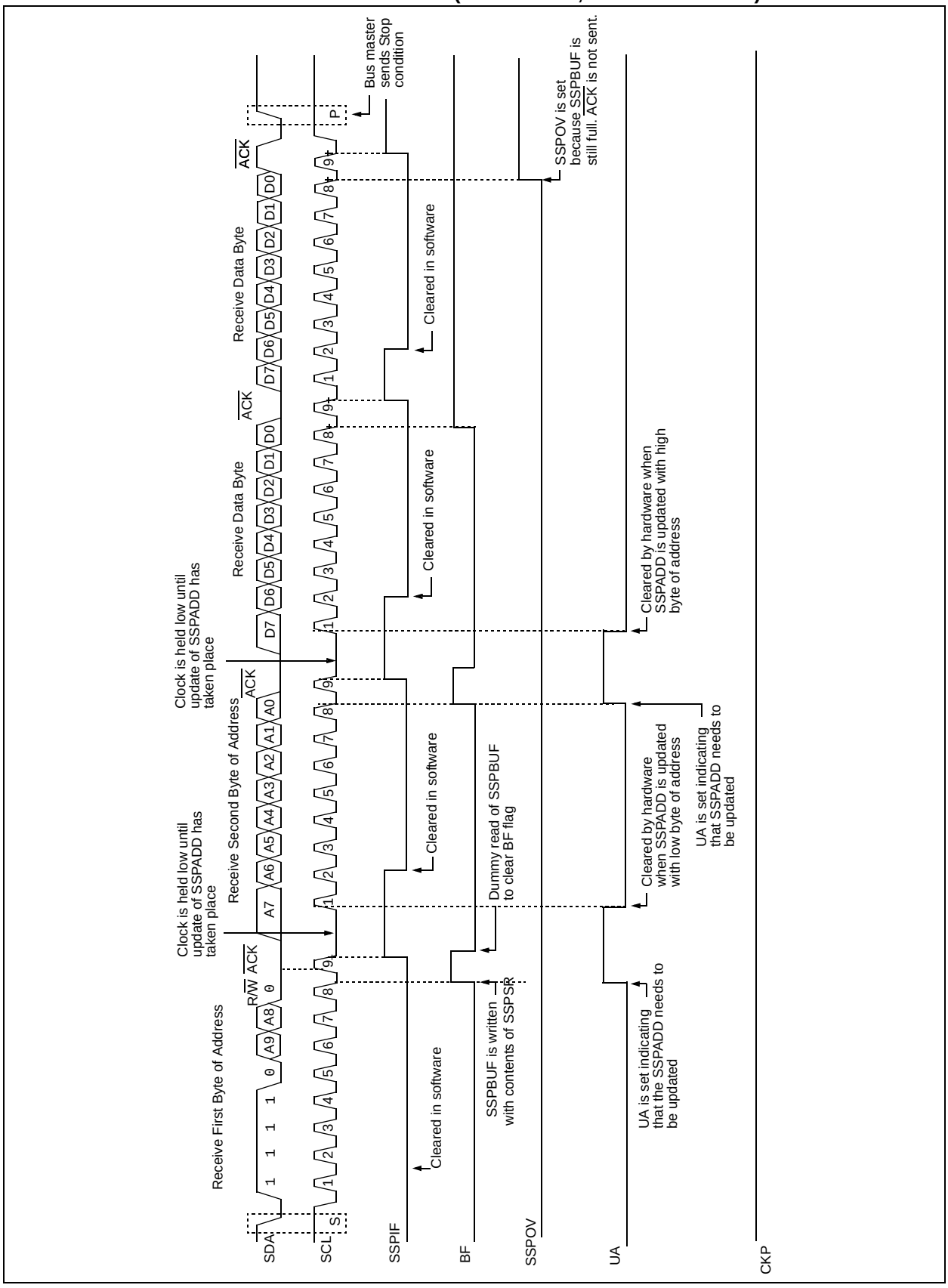
FIGURE 19-11: I²C™ SLAVE MODE TIMING (RECEPTION, 10-BIT ADDRESS)



19.2.6 TRANSMISSION

When the R/$\overline{W}$ bit of the received address byte is set and an address match occurs, the R/$\overline{W}$ bit of the SSPSTAT register is set and the slave will respond to the master by reading out data. After the address match, an $\overline{ACK}$ pulse is generated by the slave hardware and the SCL pin is held low (clock is automatically stretched) until the slave is ready to respond. See **Section 19.2.7 "Clock Stretching"**. The data the slave will transmit must be loaded into the SSPBUF register, which sets the BF bit. The SCL line is released by setting the CKP bit of the SSPCON register.

An SSP interrupt is generated for each transferred data byte. The SSPIF flag bit of the PIR1 register initiates an SSP interrupt, and must be cleared by software before the next byte is transmitted. The BF bit of the SSPSTAT register is cleared on the falling edge of the 8th received clock pulse. The SSPIF flag bit is set on the falling edge of the ninth clock pulse.

Following the 8th falling clock edge, control of the SDA line is released back to the master so that the master can acknowledge or not acknowledge the response. If the master sends a not acknowledge, the slave's transmission is complete and the slave must monitor for the next Start condition. If the master acknowledges, control of the bus is returned to the slave to transmit another byte of data. Just as with the previous byte, the clock is stretched by the slave, data must be loaded into the SSPBUF and CKP must be set to release the clock line (SCL).

FIGURE 19-12: I²C™ WAVEFORMS FOR TRANSMISSION (7-BIT ADDRESS)

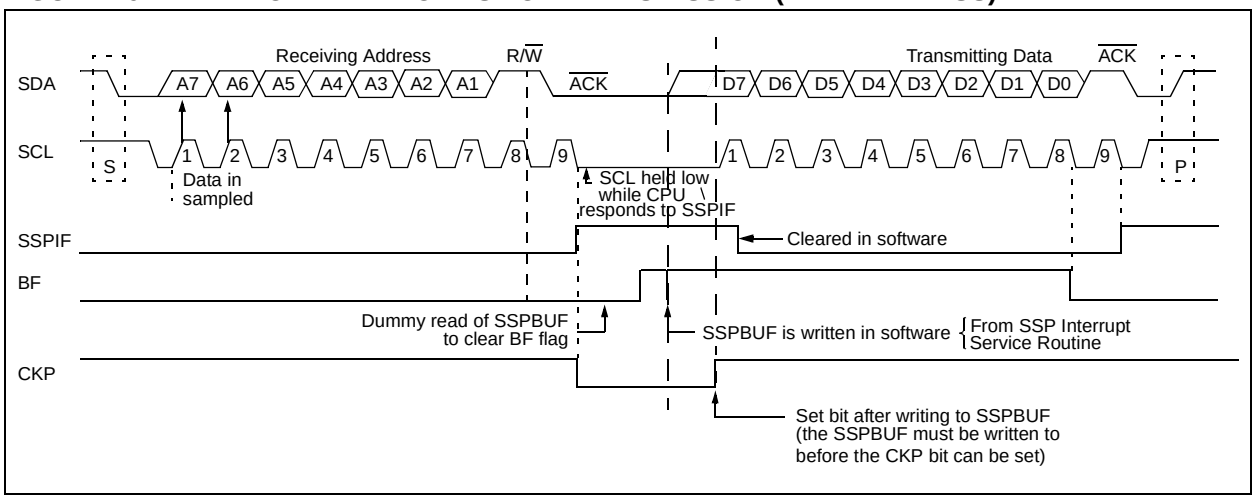


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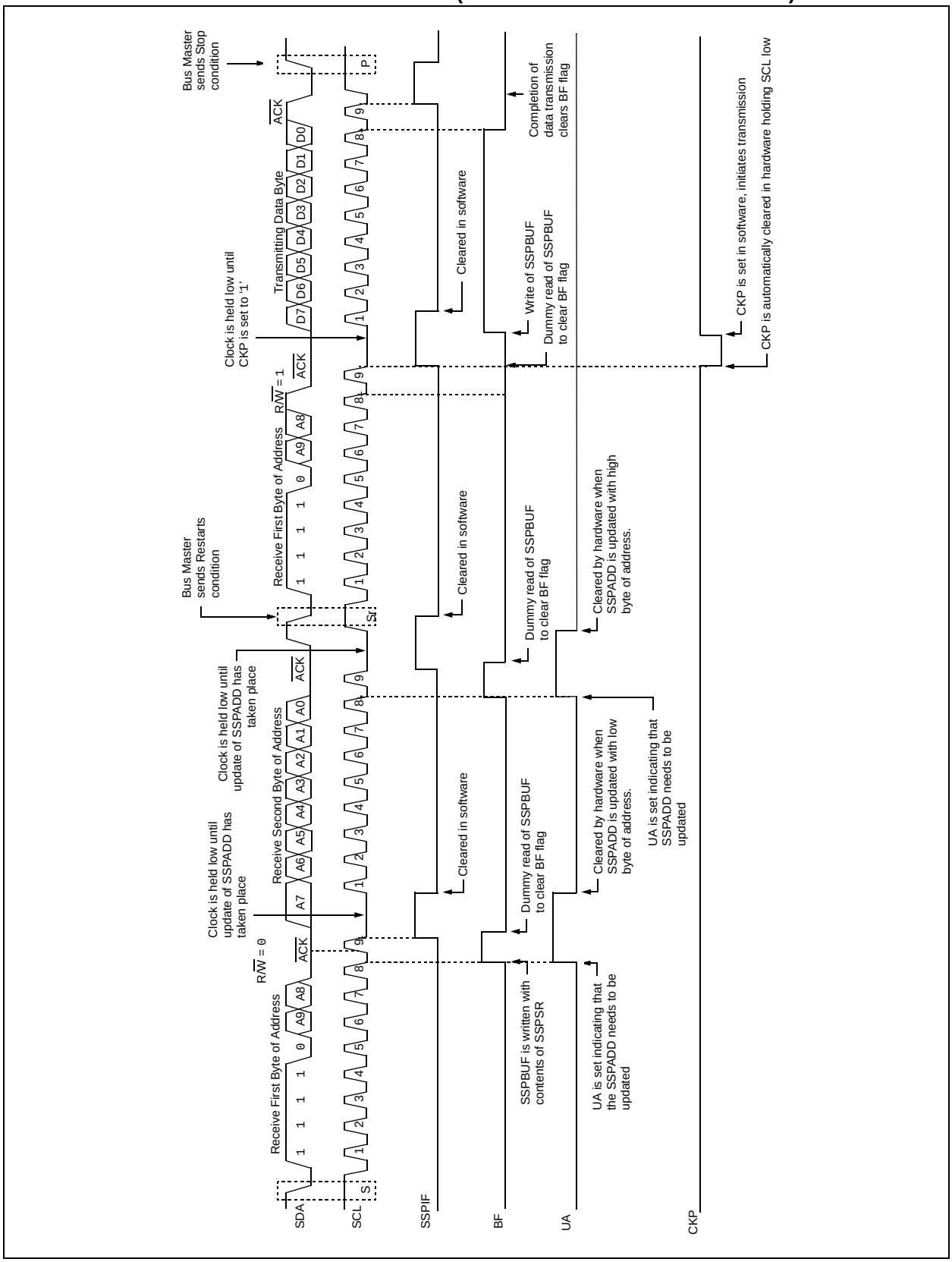
FIGURE 19-13: I²C™ SLAVE MODE TIMING (TRANSMISSION 10-BIT ADDRESS)



19.2.7 CLOCK STRETCHING

During any SCL low phase, any device on the I^2C bus may hold the SCL line low and delay, or pause, the transmission of data. This "stretching" of a transmission allows devices to slow down communication on the bus. The SCL line must be constantly sampled by the master to ensure that all devices on the bus have released SCL for more data.

Stretching usually occurs after an $\overline{ACK}$ bit of a transmission, delaying the first bit of the next byte. The SSP module hardware automatically stretches for two conditions:

- After a 10-bit address byte is received (update SSPADD register)
- Anytime the CKP bit of the SSPCON register is cleared by hardware

The module will hold SCL low until the CKP bit is set. This allows the user slave software to update SSPBUF with data that may not be readily available. In 10-bit addressing modes, the SSPADD register must be updated after receiving the first and second address bytes. The SSP module will hold the SCL line low until the SSPADD has a byte written to it. The UA bit of the SSPSTAT register will be set, along with SSPIF, indicating an address update is needed.

19.2.8 FIRMWARE MASTER MODE

Master mode of operation is supported in firmware using interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits of the SSPSTAT register are cleared from a Reset or when the SSP module is disabled (SSPEN cleared). The Stop (P) and Start (S) bits will toggle based on the Start and Stop conditions. Control of the I^2C bus may be taken when the P bit is set or the bus is Idle and both the S and P bits are clear.

In Firmware Master mode, the SCL and SDA lines are manipulated by setting/clearing the corresponding TRIS bit(s). The output level is always low, irrespective of the value(s) in the corresponding PORT register bit(s). When transmitting a '1', the TRIS bit must be set (input) and a '0', the TRIS bit must be clear (output).

The following events will cause the SSP Interrupt Flag bit, SSPIF, to be set (SSP Interrupt will occur if enabled):

- Start condition
- Stop condition
- Data transfer byte transmitted/received

Firmware Master mode of operation can be done with either the Slave mode Idle (SSPM<3:0> = 1011), or with either of the Slave modes in which interrupts are enabled. When both master and slave functionality is enabled, the software needs to differentiate the source(s) of the interrupt.

Refer to Application Note AN554, "*Software Implementation of I^2C™ Bus Master*" (DS00554) for more information.

19.2.9 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allow the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset or when the SSP module is disabled. The Stop (P) and Start (S) bits will toggle based on the Start and Stop conditions. Control of the I^2C bus may be taken when the P bit of the SSPSTAT register is set or when the bus is Idle, and both the S and P bits are clear. When the bus is busy, enabling the SSP interrupt will generate the interrupt when the Stop condition occurs.

In Multi-Master operation, the SDA line must be monitored to see if the signal level is the expected output level. This check only needs to be done when a high level is output. If a high level is expected and a low level is present, the device needs to release the SDA and SCL lines (set TRIS bits). There are two stages where this arbitration of the bus can be lost. They are the Address Transfer and Data Transfer stages.

When the slave logic is enabled, the slave continues to receive. If arbitration was lost during the address transfer stage, communication to the device may be in progress. If addressed, an $\overline{ACK}$ pulse will be generated. If arbitration was lost during the data transfer stage, the device will need to re-transfer the data at a later time.

Refer to Application Note AN578, "*Use of the SSP Module in the I^2C™ Multi-Master Environment*" (DS00578) for more information.

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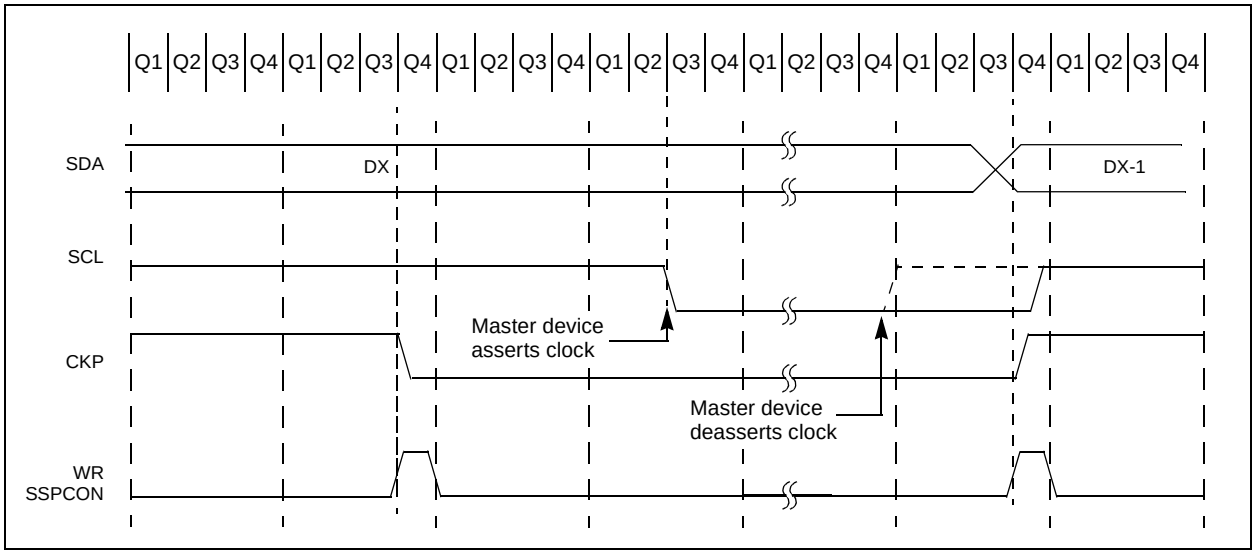
19.2.10 CLOCK SYNCHRONIZATION

When the CKP bit is cleared, the SCL output is held low once it is sampled low. Therefore, the CKP bit will not stretch the SCL line until an external I^2C master device has already asserted the SCL line low. The SCL output will remain low until the CKP bit is set and all other devices on the I^2C bus have released SCL. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCL (Figure 19-14).

19.2.11 SLEEP OPERATION

While in Sleep mode, the I^2C module can receive addresses of data, and when an address match or complete byte transfer occurs, wake the processor from Sleep (if SSP interrupt is enabled).

FIGURE 19-14: CLOCK SYNCHRONIZATION TIMING



REGISTER 19-3: SSPCON: SYNCHRONOUS SERIAL PORT CONTROL REGISTER (I²C MODE)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **WCOL**: Write Collision Detect bit
- 1 = The SSPBUF register is written while it is still transmitting the previous word (must be cleared in software)
- 0 = No collision

bit 6 **SSPOV**: Receive Overflow Indicator bit
- 1 = A byte is received while the SSPBUF register is still holding the previous byte. SSPOV is a "don't care" in Transmit mode. SSPOV must be cleared in software in either mode.
- 0 = No overflow

bit 5 **SSPEN**: Synchronous Serial Port Enable bit
- 1 = Enables the serial port and configures the SDA and SCL pins as serial port pins **(2)**
- 0 = Disables serial port and configures these pins as I/O port pins

bit 4 **CKP**: Clock Polarity Select bit
- 1 = Release control of SCL
- 0 = Holds clock low (clock stretch). (Used to ensure data setup time.)

bit 3-0 **SSPM<3:0>**: Synchronous Serial Port Mode Select bits
- 0110 = I²C Slave mode, 7-bit address
- 0111 = I²C Slave mode, 10-bit address
- 1000 = Reserved
- 1001 = Load SSPMSK register at SSPADD SFR Address **(1)**
- 1010 = Reserved
- 1011 = I²C Firmware Controlled Master mode (Slave Idle)
- 1100 = Reserved
- 1101 = Reserved
- 1110 = I²C Slave mode, 7-bit address with Start and Stop bit interrupts enabled
- 1111 = I²C Slave mode, 10-bit address with Start and Stop bit interrupts enabled

Note 1: When this mode is selected, any reads or writes to the SSPADD SFR address accesses the SSPMSK register.

2: When enabled, these pins must be properly configured as input or output using the associated TRIS bit.

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REGISTER 19-4: SSPSTAT: SYNCHRONOUS SERIAL PORT STATUS REGISTER (I²C MODE)

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0
SMP	CKE	D/$\overline{A}$	P	S	R/$\overline{W}$	UA	BF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **SMP**: SPI Data Input Sample Phase bit
1 = Slew Rate Control (limiting) disabled. Operating in I²C Standard mode (100 kHz and 1 MHz).
0 = Slew Rate Control (limiting) enabled. Operating in I²C Fast mode (400 kHz).

bit 6 **CKE**: SPI Clock Edge Select bit
This bit must be maintained clear. Used in SPI mode only.

bit 5 **D/$\overline{A}$**: DATA/$\overline{\text{ADDRESS}}$ bit (I²C mode only)
1 = Indicates that the last byte received or transmitted was data
0 = Indicates that the last byte received or transmitted was address

bit 4 **P**: Stop bit
This bit is cleared when the SSP module is disabled, or when the Start bit is detected last.
1 = Indicates that a Stop bit has been detected last (this bit is '0' on Reset)
0 = Stop bit was not detected last

bit 3 **S**: Start bit
This bit is cleared when the SSP module is disabled, or when the Stop bit is detected last.
1 = Indicates that a Start bit has been detected last (this bit is '0' on Reset)
0 = Start bit was not detected last

bit 2 **R/$\overline{W}$**: READ/$\overline{\text{WRITE}}$ bit Information
This bit holds the R/$\overline{W}$ bit information following the last address match. This bit is only valid from the address match to the next Start bit, Stop bit or $\overline{\text{ACK}}$ bit.
1 = Read
0 = Write

bit 1 **UA**: Update Address bit (10-bit I²C mode only)
1 = Indicates that the user needs to update the address in the SSPADD register
0 = Address does not need to be updated

bit 0 **BF**: Buffer Full Status bit
Receive:
1 = Receive complete, SSPBUF is full
0 = Receive not complete, SSPBUF is empty
Transmit:
1 = Transmit in progress, SSPBUF is full
0 = Transmit complete, SSPBUF is empty

REGISTER 19-5: SSPMSK: SSP MASK REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-1 **MSK<7:1>:** Mask bits
1 = The received address bit n is compared to SSPADD<n> to detect I^2C address match
0 = The received address bit n is not used to detect I^2C address match

bit 0 **MSK<0>:** Mask bit for I^2C Slave Mode, 10-bit Address
I^2C Slave Mode, 10-bit Address (SSPM<3:0> = 0111):
1 = The received address bit '0' is compared to SSPADD<0> to detect I^2C address match
0 = The received address bit '0' is not used to detect I^2C address match
All other SSP modes: this bit has no effect.

REGISTER 19-6: SSPADD: SSP I^2C™ ADDRESS REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ADD7	ADD6	ADD5	ADD4	ADD3	ADD2	ADD1	ADD0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **ADD<7:0>:** Address bits
Received address

TABLE 19-3: REGISTERS ASSOCIATED WITH I^2C™ OPERATION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
SSPBUF	Synchronous Serial Port Receive Buffer/Transmit Register								xxxx xxxx	uuuu uuuu
SSPADD	Synchronous Serial Port (I^2C mode) Address Register								0000 0000	0000 0000
SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
SSPMSK(2)	Synchronous Serial Port (I^2C mode) Address Mask Register								1111 1111	1111 1111
SSPSTAT	SMP(1)	CKE(1)	D/$\overline{A}$	P	S	R/$\overline{W}$	UA	BF	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by SSP module in I^2C mode.

Note 1: Maintain these bits clear in I^2C mode.
2: Accessible only when SSPM<3:0> = 1001.

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20.0 PROGRAM MEMORY READ

The Flash program memory is readable during normal operation over the full VDD range of the device. To read data from Program Memory, five Special Function Registers (SFRs) are used:

- PMCON1
- PMDATL
- PMDATH
- PMADRL
- PMADRH

The value written to the PMADRH:PMADRL register pair determines which program memory location is read. The read operation will be initiated by setting the RD bit of the PMCON1 register. The program memory flash controller takes two instructions to complete the read. As a consequence, after the RD bit has been set, the next two instructions will be ignored. To avoid conflict with program execution, it is recommended that the two instructions following the setting of the RD bit are NOP. When the read completes, the result is placed in the PMDATLH:PMDATL register pair. Refer to Example 20-1 for sample code.

Note:	Code-protect does not effect the CPU from performing a read operation on the program memory. For more information, refer to **Section 8.2 "Code Protection".**

EXAMPLE 20-1: PROGRAM MEMORY READ

```
                BANKSEL  PMADRL          ;
                MOVF     MS_PROG_ADDR, W;
                MOVWF    PMADRH          ;MS Byte of Program Address to read
                MOVF     LS_PROG_ADDR, W;
                MOVWF    PMADRL          ;LS Byte of Program Address to read
                BANKSEL  PMCON1          ;
Required        BSF      PMCON1, RD      ;Initiate Read
Sequence        NOP
                NOP                      ;Any instructions here are ignored as program
                                         ;memory is read in second cycle after BSF
                BANKSEL  PMDATL          ;
                MOVF     PMDATL, W       ;W = LS Byte of Program Memory Read
                MOVWF    LOWPMBYTE       ;
                MOVF     PMDATH, W       ;W = MS Byte of Program Memory Read
                MOVWF    HIGHPMBYTE      ;
```

REGISTER 20-1: PMCON1: PROGRAM MEMORY CONTROL 1 REGISTER

U-1	U-0	U-0	U-0	U-0	U-0	U-0	R/S-0
—	—	—	—	—	—	—	RD
bit 7							bit 0

Legend:		S = Setable bit, cleared in hardware	
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7 **Unimplemented:** Read as ‘1’

bit 6-1 **Unimplemented:** Read as ‘0’

bit 0 **RD:** Read Control bit

1 = Initiates a program memory read (The RD is cleared in hardware; the RD bit can only be set (not cleared) in software).

0 = Does not initiate a program memory read

REGISTER 20-2: PMDATH: PROGRAM MEMORY DATA HIGH REGISTER

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	PMD13	PMD12	PMD11	PMD10	PMD9	PMD8
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Read as ‘0’

bit 5-0 **PMD<13:8>:** The value of the program memory word pointed to by PMADRH and PMADRL after a program memory read command.

REGISTER 20-3: PMDATL: PROGRAM MEMORY DATA LOW REGISTER

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
PMD7	PMD6	PMD5	PMD4	PMD3	PMD2	PMD1	PMD0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-0 **PMD<7:0>:** The value of the program memory word pointed to by PMADRH and PMADRL after a program memory read command.

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REGISTER 20-4: PMADRH: PROGRAM MEMORY ADDRESS HIGH REGISTER

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	—	PMA12	PMA11	PMA10	PMA9	PMA8
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-5 **Unimplemented:** Read as ‘0’

bit 4-0 **PMA<12:8>:** Program Memory Read Address bits

REGISTER 20-5: PMADRL: PROGRAM MEMORY ADDRESS LOW REGISTER

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
PMA7	PMA6	PMA5	PMA4	PMA3	PMA2	PMA1	PMA0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-0 **PMA<7:0>:** Program Memory Read Address bits

TABLE 20-1: SUMMARY OF REGISTERS ASSOCIATED WITH PROGRAM MEMORY READ

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
PMCON1	—	—	—	—	—	—	—	RD	1--- ---0	1--- ---0
PMADRH	—	—	—	Program Memory Read Address Register High Byte					---x xxxx	---x xxxx
PMADRL	Program Memory Read Address Register Low Byte								xxxx xxxx	xxxx xxxx
PMDATH	—	—	Program Memory Read Data Register High Byte						--xx xxxx	--xx xxxx
PMDATL	Program Memory Read Data Register Low Byte								xxxx xxxx	xxxx xxxx

Legend: x = unknown, u = unchanged, – = unimplemented, read as ‘0’. Shaded cells are not used by the program memory read.

NOTES:

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21.0 POWER-DOWN MODE (SLEEP)

The Power-down mode is entered by executing a SLEEP instruction.

Upon entering Sleep mode, the following conditions exist:

1. WDT will be cleared but keeps running, if enabled.
2. $\overline{PD}$ bit of the STATUS register is cleared.
3. $\overline{TO}$ bit of the STATUS register is set.
4. CPU clock is disabled.
5. 31 kHz LFINTOSC is unaffected and peripherals that operate from it may continue operation in Sleep.
6. Timer1/3 oscillator is unaffected and peripherals that operate from it may continue operation in Sleep.
7. ADC is unaffected, if the dedicated FRC clock is selected.
8. Capacitive Sensing oscillators are unaffected.
9. I/O ports maintain the status they had before SLEEP was executed (driving high, low or high-impedance).
10. Resets other than WDT are not affected by Sleep mode.

Refer to individual chapters for more details on peripheral operation during Sleep.

To minimize current consumption, the following conditions should be considered:

- I/O pins should not be floating
- External circuitry sinking current from I/O pins
- Internal circuitry sourcing current from I/O pins
- Current draw from pins with internal weak pull-ups
- Modules using 31 kHz LFINTOSC
- Modules using Timer1/3 oscillator

I/O pins that are high-impedance inputs should be pulled to V_{DD} or V_{SS} externally to avoid switching currents caused by floating inputs.

Examples of internal circuitry that might be sourcing current include modules such as the DAC and FVR modules. See **Section 11.0 “Digital-to-Analog Converter (DAC) Module”** and **Section 10.0 “Fixed Voltage Reference”** for more information on these modules.

21.1 Wake-up from Sleep

The device can wake-up from Sleep through one of the following events:

1. External Reset input on $\overline{MCLR}$ pin, if enabled
2. BOR Reset, if enabled
3. POR Reset
4. Watchdog Timer, if enabled
5. Any external interrupt
6. Interrupts by peripherals capable of running during Sleep (see individual peripheral for more information)

The first three events will cause a device Reset. The last three events are considered a continuation of program execution.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is prefetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be enabled. Wake-up will occur regardless of the state of the GIE bit. If the GIE bit is disabled, the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is enabled, the device executes the instruction after the SLEEP instruction, the device will call the Interrupt Service Routine. In cases where the execution of the instruction following SLEEP is not desirable, the user should have a NOP after the SLEEP instruction.

The WDT is cleared when the device wakes up from Sleep, regardless of the source of wake-up.

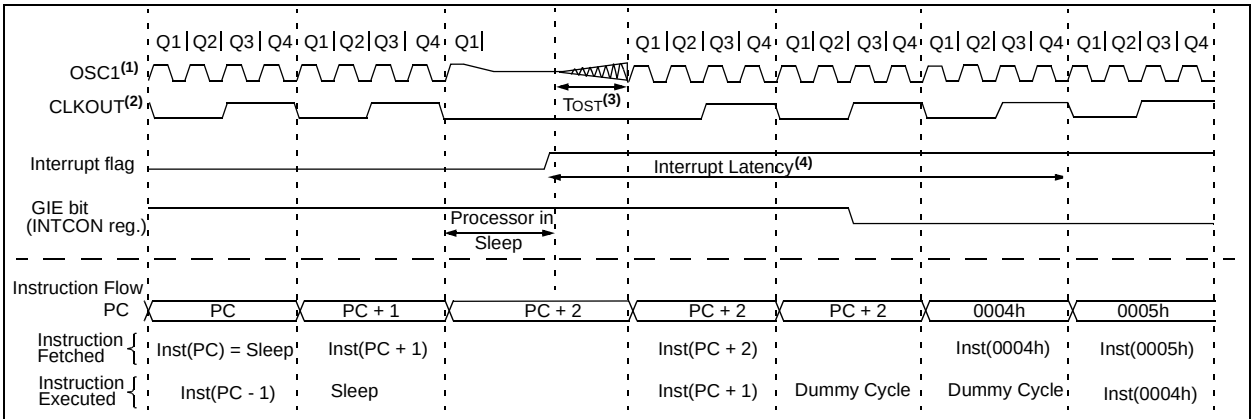
21.1.1 WAKE-UP USING INTERRUPTS

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs **before** the execution of a SLEEP instruction
  - SLEEP instruction will execute as a NOP.
  - WDT and WDT prescaler will not be cleared
  - $\overline{TO}$ bit of the STATUS register will not be set
  - $\overline{PD}$ bit of the STATUS register will not be cleared.
- If the interrupt occurs **during or after** the execution of a SLEEP instruction
  - SLEEP instruction will be completely executed
  - Device will immediately wake-up from Sleep
  - WDT and WDT prescaler will be cleared
  - $\overline{TO}$ bit of the STATUS register will be set
  - $\overline{PD}$ bit of the STATUS register will be cleared.

Even if the flag bits were checked before executing a SLEEP instruction, it may be possible for flag bits to become set before the SLEEP instruction completes. To determine whether a SLEEP instruction executed, test the $\overline{PD}$ bit. If the $\overline{PD}$ bit is set, the SLEEP instruction was executed as a NOP.

FIGURE 21-1: WAKE-UP FROM SLEEP THROUGH INTERRUPT



Note 1: XT, HS or LP Oscillator mode assumed.
2: CLKOUT is not available in XT, HS, or LP Oscillator modes, but shown here for timing reference.
3: TOST = 1024 TOSC (drawing not to scale). This delay applies only to XT, HS or LP Oscillator modes.
4: GIE = 1 assumed. In this case after wake-up, the processor calls the ISR at 0004h. If GIE = 0, execution will continue in-line.

TABLE 21-1: SUMMARY OF REGISTERS ASSOCIATED WITH POWER-DOWN MODE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	0000 000x	0000 000x
IOCBF	IOCBF7	IOCBF6	IOCBF5	IOCBF4	IOCBF3	IOCBF2	IOCBF1	IOCBF0	0000 0000	0000 0000
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIE2	TMR3GIE	TMR3IE	TMRBIE	TMRAIE	—	—	—	CCP2IE	0000 ---0	0000 ---0
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIR2	TMR3GIF	TMR3IF	TMRBIF	TMRAIF	—	—	—	CCP2IF	0000 ---0	0000 ---0
STATUS	IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC	C	0001 1xxx	000q quuu

Legend: — = unimplemented location, read as '0'. Shaded cells are not used in Power-Down mode.

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22.0 IN-CIRCUIT SERIAL PROGRAMMING™ (ICSP™)

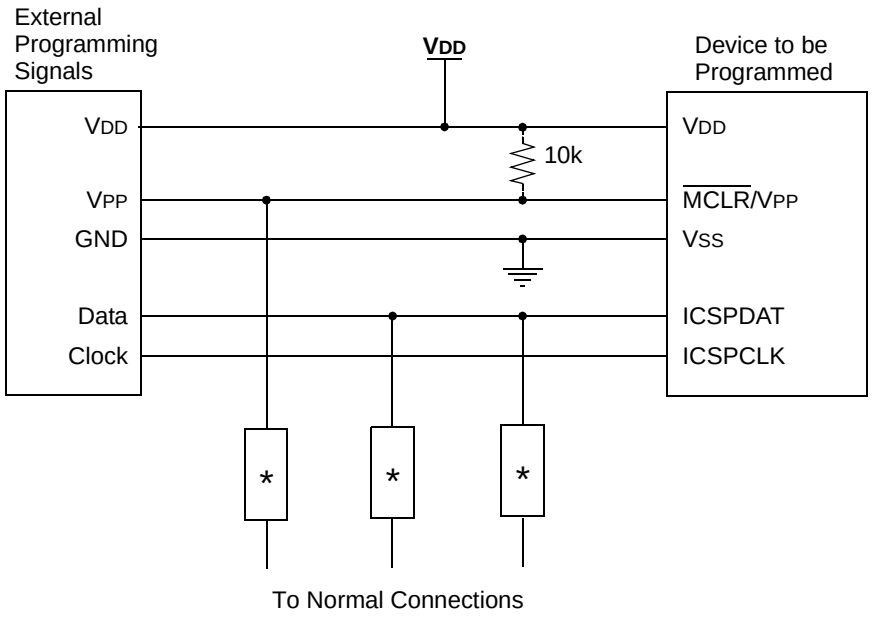
ICSP™ programming allows customers to manufacture circuit boards with unprogrammed devices. Programming can be done after the assembly process allowing the device to be programmed with the most recent firmware or a custom firmware. Five pins are needed for ICSP™ programming:

- ICSPCLK
- ICSPDAT
- MCLR/VPP
- VDD
- VSS

The device is placed into Program/Verify mode by holding the ICSPCLK and ICSPDAT pins low then raising the voltage on MCLR/VPP from 0v to VPP. In Program/Verify mode the program memory, user IDs and the Configuration Words are programmed through serial communications. The ICSPDAT pin is a bidirectional I/O used for transferring the serial data and the ISCPCLK pin is the clock input. For more information on ICSP™ refer to the *"PIC16F707/PIC16LF707 Programming Specification"* (DS41405A).

Note:	The ICD 2 produces a VPP voltage greater than the maximum VPP specification of the PIC16(L)F707. When using this programmer, an external circuit, such as the AC164112 MPLAB® ICD 2 VPP voltage limiter, is required to keep the VPP voltage within the device specifications.

FIGURE 22-1: TYPICAL CONNECTION FOR ICSP™ PROGRAMMING



* Isolation devices (as required).

NOTES:

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23.0 INSTRUCTION SET SUMMARY

The PIC16(L)F707 instruction set is highly orthogonal and is comprised of three basic categories:

- **Byte-oriented** operations
- **Bit-oriented** operations
- **Literal and control** operations

Each PIC16 instruction is a 14-bit word divided into an **opcode,** which specifies the instruction type and one or more **operands,** which further specify the operation of the instruction. The formats for each of the categories is presented in Figure 23-1, while the various opcode fields are summarized in Table 23-1.

Table 23-2 lists the instructions recognized by the MPASM™ assembler.

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator, which selects the bit affected by the operation, while 'f' represents the address of the file in which the bit is located.

For **literal and control** operations, 'k' represents an 8-bit or 11-bit constant, or literal value.

One instruction cycle consists of four oscillator periods; for an oscillator frequency of 4 MHz, this gives a nominal instruction execution time of 1 µs. All instructions are executed within a single instruction cycle, unless a conditional test is true, or the program counter is changed as a result of an instruction. When this occurs, the execution takes two instruction cycles, with the second cycle executed as a NOP.

All instruction examples use the format '0xhh' to represent a hexadecimal number, where 'h' signifies a hexadecimal digit.

23.1 Read-Modify-Write Operations

Any instruction that specifies a file register as part of the instruction performs a Read-Modify-Write (R-M-W) operation. The register is read, the data is modified, and the result is stored according to either the instruction, or the destination designator 'd'. A read operation is performed on a register even if the instruction writes to that register.

For example, a CLRF PORTB instruction will read PORTB, clear all the data bits, then write the result back to PORTB. This example would have the unintended consequence of clearing the condition that set the RBIF flag.

TABLE 23-1: OPCODE FIELD DESCRIPTIONS

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= 0 or 1). The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; d = 0: store result in W, d = 1: store result in file register f. Default is d = 1.
PC	Program Counter
$\overline{TO}$	Time-out bit
C	Carry bit
DC	Digit carry bit
Z	Zero bit
$\overline{PD}$	Power-down bit

FIGURE 23-1: GENERAL FORMAT FOR INSTRUCTIONS

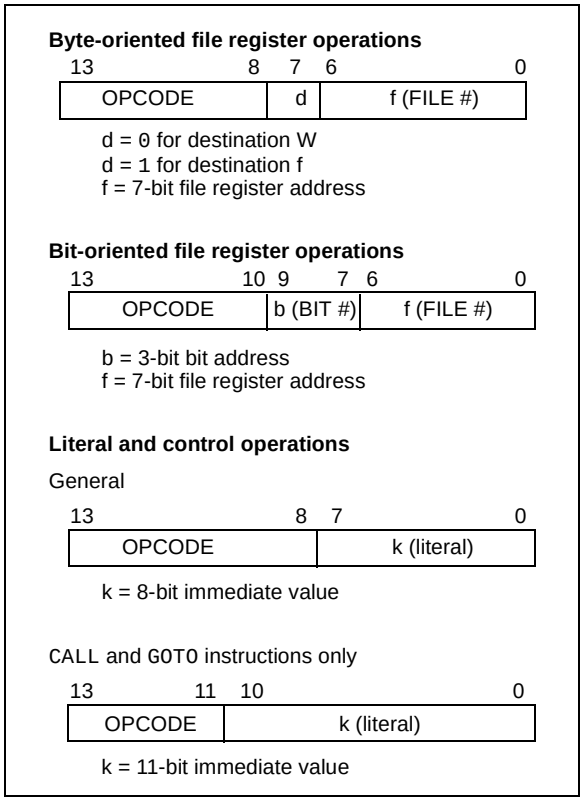

TABLE 23-2: PIC16(L)F707 INSTRUCTION SET

Mnemonic, Operands		Description	Cycles	14-Bit Opcode MSb			LSb	Status Affected	Notes
BYTE-ORIENTED FILE REGISTER OPERATIONS									
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C, DC, Z	**1, 2**
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	**1, 2**
CLRF	f	Clear f	1	00	0001	lfff	ffff	Z	**2**
CLRW	–	Clear W	1	00	0001	0xxx	xxxx	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	**1, 2**
DECF	f, d	Decrement f	1	00	0011	dfff	ffff	Z	**1, 2**
DECFSZ	f, d	Decrement f, Skip if 0	1**(2)**	00	1011	dfff	ffff		**1, 2, 3**
INCF	f, d	Increment f	1	00	1010	dfff	ffff	Z	**1, 2**
INCFSZ	f, d	Increment f, Skip if 0	1**(2)**	00	1111	dfff	ffff		**1, 2, 3**
IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	**1, 2**
MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	**1, 2**
MOVWF	f	Move W to f	1	00	0000	lfff	ffff		
NOP	–	No Operation	1	00	0000	0xx0	0000		
RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	C	**1, 2**
RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	C	**1, 2**
SUBWF	f, d	Subtract W from f	1	00	0010	dfff	ffff	C, DC, Z	**1, 2**
SWAPF	f, d	Swap nibbles in f	1	00	1110	dfff	ffff		**1, 2**
XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	**1, 2**
BIT-ORIENTED FILE REGISTER OPERATIONS									
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff		**1, 2**
BSF	f, b	Bit Set f	1	01	01bb	bfff	ffff		**1, 2**
BTFSC	f, b	Bit Test f, Skip if Clear	1 **(2)**	01	10bb	bfff	ffff		**3**
BTFSS	f, b	Bit Test f, Skip if Set	1 **(2)**	01	11bb	bfff	ffff		**3**
LITERAL AND CONTROL OPERATIONS									
ADDLW	k	Add literal and W	1	11	111x	kkkk	kkkk	C, DC, Z	
ANDLW	k	AND literal with W	1	11	1001	kkkk	kkkk	Z	
CALL	k	Call Subroutine	2	10	0kkk	kkkk	kkkk		
CLRWDT	–	Clear Watchdog Timer	1	00	0000	0110	0100	$\overline{TO}$, $\overline{PD}$	
GOTO	k	Go to address	2	10	1kkk	kkkk	kkkk		
IORLW	k	Inclusive OR literal with W	1	11	1000	kkkk	kkkk	Z	
MOVLW	k	Move literal to W	1	11	00xx	kkkk	kkkk		
RETFIE	–	Return from interrupt	2	00	0000	0000	1001		
RETLW	k	Return with literal in W	2	11	01xx	kkkk	kkkk		
RETURN	–	Return from Subroutine	2	00	0000	0000	1000		
SLEEP	–	Go into Standby mode	1	00	0000	0110	0011	$\overline{TO}$, $\overline{PD}$	
SUBLW	k	Subtract W from literal	1	11	110x	kkkk	kkkk	C, DC, Z	
XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z	

Note 1: When an I/O register is modified as a function of itself (e.g., `MOVF PORTA, 1`), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 module.

3: If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a `NOP`.

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23.2 Instruction Descriptions

ADDLW	Add literal and W
Syntax:	[*label*] ADDLW k
Operands:	$0 \le k \le 255$
Operation:	(W) + k → (W)
Status Affected:	C, DC, Z
Description:	The contents of the W register are added to the eight-bit literal 'k' and the result is placed in the W register.

ADDWF	Add W and f
Syntax:	[*label*] ADDWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) + (f) → (destination)
Status Affected:	C, DC, Z
Description:	Add the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

ANDLW	AND literal with W
Syntax:	[*label*] ANDLW k
Operands:	$0 \le k \le 255$
Operation:	(W) .AND. (k) → (W)
Status Affected:	Z
Description:	The contents of W register are AND'ed with the eight-bit literal 'k'. The result is placed in the W register.

ANDWF	AND W with f
Syntax:	[*label*] ANDWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) .AND. (f) → (destination)
Status Affected:	Z
Description:	AND the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

BCF	Bit Clear f
Syntax:	[*label*] BCF f,b
Operands:	$0 \le f \le 127$ $0 \le b \le 7$
Operation:	0 → (f<b>)
Status Affected:	None
Description:	Bit 'b' in register 'f' is cleared.

BSF	Bit Set f
Syntax:	[*label*] BSF f,b
Operands:	$0 \le f \le 127$ $0 \le b \le 7$
Operation:	1 → (f<b>)
Status Affected:	None
Description:	Bit 'b' in register 'f' is set.

BTFSC	Bit Test f, Skip if Clear
Syntax:	[*label*] BTFSC f,b
Operands:	$0 \le f \le 127$ $0 \le b \le 7$
Operation:	skip if (f<b>) = 0
Status Affected:	None
Description:	If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b', in register 'f', is '0', the next instruction is discarded, and a NOP is executed instead, making this a 2-cycle instruction.

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BTFSS **Bit Test f, Skip if Set**

Syntax:	[*label*] BTFSS f,b
Operands:	$0 \le f \le 127$ $0 \le b < 7$
Operation:	skip if (f<b>) = 1
Status Affected:	None
Description:	If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.

CALL **Call Subroutine**

Syntax:	[*label*] CALL k
Operands:	$0 \le k \le 2047$
Operation:	(PC)+ 1→ TOS, k → PC<10:0>, (PCLATH<4:3>) → PC<12:11>
Status Affected:	None
Description:	Call Subroutine. First, return address (PC + 1) is pushed onto the stack. The eleven-bit immediate address is loaded into PC bits <10:0>. The upper bits of the PC are loaded from PCLATH. CALL is a two-cycle instruction.

CLRF **Clear f**

Syntax:	[*label*] CLRF f
Operands:	$0 \le f \le 127$
Operation:	00h → (f) 1 → Z
Status Affected:	Z
Description:	The contents of register 'f' are cleared and the Z bit is set.

CLRW **Clear W**

Syntax:	[*label*] CLRW
Operands:	None
Operation:	00h → (W) 1 → Z
Status Affected:	Z
Description:	W register is cleared. Zero bit (Z) is set.

CLRWDT **Clear Watchdog Timer**

Syntax:	[*label*] CLRWDT
Operands:	None
Operation:	00h → WDT 0 → WDT prescaler, 1 → $\overline{TO}$ 1 → $\overline{PD}$
Status Affected:	$\overline{TO}$, $\overline{PD}$
Description:	CLRWDT instruction resets the Watchdog Timer. It also resets the prescaler of the WDT. Status bits $\overline{TO}$ and $\overline{PD}$ are set.

COMF **Complement f**

Syntax:	[*label*] COMF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	$(\bar{f}) \rightarrow$ (destination)
Status Affected:	Z
Description:	The contents of register 'f' are complemented. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

DECF **Decrement f**

Syntax:	[*label*] DECF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(f) - 1 → (destination)
Status Affected:	Z
Description:	Decrement register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

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DECFSZ	Decrement f, Skip if 0
Syntax:	[*label*] DECFSZ f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(f) - 1 → (destination); skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are decremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', then a NOP is executed instead, making it a 2-cycle instruction.

GOTO	Unconditional Branch
Syntax:	[*label*] GOTO k
Operands:	$0 \le k \le 2047$
Operation:	k → PC<10:0> PCLATH<4:3> → PC<12:11>
Status Affected:	None
Description:	GOTO is an unconditional branch. The eleven-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<4:3>. GOTO is a two-cycle instruction.

INCF	Increment f
Syntax:	[*label*] INCF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(f) + 1 → (destination)
Status Affected:	Z
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

INCFSZ	Increment f, Skip if 0
Syntax:	[*label*] INCFSZ f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(f) + 1 → (destination), skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a 2-cycle instruction.

IORLW	Inclusive OR literal with W
Syntax:	[*label*] IORLW k
Operands:	$0 \le k \le 255$
Operation:	(W) .OR. k → (W)
Status Affected:	Z
Description:	The contents of the W register are OR'ed with the eight-bit literal 'k'. The result is placed in the W register.

IORWF	Inclusive OR W with f
Syntax:	[*label*] IORWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) .OR. (f) → (destination)
Status Affected:	Z
Description:	Inclusive OR the W register with register 'f'. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

MOVF	Move f
Syntax:	[*label*] MOVF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(f) → (dest)
Status Affected:	Z
Description:	The contents of register f is moved to a destination dependent upon the status of d. If d = 0, destination is W register. If d = 1, the destination is file register f itself. d = 1 is useful to test a file register since status flag Z is affected.
Words:	1
Cycles:	1
Example:	MOVF FSR, 0 After Instruction W = value in FSR register Z = 1

MOVWF	Move W to f
Syntax:	[*label*] MOVWF f
Operands:	$0 \le f \le 127$
Operation:	(W) → (f)
Status Affected:	None
Description:	Move data from W register to register 'f'.
Words:	1
Cycles:	1
Example:	MOVW OPTION F Before Instruction OPTION = 0xFF W = 0x4F After Instruction OPTION = 0x4F W = 0x4F

MOVLW	Move literal to W
Syntax:	[*label*] MOVLW k
Operands:	$0 \le k \le 255$
Operation:	k → (W)
Status Affected:	None
Description:	The eight-bit literal 'k' is loaded into W register. The "don't cares" will assemble as '0's.
Words:	1
Cycles:	1
Example:	MOVLW 0x5A After Instruction W = 0x5A

NOP	No Operation
Syntax:	[*label*] NOP
Operands:	None
Operation:	No operation
Status Affected:	None
Description:	No operation.
Words:	1
Cycles:	1
Example:	NOP

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RETFIE	Return from Interrupt
Syntax:	[*label*] RETFIE
Operands:	None
Operation:	TOS → PC, 1 → GIE
Status Affected:	None
Description:	Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE (INTCON<7>). This is a two-cycle instruction.
Words:	1
Cycles:	2
Example:	`RETFIE` After Interrupt PC = TOS GIE = 1

RETLW	Return with literal in W
Syntax:	[*label*] RETLW k
Operands:	0 ≤ k ≤ 255
Operation:	k → (W); TOS → PC
Status Affected:	None
Description:	The W register is loaded with the eight bit literal ‘k’. The program counter is loaded from the top of the stack (the return address). This is a two-cycle instruction.
Words:	1
Cycles:	2

Example:	`CALL TABLE;W contains table`
	`;offset value`
	`TABLE ;W now has table value`
	`•`
	`•`
	`•`
	`ADDWF PC ;W = offset`
	`RETLW k1 ;Begin table`
	`RETLW k2 ;`
	`•`
	`•`
	`•`
	`RETLW kn ; End of table`
	
	Before Instruction
	W = 0x07
	After Instruction
	W = value of k8

RETURN	Return from Subroutine
Syntax:	[*label*] RETURN
Operands:	None
Operation:	TOS → PC
Status Affected:	None
Description:	Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a two-cycle instruction.

RLF **Rotate Left f through Carry**

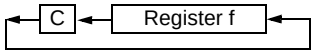
Syntax: [*label*] RLF f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: See description below

Status Affected: C

Description: The contents of register 'f' are rotated one bit to the left through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is stored back in register 'f'.



Words: 1

Cycles: 1

Example:

```
RLF     REG1,0
```

Before Instruction
```
REG1    =   1110 0110
C       =   0
```
After Instruction
```
REG1    =   1110 0110
W       =   1100 1100
C       =   1
```

RRF **Rotate Right f through Carry**

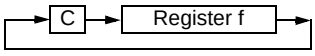
Syntax: [*label*] RRF f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: See description below

Status Affected: C

Description: The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.



SLEEP **Enter Sleep mode**

Syntax: [*label*] SLEEP

Operands: None

Operation: 00h → WDT,
0 → WDT prescaler,
1 → $\overline{TO}$,
0 → $\overline{PD}$

Status Affected: $\overline{TO}$, $\overline{PD}$

Description: The power-down Status bit, $\overline{PD}$ is cleared. Time-out Status bit, $\overline{TO}$ is set. Watchdog Timer and its prescaler are cleared.
The processor is put into Sleep mode with the oscillator stopped.

SUBLW **Subtract W from literal**

Syntax: [*label*] SUBLW k

Operands: $0 \le k \le 255$

Operation: k - (W) → (W)

Status Affected: C, DC, Z

Description: The W register is subtracted (2's complement method) from the eight-bit literal 'k'. The result is placed in the W register.

C = 0	W > k
C = 1	W ≤ k
DC = 0	W<3:0> > k<3:0>
DC = 1	W<3:0> ≤ k<3:0>

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SUBWF **Subtract W from f**

Syntax: [*label*] SUBWF f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: (f) - (W) → (destination)

Status Affected: C, DC, Z

Description: Subtract (2's complement method) W register from register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f.

C = 0	W > f
C = 1	W ≤ f
DC = 0	W<3:0> > f<3:0>
DC = 1	W<3:0> ≤ f<3:0>

SWAPF **Swap Nibbles in f**

Syntax: [*label*] SWAPF f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: (f<3:0>) → (destination<7:4>),
(f<7:4>) → (destination<3:0>)

Status Affected: None

Description: The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed in register 'f'.

XORLW **Exclusive OR literal with W**

Syntax: [*label*] XORLW k

Operands: $0 \le k \le 255$

Operation: (W) .XOR. k → (W)

Status Affected: Z

Description: The contents of the W register are XOR'ed with the eight-bit literal 'k'. The result is placed in the W register.

XORWF **Exclusive OR W with f**

Syntax: [*label*] XORWF f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: (W) .XOR. (f) → (destination)

Status Affected: Z

Description: Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

NOTES:

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24.0 DEVELOPMENT SUPPORT

The PIC® microcontrollers and dsPIC® digital signal controllers are supported with a full range of software and hardware development tools:

- Integrated Development Environment
  - MPLAB® IDE Software
- Compilers/Assemblers/Linkers
  - MPLAB C Compiler for Various Device Families
  - HI-TECH C for Various Device Families
  - MPASM™ Assembler
  - MPLINK™ Object Linker/ MPLIB™ Object Librarian
  - MPLAB Assembler/Linker/Librarian for Various Device Families
- Simulators
  - MPLAB SIM Software Simulator
- Emulators
  - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers
  - MPLAB ICD 3
  - PICkit™ 3 Debug Express
- Device Programmers
  - PICkit™ 2 Programmer
  - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards, Evaluation Kits, and Starter Kits

24.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16/32-bit microcontroller market. The MPLAB IDE is a Windows® operating system-based application that contains:

- A single graphical interface to all debugging tools
  - Simulator
  - Programmer (sold separately)
  - In-Circuit Emulator (sold separately)
  - In-Circuit Debugger (sold separately)
- A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High-level source code debugging
- Mouse over variable inspection
- Drag and drop variables from source to watch windows
- Extensive on-line help
- Integration of select third party tools, such as IAR C Compilers

The MPLAB IDE allows you to:

- Edit your source files (either C or assembly)
- One-touch compile or assemble, and download to emulator and simulator tools (automatically updates all project information)
- Debug using:
  - Source files (C or assembly)
  - Mixed C and assembly
  - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

24.2 MPLAB C Compilers for Various Device Families

The MPLAB C Compiler code development systems are complete ANSI C compilers for Microchip's PIC18, PIC24 and PIC32 families of microcontrollers and the dsPIC30 and dsPIC33 families of digital signal controllers. These compilers provide powerful integration capabilities, superior code optimization and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

24.3 HI-TECH C for Various Device Families

The HI-TECH C Compiler code development systems are complete ANSI C compilers for Microchip's PIC family of microcontrollers and the dsPIC family of digital signal controllers. These compilers provide powerful integration capabilities, omniscient code generation and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

The compilers include a macro assembler, linker, preprocessor, and one-step driver, and can run on multiple platforms.

24.4 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel® standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

24.5 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler and the MPLAB C18 C Compiler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

24.6 MPLAB Assembler, Linker and Librarian for Various Device Families

MPLAB Assembler produces relocatable machine code from symbolic assembly language for PIC24, PIC32 and dsPIC devices. MPLAB C Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- Support for the entire device instruction set
- Support for fixed-point and floating-point data
- Command line interface
- Rich directive set
- Flexible macro language
- MPLAB IDE compatibility

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24.7 MPLAB SIM Software Simulator

The MPLAB SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC® DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB SIM Software Simulator fully supports symbolic debugging using the MPLAB C Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

24.8 MPLAB REAL ICE In-Circuit Emulator System

MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs PIC® Flash MCUs and dsPIC® Flash DSCs with the easy-to-use, powerful graphical user interface of the MPLAB Integrated Development Environment (IDE), included with each kit.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with in-circuit debugger systems (RJ11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB IDE. In upcoming releases of MPLAB IDE, new devices will be supported, and new features will be added. MPLAB REAL ICE offers significant advantages over competitive emulators including low-cost, full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, a ruggedized probe interface and long (up to three meters) interconnection cables.

24.9 MPLAB ICD 3 In-Circuit Debugger System

MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost effective high-speed hardware debugger/programmer for Microchip Flash Digital Signal Controller (DSC) and microcontroller (MCU) devices. It debugs and programs PIC® Flash microcontrollers and dsPIC® DSCs with the powerful, yet easy-to-use graphical user interface of MPLAB Integrated Development Environment (IDE).

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

24.10 PICkit 3 In-Circuit Debugger/ Programmer and PICkit 3 Debug Express

The MPLAB PICkit 3 allows debugging and programming of PIC® and dsPIC® Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB Integrated Development Environment (IDE). The MPLAB PICkit 3 is connected to the design engineer's PC using a full speed USB interface and can be connected to the target via an Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the reset line to implement in-circuit debugging and In-Circuit Serial Programming™.

The PICkit 3 Debug Express include the PICkit 3, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

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24.11 PICkit 2 Development Programmer/Debugger and PICkit 2 Debug Express

The PICkit™ 2 Development Programmer/Debugger is a low-cost development tool with an easy to use interface for programming and debugging Microchip's Flash families of microcontrollers. The full featured Windows® programming interface supports baseline (PIC10F, PIC12F5xx, PIC16F5xx), midrange (PIC12F6xx, PIC16F), PIC18F, PIC24, dsPIC30, dsPIC33, and PIC32 families of 8-bit, 16-bit, and 32-bit microcontrollers, and many Microchip Serial EEPROM products. With Microchip's powerful MPLAB Integrated Development Environment (IDE) the PICkit™ 2 enables in-circuit debugging on most PIC® microcontrollers. In-Circuit-Debugging runs, halts and single steps the program while the PIC microcontroller is embedded in the application. When halted at a breakpoint, the file registers can be examined and modified.

The PICkit 2 Debug Express include the PICkit 2, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

24.12 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at V_{DDMIN} and V_{DDMAX} for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages and a modular, detachable socket assembly to support various package types. The ICSP™ cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an MMC card for file storage and data applications.

24.13 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM™ and dsPICDEM™ demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ® security ICs, CAN, IrDA®, PowerSmart battery management, SEEVAL® evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

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25.0 ELECTRICAL SPECIFICATIONS

Absolute Maximum Ratings(†)

Ambient temperature under bias ..... -40°C to +125°C
Storage temperature ..... -65°C to +150°C
Voltage on V_{DD} with respect to V_{SS}, PIC16F707 ..... -0.3V to +6.5V
Voltage on V_{CAP} pin with respect to V_{SS}, PIC16F707 ..... -0.3V to +4.0V
Voltage on V_{DD} with respect to V_{SS}, PIC16LF707 ..... -0.3V to +4.0V
Voltage on $\overline{MCLR}$ with respect to Vss ..... -0.3V to +9.0V
Voltage on all other pins with respect to V_{SS} ..... -0.3V to (V_{DD} + 0.3V)
Total power dissipation(1) ..... 800 mW
Maximum current out of V_{SS} pin ..... 95 mA
Maximum current into V_{DD} pin ..... 70 mA
Clamp current, I_K (V_{PIN} < 0 or V_{PIN} > V_{DD}) ..... ± 20 mA
Maximum output current sunk by any I/O pin ..... 25 mA
Maximum output current sourced by any I/O pin ..... 25 mA
Maximum current sunk by all ports(2), -40°C ≤ T_A ≤ +85°C for industrial ..... 200 mA
Maximum current sunk by all ports(2), -40°C ≤ T_A ≤ +125°C for extended ..... 90 mA
Maximum current sourced by all ports(2), 40°C ≤ T_A ≤ +85°C for industrial ..... 140 mA
Maximum current sourced by all ports(2), -40°C ≤ T_A ≤ +125°C for extended ..... 65 mA

Note 1: Power dissipation is calculated as follows: $P_{DIS} = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD} - V_{OH}) \times I_{OH}\} + \sum(V_{OL} \times I_{OL})$.

† NOTICE: Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure above maximum rating conditions for extended periods may affect device reliability.

25.1 DC Characteristics: PIC16(L)F707-I/E (Industrial, Extended)

PIC16LF707			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
PIC16F707			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
Param. No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
D001	VDD	**Supply Voltage**					
		PIC16LF707	1.8 1.8 2.3 2.5	— — — —	3.6 3.6 3.6 3.6	V V V V	FOSC ≤ 16 MHz: HFINTOSC, EC FOSC ≤ 4 MHz FOSC ≤ 20 MHz, EC FOSC ≤ 20 MHz, HS
D001		**PIC16F707**	1.8 1.8 2.3 2.5	— — — —	5.5 5.5 5.5 5.5	V V V V	FOSC ≤ 16 MHz: HFINTOSC, EC FOSC ≤ 4 MHz FOSC ≤ 20 MHz, EC FOSC ≤ 20 MHz, HS
D002*	VDR	**RAM Data Retention Voltage(1)**					
		PIC16LF707	1.5	—	—	V	Device in Sleep mode
D002*		**PIC16F707**	1.7	—	—	V	Device in Sleep mode
	VPOR*	**Power-on Reset Release Voltage**	—	1.6	—	V	
	VPORR*	**Power-on Reset Rearm Voltage**					
		PIC16LF707	—	0.8	—	V	Device in Sleep mode
		PIC16F707	—	1.7	—	V	Device in Sleep mode
D003	VFVR	**Fixed Voltage Reference Voltage, Initial Accuracy**	-5.5 -5.5 -5.5	— — —	5.5 5.5 5.5	% % %	VFVR = 1.024V, VDD ≥ 2.5V VFVR = 2.048V, VDD ≥ 2.5V VFVR = 4.096V, VDD ≥ 4.75V; -40 ≤ TA ≤ 85°C
			-6 -6 -6	— — —	6 6 6	% % %	VFVR = 1.024V, VDD ≥ 2.5V VFVR = 2.048V, VDD ≥ 2.5V VFVR = 4.096V, VDD ≥ 4.75V; -40 ≤ TA ≤ 125°C
D004*	SVDD	**VDD Rise Rate** to ensure internal Power-on Reset signal	0.05	—	—	V/ms	See **Section 3.2 "Power-on Reset (POR)"** for details.

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in Sleep mode without losing RAM data.

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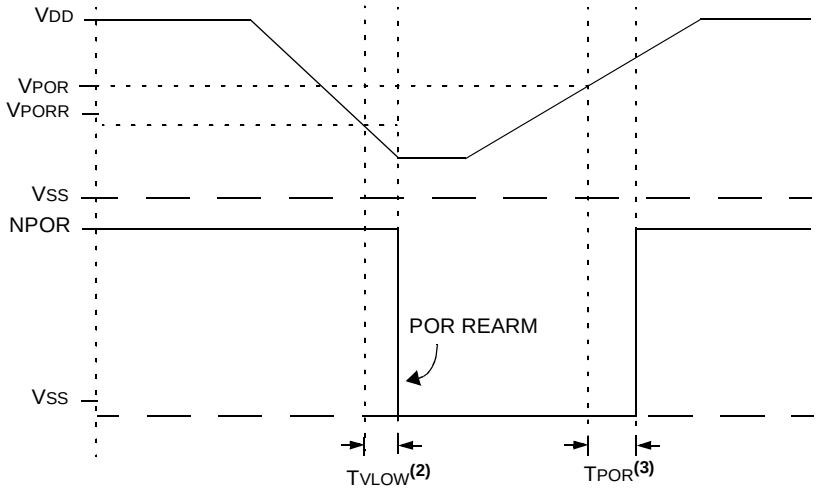
FIGURE 25-1: POR AND POR REARM WITH SLOW RISING VDD



Note 1: When NPOR is low, the device is held in Reset.
2: TPOR 1 μs typical.
3: TVLOW 2.7 μs typical.

25.2 DC Characteristics: PIC16(L)F707-I/E (Industrial, Extended)

PIC16LF707			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
PIC16F707			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
Param No.	**Device Characteristics**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**	
						VDD	**Note**
	Supply Current (IDD)(1, 2)						
D009	LDO Regulator	—	350	—	µA	—	HS, EC OR INTOSC/INTOSCIO (8-16 MHz) Clock modes with all VCAP pins disabled
		—	50	—	µA	—	All VCAP pins disabled
		—	30	—	µA	—	VCAP enabled on RA0, RA5 or RA6
		—	5	—	µA	—	LP Clock mode and Sleep (requires FVR and BOR to be disabled)
D010		—	7.0	12	µA	1.8	FOSC = 32 kHz LP Oscillator mode **(Note 4)**, -40°C ≤ TA ≤ +85°C
		—	9.0	14	µA	3.0	
D010		—	11	20	µA	1.8	FOSC = 32 kHz LP Oscillator mode **(Note 4)**, -40°C ≤ TA ≤ +85°C
		—	14	22	µA	3.0	
		—	15	24	µA	5.0	
D011		—	7.0	12	µA	1.8	FOSC = 32 kHz LP Oscillator mode -40°C ≤ TA ≤ +125°C
		—	9.0	18	µA	3.0	
D011		—	11	21	µA	1.8	FOSC = 32 kHz LP Oscillator mode **(Note 4)** -40°C ≤ TA ≤ +125°C
		—	14	25	µA	3.0	
		—	15	27	µA	5.0	
D011		—	110	150	µA	1.8	FOSC = 1 MHz XT Oscillator mode
		—	150	215	µA	3.0	
D011		—	120	175	µA	1.8	FOSC = 1 MHz XT Oscillator mode **(Note 5)**
		—	180	250	µA	3.0	
		—	240	300	µA	5.0	
D012		—	230	300	µA	1.8	FOSC = 4 MHz XT Oscillator mode
		—	400	600	µA	3.0	
D012		—	250	350	µA	1.8	FOSC = 4 MHz XT Oscillator mode **(Note 5)**
		—	420	650	µA	3.0	
		—	500	750	µA	5.0	
D013		—	125	180	µA	1.8	FOSC = 1 MHz EC Oscillator mode
		—	230	270	µA	3.0	
D013		—	150	205	µA	1.8	FOSC = 1 MHz EC Oscillator mode **(Note 5)**
		—	225	320	µA	3.0	
		—	250	410	µA	5.0	

Note 1: The test conditions for all IDD measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; $\overline{MCLR}$ = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be extended by the formula IR = VDD/2REXT (mA) with REXT in kΩ.

4: FVR and BOR are disabled.

5: 0.1 µF capacitor on VCAP (RA0).

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25.2 DC Characteristics: PIC16(L)F707-I/E (Industrial, Extended) (Continued)

PIC16LF707			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
PIC16F707			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
Param No.	**Device Characteristics**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**	
						VDD	**Note**
	Supply Current (IDD)(1, 2)						
D014		—	290	330	µA	1.8	FOSC = 4 MHz EC Oscillator mode
		—	460	500	µA	3.0	
D014		—	300	430	µA	1.8	FOSC = 4 MHz EC Oscillator mode **(Note 5)**
		—	450	655	µA	3.0	
		—	500	730	µA	5.0	
D015		—	100	130	µA	1.8	FOSC = 500 kHz MFINTOSC mode
		—	120	150	µA	3.0	
D015		—	115	195	µA	1.8	FOSC = 500 kHz MFINTOSC mode **(Note 5)**
		—	135	200	µA	3.0	
		—	150	220	µA	5.0	
D016		—	650	800	µA	1.8	FOSC = 8 MHz HFINTOSC mode
		—	1000	1200	µA	3.0	
D016		—	625	850	µA	1.8	FOSC = 8 MHz HFINTOSC mode **(Note 5)**
		—	1000	1200	µA	3.0	
		—	1100	1500	µA	5.0	
D017		—	1.0	1.2	mA	1.8	FOSC = 16 MHz HFINTOSC mode
		—	1.5	1.85	mA	3.0	
D017		—	1	1.2	mA	1.8	FOSC = 16 MHz HFINTOSC mode **(Note 5)**
		—	1.5	1.7	mA	3.0	
		—	1.7	2.1	mA	5.0	
D018		—	210	240	µA	1.8	FOSC = 4 MHz EXTRC mode **(Note 3, Note 5)**
		—	340	380	µA	3.0	
D018		—	225	320	µA	1.8	FOSC = 4 MHz EXTRC mode **(Note 3, Note 5)**
		—	360	445	µA	3.0	
		—	410	650	µA	5.0	
D019		—	1.6	1.9	mA	3.0	FOSC = 20 MHz HS Oscillator mode
		—	2.0	2.8	mA	3.6	
D019		—	1.6	2	mA	3.0	FOSC = 20 MHz HS Oscillator mode **(Note 5)**
		—	1.9	3.2	mA	5.0	

Note 1: The test conditions for all IDD measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; $\overline{MCLR}$ = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be extended by the formula IR = VDD/2REXT (mA) with REXT in kΩ.

4: FVR and BOR are disabled.

5: 0.1 µF capacitor on VCAP (RA0).

25.3 DC Characteristics: PIC16(L)F707-I/E (Power-Down)

PIC16LF707			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended					
PIC16F707			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended					
Param No.	**Device Characteristics**	**Min.**	**Typ†**	**Max. +85°C**	**Max. +125°C**	**Units**	**Conditions**	
							VDD	**Note**
	Power-down Base Current (IPD)(2)							
D020		—	0.02	0.7	3.9	µA	1.8	WDT, BOR, FVR, and T1OSC disabled, all Peripherals Inactive
		—	0.08	1.0	4.3	µA	3.0	
D020		—	4.3	10.2	17	µA	1.8	WDT, BOR, FVR, and T1OSC disabled, all Peripherals Inactive
		—	5	10.5	18	µA	3.0	
		—	5.5	11.8	21	µA	5.0	
D021		—	0.5	1.7	4.1	µA	1.8	LPWDT Current **(Note 1)**
		—	0.8	2.5	4.8	µA	3.0	
D021		—	6	13.5	16.4	µA	1.8	LPWDT Current **(Note 1)**
		—	6.5	14.5	16.8	µA	3.0	
		—	7.5	16	18.7	µA	5.0	
D021A		—	8.5	18	22	µA	1.8	FVR current **(Note 1, Note 3)**
		—	8.5	18	22	µA	3.0	
D021A		—	23	44	48	µA	1.8	FVR current **(Note 1, Note 3, Note 5)**
		—	25	45	55	µA	3.0	
		—	26	60	70	µA	5.0	
D022		—	—	—	—	µA	1.8	BOR Current **(Note 1, Note 3)**
		—	7.5	12	22	µA	3.0	
D022		—	—	—	—	µA	1.8	BOR Current **(Note 1, Note 3, Note 5)**
		—	23	42	49	µA	3.0	
		—	25	46	50	µA	5.0	
D026		—	0.6	3	7	µA	1.8	T1OSC Current **(Note 1)**
		—	1.8	6	8.75	µA	3.0	
D026		—	4.5	11.1	—	µA	1.8	T1OSC Current **(Note 1)**
		—	6	12.5	—	µA	3.0	
		—	7	13.5	—	µA	5.0	

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD.

3: Fixed Voltage Reference is automatically enabled whenever the BOR is enabled.

4: A/D oscillator source is FRC.

5: 0.1 µF capacitor on VCAP (RA0).

6: Includes FVR IPD and DAC IPD.

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25.3 DC Characteristics: PIC16(L)F707-I/E (Power-Down) (Continued)

PIC16LF707			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended					
PIC16F707			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended					
Param No.	**Device Characteristics**	**Min.**	**Typ†**	**Max. +85°C**	**Max. +125°C**	**Units**	**Conditions**	
							VDD	**Note**
	Power-down Base Current (IPD)(2)							
D027		—	0.06	0.7	5.0	µA	1.8	A/D Current **(Note 1, Note 4)**, no conversion in progress
		—	0.08	1.0	5.5	µA	3.0	
D027		—	6	10.7	18	µA	1.8	A/D Current **(Note 1, Note 4)**, no conversion in progress
		—	7	10.6	20	µA	3.0	
		—	7.2	11.9	22	µA	5.0	
D027A		—	250	400	—	µA	1.8	A/D Current **(Note 1, Note 4)**, conversion in progress
		—	250	400	—	µA	3.0	
D027A		—	280	430	—	µA	1.8	A/D Current **(Note 1, Note 4, Note 5)**, conversion in progress
		—	280	430	—	µA	3.0	
		—	280	430	—	µA	5.0	
D028		—	2.2	3.2	14.4	µA	1.8	Cap Sense Low Range Low Power
		—	3.3	4.4	15.6	µA	3.0	
D028		—	6.5	13	21	µA	1.8	Cap Sense Low Range Low Power
		—	8	14	23	µA	3.0	
		—	8	14	25	µA	5.0	
D028A		—	4.2	6	17	µA	1.8	Cap Sense Low Range Medium Power
		—	6	7	18	µA	3.0	
D028A		—	8.5	15.5	23	µA	1.8	Cap Sense Low Range Medium Power
		—	11	17	24	µA	3.0	
		—	11	18	27	µA	5.0	
D028B		—	12	14	25	µA	1.8	Cap Sense Low Range High Power
		—	32	35	44	µA	3.0	
D028B		—	16	20	31	µA	1.8	Cap Sense Low Range High Power
		—	36	41	50	µA	3.0	
		—	42	49	58	µA	5.0	
D028C		—	115	—	—	µA	1.8	Cap Sense HighRange Low Power **(Note 6)**
		—	120	—	—	µA	3.0	
D028C		—	135	—	—	µA	1.8	Cap Sense High Range Low Power **(Note 6)**
		—	140	—	—	µA	3.0	
		—	150	—	—	µA	5.0	

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD.

3: Fixed Voltage Reference is automatically enabled whenever the BOR is enabled.

4: A/D oscillator source is FRC.

5: 0.1 µF capacitor on VCAP (RA0).

6: Includes FVR IPD and DAC IPD.

25.3 DC Characteristics: PIC16(L)F707-I/E (Power-Down) (Continued)

PIC16LF707			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended					
PIC16F707			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended					
Param No.	**Device Characteristics**	**Min.**	**Typ†**	**Max. +85°C**	**Max. +125°C**	**Units**	**Conditions**	
							VDD	**Note**
D028D		—	125	—	—	µA	1.8	Cap Sense HighRange Medium Power **(Note 6)**
		—	130	—	—	µA	3.0	
D028D		—	145	—	—	µA	1.8	Cap Sense High Range Medium Power **(Note 6)**
		—	150	—	—	µA	3.0	
		—	160	—	—	µA	5.0	
D028E		—	150	—	—	µA	1.8	Cap Sense HighRange High Power **(Note 6)**
		—	170	—	—	µA	3.0	
D028E		—	180	—	—	µA	1.8	Cap Sense High Range High Power **(Note 6)**
		—	190	—	—	µA	3.0	
		—	200	—	—	µA	5.0	

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD.

3: Fixed Voltage Reference is automatically enabled whenever the BOR is enabled.

4: A/D oscillator source is FRC.

5: 0.1 µF capacitor on VCAP (RA0).

6: Includes FVR IPD and DAC IPD.

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25.4 DC Characteristics: PIC16(L)F707-I/E

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
	VIL	**Input Low Voltage**					
		I/O PORT:					
D030		with TTL buffer	—	—	0.8	V	4.5V ≤ VDD ≤ 5.5V
D030A			—	—	0.15 VDD	V	1.8V ≤ VDD ≤ 4.5V
D031		with Schmitt Trigger buffer	—	—	0.2 VDD	V	2.0V ≤ VDD ≤ 5.5V
		with I²C™ levels	—	—	0.3 VDD	V	
D032		$\overline{MCLR}$, OSC1 (RC mode)(1)	—	—	0.2 VDD	V	
D033A		OSC1 (HS mode)	—	—	0.3 VDD	V	
	VIH	**Input High Voltage**					
		I/O ports:		—	—		
D040		with TTL buffer	2.0	—	—	V	4.5V ≤ VDD ≤ 5.5V
D040A			0.25 VDD + 0.8	—	—	V	1.8V ≤ VDD ≤ 4.5V
D041		with Schmitt Trigger buffer	0.8 VDD	—	—	V	2.0V ≤ VDD ≤ 5.5V
		with I²C™ levels	0.7 VDD	—	—	V	
D042		$\overline{MCLR}$	0.8 VDD	—	—	V	
D043A		OSC1 (HS mode)	0.7 VDD	—	—	V	
D043B		OSC1 (RC mode)	0.9 VDD	—	—	V	**(Note 1)**
	IIL	**Input Leakage Current(2)**					
D060		I/O ports	—	± 5	± 125	nA	VSS ≤ VPIN ≤ VDD, Pin at high-impedance, 85°C
				± 5	± 1000	nA	125°C
D061		$\overline{MCLR}$(3)	—	± 50	± 200	nA	VSS ≤ VPIN ≤ VDD, 85°C
	IPUR	**PORTB Weak Pull-up Current**					
D070*			25 25	100 140	200 300	 μA	VDD = 3.3V, VPIN = VSS VDD = 5.0V, VPIN = VSS
	VOL	**Output Low Voltage(4)**					
D080		I/O ports	—	—	0.6	V	IOL = 8mA, VDD = 5V IOL = 6mA, VDD = 3.3V IOL = 1.8mA, VDD = 1.8V
	VOH	**Output High Voltage(4)**					
D090		I/O ports	VDD - 0.7	—	—	V	IOH = 3.5mA, VDD = 5V IOH = 3mA, VDD = 3.3V IOH = 1mA, VDD = 1.8V
		Capacitive Loading Specs on Output Pins					
D101*	COSC2	OSC2 pin	—	—	15	pF	In XT, HS and LP modes when external clock is used to drive OSC1
D101A*	CIO	All I/O pins	—	—	50	pF	
		Program Flash Memory					

Legend: TBD = To Be Determined

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in RC mode.

2: Negative current is defined as current sourced by the pin.

3: The leakage current on the $\overline{MCLR}$ pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

4: Including OSC2 in CLKOUT mode.

25.4 DC Characteristics: PIC16(L)F707-I/E (Continued)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
D130	EP	Cell Endurance	100	1k	—	E/W	Temperature during programming: 10°C ≤ TA ≤ 40°C
D131		VDD for Read	VMIN	—	—	V	
		Voltage on $\overline{MCLR}$/VPP during Erase/Program	8.0	—	9.0	V	Temperature during programming: 10°C ≤ TA ≤ 40°C
		VDD for Bulk Erase	2.7	3	—	V	Temperature during programming: 10°C ≤ TA ≤ 40°C
D132	VPEW	VDD for Write or Row Erase	2.7	—	—	V	VMIN = Minimum operating voltage VMAX = Maximum operating voltage
	IPPPGM	Current on $\overline{MCLR}$/VPP during Erase/Write	—	—	5.0	mA	Temperature during programming: 10°C ≤ TA ≤ 40°C
	IDDPGM	Current on VDD during Erase/ Write	—		5.0	mA	Temperature during programming: 10°C ≤ TA ≤ 40°C
D133	TPEW	Erase/Write cycle time	—		2.8	ms	Temperature during programming: 10°C ≤ TA ≤ 40°C
D134	TRETD	Characteristic Retention	40	—	—	Year	Provided no other specifications are violated
		VCAP Capacitor Charging					
D135		Charging current	—	200	—	µA	
D135A		Source/sink capability when charging complete	—	0.0	—	mA	

Legend: TBD = To Be Determined

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in RC mode.

2: Negative current is defined as current sourced by the pin.

3: The leakage current on the $\overline{MCLR}$ pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

4: Including OSC2 in CLKOUT mode.

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25.5 Thermal Considerations

Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +125°C					
Param No.	**Sym**	**Characteristic**	**Typ**	**Units**	**Conditions**
TH01	θJA	Thermal Resistance Junction to Ambient	47.2	°C/W	40-pin PDIP package
			46	°C/W	44-pin TQFP package
			24.4	°C/W	44-pin QFN 8x8mm package
			TBD	°C/W	40-pin UQFN 5x5mm package
TH02	θJC	Thermal Resistance Junction to Case	24.7	°C/W	40-pin PDIP package
			14.5	°C/W	44-pin TQFP package
			20	°C/W	44-pin QFN 8x8mm package
			TBD	°C/W	40-pin UQFN 5x5mm package
TH03	TJMAX	Maximum Junction Temperature	150	°C	
TH04	PD	Power Dissipation	—	W	PD = PINTERNAL + PI/O
TH05	PINTERNAL	Internal Power Dissipation	—	W	PINTERNAL = IDD x VDD(1)
TH06	PI/O	I/O Power Dissipation	—	W	PI/O = Σ (IOL * VOL) + Σ (IOH * (VDD - VOH))
TH07	PDER	Derated Power	—	W	PDER = PDMAX (TJ - TA)/θJA(2, 3)

Note 1: IDD is current to run the chip alone without driving any load on the output pins.
2: TA = Ambient Temperature
3: TJ = Junction Temperature

25.6 Timing Parameter Symbology

The timing parameter symbols have been created with one of the following formats:

1. TppS2ppS
2. TppS

T			
F	Frequency	T	Time

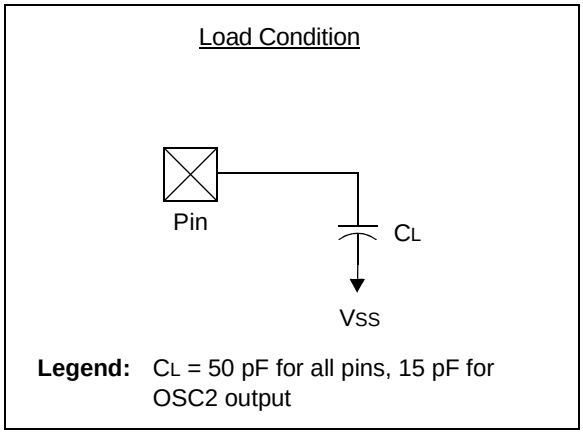
Lowercase letters (pp) and their meanings:

pp			
cc	CCP1	osc	OSC1
ck	CLKOUT	rd	$\overline{RD}$
cs	$\overline{CS}$	rw	$\overline{RD}$ or $\overline{WR}$
di	SDI	sc	SCK
do	SDO	ss	$\overline{SS}$
dt	Data in	t0	T0CKI
io	I/O PORT	t1	T1CKI
mc	$\overline{MCLR}$	wr	$\overline{WR}$

Uppercase letters and their meanings:

S			
F	Fall	P	Period
H	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance

FIGURE 25-2: LOAD CONDITIONS



Legend: C_L = 50 pF for all pins, 15 pF for OSC2 output

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25.7 AC Characteristics: PIC16F707-I/E

FIGURE 25-3: CLOCK TIMING

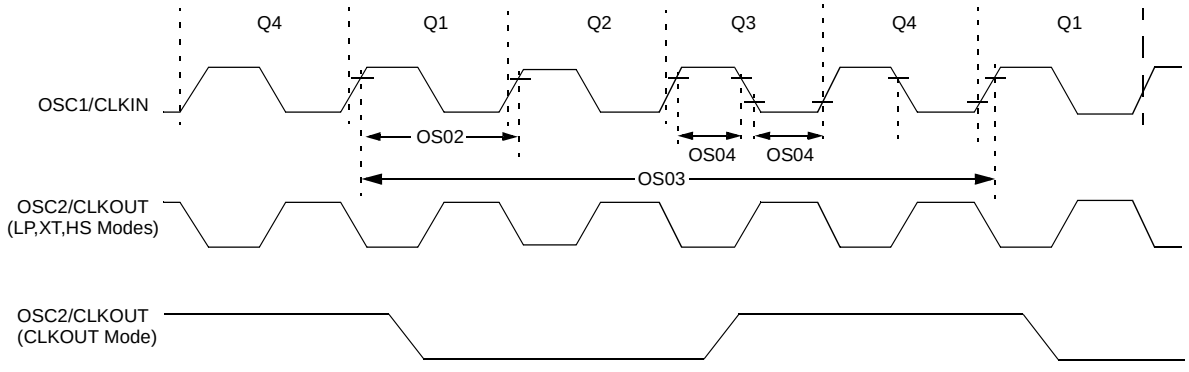
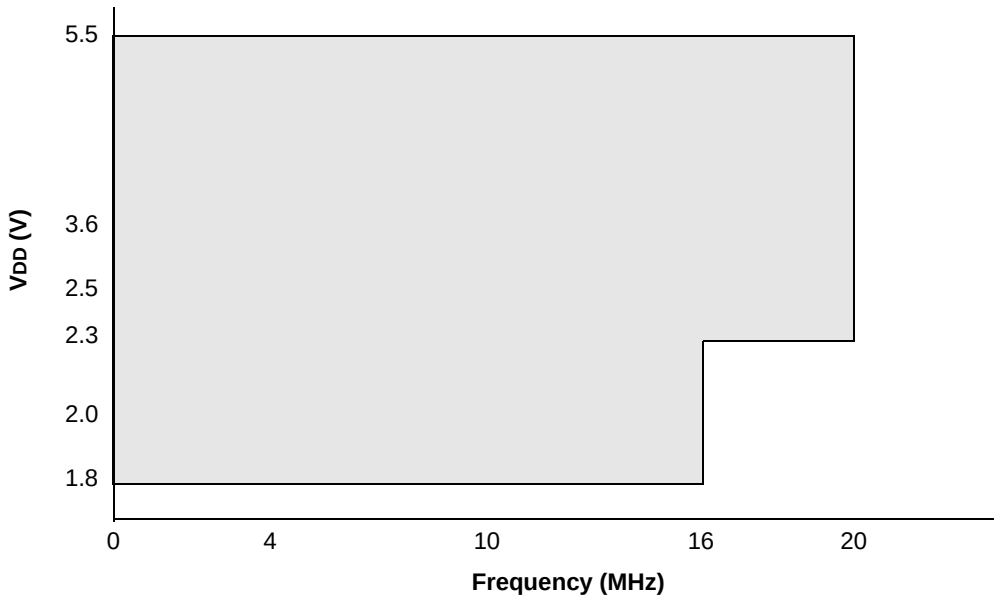
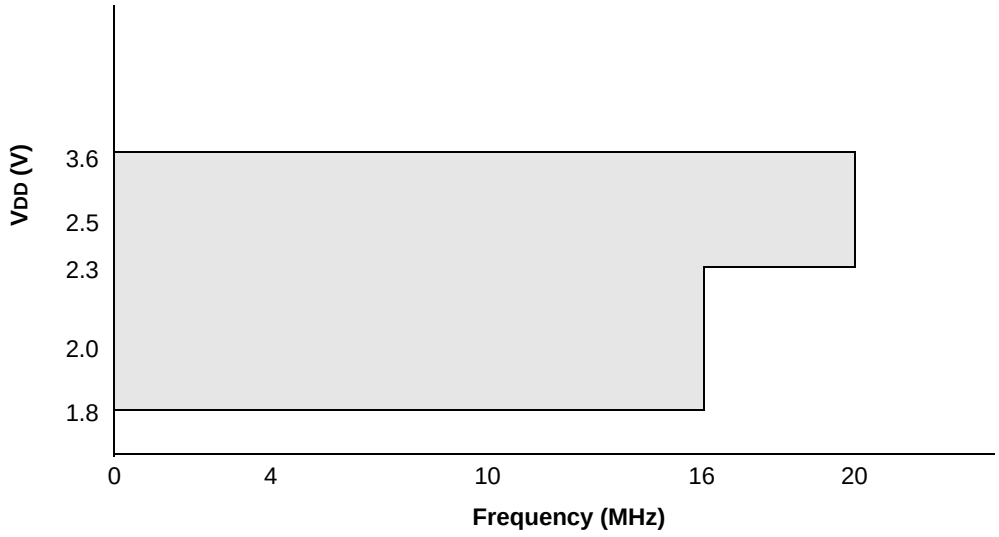


FIGURE 25-4: PIC16F707 VOLTAGE FREQUENCY GRAPH, -40°C ≤ TA ≤ +125°C



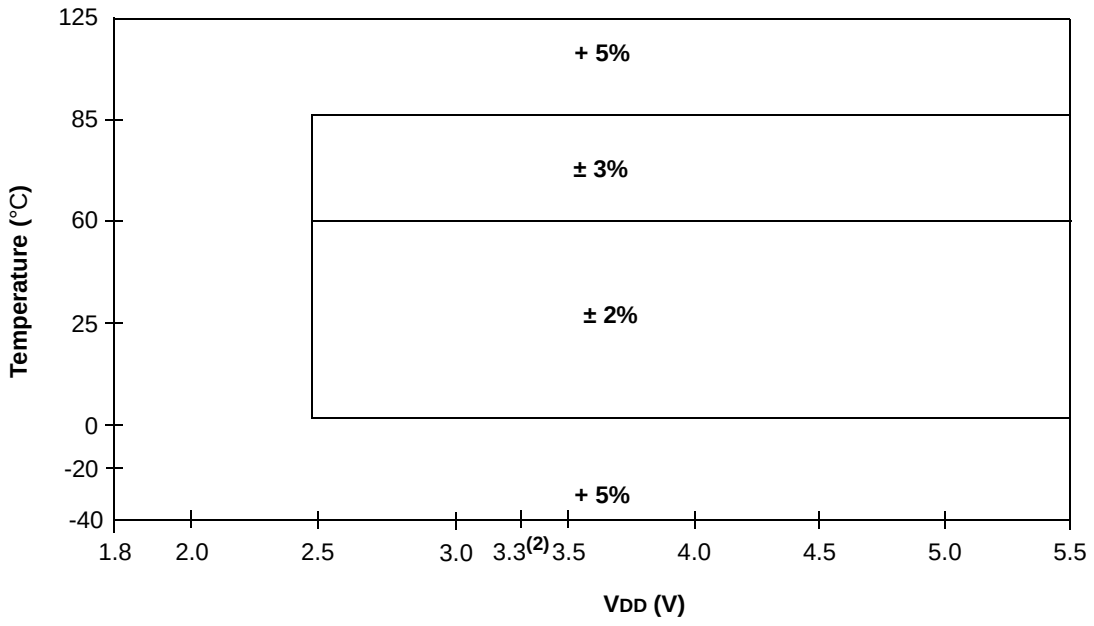
Note 1: The shaded region indicates the permissible combinations of voltage and frequency.
2: Refer to Table 25-1 for each Oscillator mode's supported frequencies.

FIGURE 25-5: PIC16LF707 VOLTAGE FREQUENCY GRAPH, -40°C ≤ TA ≤ +125°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.
2: Refer to Table 25-1 for each Oscillator mode's supported frequencies.

FIGURE 25-6: HFINTOSC FREQUENCY ACCURACY OVER DEVICE VDD AND TEMPERATURE



Note 1: This chart covers both regulator enabled and regulator disabled states.
2: Regulator Nominal voltage.

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TABLE 25-1: CLOCK OSCILLATOR TIMING REQUIREMENTS

Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +125°C							
Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
OS01	FOSC	External CLKIN Frequency(1)	DC	—	37	kHz	LP Oscillator mode
			DC	—	4	MHz	XT Oscillator mode
			DC	—	20	MHz	HS Oscillator mode
			DC	—	20	MHz	EC Oscillator mode
		Oscillator Frequency(1)	—	32.768	—	kHz	LP Oscillator mode
			0.1	—	4	MHz	XT Oscillator mode
			1	—	4	MHz	HS Oscillator mode, VDD ≤ 2.7V
			1	—	20	MHz	HS Oscillator mode, VDD > 2.7V
			DC	—	4	MHz	RC Oscillator mode
OS02	TOSC	External CLKIN Period(1)	27	—	∞	µs	LP Oscillator mode
			250	—	∞	ns	XT Oscillator mode
			50	—	∞	ns	HS Oscillator mode
			50	—	∞	ns	EC Oscillator mode
		Oscillator Period(1)	—	30.5	—	µs	LP Oscillator mode
			250	—	10,000	ns	XT Oscillator mode
			250	—	1,000	ns	HS Oscillator mode, VDD ≤ 2.7V
			50	—	1,000	ns	HS Oscillator mode, VDD > 2.7V
			250	—	—	ns	RC Oscillator mode
OS03	TCY	Instruction Cycle Time(1)	200	TCY	DC	ns	TCY = 4/FOSC
OS04*	TosH, TosL	External CLKIN High, External CLKIN Low	2	—	—	µs	LP oscillator
			100	—	—	ns	XT oscillator
			20	—	—	ns	HS oscillator
OS05*	TosR, TosF	External CLKIN Rise, External CLKIN Fall	0	—	∞	ns	LP oscillator
			0	—	∞	ns	XT oscillator
			0	—	∞	ns	HS oscillator

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

TABLE 25-2: OSCILLATOR PARAMETERS

Standard Operating Conditions (unless otherwise stated) Operating Temperature -40°C ≤ TA ≤ +125°C								
Param No.	**Sym.**	**Characteristic**	**Freq. Tolerance**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
OS08	HFOSC	Internal Calibrated HFINTOSC Frequency(2)	±2%	—	16.0	—	MHz	0°C ≤ TA ≤ +85°C, VDD ≥ 2.5V
			±5%	—	16.0	—	MHz	-40°C ≤ TA ≤ +125°C
OS08A	MFOSC	Internal Calibrated MFINTOSC Frequency(2)	±2%	—	500	—	kHz	0°C ≤ TA ≤ +85°C VDD ≥ 2.5V
			±5%	—	500	10	kHz	-40°C ≤ TA ≤ +125°C
OS10*	TIOSC ST	HFINTOSC Wake-up from Sleep Start-up Time	—	—	5	8	µs	
		MFINTOSC Wake-up from Sleep Start-up Time	—	—	20	30	µs	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to the OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

2: To ensure these oscillator frequency tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 µF and 0.01 µF values in parallel are recommended.

3: By design.

FIGURE 25-7: CLKOUT AND I/O TIMING

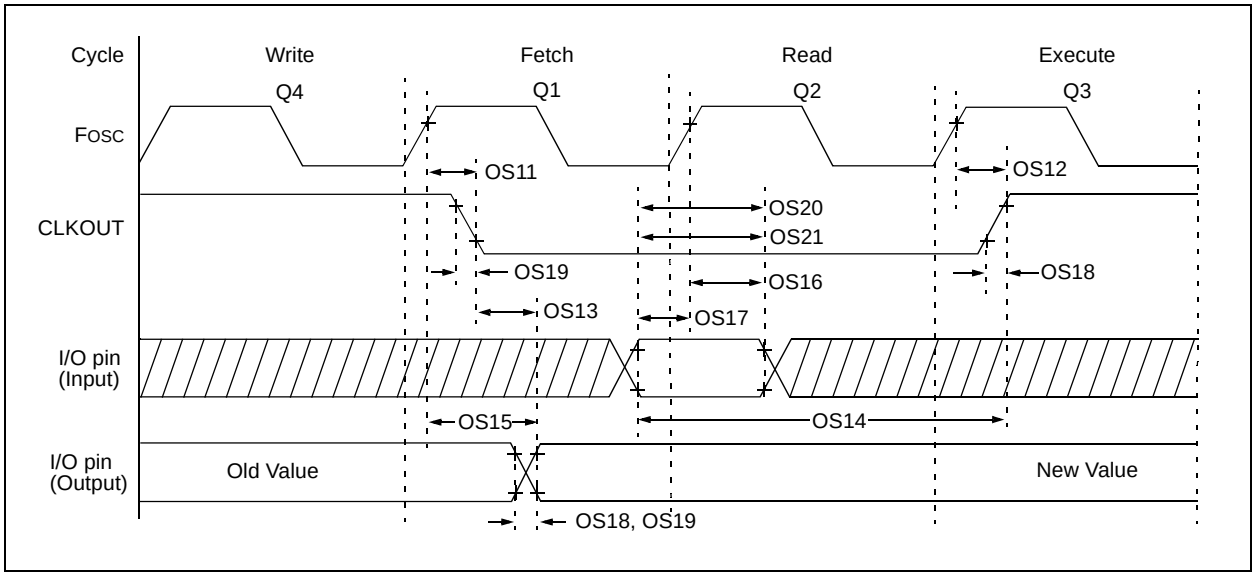

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TABLE 25-3: CLKOUT AND I/O TIMING PARAMETERS

Standard Operating Conditions (unless otherwise stated) Operating Temperature -40°C ≤ TA ≤ +125°C							
Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
OS11	TosH2ckL	Fosc↑ to CLKOUT↓ (1)	—	—	70	ns	VDD = 3.3-5.0V
OS12	TosH2ckH	Fosc↑ to CLKOUT↑ (1)	—	—	72	ns	VDD = 3.3-5.0V
OS13	TckL2ioV	CLKOUT↓ to Port out valid(1)	—	—	20	ns	
OS14	TioV2ckH	Port input valid before CLKOUT↑(1)	TOSC + 200 ns	—	—	ns	
OS15	TosH2ioV	Fosc↑ (Q1 cycle) to Port out valid	—	50	70*	ns	VDD = 3.3-5.0V
OS16	TosH2ioI	Fosc↑ (Q2 cycle) to Port input invalid (I/O in hold time)	50	—	—	ns	VDD = 3.3-5.0V
OS17	TioV2osH	Port input valid to Fosc↑ (Q2 cycle) (I/O in setup time)	20	—	—	ns	
OS18	TioR	Port output rise time(2)	— —	40 15	72 32	ns	VDD = 2.0V VDD = 3.3-5.0V
OS19	TioF	Port output fall time(2)	— —	28 15	55 30	ns	VDD = 2.0V VDD = 3.3-5.0V
OS20*	Tinp	INT pin input high or low time	25	—	—	ns	
OS21*	Trbp	PORTB interrupt-on-change new input level time	TCY	—	—	ns	

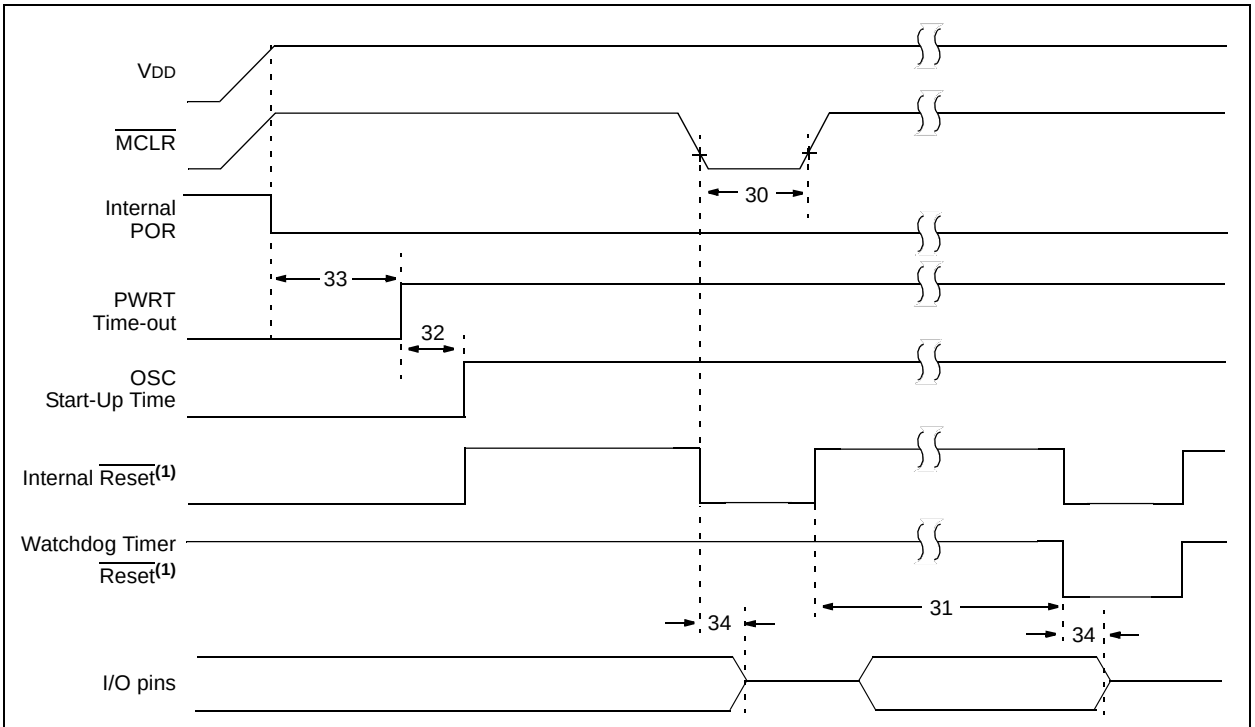
* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated.

Note 1: Measurements are taken in RC mode where CLKOUT output is 4 x TOSC.

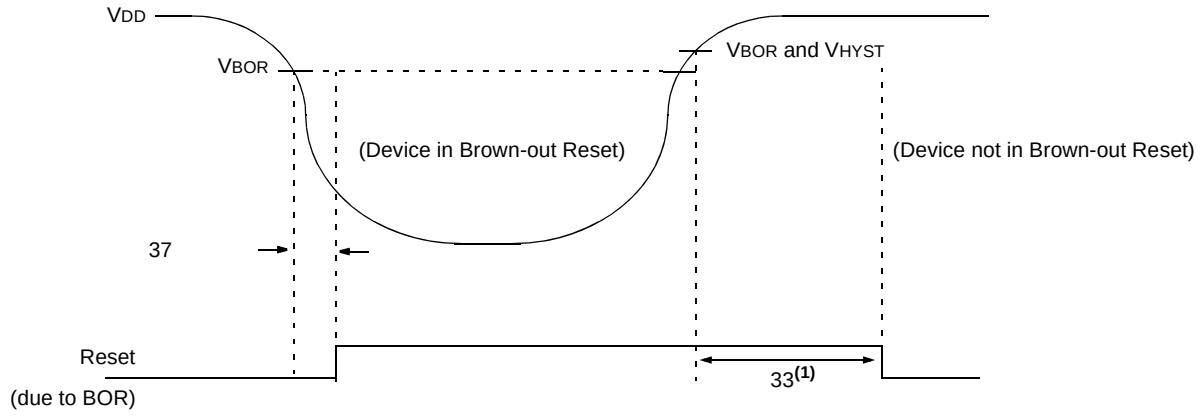
2: Includes OSC2 in CLKOUT mode.

FIGURE 25-8: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING



Note 1: Asserted low.

FIGURE 25-9: BROWN-OUT RESET TIMING AND CHARACTERISTICS



Note 1: 64 ms delay only if $\overline{\text{PWRTE}}$ bit in the Configuration Word register is programmed to '0'. 2 ms delay if $\overline{\text{PWRTE}}$ = 0 and VREGEN = 1.

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TABLE 25-4: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER, AND BROWN-OUT RESET PARAMETERS

Standard Operating Conditions (unless otherwise stated) Operating Temperature -40°C ≤ TA ≤ +125°C							
Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
30	TMCL	$\overline{\text{MCLR}}$ Pulse Width (low)	2 5	— —	— —	µs µs	VDD = 3.3-5V, -40°C to +85°C VDD = 3.3-5V
31	TWDTLP	Low Power Watchdog Timer Time-out Period (No Prescaler)	10	18	27	ms	VDD = 3.3V-5V
32	TOST	Oscillator Start-up Timer Period(1), (2)	—	1024	—	Tosc	**(Note 3)**
33*	TPWRT	Power-up Timer Period, $\overline{\text{PWRTE}}$ = 0	40	65	140	ms	
34*	TIOZ	I/O high-impedance from $\overline{\text{MCLR}}$ Low or Watchdog Timer Reset	—	—	2.0	µs	
35	VBOR	Brown-out Reset Voltage	2.38 1.80	2.5 1.9	2.73 2.11	V	BORV=2.5V BORV=1.9V
36*	VHYST	Brown-out Reset Hysteresis	0	25	50	mV	-40°C to +85°C
37*	TBORDC	Brown-out Reset DC Response Time	1	3	5 10	µs	VDD ≤ VBOR, -40°C to +85°C VDD ≤ VBOR

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to the OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

2: By design.

3: Period of the slower clock.

4: To ensure these voltage tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 µF and 0.01 µF values in parallel are recommended.

FIGURE 25-10: TIMER0/A/B AND TIMER1/3 EXTERNAL CLOCK TIMINGS

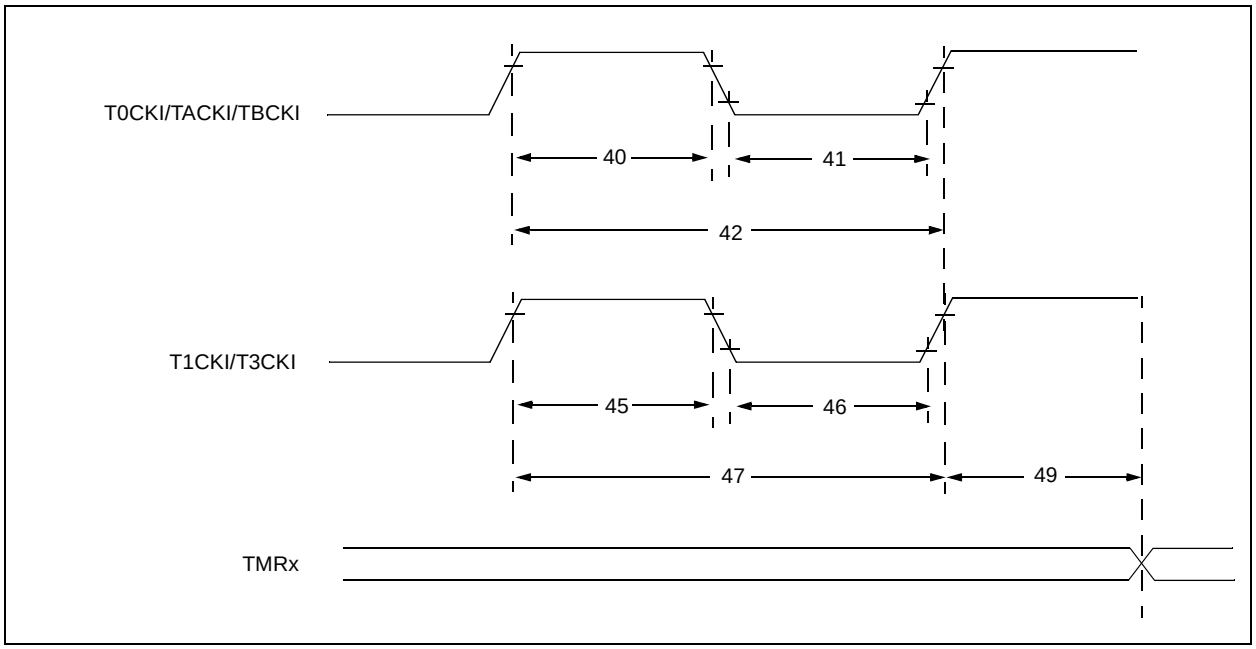

TABLE 25-5: TIMER0/A/B AND TIMER1/3 EXTERNAL CLOCK REQUIREMENTS

Standard Operating Conditions (unless otherwise stated) Operating Temperature -40°C ≤ TA ≤ +125°C								
Param No.	**Sym.**	**Characteristic**		**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
40*	TT0H	T0CKI/TACKI/TBCKI High Pulse Width	No Prescaler	0.5 TCY + 20	—	—	ns	
			With Prescaler	10	—	—	ns	
41*	TT0L	T0CKI/TACKI/TBCKI Low Pulse Width	No Prescaler	0.5 TCY + 20	—	—	ns	
			With Prescaler	10	—	—	ns	
42*	TT0P	T0CKI/TACKI/TBCKI Period		Greater of: 20 or $\frac{T_{CY} + 40}{N}$	—	—	ns	N = prescale value (2, 4, ..., 256)
45*	TT1H	T1CKI/T3CKI High Time	Synchronous, No Prescaler	0.5 TCY + 20	—	—	ns	
			Synchronous, with Prescaler	15	—	—	ns	
			Asynchronous	30	—	—	ns	
46*	TT1L	T1CKI/T3CKI Low Time	Synchronous, No Prescaler	0.5 TCY + 20	—	—	ns	
			Synchronous, with Prescaler	15	—	—	ns	
			Asynchronous	30	—	—	ns	
47*	TT1P	T1CKI/T3CKI Input Period	Synchronous	Greater of: 30 or $\frac{T_{CY} + 40}{N}$	—	—	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous	60	—	—	ns	
48	FT1	Timer1 Oscillator Input Frequency Range (oscillator enabled by setting bit T1OSCEN)		32.4	32.768	33.1	kHz	
49*	TCKEZTMR1	Delay from External Clock Edge to Timer Increment		2 TOSC	—	7 TOSC	—	Timers in Sync mode

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

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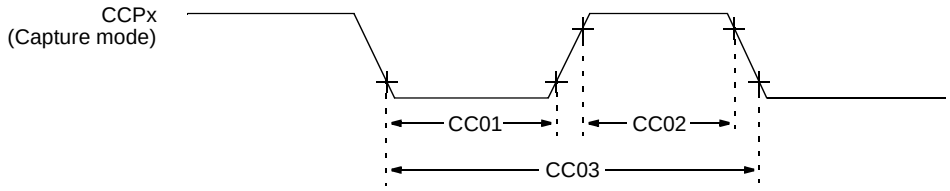
FIGURE 25-11: CAPTURE/COMPARE/PWM TIMINGS (CCP)



Note: Refer to Figure 25-2 for load conditions.

TABLE 25-6: CAPTURE/COMPARE/PWM REQUIREMENTS (CCP)

Standard Operating Conditions (unless otherwise stated) Operating Temperature -40°C ≤ TA ≤ +125°C								
Param No.	**Sym.**	**Characteristic**		**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
CC01*	TccL	CCPx Input Low Time	No Prescaler	0.5TCY + 20	—	—	ns	
			With Prescaler	20	—	—	ns	
CC02*	TccH	CCPx Input High Time	No Prescaler	0.5TCY + 20	—	—	ns	
			With Prescaler	20	—	—	ns	
CC03*	TccP	CCPx Input Period		$\frac{3T_{CY} + 40}{N}$	—	—	ns	N = prescale value (1, 4 or 16)

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 25-7: PIC16F707 A/D CONVERTER (ADC) CHARACTERISTICS

Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +125°C							
Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
AD01	NR	Resolution	—	—	8	bit	
AD02	EIL	Integral Error	—	—	±1.7	LSb	VREF = 3.0V
AD03	EDL	Differential Error	—	—	±1	LSb	No missing codes VREF = 3.0V
AD04	EOFF	Offset Error	—	—	±2.2	LSb	VREF = 3.0V
AD05	EGN	Gain Error	—	—	±1.5	LSb	VREF = 3.0V
AD06	VREF	Reference Voltage(3)	1.8	—	VDD	V	
AD07	VAIN	Full-Scale Range	VSS	—	VREF	V	
AD08	ZAIN	Recommended Impedance of Analog Voltage Source	—	—	50	kΩ	Can go higher if external 0.01µF capacitor is present on input pin.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Total Absolute Error includes integral, differential, offset and gain errors.

2: The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.

3: When ADC is off, it will not consume any current other than leakage current. The power-down current specification includes any such leakage from the ADC module.

TABLE 25-8: PIC16F707 A/D CONVERSION REQUIREMENTS

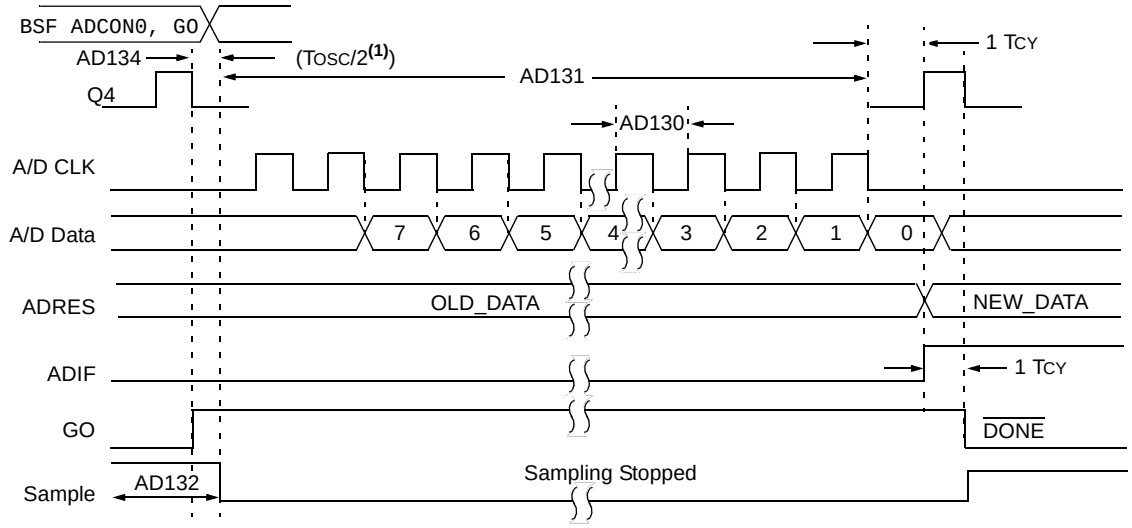
Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +125°C							
Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
AD130*	TAD	A/D Clock Period	1.0	—	9.0	µs	TOSC-based
		A/D Internal RC Oscillator Period	1.0	2.0	6.0	µs	ADCS<1:0> = 11 (ADRC mode)
AD131	TCNV	Conversion Time (not including Acquisition Time)(1)	—	10.5	—	TAD	Set GO/$\overline{\text{DONE}}$ bit to conversion complete
AD132*	TACQ	Acquisition Time	—	1.0	—	µs	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The ADRES register may be read on the following TCY cycle.

FIGURE 25-12: PIC16F707 A/D CONVERSION TIMING (NORMAL MODE)



Note 1: If the A/D clock source is selected as RC, a time of TCY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.

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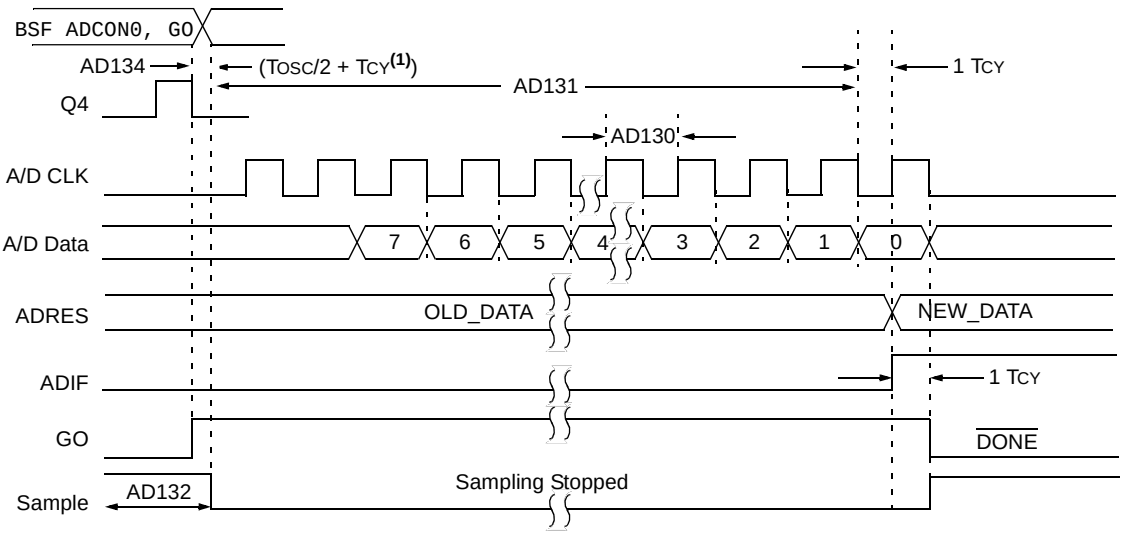
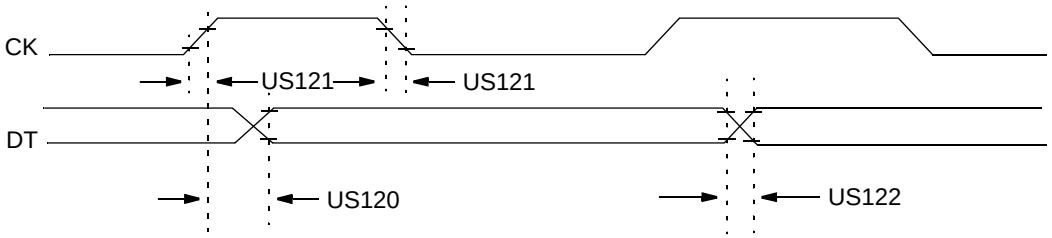
FIGURE 25-13: PIC16F707 A/D CONVERSION TIMING (SLEEP MODE)



Note 1: If the A/D clock source is selected as RC, a time of TCY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.

FIGURE 25-14: USART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING

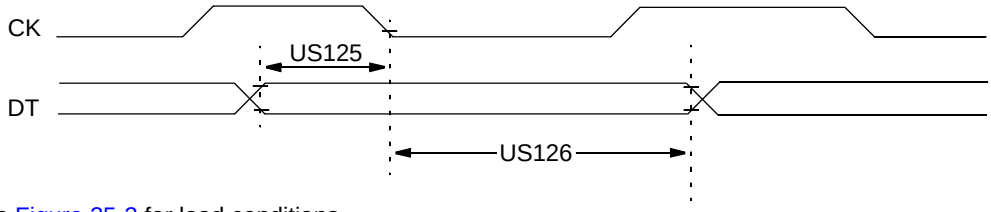


Note: Refer to Figure 25-2 for load conditions.

TABLE 25-9: USART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Standard Operating Conditions (unless otherwise stated) Operating Temperature -40°C ≤ TA ≤ +125°C							
Param. No.	**Symbol**	**Characteristic**		**Min.**	**Max.**	**Units**	**Conditions**
US120	TCKH2DTV	SYNC XMIT (Master and Slave) Clock high to data-out valid	3.0-5.5V	—	80	ns	
			1.8-5.5V	—	100	ns	
US121	TCKRF	Clock out rise time and fall time (Master mode)	3.0-5.5V	—	45	ns	
			1.8-5.5V	—	50	ns	
US122	TDTRF	Data-out rise time and fall time	3.0-5.5V	—	45	ns	
			1.8-5.5V	—	50	ns	

FIGURE 25-15: USART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING

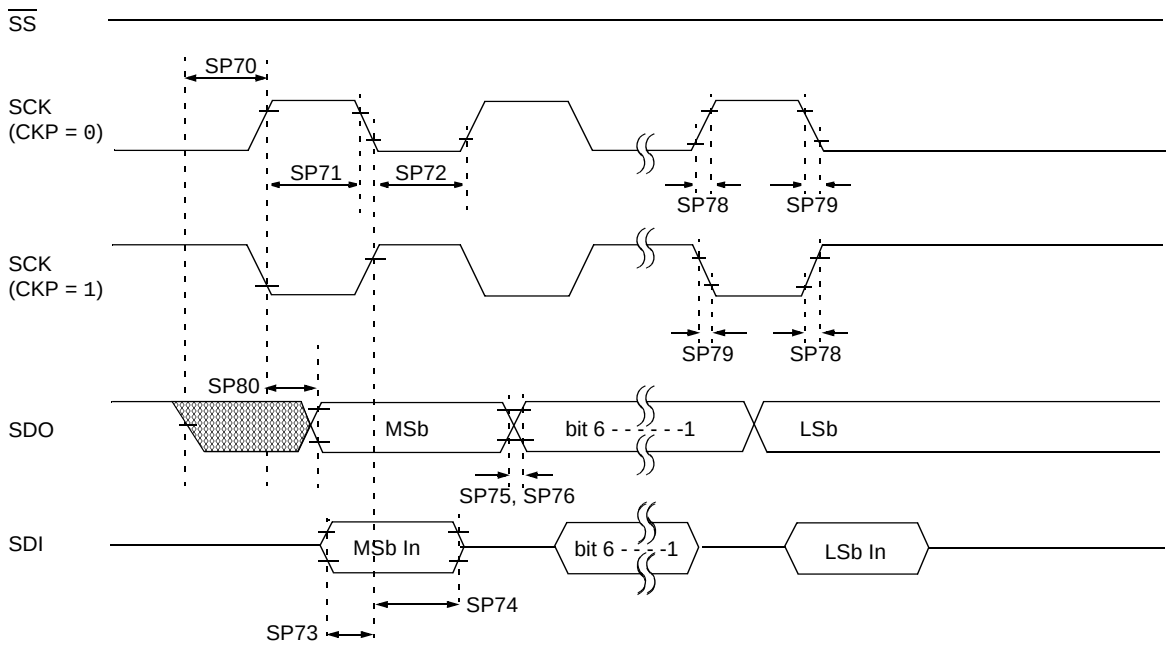


Note: Refer to Figure 25-2 for load conditions.

TABLE 25-10: USART SYNCHRONOUS RECEIVE REQUIREMENTS

Standard Operating Conditions (unless otherwise stated) Operating Temperature -40°C ≤ TA ≤ +125°C						
Param. No.	**Symbol**	**Characteristic**	**Min.**	**Max.**	**Units**	**Conditions**
US125	TDTV2CKL	SYNC RCV (Master and Slave) Data-hold before CK ↓ (DT hold time)	10	—	ns	
US126	TCKL2DTL	Data-hold after CK ↓ (DT hold time)	15	—	ns	

FIGURE 25-16: SPI MASTER MODE TIMING (CKE = 0, SMP = 0)



Note: Refer to Figure 25-2 for load conditions.

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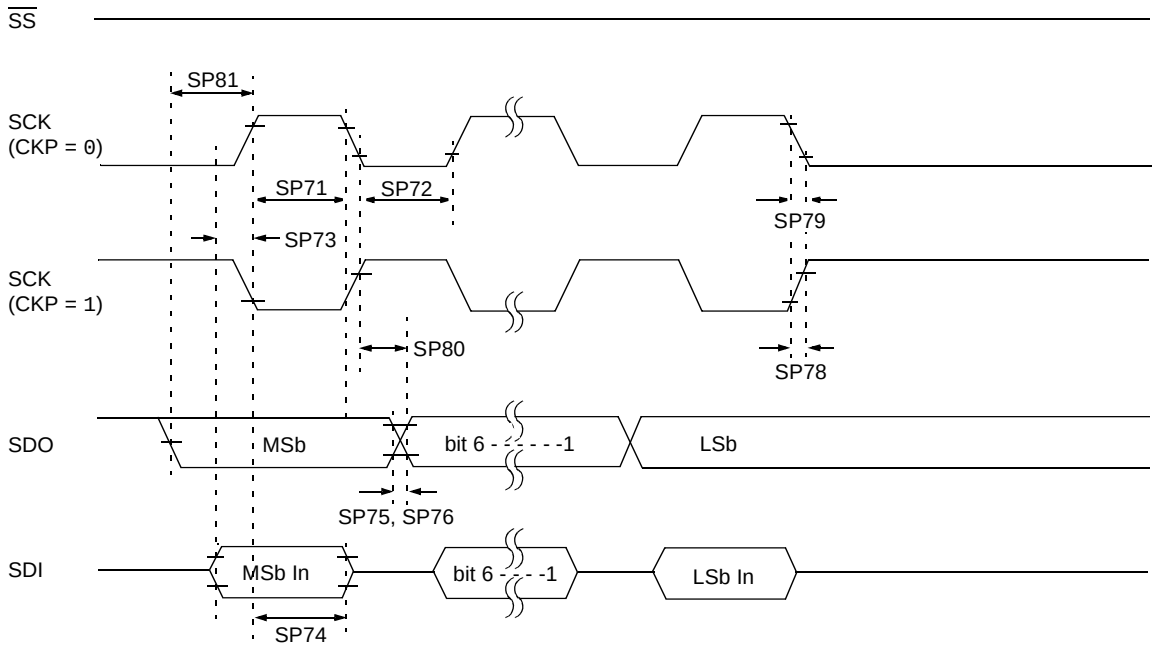
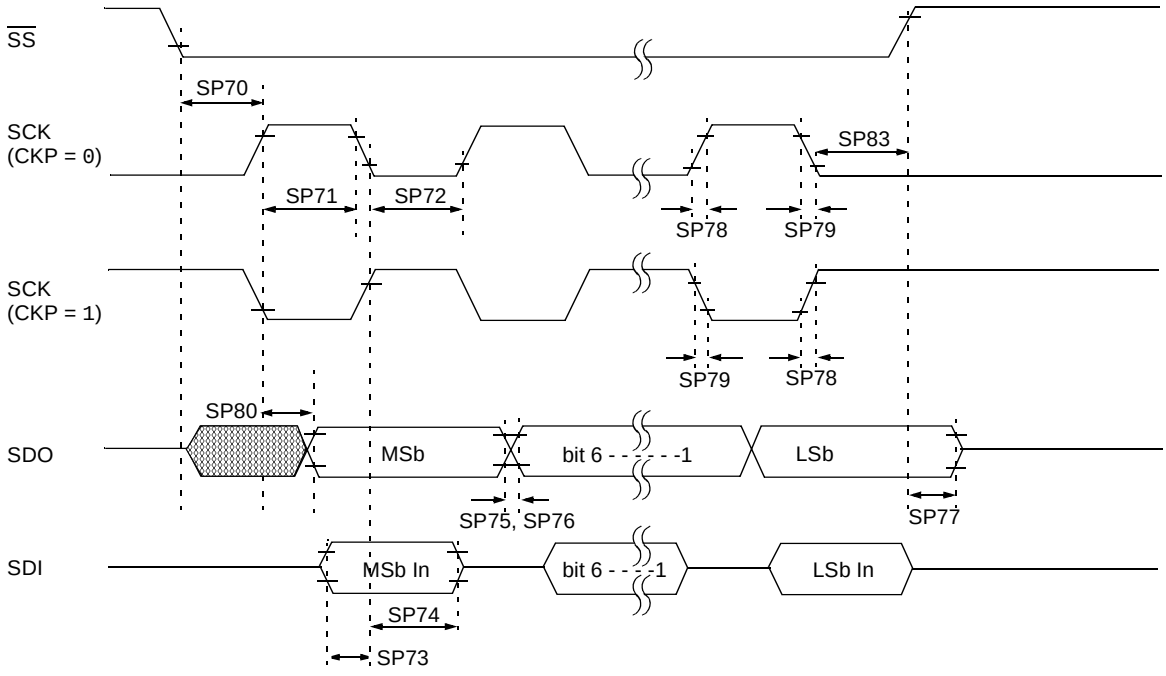
FIGURE 25-17: SPI MASTER MODE TIMING (CKE = 1, SMP = 1)



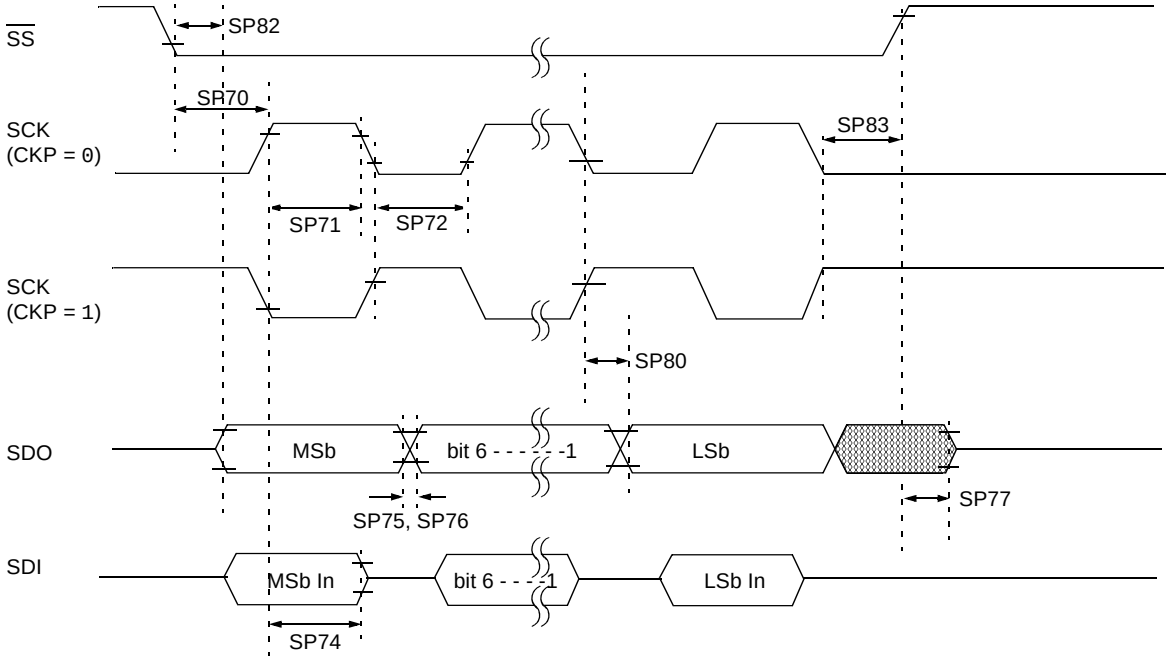
Note: Refer to Figure 25-2 for load conditions.

FIGURE 25-18: SPI SLAVE MODE TIMING (CKE = 0)



Note: Refer to Figure 25-2 for load conditions.

FIGURE 25-19: SPI SLAVE MODE TIMING (CKE = 1)



Note: Refer to Figure 25-2 for load conditions.

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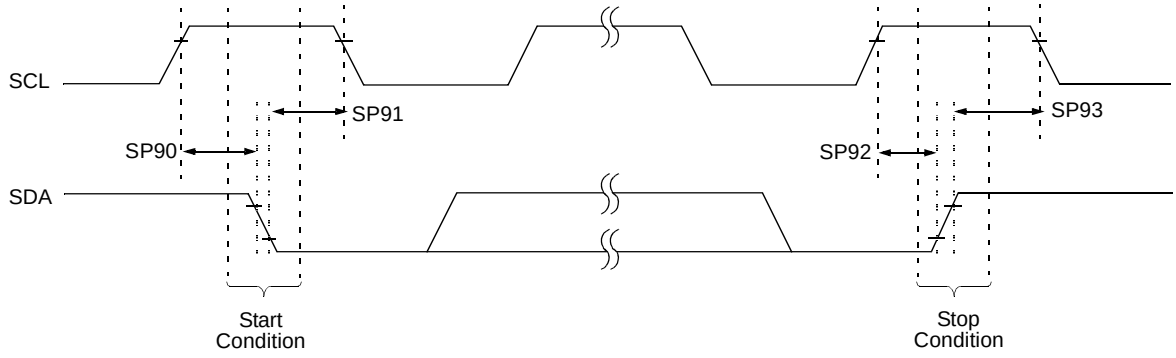
TABLE 25-11: SPI MODE REQUIREMENTS

Param No.	Symbol	Characteristic		Min.	Typ†	Max.	Units	Conditions
SP70*	TssL2scH, TssL2scL	$\overline{SS}$↓ to SCK↓ or SCK↑ input		Tcy	—	—	ns	
SP71*	TscH	SCK input high time (Slave mode)		Tcy + 20	—	—	ns	
SP72*	TscL	SCK input low time (Slave mode)		Tcy + 20	—	—	ns	
SP73*	TdiV2scH, TdiV2scL	Setup time of SDI data input to SCK edge		100	—	—	ns	
SP74*	TscH2diL, TscL2diL	Hold time of SDI data input to SCK edge		100	—	—	ns	
SP75*	TdoR	SDO data output rise time	3.0-5.5V	—	10	25	ns	
			1.8-5.5V	—	25	50	ns	
SP76*	TdoF	SDO data output fall time		—	10	25	ns	
SP77*	TssH2doZ	$\overline{SS}$↑ to SDO output high-impedance		10	—	50	ns	
SP78*	TscR	SCK output rise time (Master mode)	3.0-5.5V	—	10	25	ns	
			1.8-5.5V	—	25	50	ns	
SP79*	TscF	SCK output fall time (Master mode)		—	10	25	ns	
SP80*	TscH2doV, TscL2doV	SDO data output valid after SCK edge	3.0-5.5V	—	—	50	ns	
			1.8-5.5V	—	—	145	ns	
SP81*	TdoV2scH , TdoV2scL	SDO data output setup to SCK edge		Tcy	—	—	ns	
SP82*	TssL2doV	SDO data output valid after $\overline{SS}$↓ edge		—	—	50	ns	
SP83*	TscH2ssH, TscL2ssH	$\overline{SS}$ ↑ after SCK edge		1.5Tcy + 40	—	—	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 25-20: I²C™ BUS START/STOP BITS TIMING



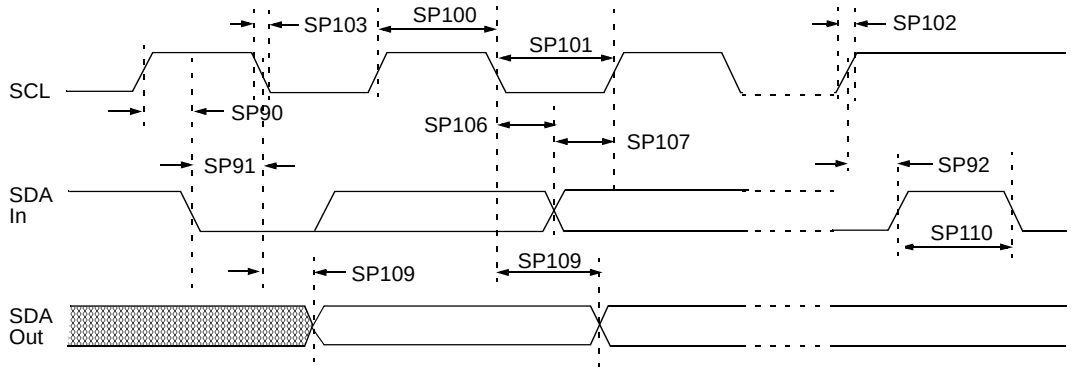
Note: Refer to Figure 25-2 for load conditions.

TABLE 25-12: I²C™ BUS START/STOP BITS REQUIREMENTS

Param No.	Symbol	Characteristic		Min.	Typ	Max.	Units	Conditions
SP90*	TSU:STA	Start condition	100 kHz mode	4700	—	—	ns	Only relevant for Repeated Start condition
		Setup time	400 kHz mode	600	—	—		
SP91*	THD:STA	Start condition	100 kHz mode	4000	—	—	ns	After this period, the first clock pulse is generated
		Hold time	400 kHz mode	600	—	—		
SP92*	TSU:STO	Stop condition	100 kHz mode	4700	—	—	ns	
		Setup time	400 kHz mode	600	—	—		
SP93	THD:STO	Stop condition	100 kHz mode	4000	—	—	ns	
		Hold time	400 kHz mode	600	—	—		

* These parameters are characterized but not tested.

FIGURE 25-21: I²C™ BUS DATA TIMING



Note: Refer to Figure 25-2 for load conditions.

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TABLE 25-13: I²C™ BUS DATA REQUIREMENTS

Param. No.	Symbol	Characteristic		Min.	Max.	Units	Conditions
SP100*	THIGH	Clock high time	100 kHz mode	4.0	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	0.6	—	μs	Device must operate at a minimum of 10 MHz
			SSP Module	1.5TCY	—		
SP101*	TLOW	Clock low time	100 kHz mode	4.7	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	1.3	—	μs	Device must operate at a minimum of 10 MHz
			SSP Module	1.5TCY	—		
SP102*	TR	SDA and SCL rise time	100 kHz mode	—	1000	ns	
			400 kHz mode	20 + 0.1CB	300	ns	CB is specified to be from 10-400 pF
SP103*	TF	SDA and SCL fall time	100 kHz mode	—	250	ns	
			400 kHz mode	20 + 0.1CB	250	ns	CB is specified to be from 10-400 pF
SP106*	THD:DAT	Data input hold time	100 kHz mode	0	—	ns	
			400 kHz mode	0	0.9	μs	
SP107*	TSU:DAT	Data input setup time	100 kHz mode	250	—	ns	**(Note 2)**
			400 kHz mode	100	—	ns	
SP109*	TAA	Output valid from clock	100 kHz mode	—	3500	ns	**(Note 1)**
			400 kHz mode	—	—	ns	
SP110*	TBUF	Bus free time	100 kHz mode	4.7	—	μs	Time the bus must be free before a new transmission can start
			400 kHz mode	1.3	—	μs	
SP111	CB	Bus capacitive loading		—	400	pF	

* These parameters are characterized but not tested.

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of Start or Stop conditions.

2: A Fast mode (400 kHz) I²C™ bus device can be used in a Standard mode (100 kHz) I²C bus system, but the requirement TSU:DAT ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the low period of the SCL signal. If such a device does stretch the low period of the SCL signal, it must output the next data bit to the SDA line TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I²C bus specification), before the SCL line is released.

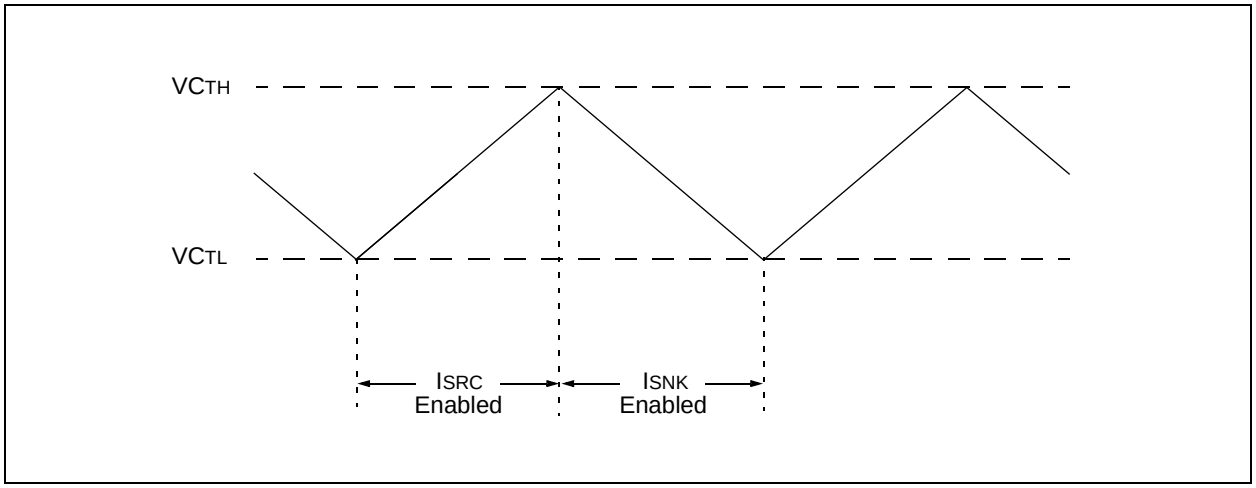
TABLE 25-14: CAP SENSE OSCILLATOR SPECIFICATIONS

Param. No.	Symbol	Characteristic		Min.	Typ†	Max.	Units	Conditions
CS01	I_{SRC}	Current Source	High	—	-5.8	-6	μA	-40, -85°C
			Medium	—	-1.1	-3.2	μA	
			Low	—	-0.2	-0.9	μA	
CS02	I_{SNK}	Current Sink	High	—	6.6	6	μA	-40, -85°C
			Medium	—	1.3	3.2	μA	
			Low	—	0.24	0.9	μA	
CS03	VC_{HYST}	Cap Hysteresis	High	—	525	—	mV	VC_{TH}-VC_{TL}
			Medium	—	375	—	mV	
			Low	—	280	—	mV	

* These parameters are characterized but not tested.

† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 25-22: CAP SENSE OSCILLATOR



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26.0 DC AND AC CHARACTERISTICS GRAPHS AND CHARTS

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore, outside the warranted range.

"Typical" represents the mean of the distribution at 25°C. "Maximum" or "minimum" represents (mean + 3σ) or (mean - 3σ) respectively, where σ is a standard deviation, over the whole temperature range.

FIGURE 26-1: PIC16F707 MAXIMUM IDD vs. FOSC OVER VDD, EC MODE, VCAP = 0.1µF

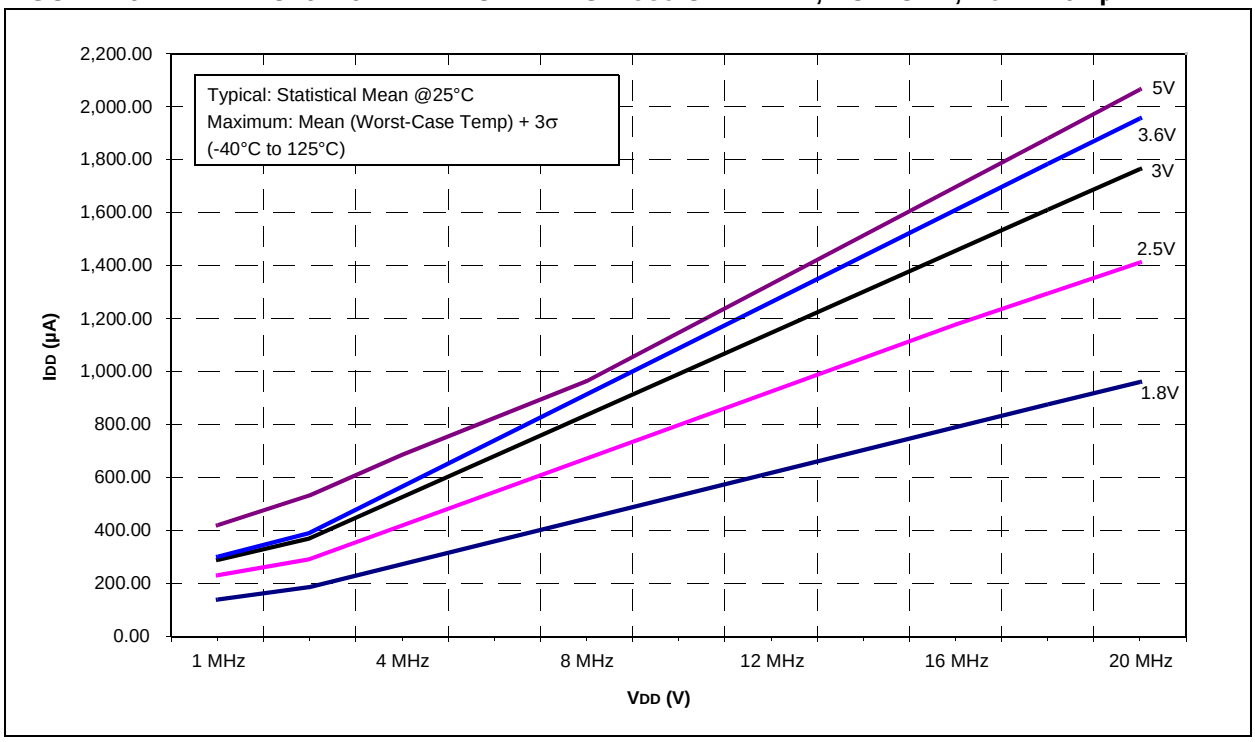

FIGURE 26-2: PIC16LF707 MAXIMUM IDD vs. FOSC OVER VDD, EC MODE

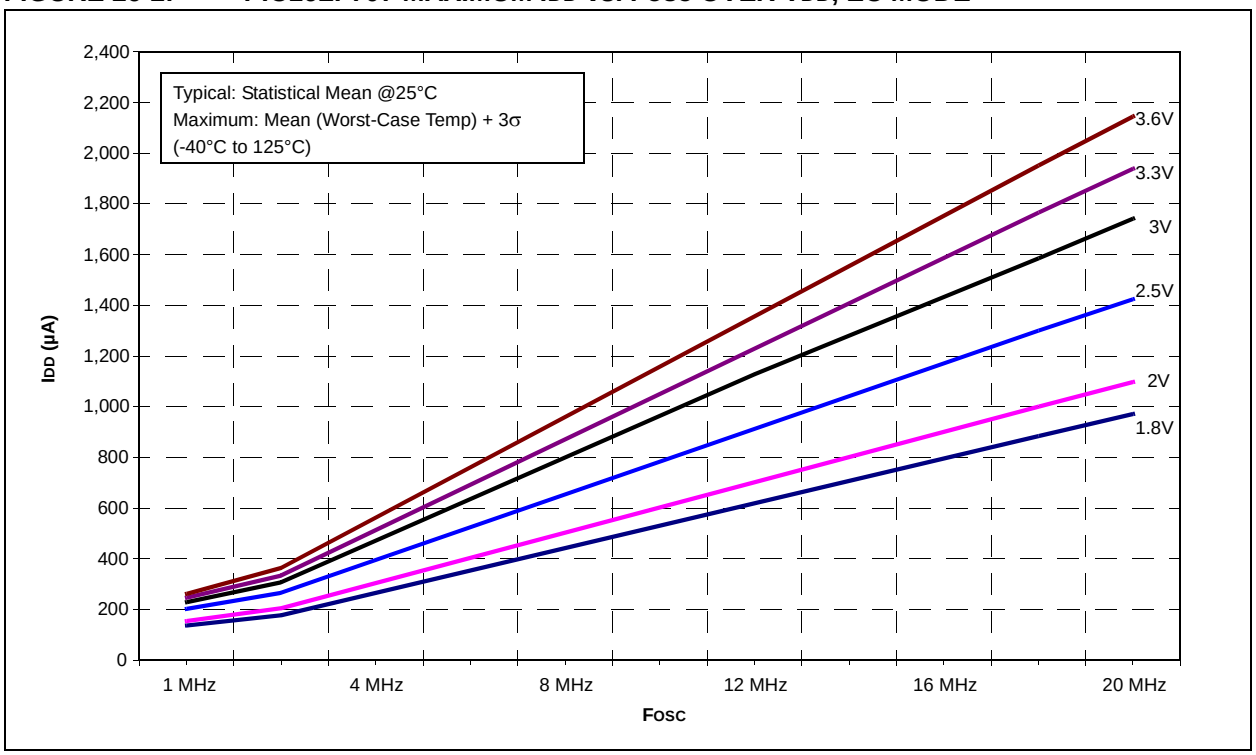

FIGURE 26-3: PIC16F707 TYPICAL IDD vs. FOSC OVER VDD, EC MODE, VCAP = 0.1µF

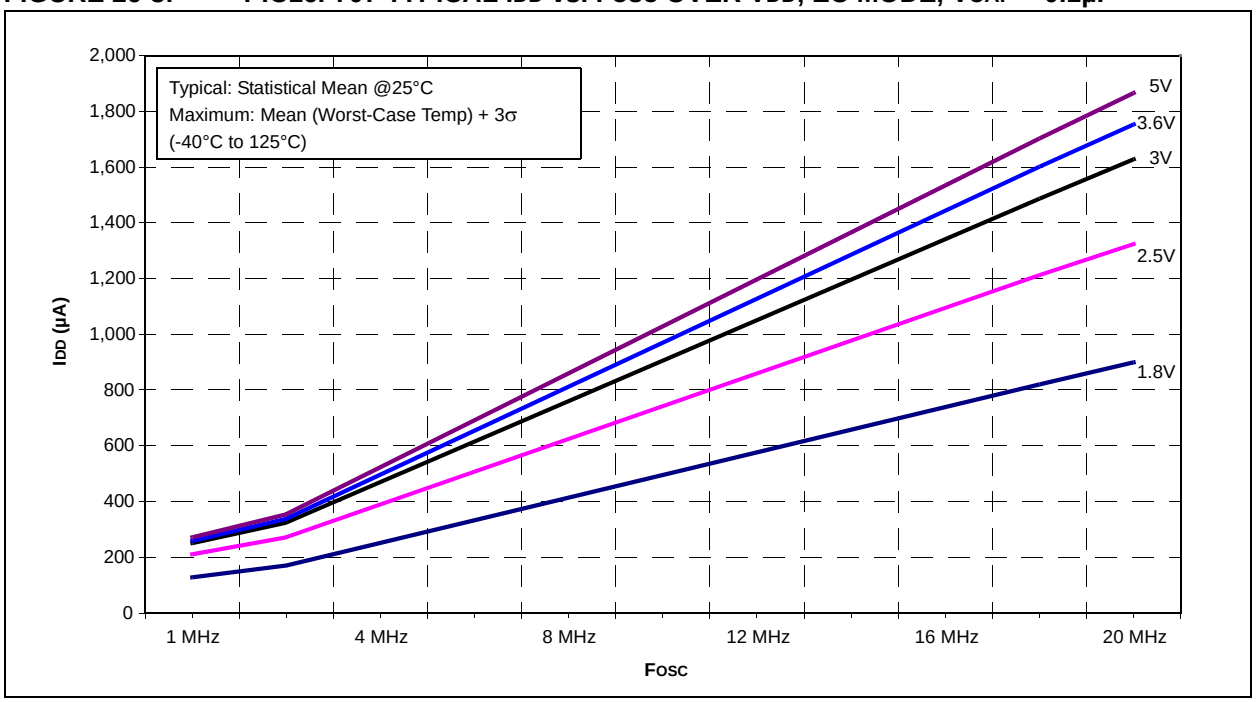

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FIGURE 26-4: PIC16LF707 TYPICAL IDD vs. FOSC OVER VDD, EC MODE

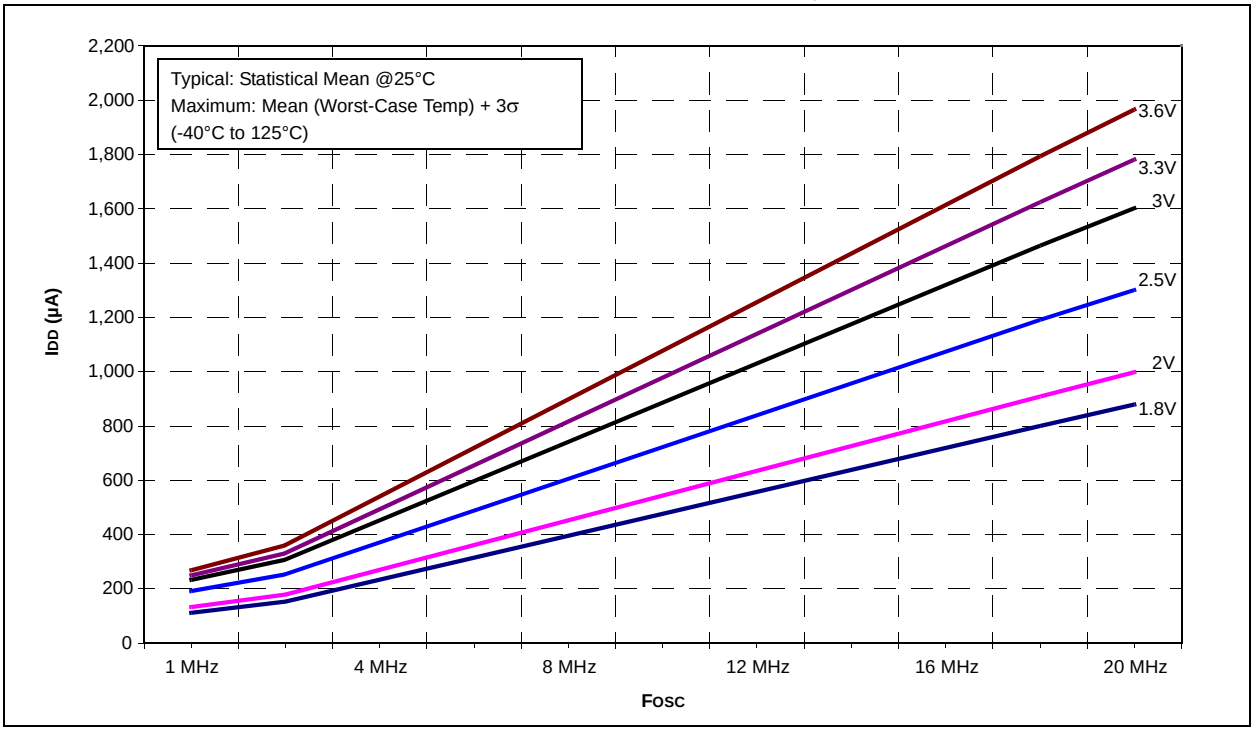

FIGURE 26-5: PIC16F707 MAXIMUM IDD vs. VDD OVER FOSC, EXTRC MODE, VCAP = 0.1µF

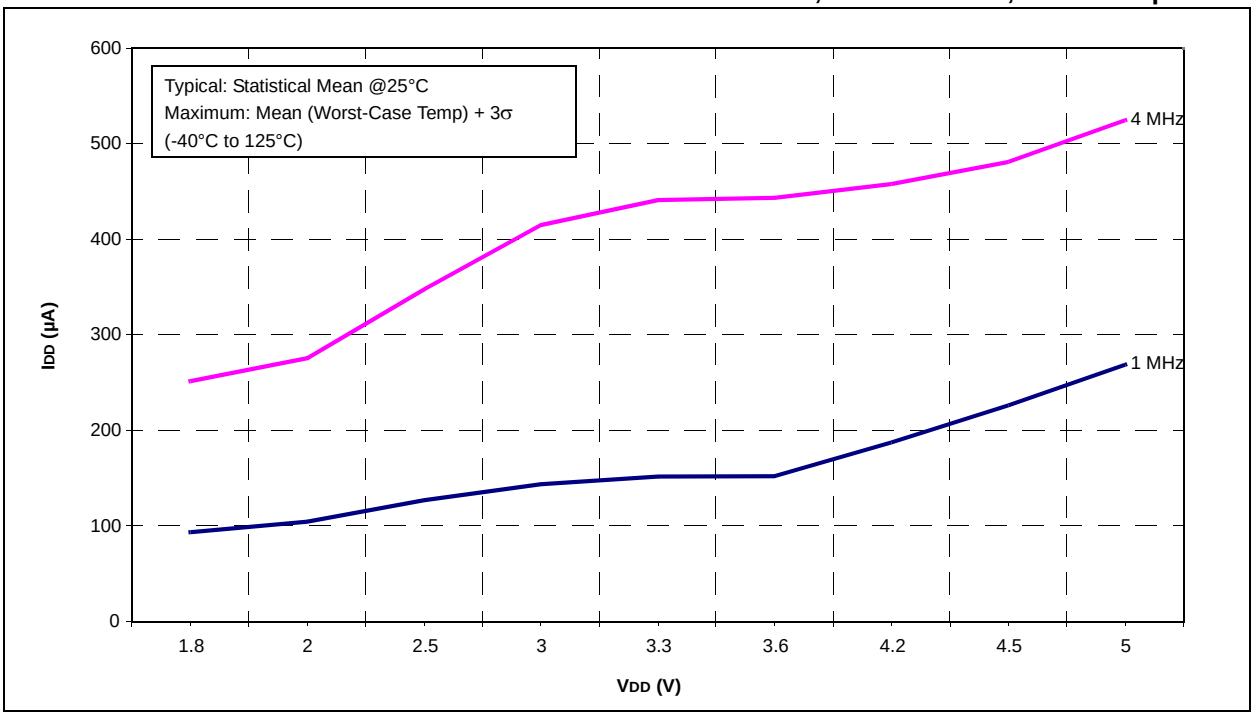

FIGURE 26-6: PIC16LF707 MAXIMUM IDD vs. VDD OVER FOSC, EXTRC MODE

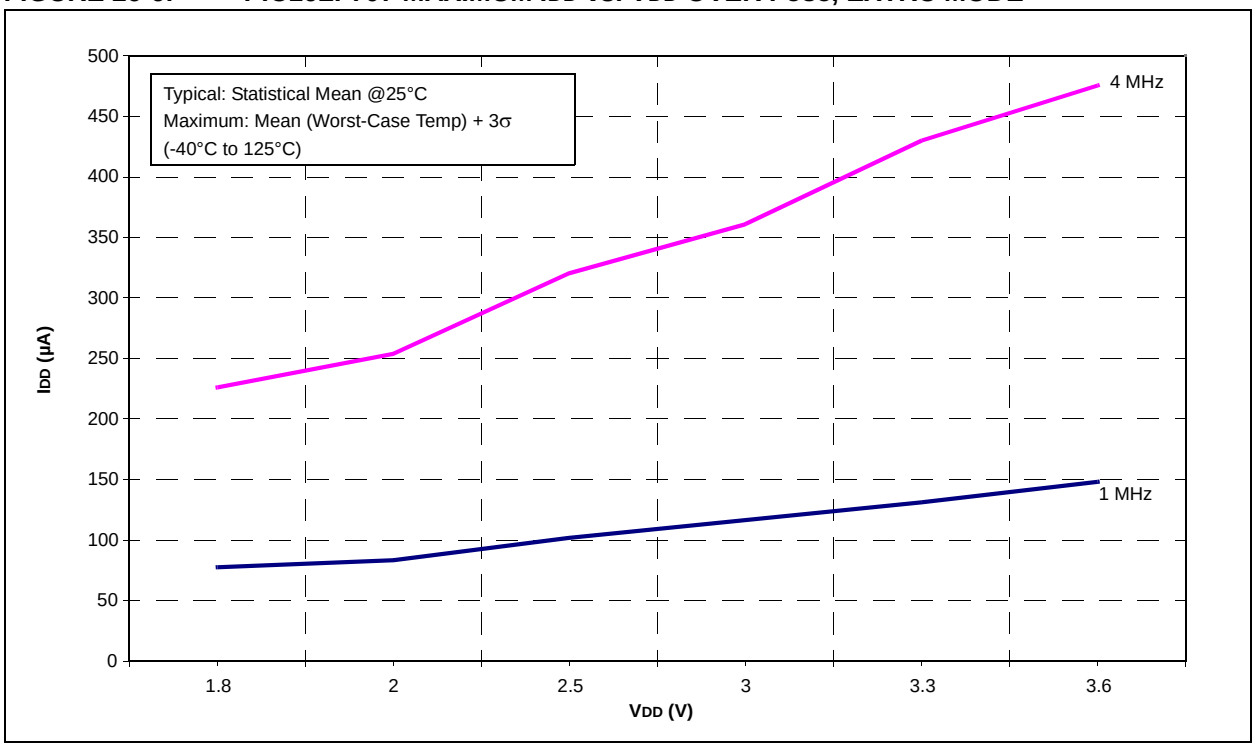

FIGURE 26-7: PIC16F707 TYPICAL IDD vs. VDD OVER FOSC, EXTRC MODE, VCAP = 0.1µF

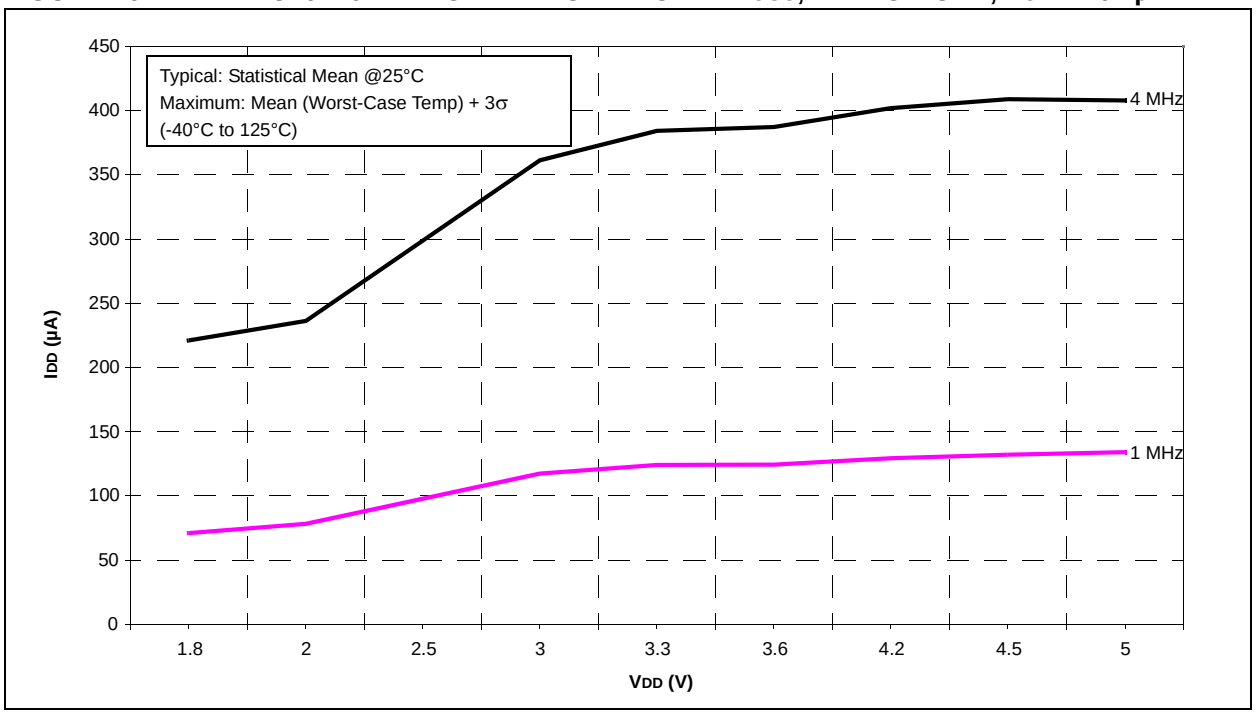

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FIGURE 26-8: PIC16LF707 TYPICAL IDD vs. VDD OVER FOSC, EXTRC MODE

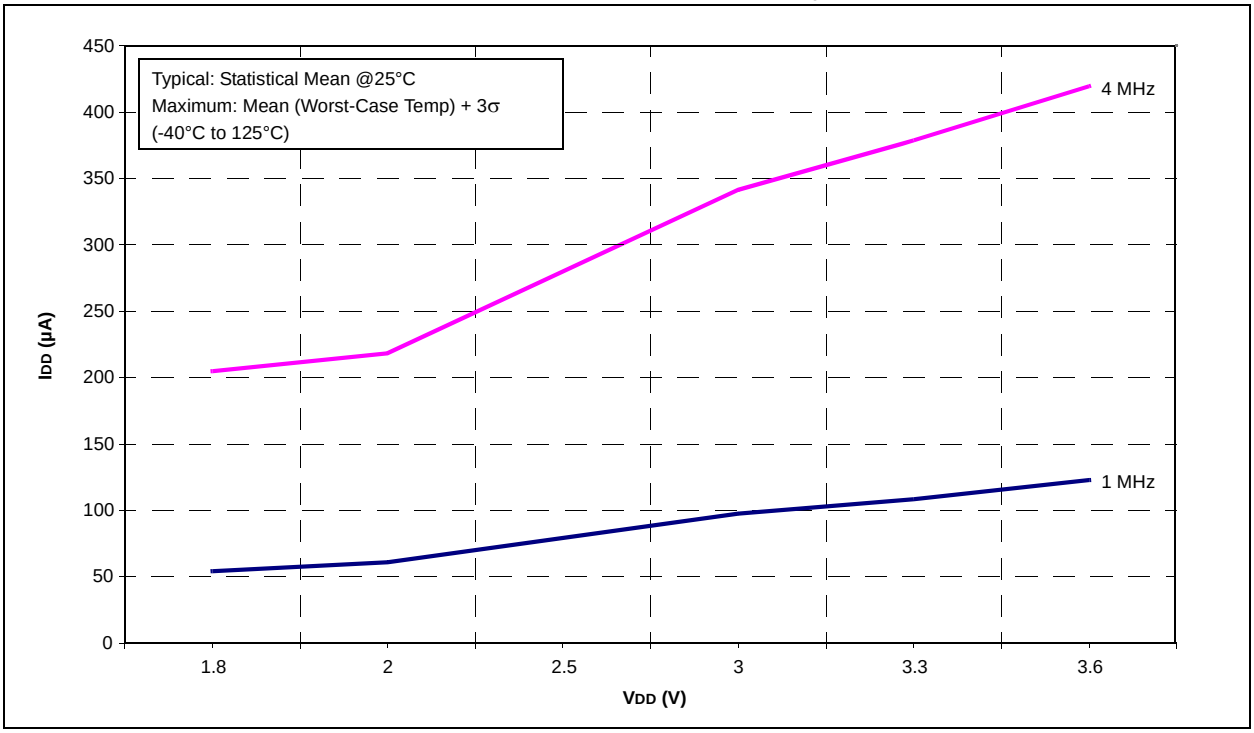

FIGURE 26-9: PIC16F707 MAXIMUM IDD vs. FOSC OVER VDD, HS MODE, VCAP = 0.1µF

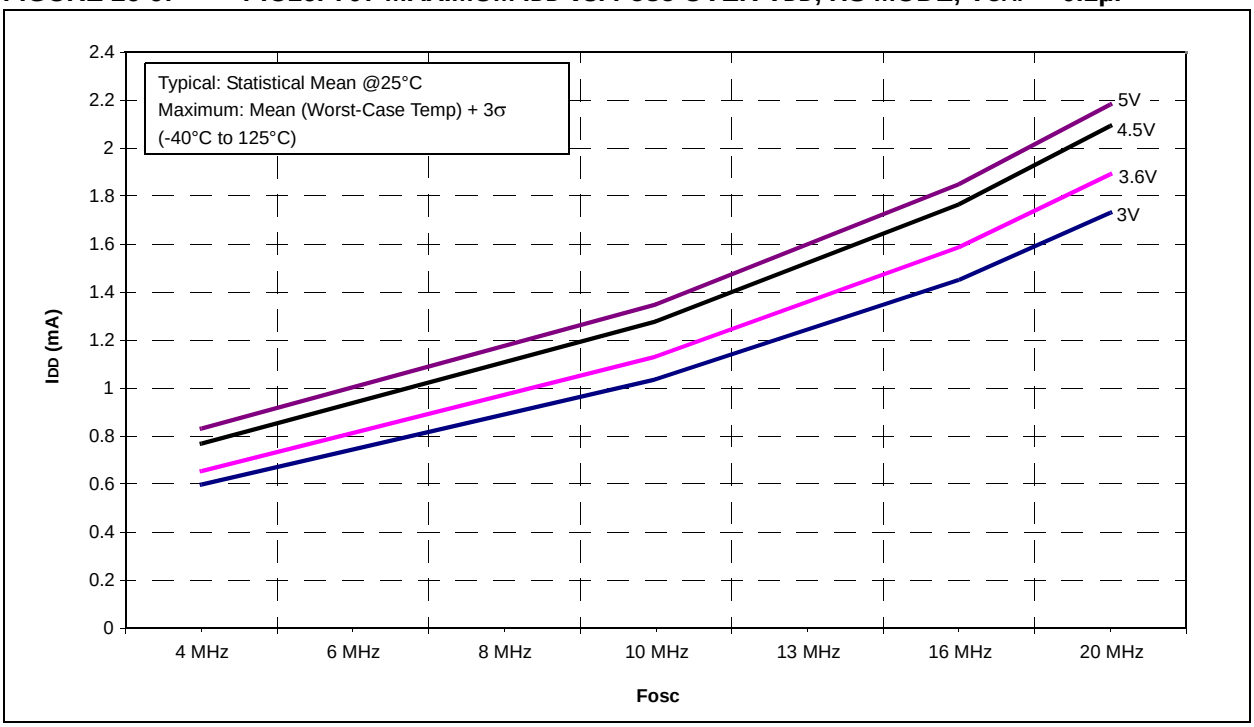

FIGURE 26-10: PIC16LF707 MAXIMUM IDD vs. FOSC OVER VDD, HS MODE

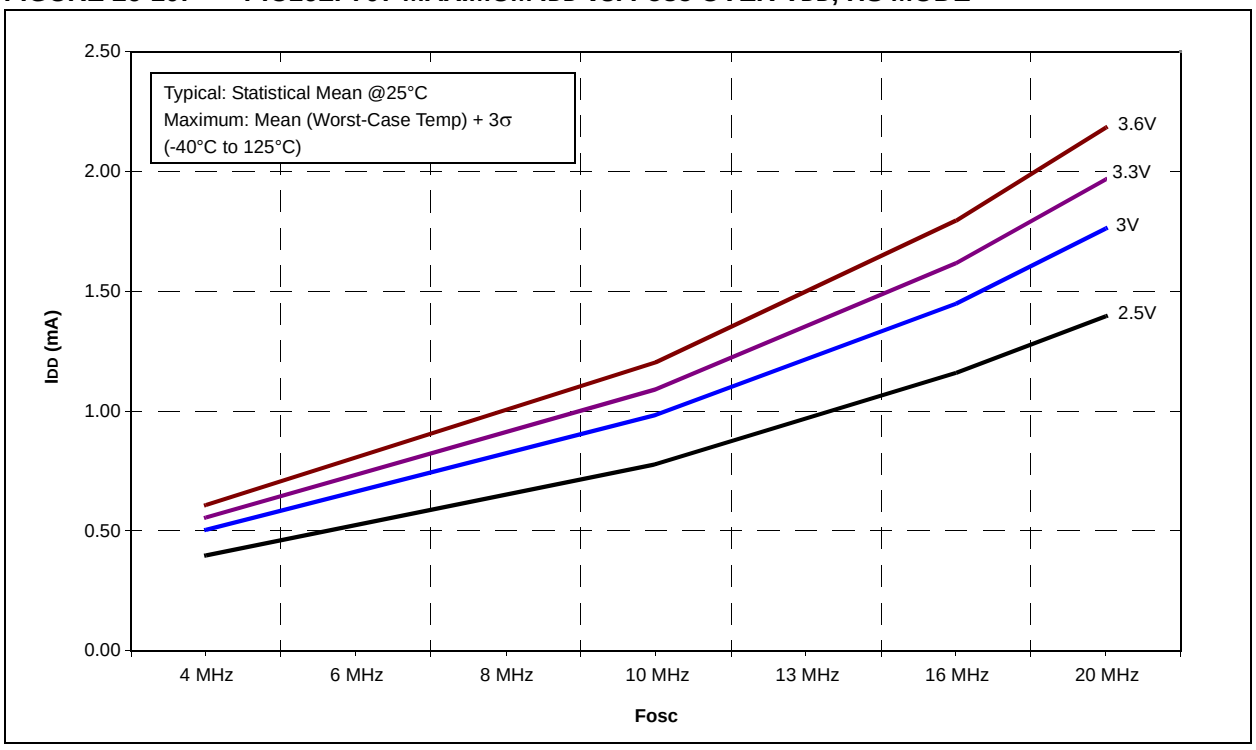

FIGURE 26-11: PIC16F707 TYPICAL IDD vs. FOSC OVER VDD, HS MODE, VCAP = 0.1µF

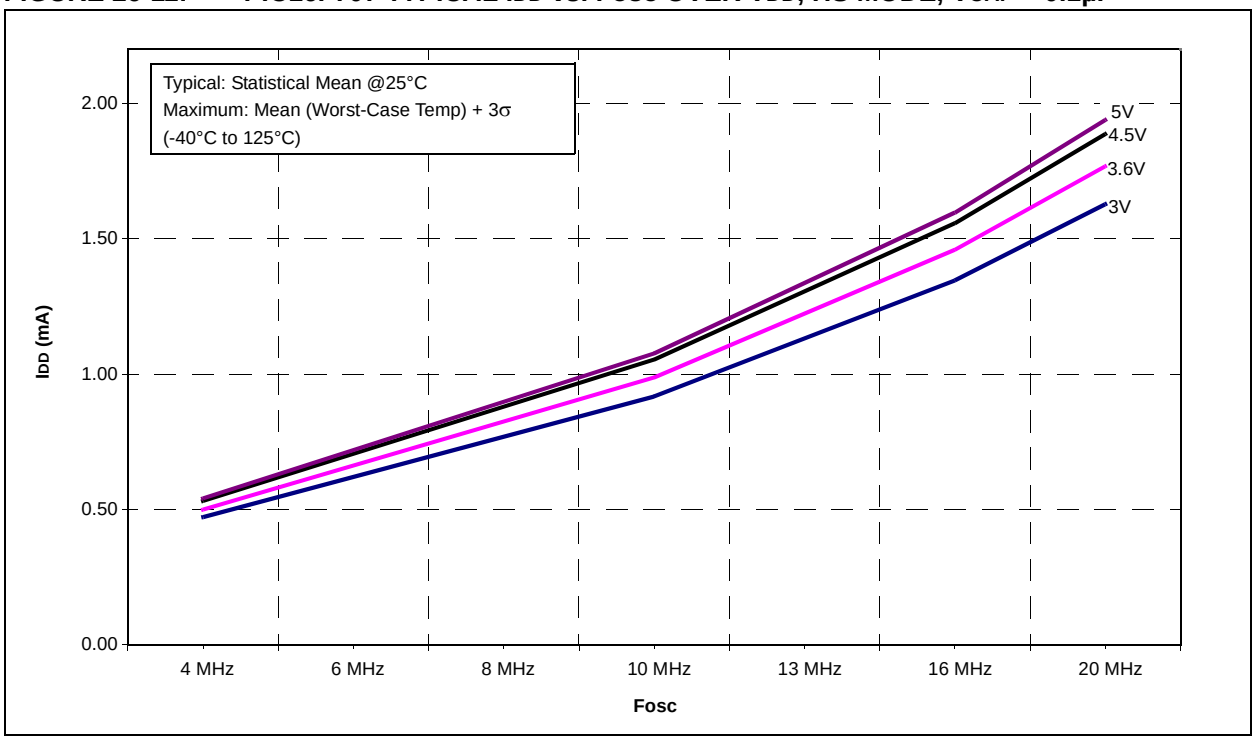

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FIGURE 26-12: PIC16LF707 TYPICAL IDD vs. FOSC OVER VDD, HS MODE

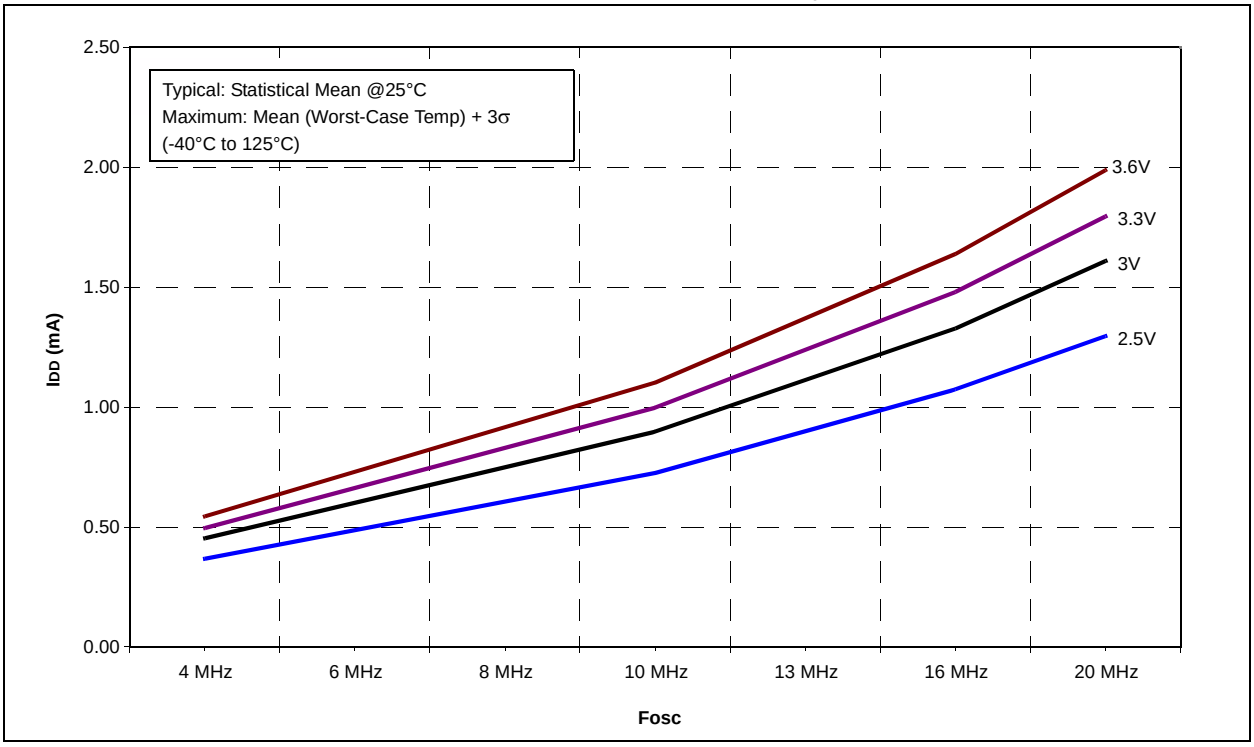

FIGURE 26-13: PIC16F707 MAXIMUM IDD vs. VDD OVER FOSC, XT MODE, VCAP = 0.1µF

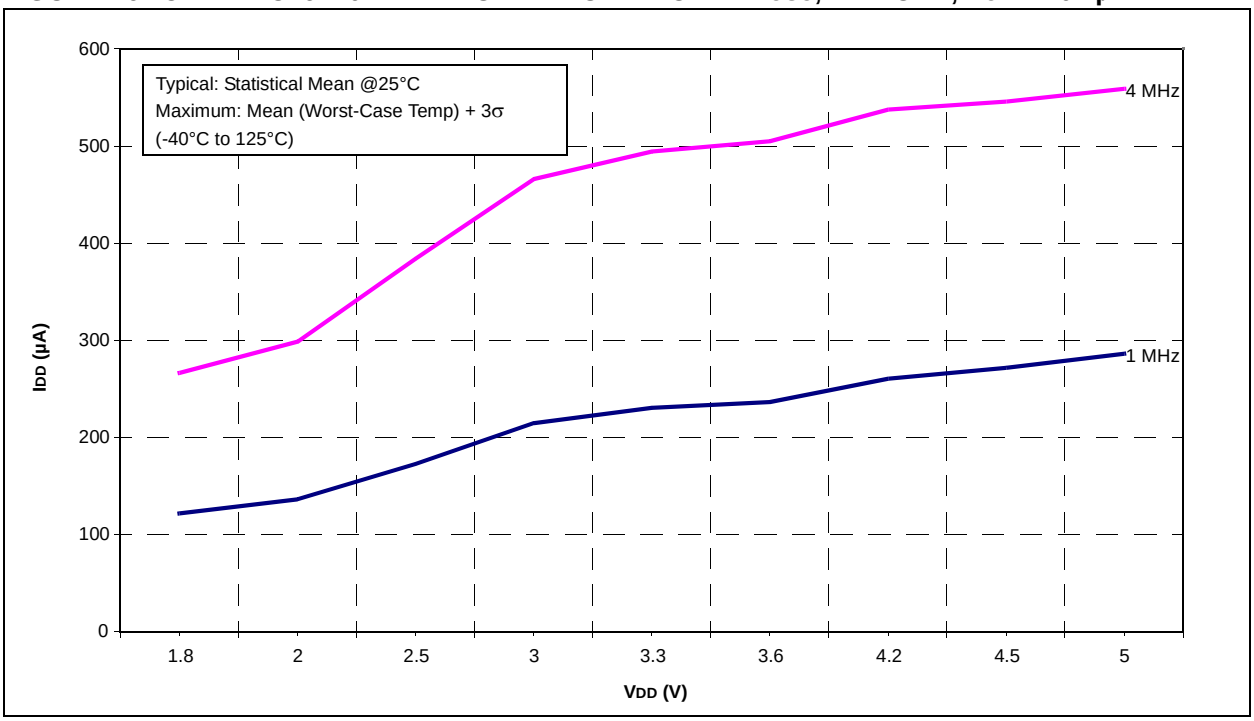

FIGURE 26-14: PIC16LF707 MAXIMUM IDD vs. VDD OVER FOSC, XT MODE

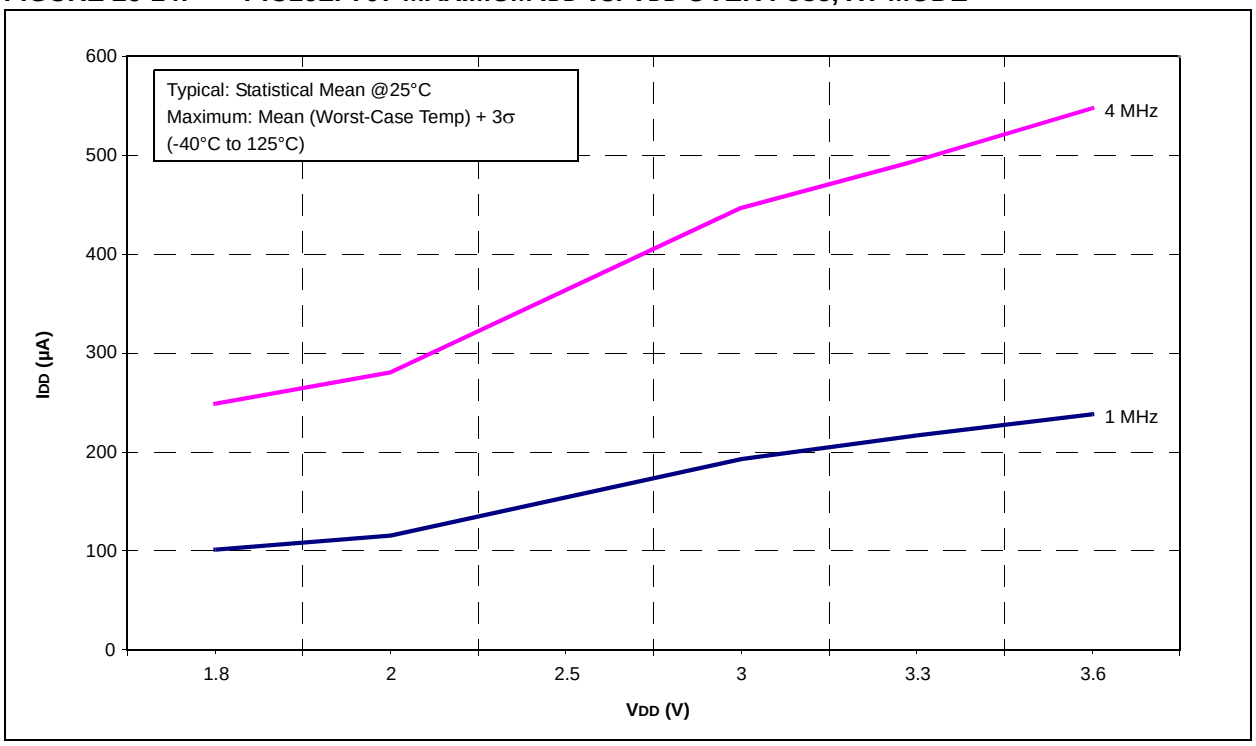

FIGURE 26-15: PIC16F707 TYPICAL IDD vs. VDD OVER FOSC, XT MODE, VCAP = 0.1µF

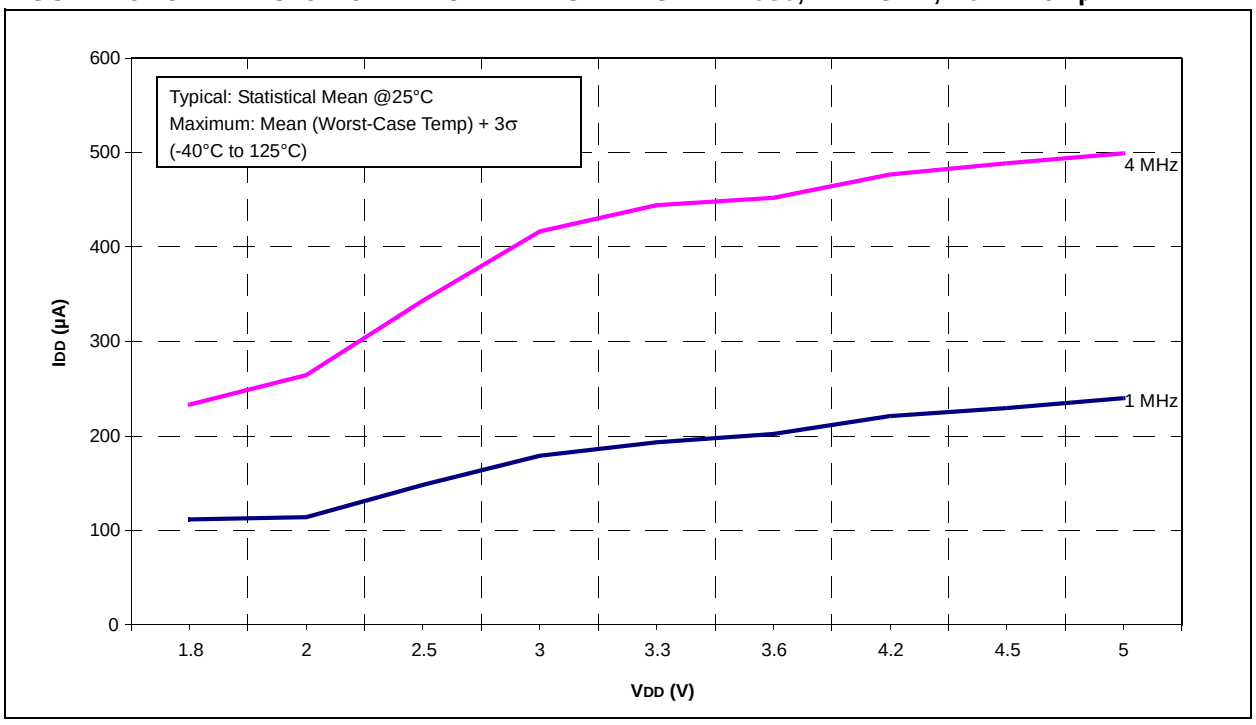

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FIGURE 26-16: PIC16LF707 TYPICAL IDD vs. VDD OVER FOSC, XT MODE

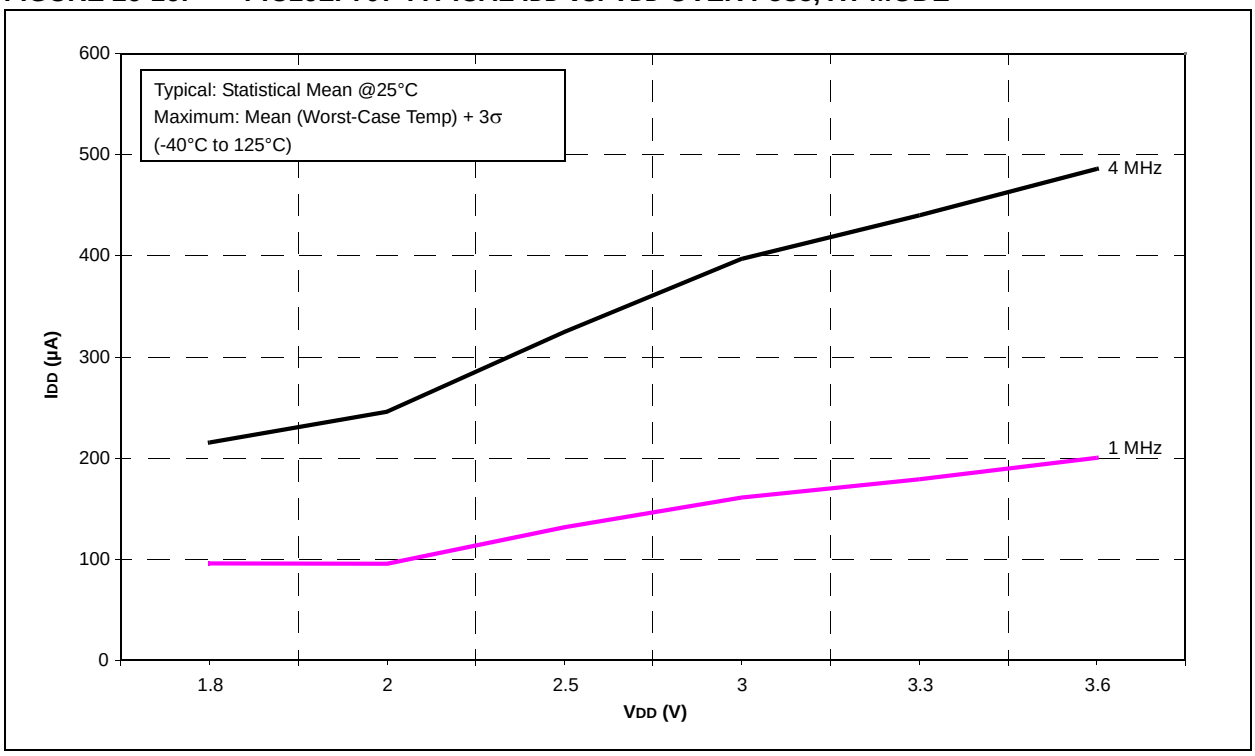

FIGURE 26-17: PIC16F707 IDD vs. VDD, LP MODE, VCAP = 0.1µF

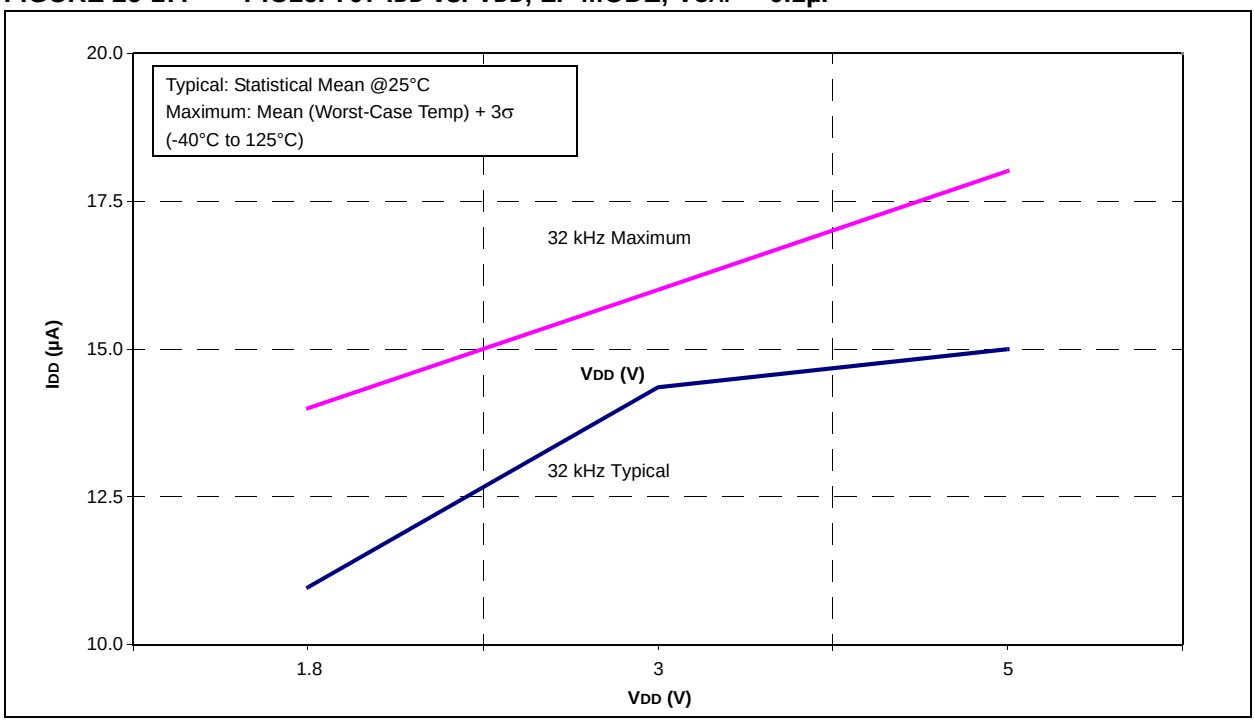

FIGURE 26-18: PIC16LF707 IDD vs. VDD, LP MODE

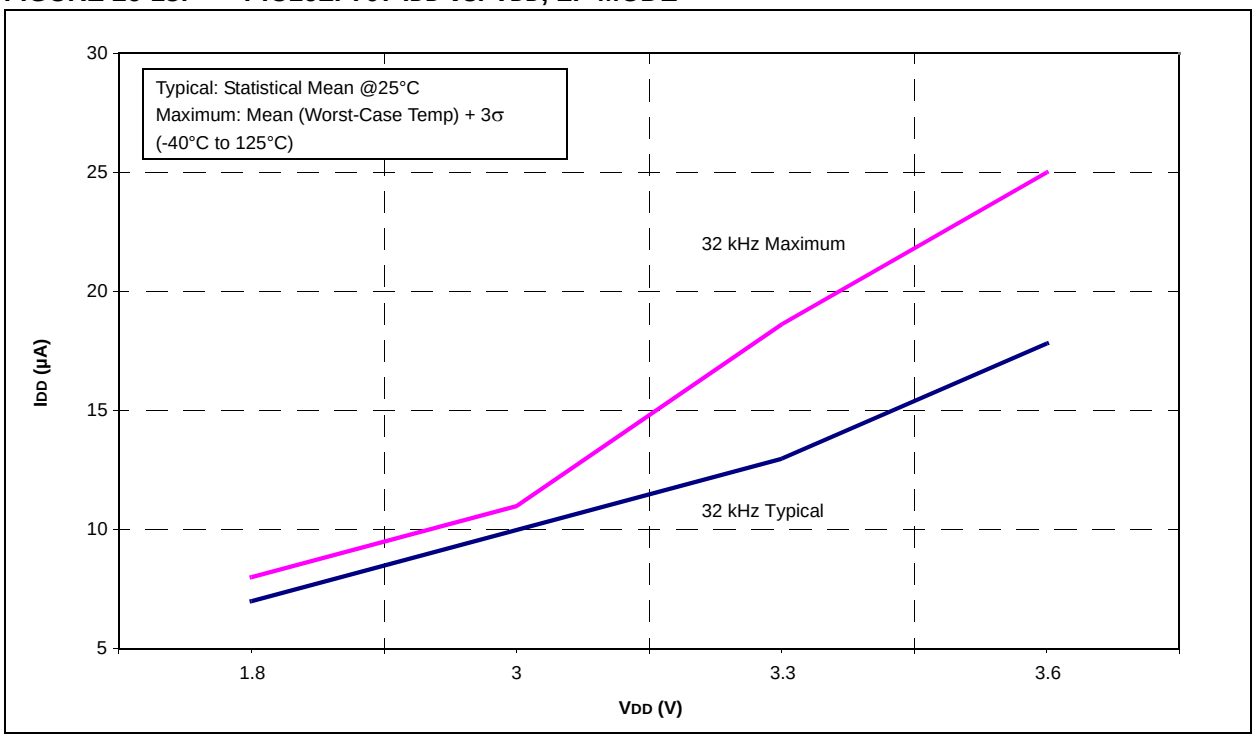

FIGURE 26-19: PIC16F707 MAXIMUM IDD vs. FOSC OVER VDD, INTOSC MODE, VCAP = 0.1µF

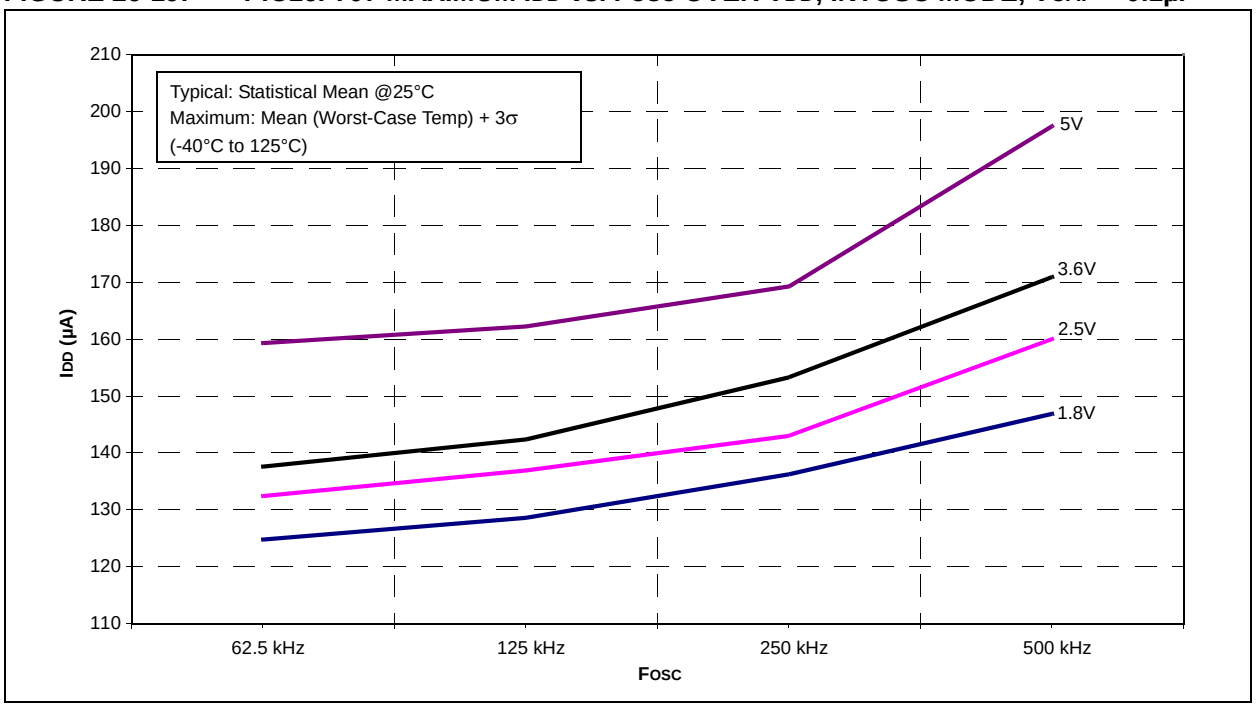

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FIGURE 26-20: PIC16LF707 MAXIMUM IDD vs. FOSC OVER VDD, INTOSC MODE

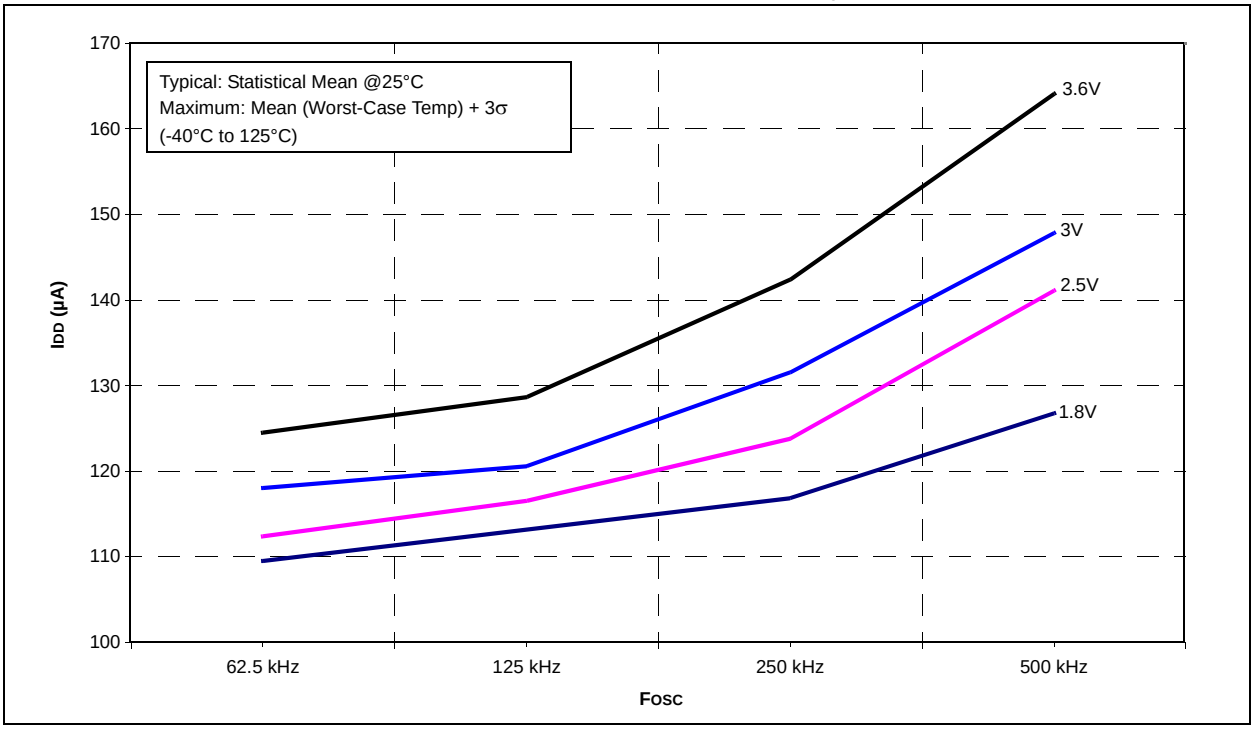

FIGURE 26-21: PIC16F707 MAXIMUM IDD vs. FOSC OVER VDD, INTOSC MODE, VCAP = 0.1µF

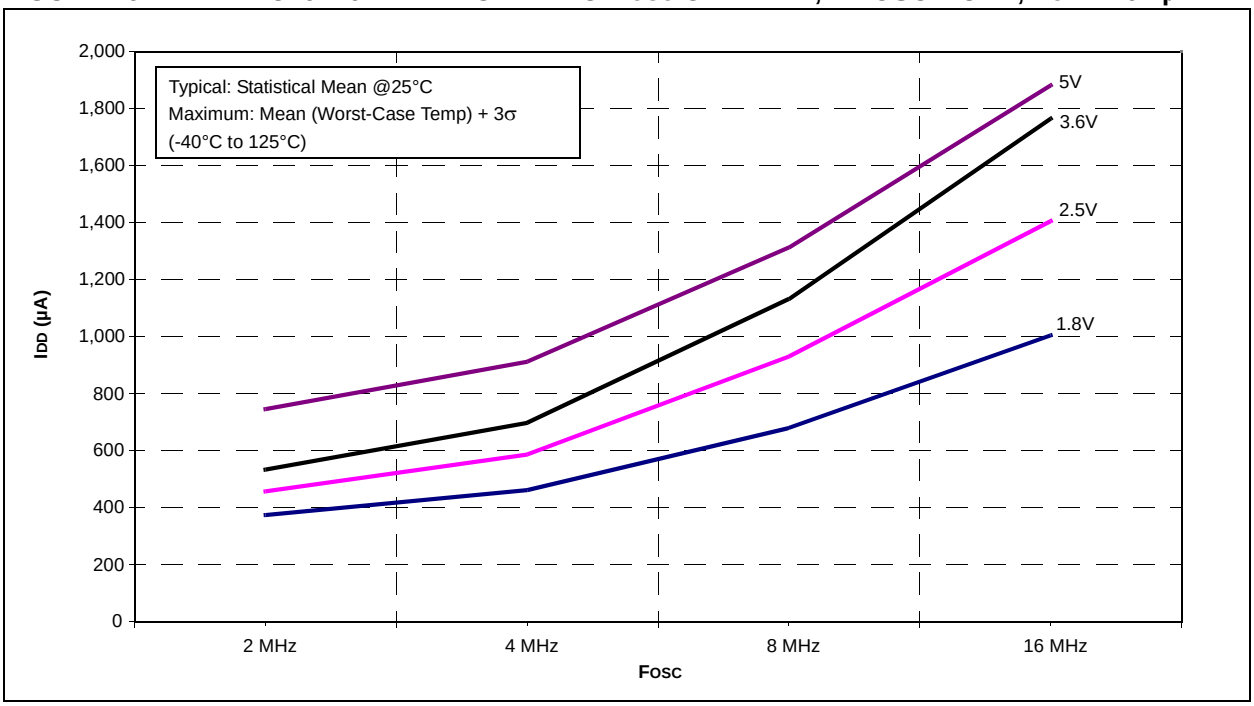

FIGURE 26-22: PIC16LF707 MAXIMUM IDD vs. FOSC OVER VDD, INTOSC MODE

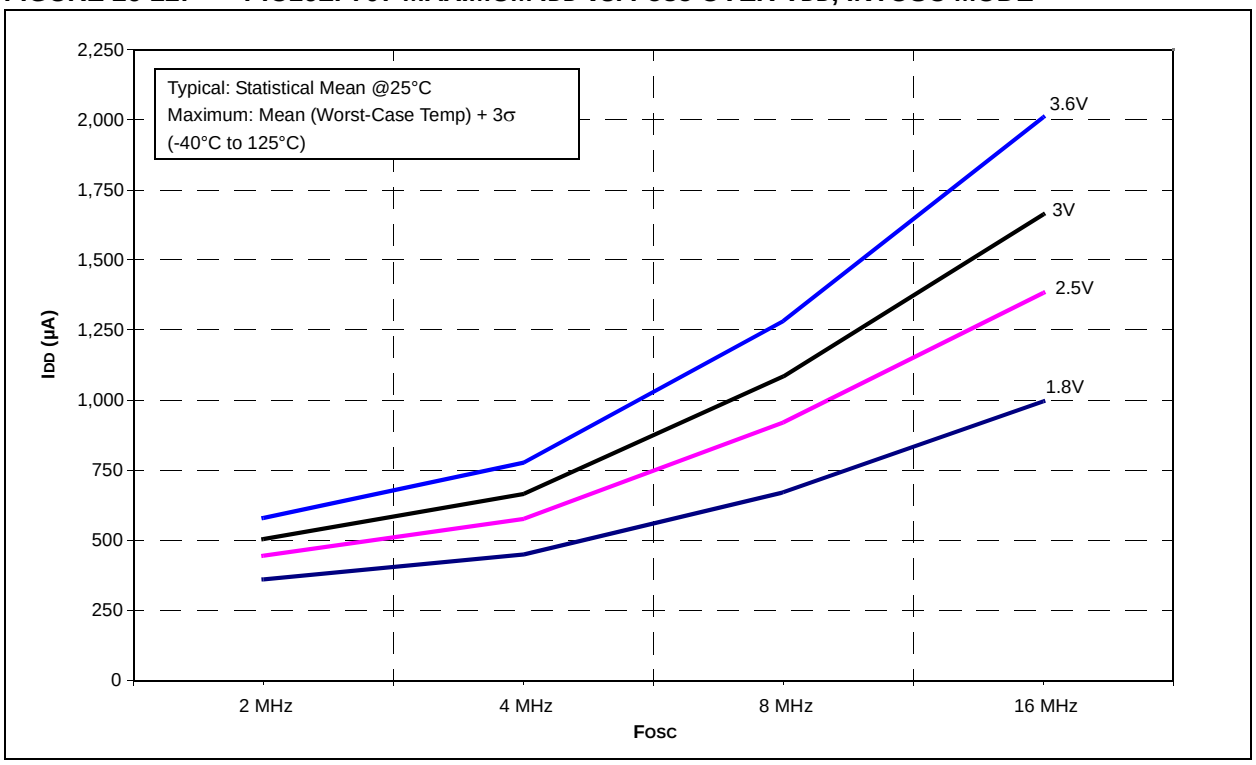

FIGURE 26-23: PIC16F707 TYPICAL IDD vs. FOSC OVER VDD, INTOSC MODE, VCAP = 0.1µF

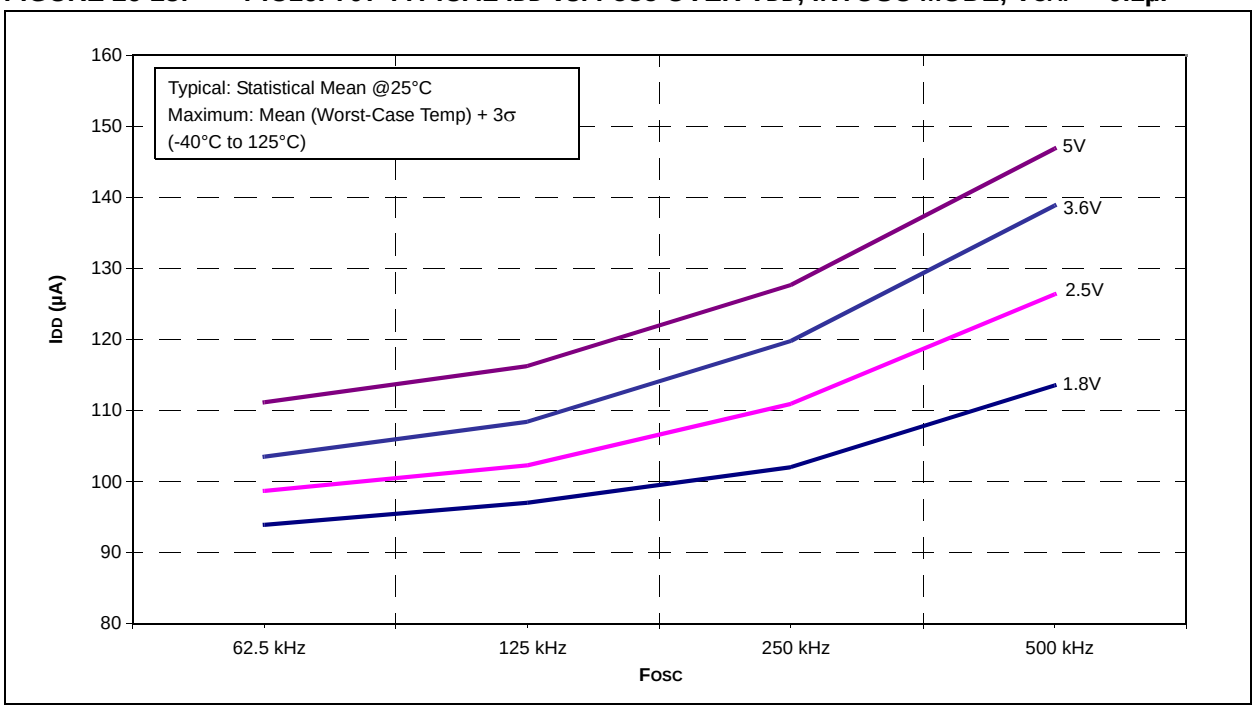

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FIGURE 26-24: PIC16LF707 TYPICAL IDD vs. FOSC OVER VDD, INTOSC MODE

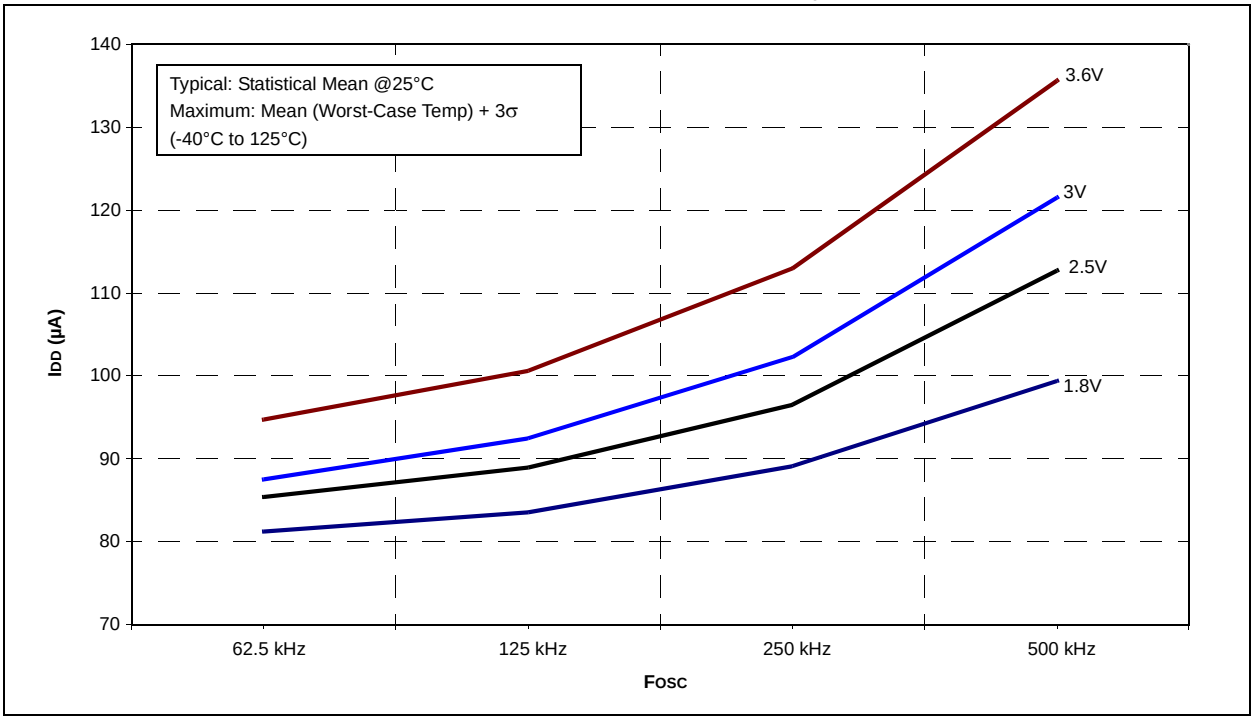

FIGURE 26-25: PIC16F707 TYPICAL IDD vs. FOSC OVER VDD, INTOSC MODE, VCAP = 0.1µF

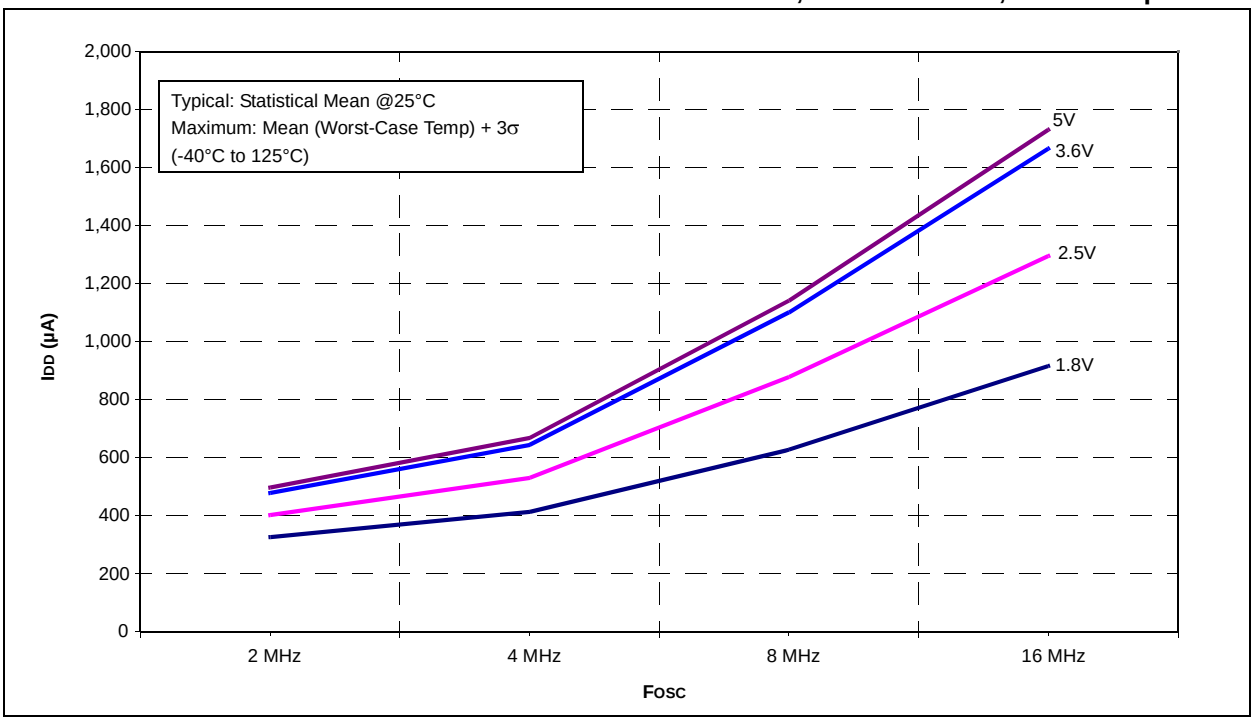

FIGURE 26-26: PIC16LF707 TYPICAL IDD vs. FOSC OVER VDD, INTOSC MODE

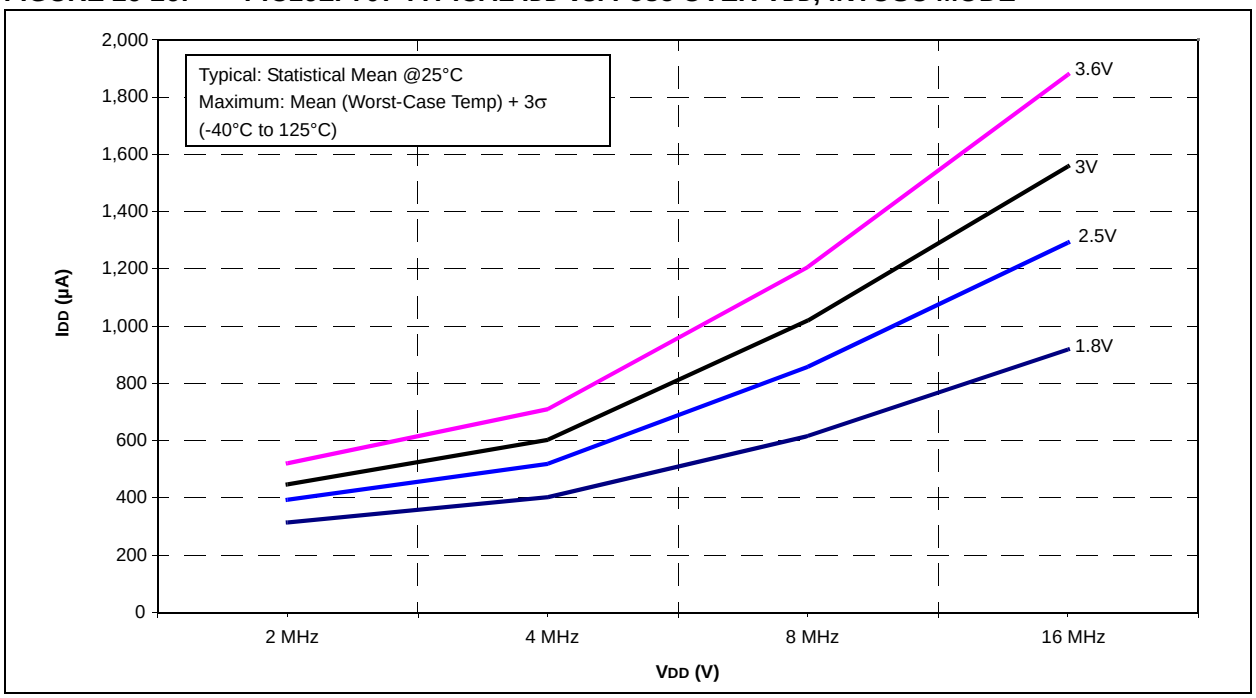

FIGURE 26-27: PIC16F707 MAXIMUM BASE IPD vs. VDD, VCAP = 0.1µF

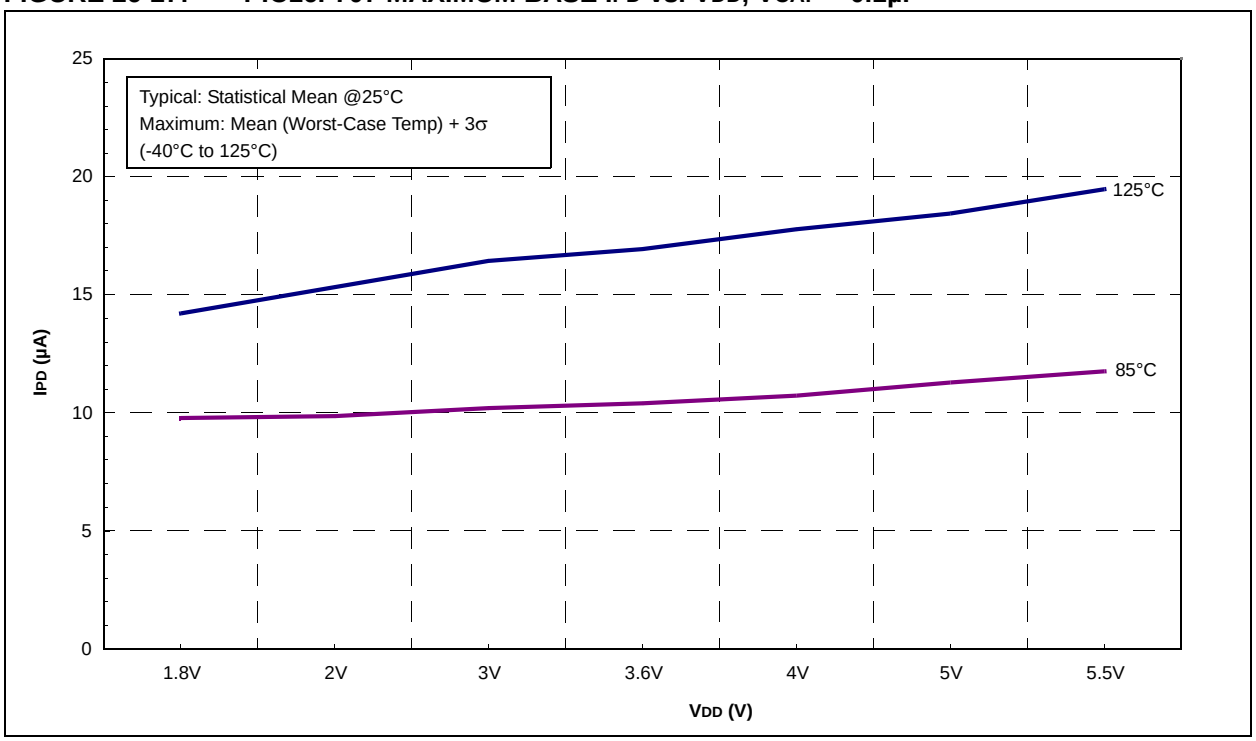

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FIGURE 26-28: PIC16LF707 MAXIMUM BASE I_{PD} VS. V_{DD}

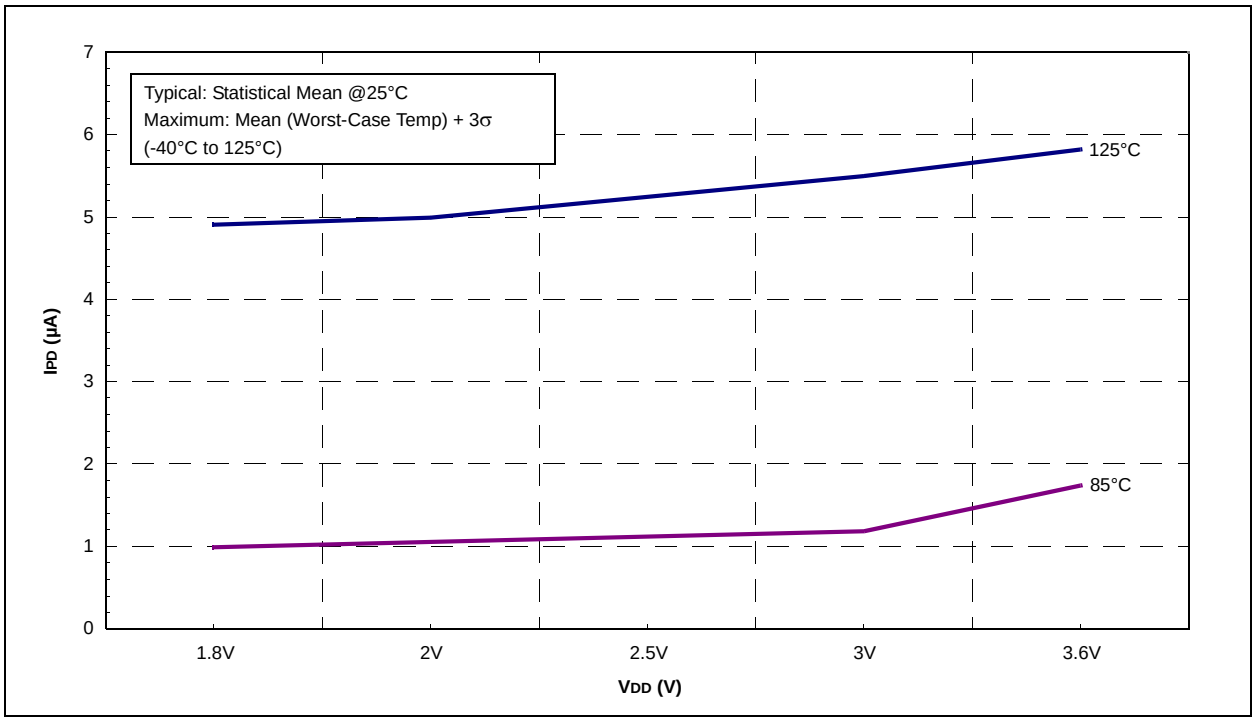

FIGURE 26-29: PIC16F707 TYPICAL BASE I_{PD} VS. V_{DD}, V_{CAP} = 0.1µF

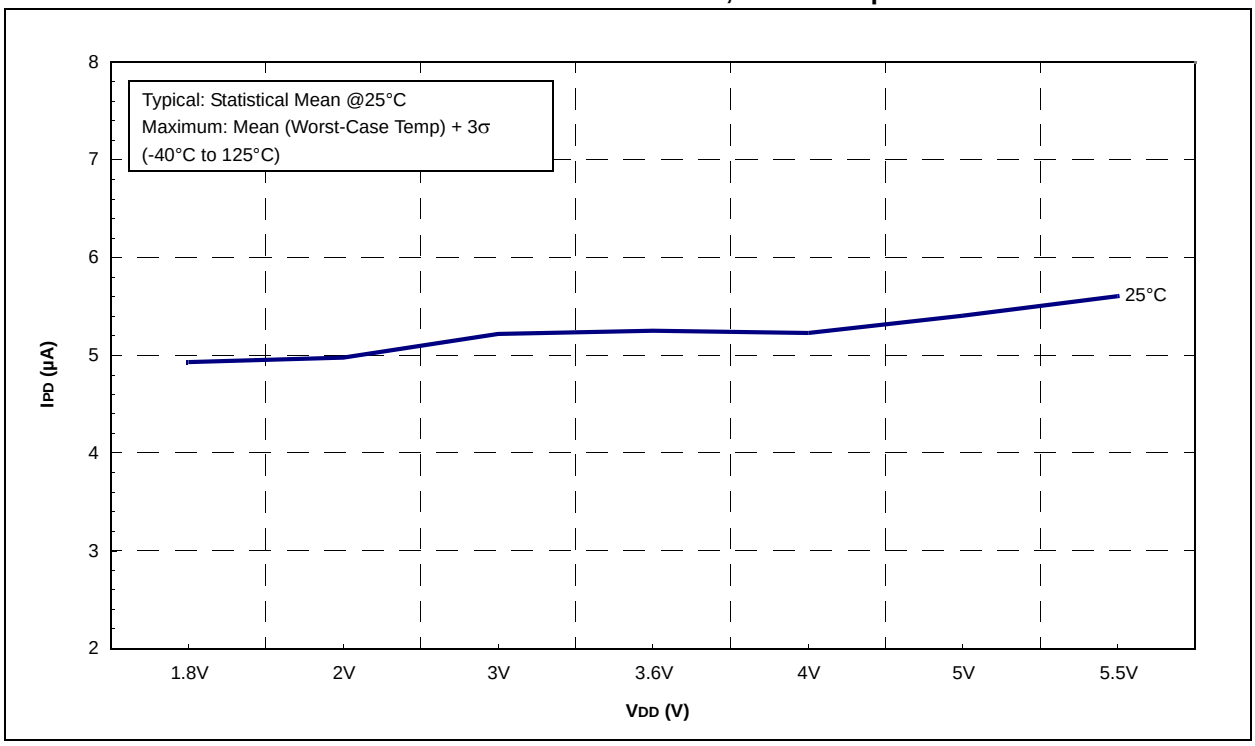

FIGURE 26-30: PIC16LF707 TYPICAL BASE IPD VS. VDD

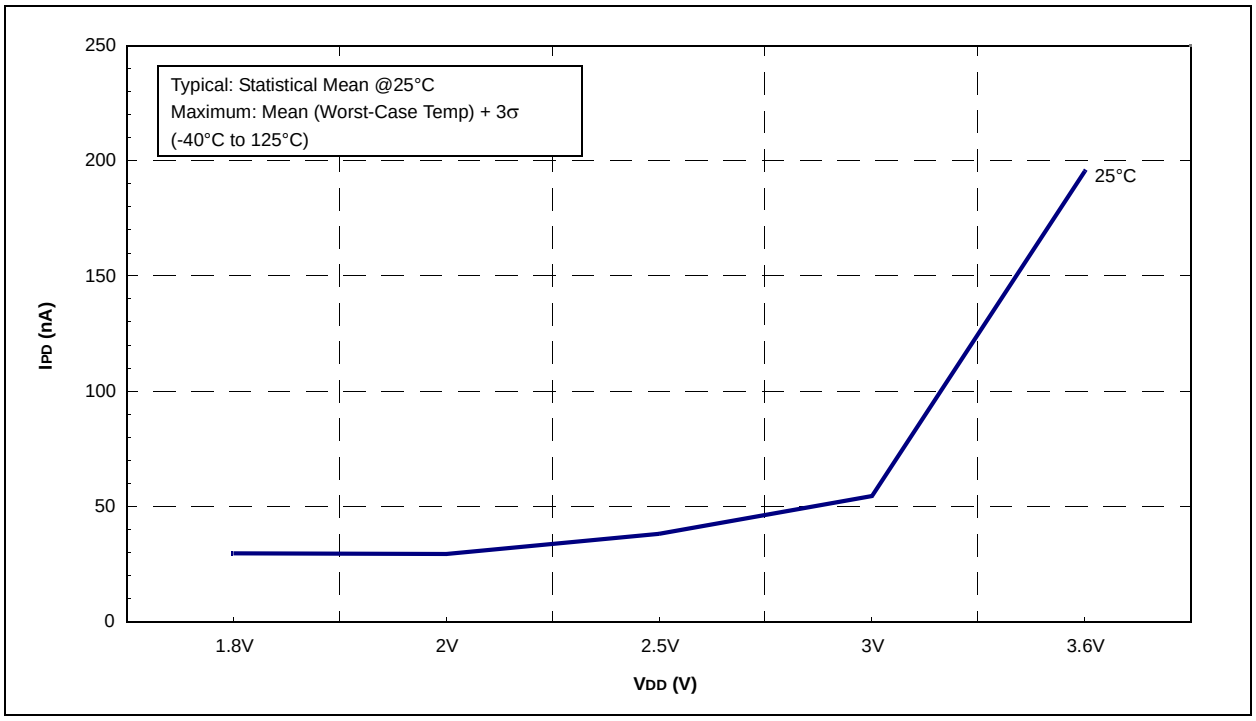

FIGURE 26-31: PIC16F707 FIXED VOLTAGE REFERENCE IPD VS. VDD, VCAP = 0.1µF

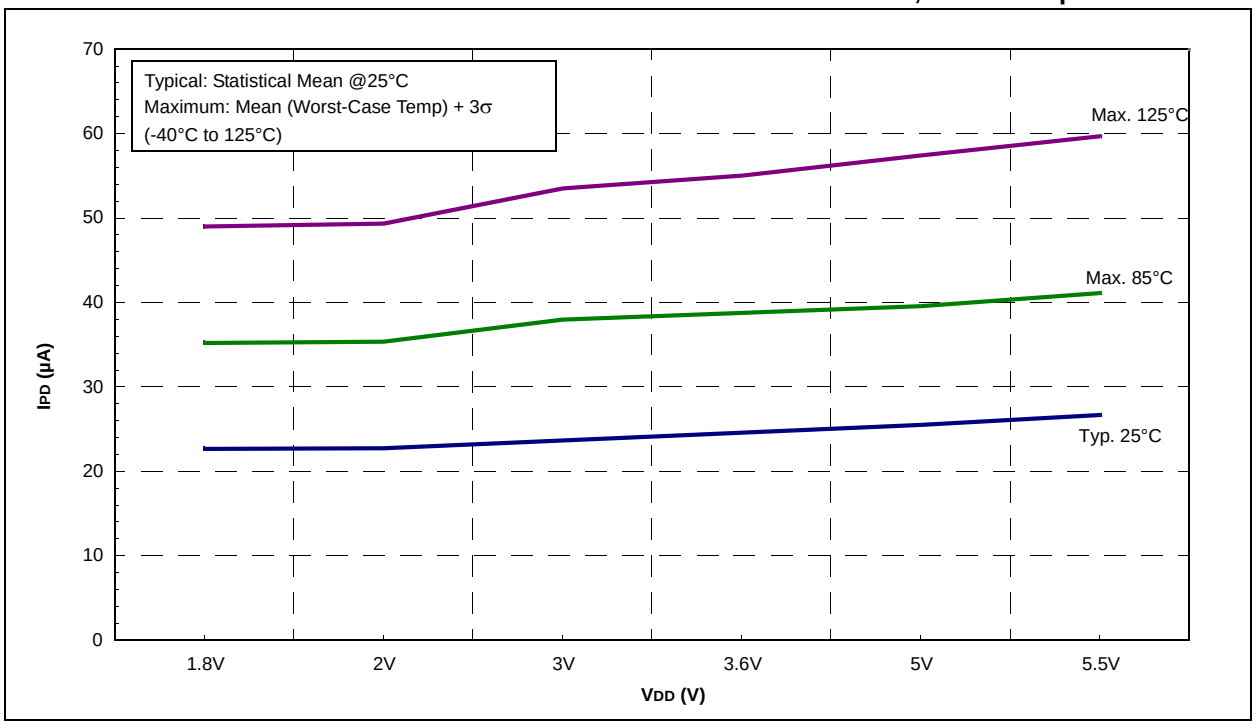

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FIGURE 26-32: PIC16LF707 FIXED VOLTAGE REFERENCE I_{PD} vs. V_{DD}

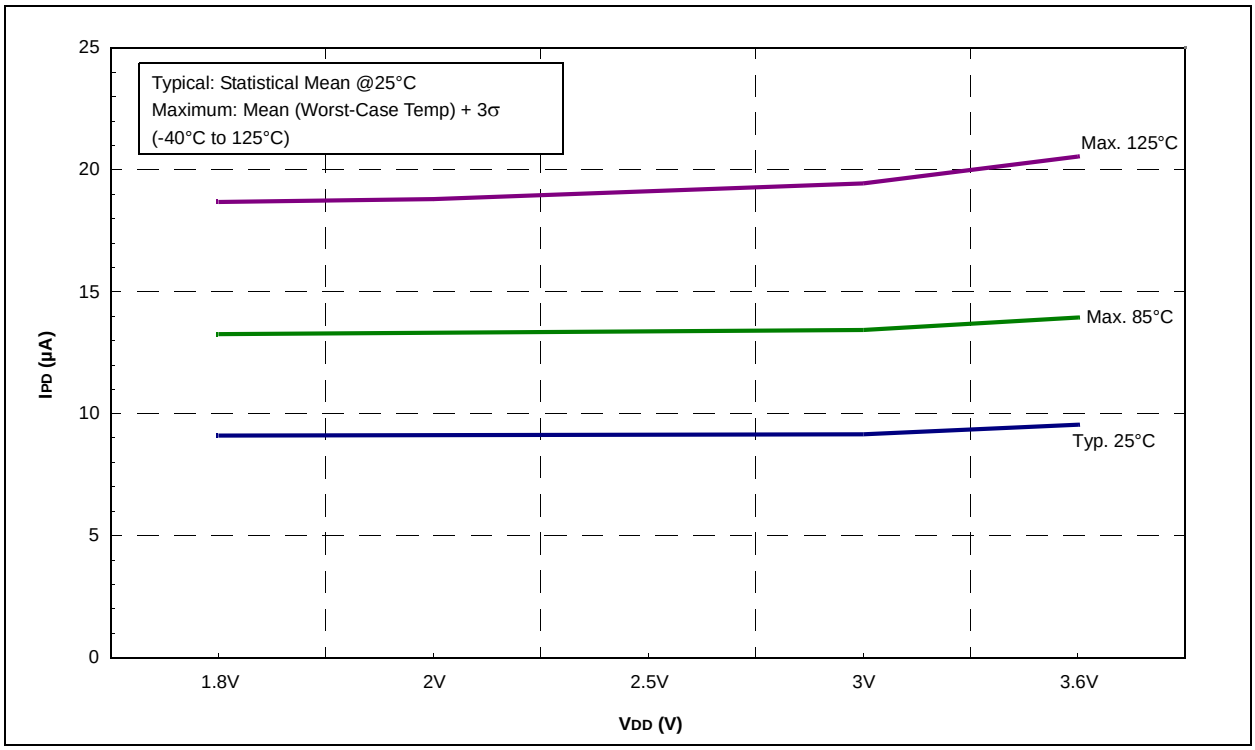

FIGURE 26-33: PIC16F707 BOR I_{PD} vs. V_{DD}, V_{CAP} = 0.1µF

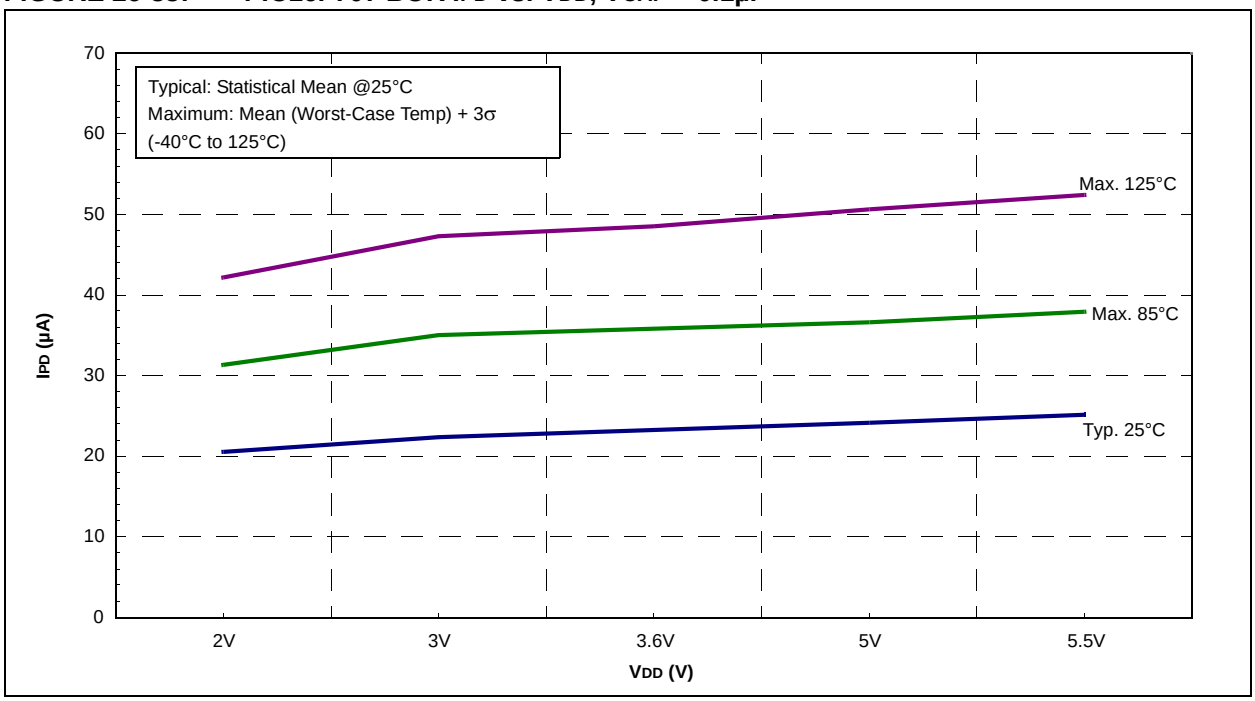

FIGURE 26-34: PIC16LF707 BOR IPD VS. VDD

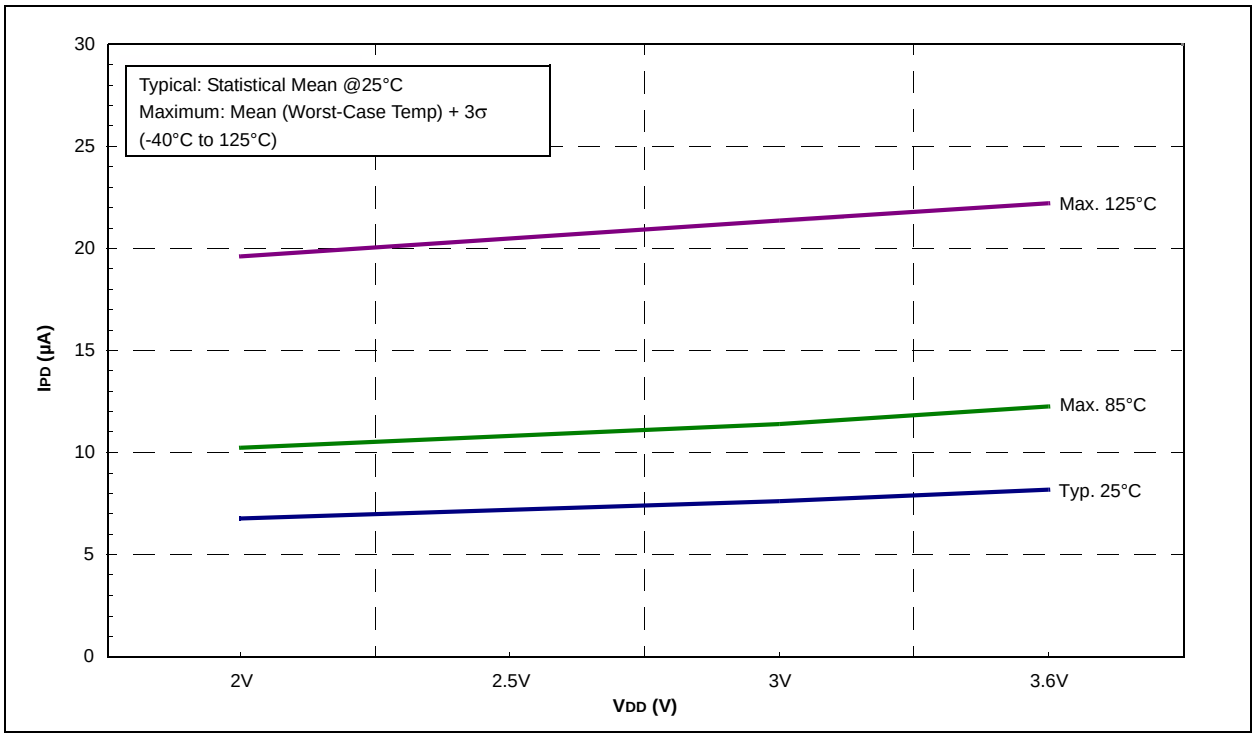

FIGURE 26-35: PIC16F707 CAP SENSE HIGH POWER IPD VS. VDD, VCAP = 0.1µF

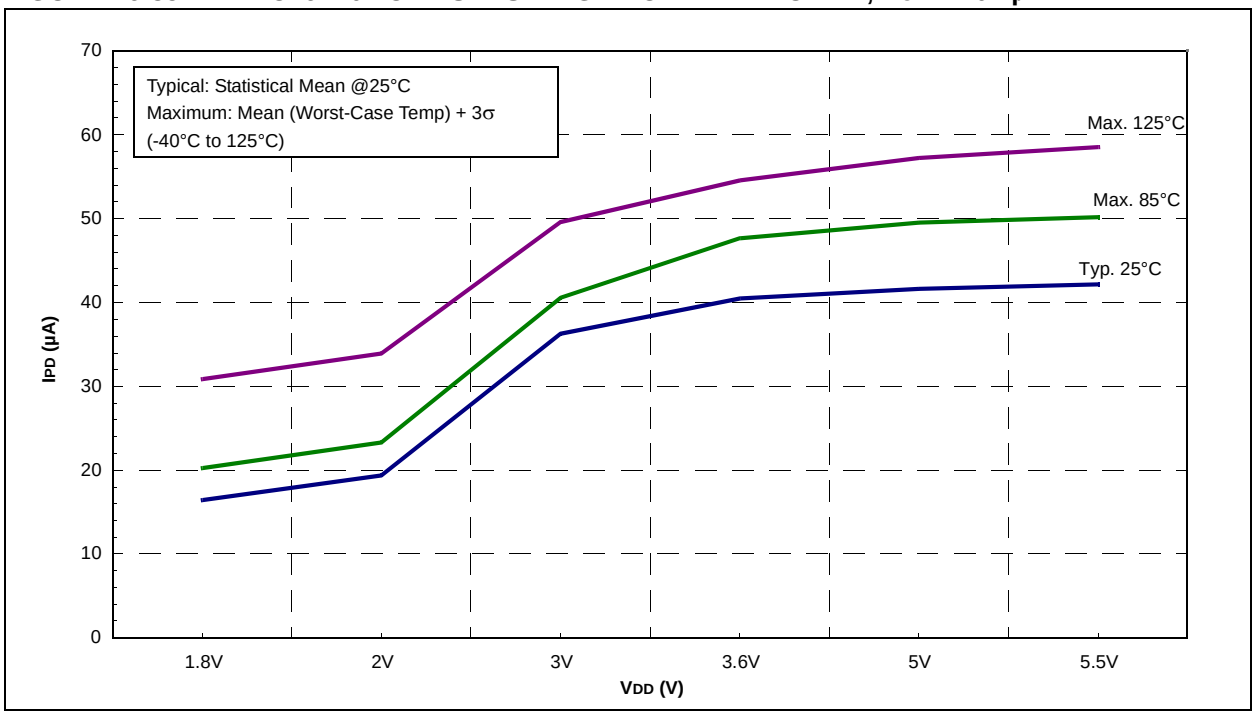

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FIGURE 26-36: PIC16LF707 CAP SENSE HIGH POWER IPD vs. VDD

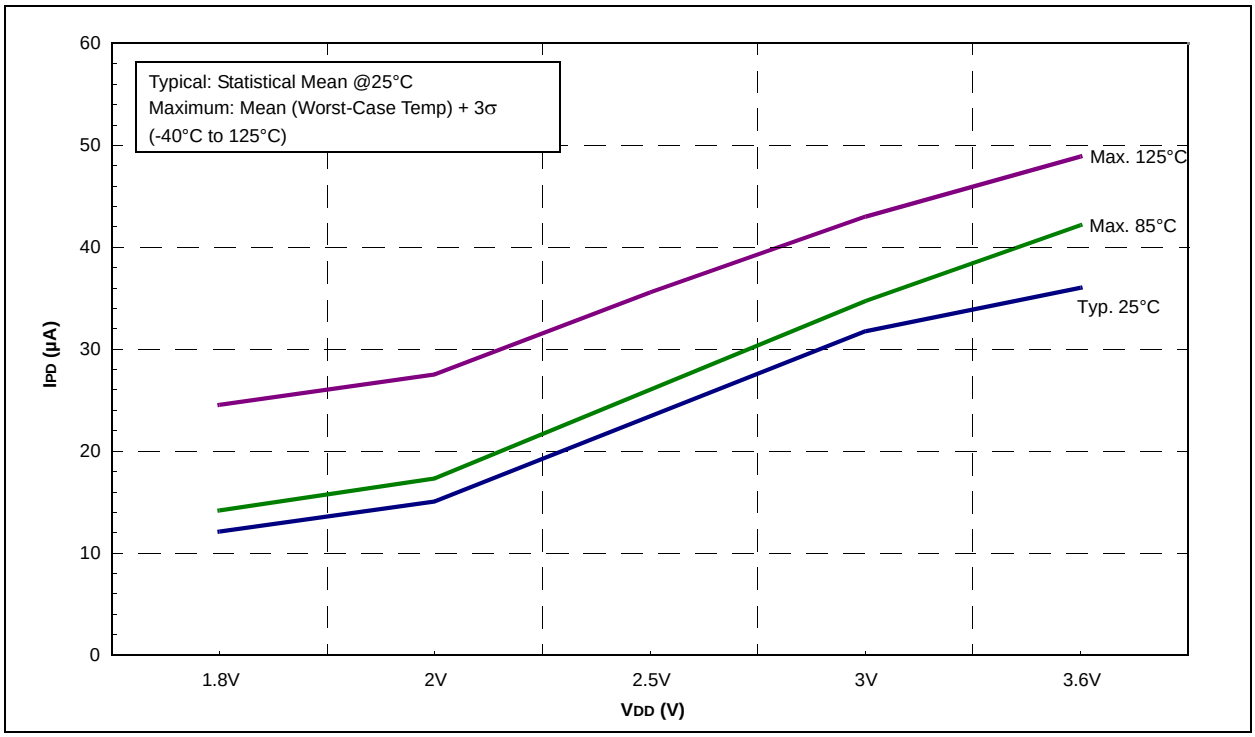

FIGURE 26-37: PIC16F707 CAP SENSE MEDIUM POWER IPD vs. VDD, VCAP = 0.1µF

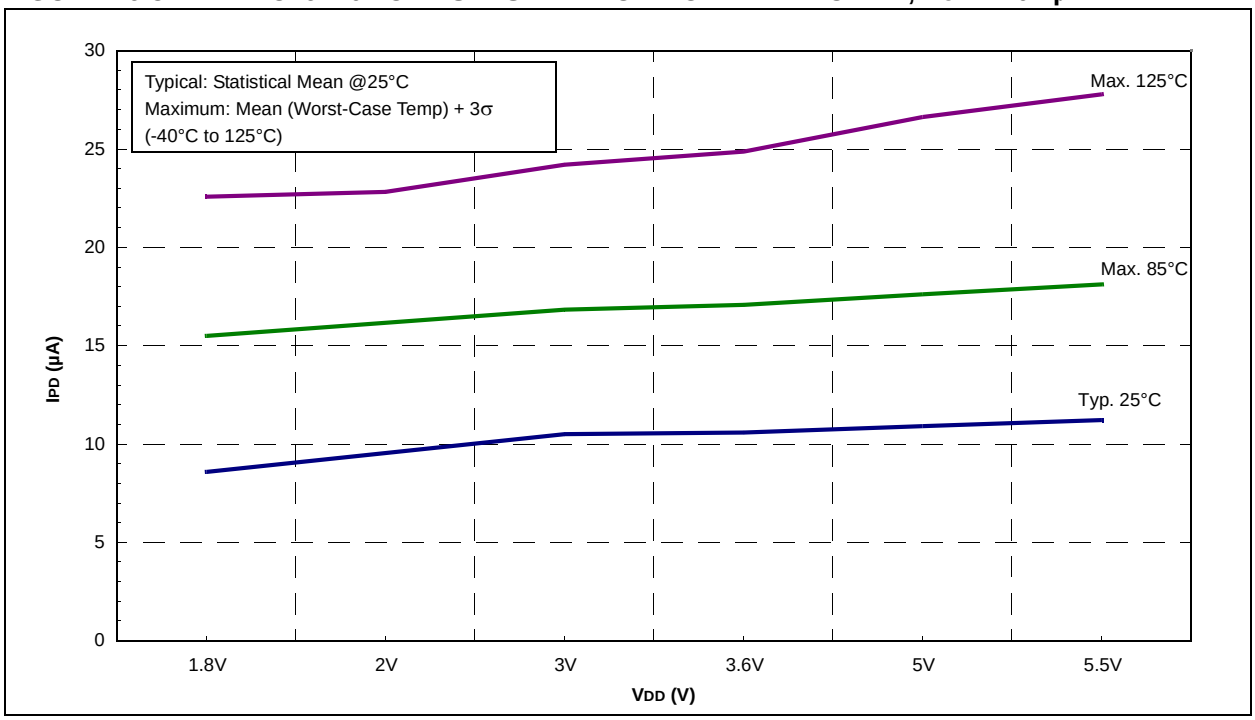

FIGURE 26-38: PIC16LF707 CAP SENSE MEDIUM POWER IPD vs. VDD

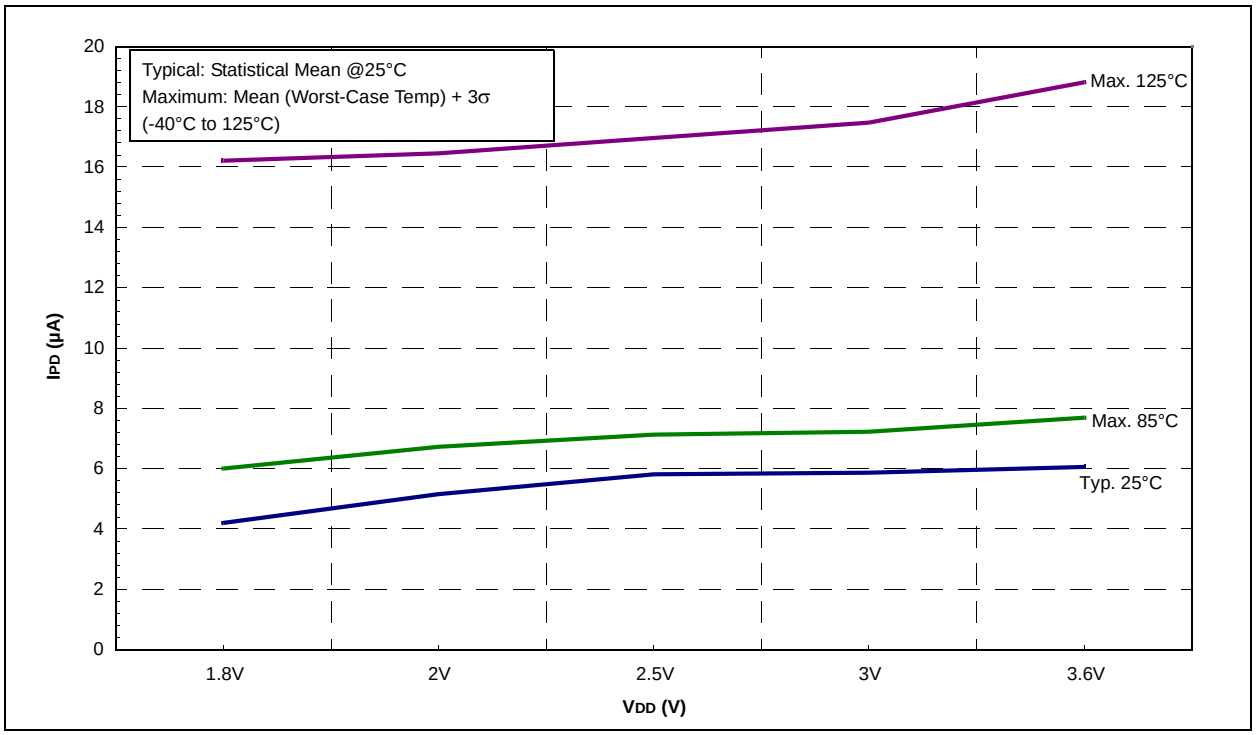

FIGURE 26-39: PIC16F707 CAP SENSE LOW POWER IPD vs. VDD, VCAP = 0.1µF

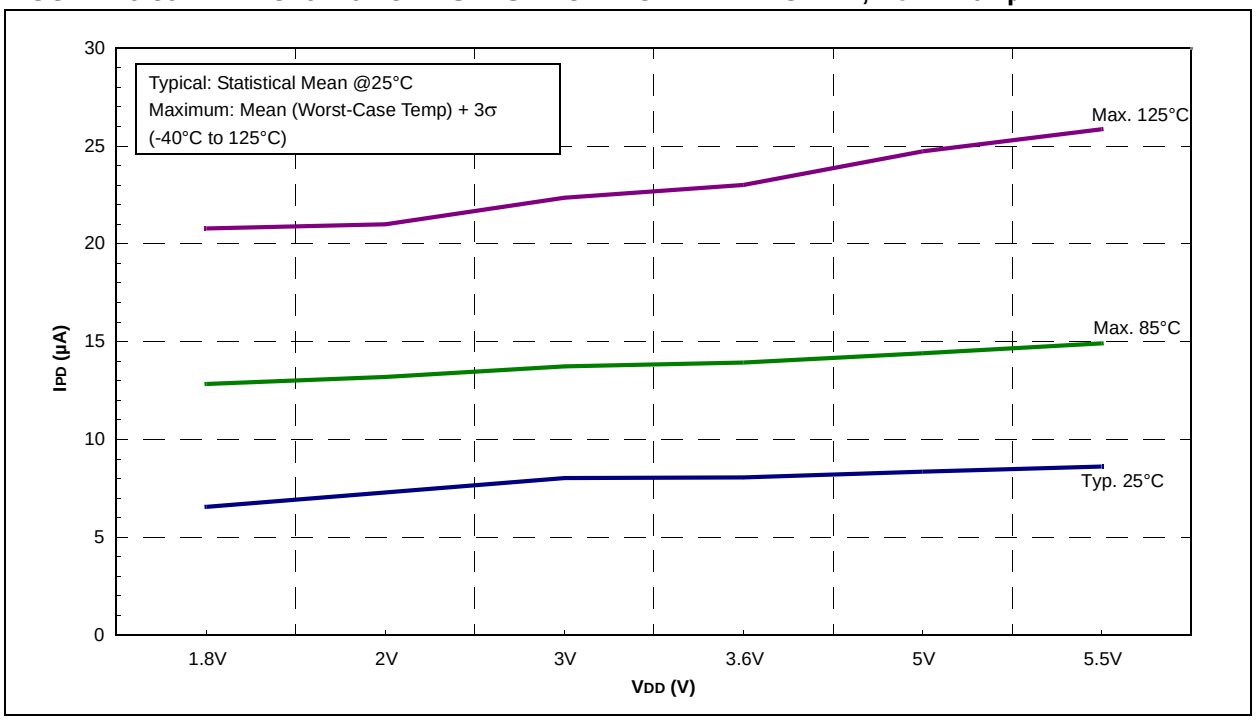

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FIGURE 26-40: PIC16LF707 CAP SENSE LOW POWER IPD vs. VDD

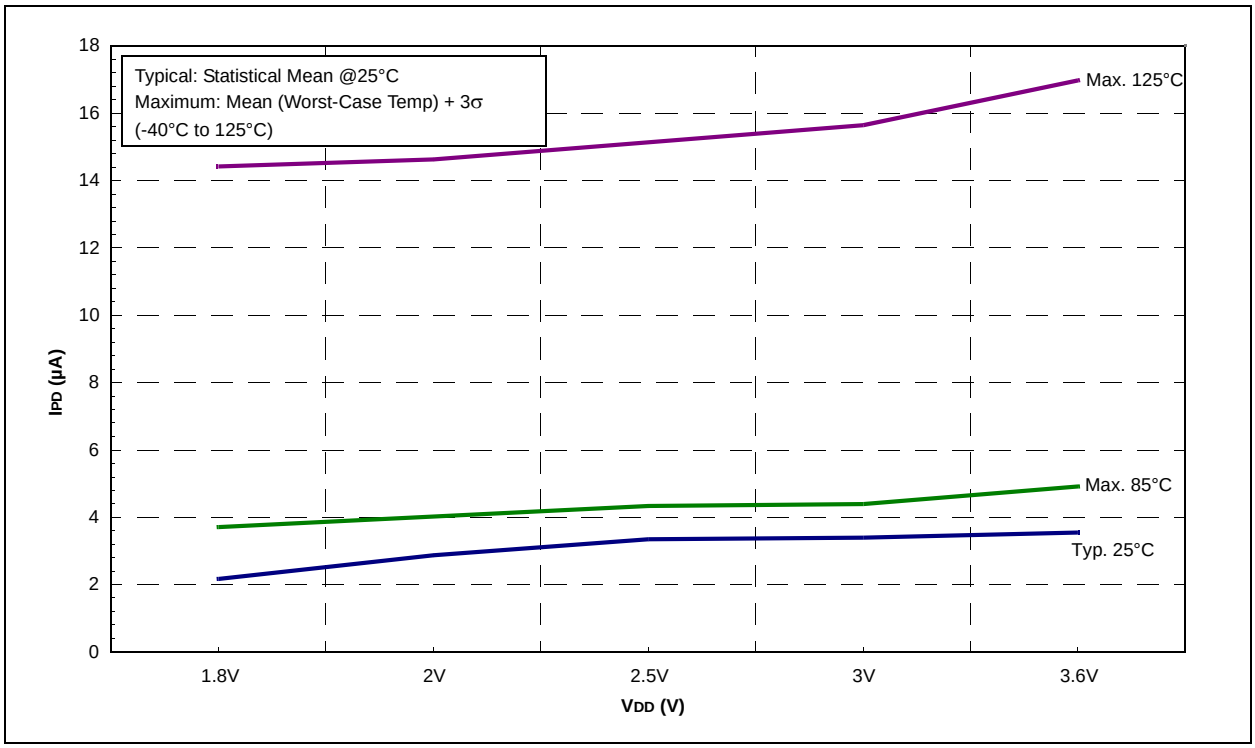

FIGURE 26-41: PIC16F707 T1OSC 32 kHz IPD vs. VDD, VCAP = 0.1µF

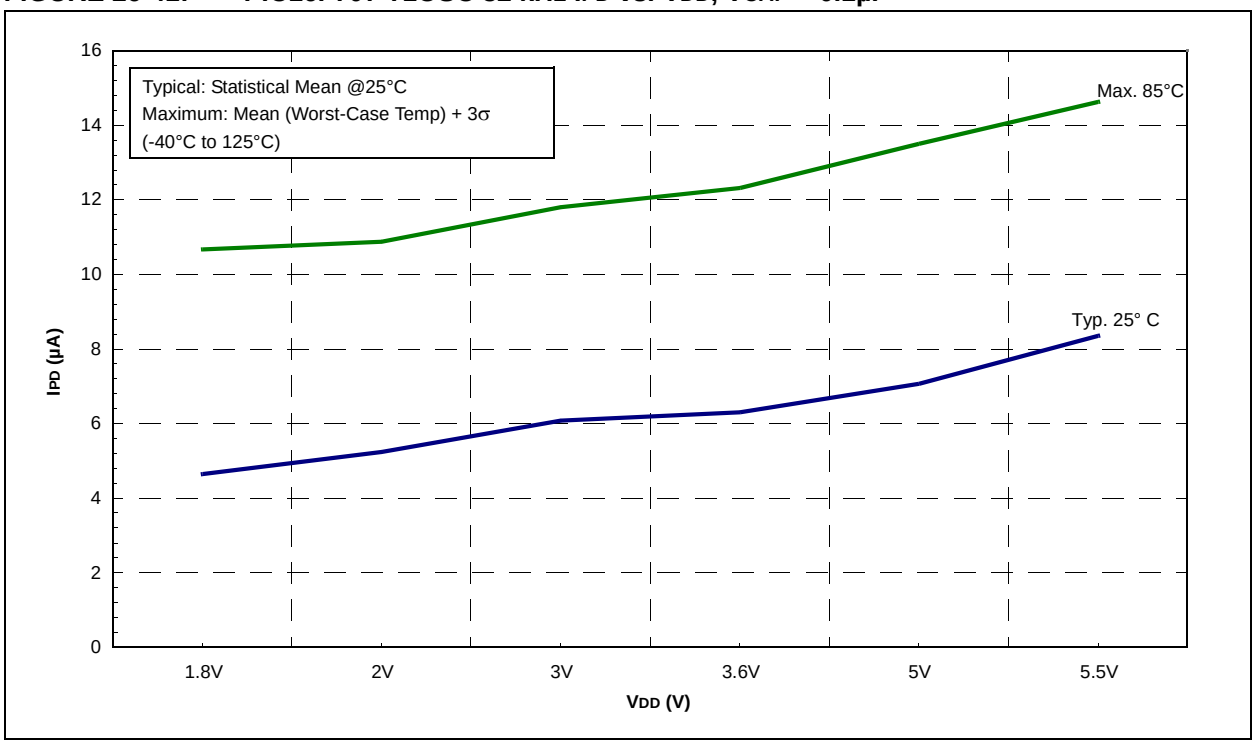

FIGURE 26-42: PIC16LF707 T1OSC 32 kHz IPD vs. VDD

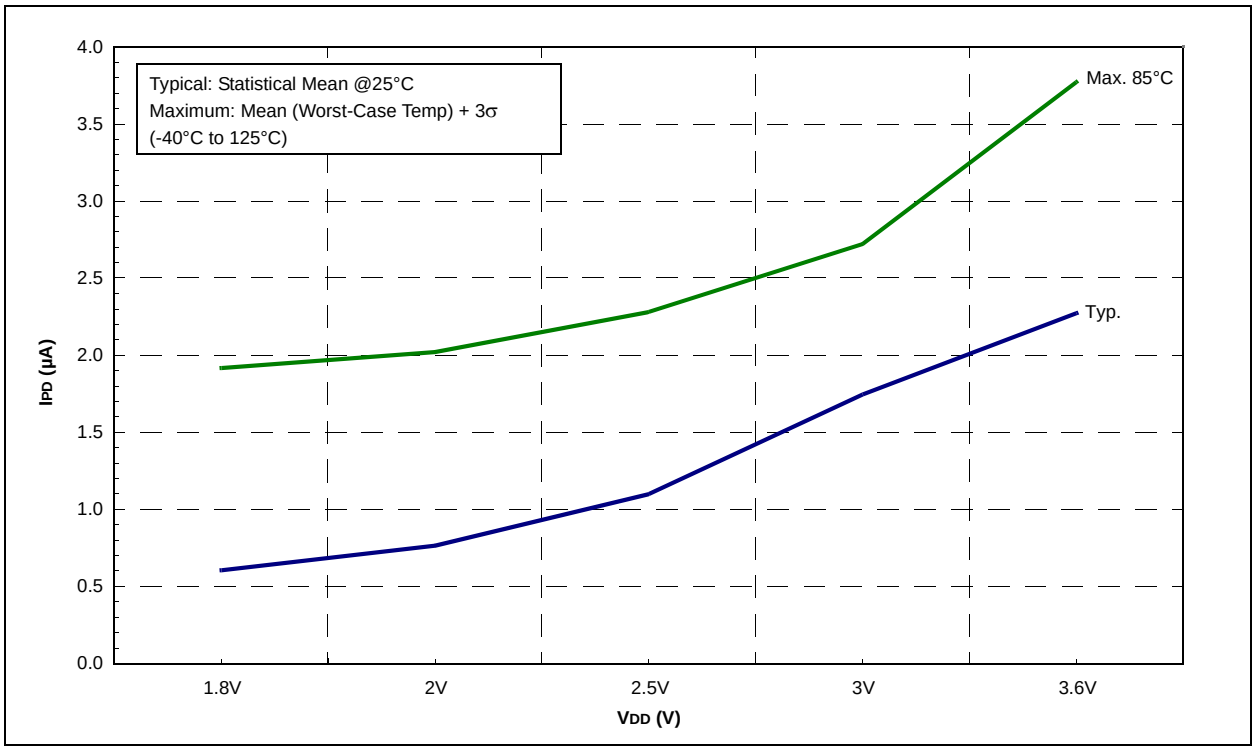

FIGURE 26-43: PIC16F707 TYPICAL ADC IPD vs. VDD, VCAP = 0.1µF

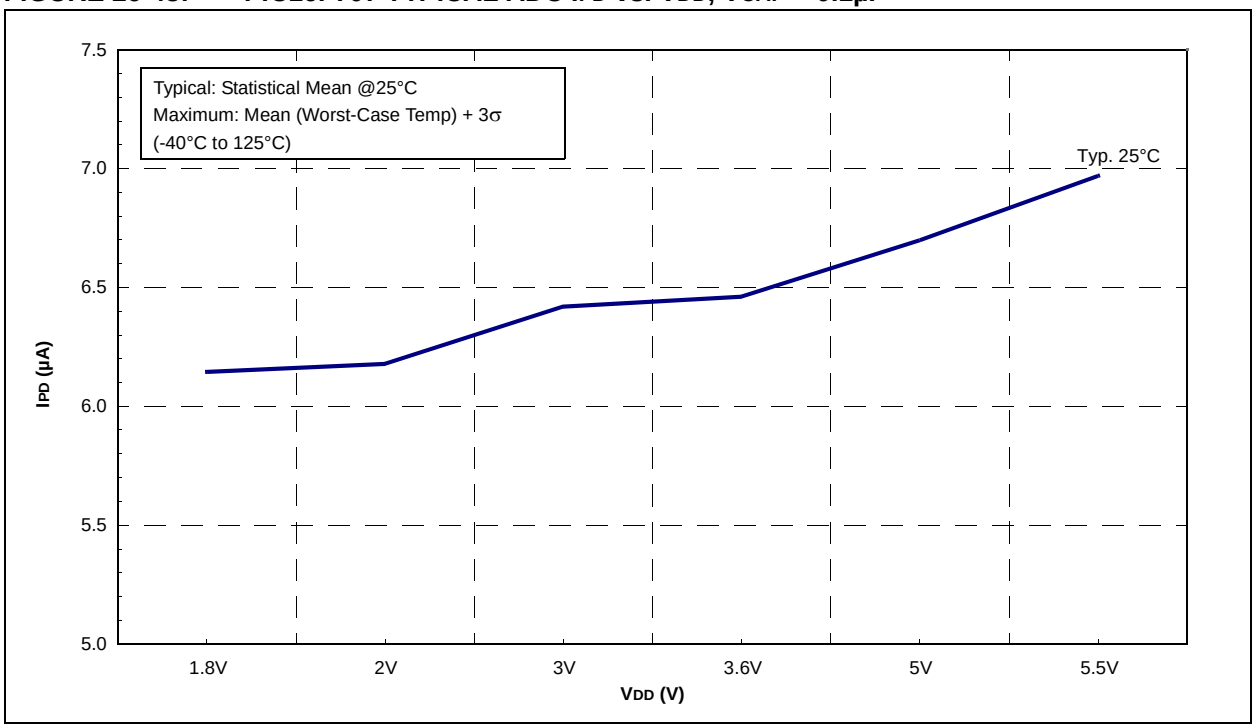

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FIGURE 26-44: PIC16LF707 TYPICAL ADC IPD VS. VDD

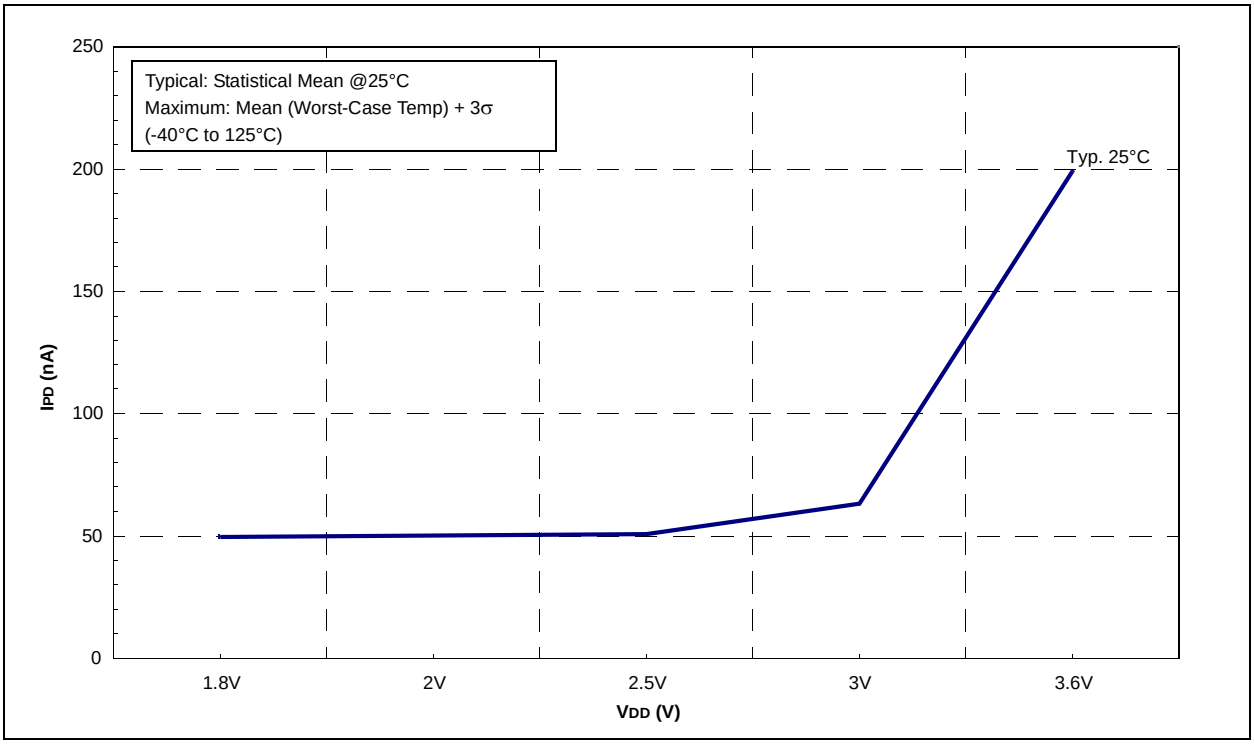

FIGURE 26-45: PIC16F707 ADC IPD VS. VDD, VCAP = 0.1µF

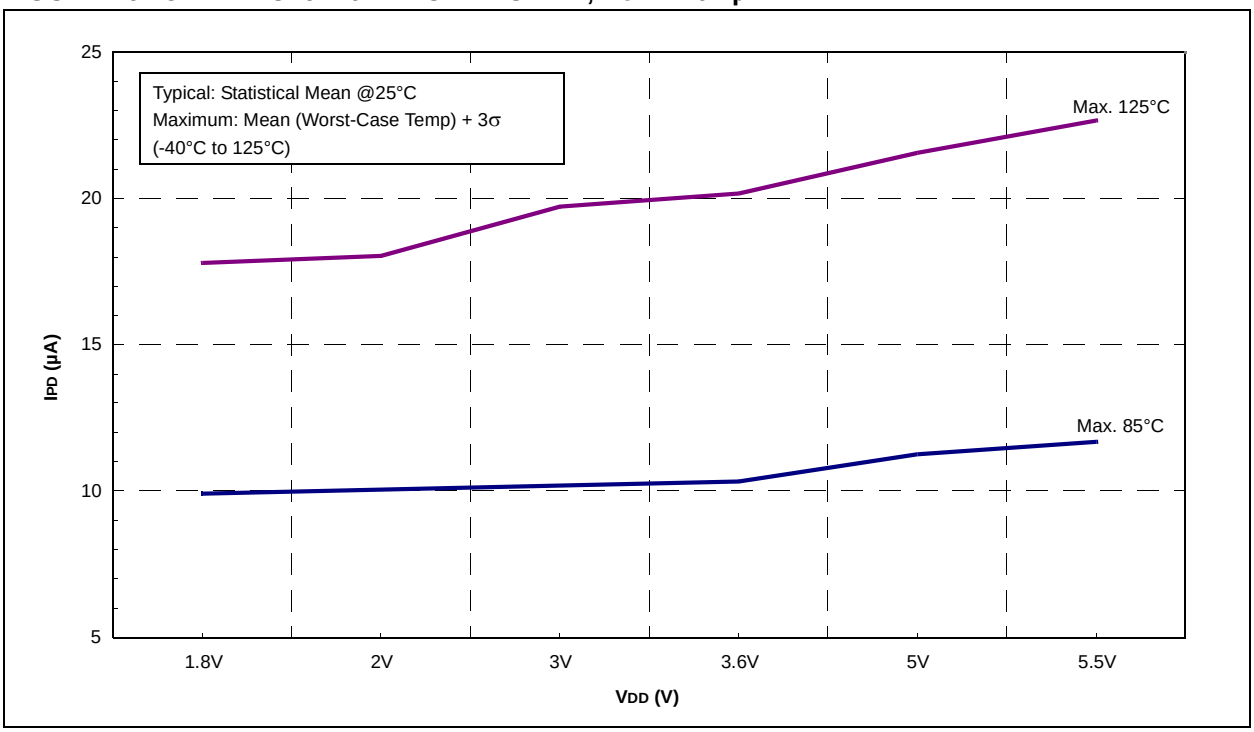

FIGURE 26-46: PIC16LF707 ADC IPD vs. VDD

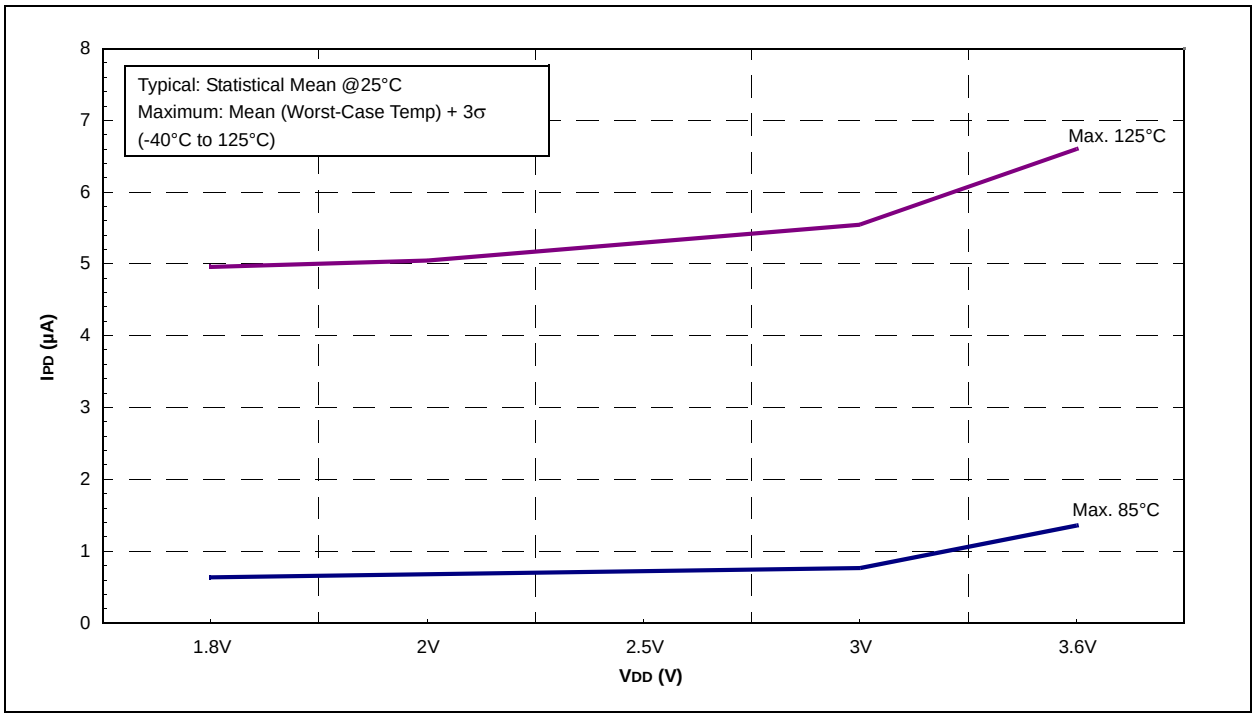

FIGURE 26-47: PIC16F707 WDT IPD vs. VDD, VCAP = 0.1µF

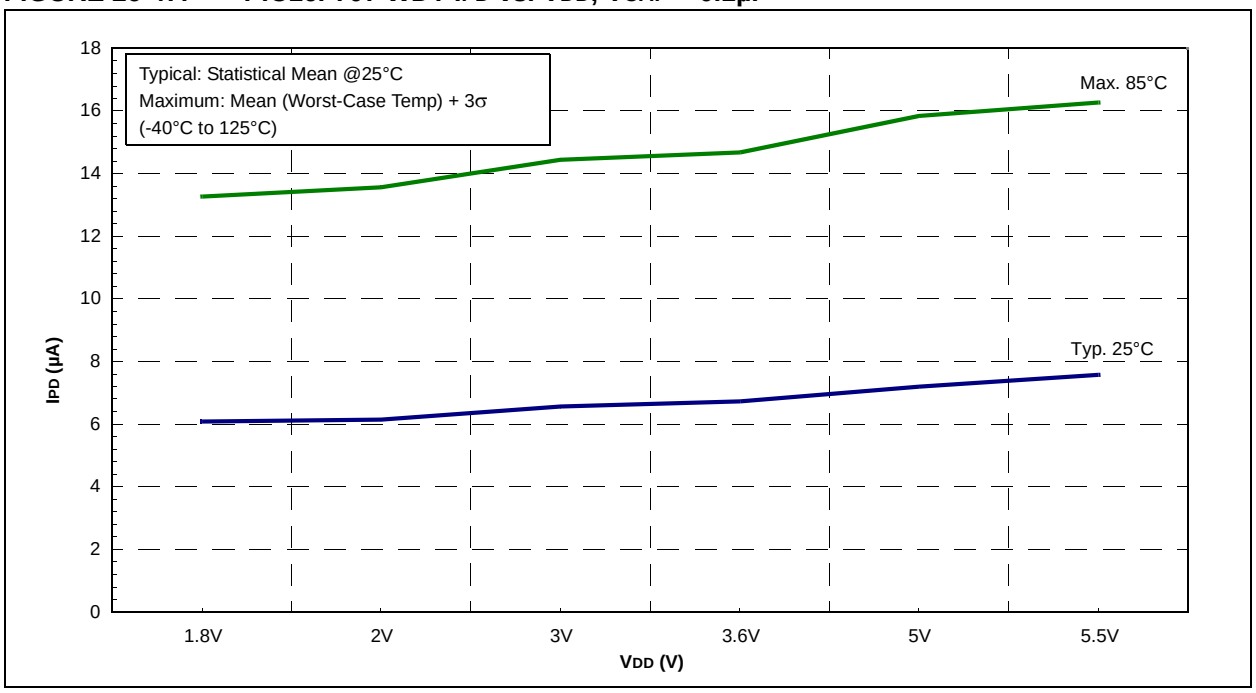

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FIGURE 26-48: PIC16LF707 WDT IPD VS. VDD

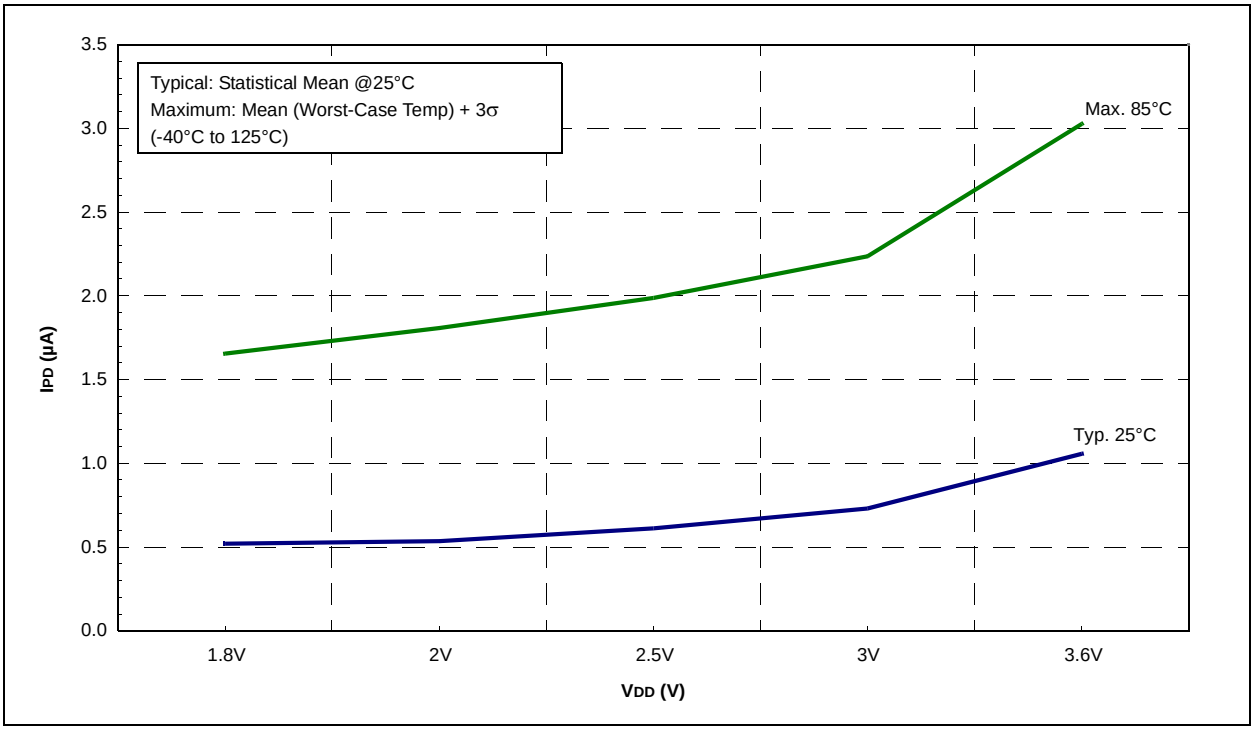

FIGURE 26-49: TTL INPUT THRESHOLD VIN VS. VDD OVER TEMPERATURE

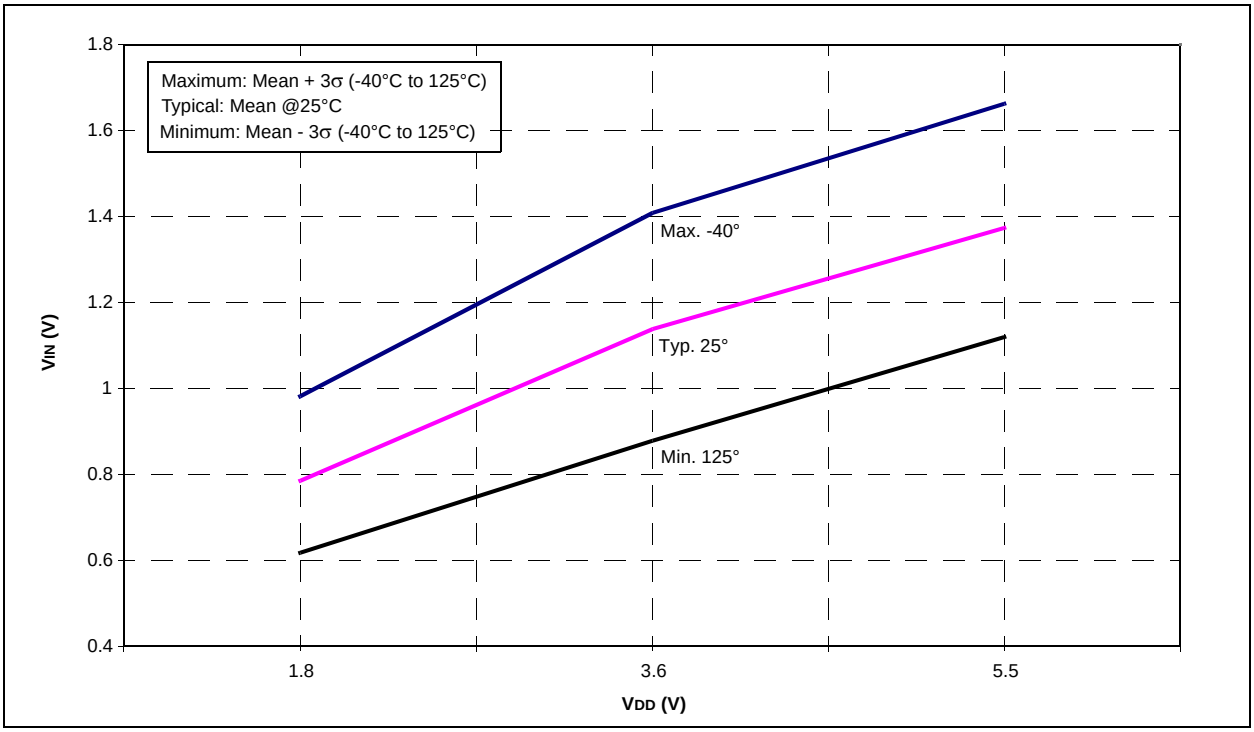

FIGURE 26-50: SCHMITT TRIGGER INPUT THRESHOLD VIN VS. VDD OVER TEMPERATURE

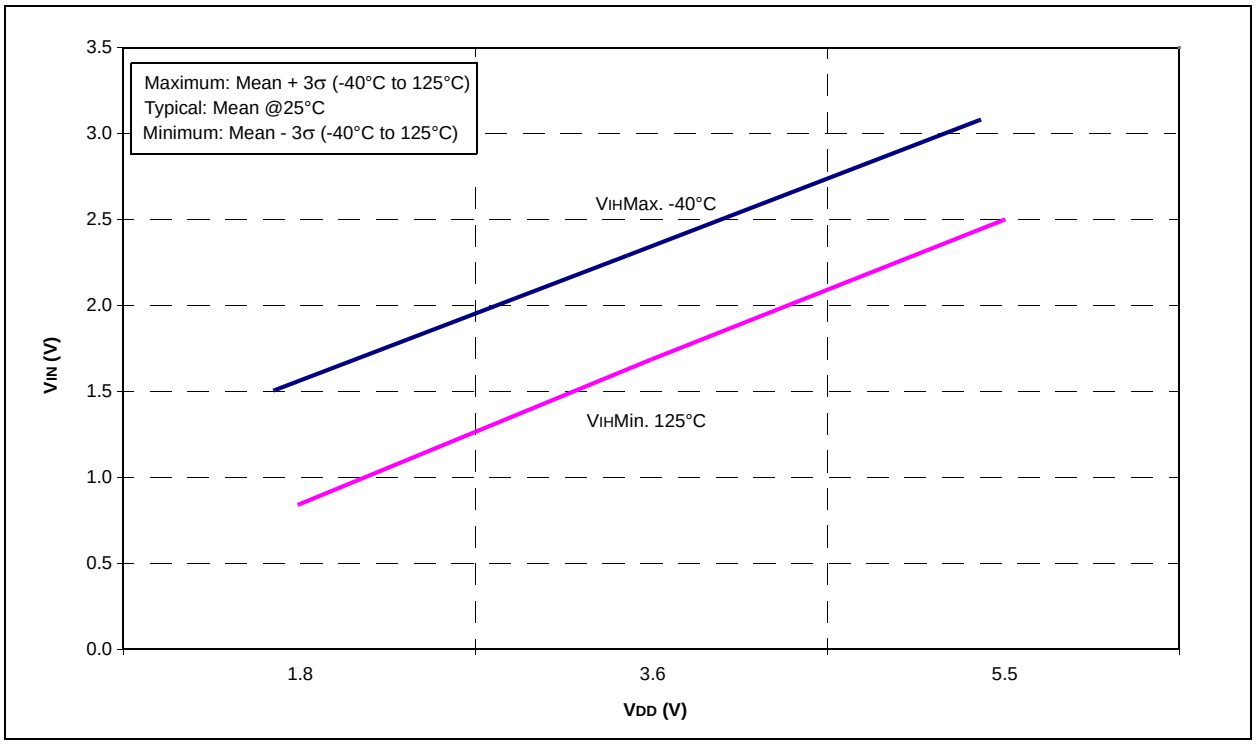

FIGURE 26-51: SCHMITT TRIGGER INPUT THRESHOLD VIN VS. VDD OVER TEMPERATURE

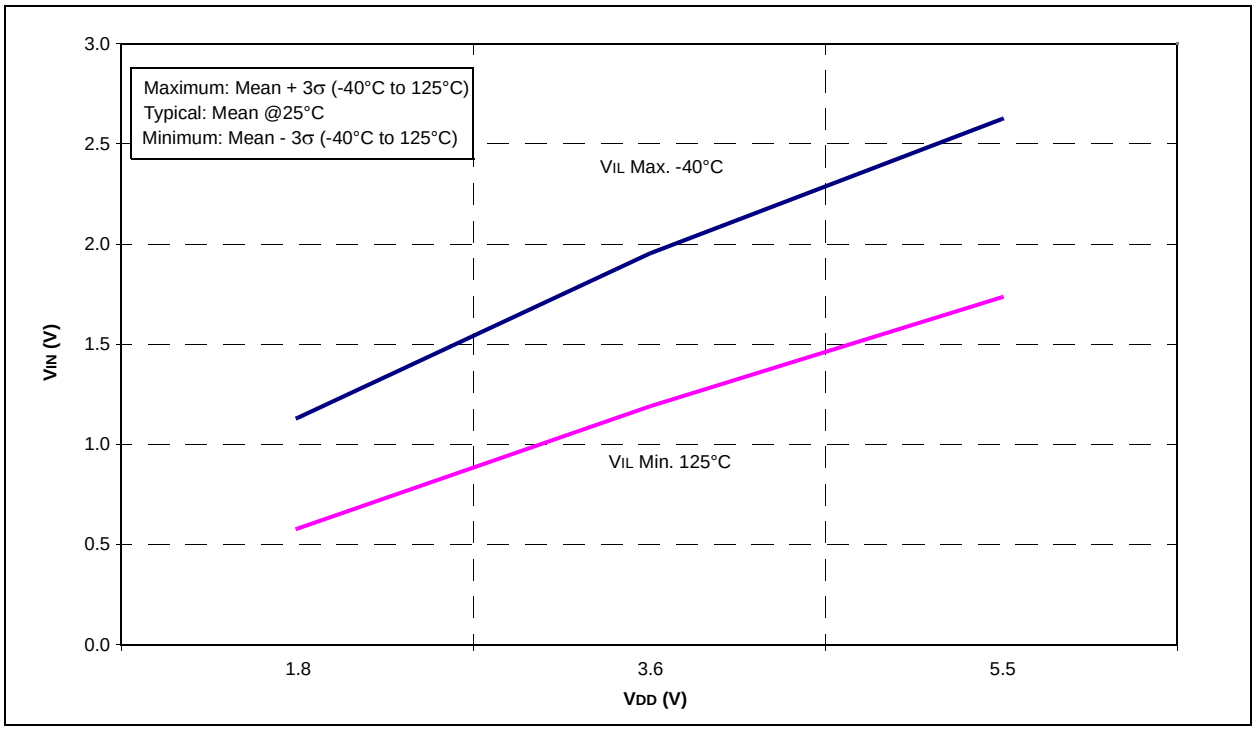

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FIGURE 26-52: VOH vs. IOH OVER TEMPERATURE, VDD = 5.5V

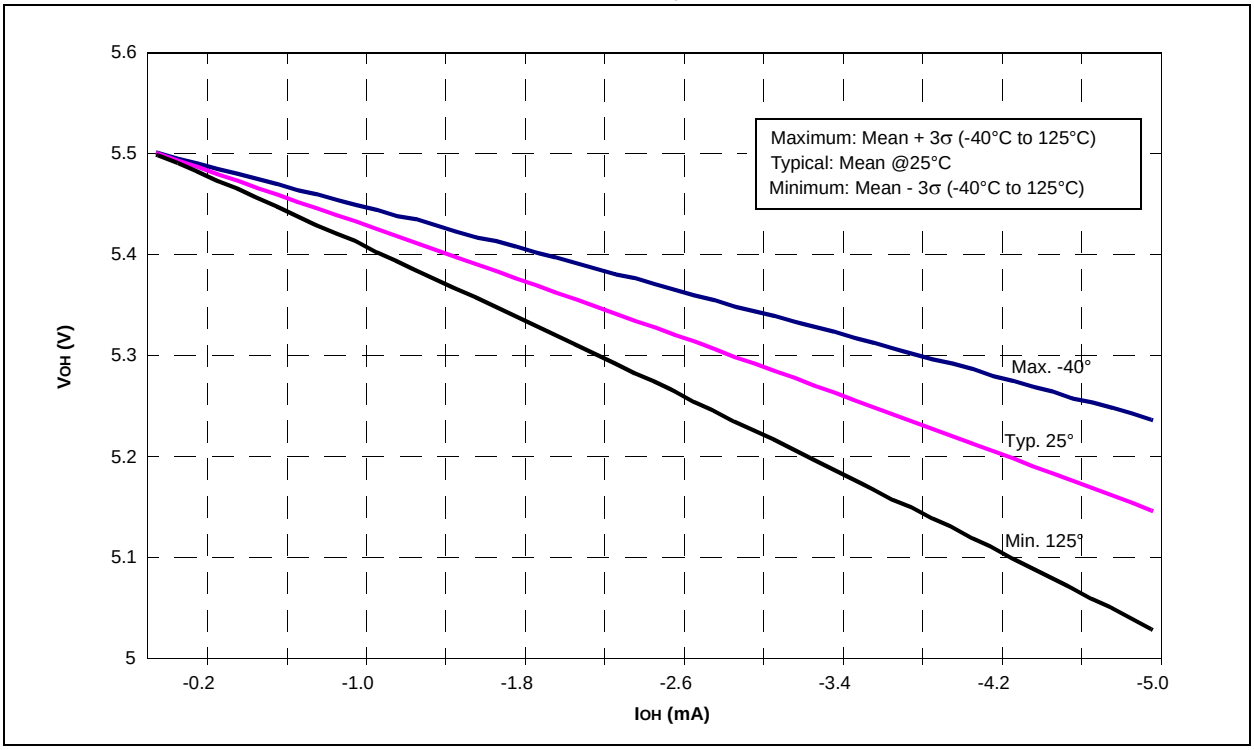

FIGURE 26-53: VOH vs. IOH OVER TEMPERATURE, VDD = 3.6V

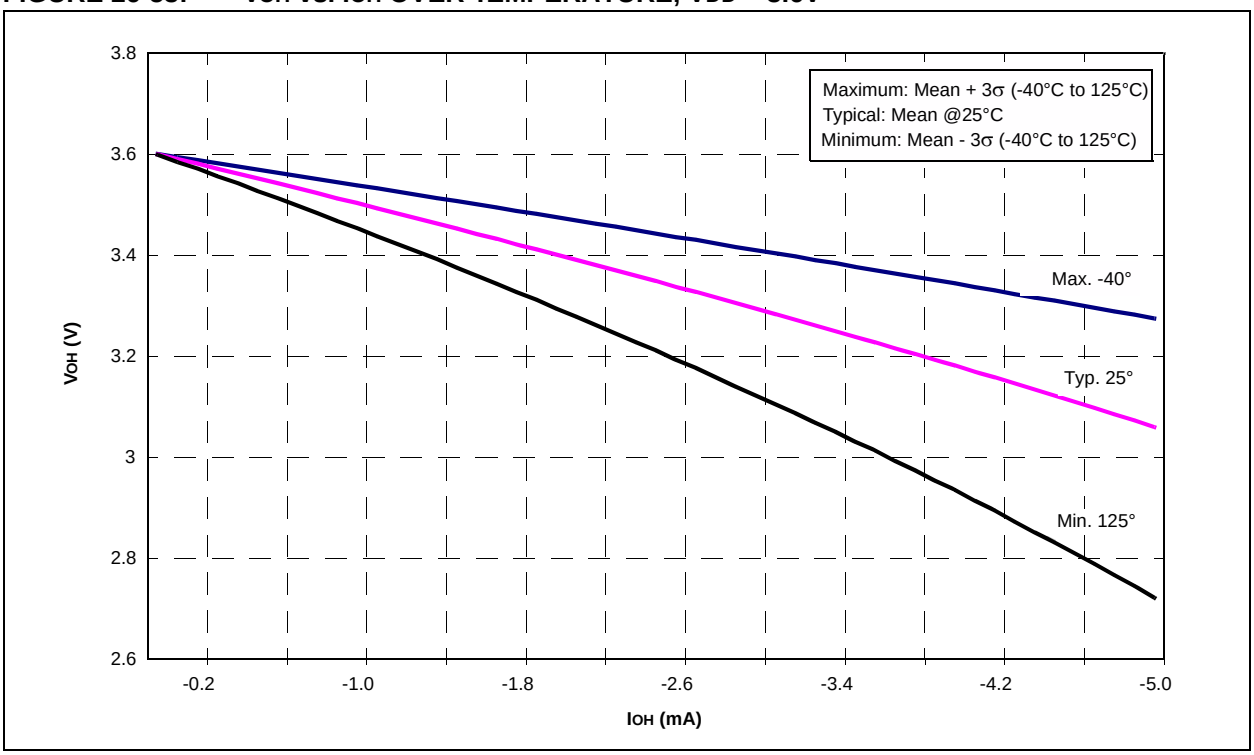

FIGURE 26-54: VOH vs. IOH OVER TEMPERATURE, VDD = 1.8V

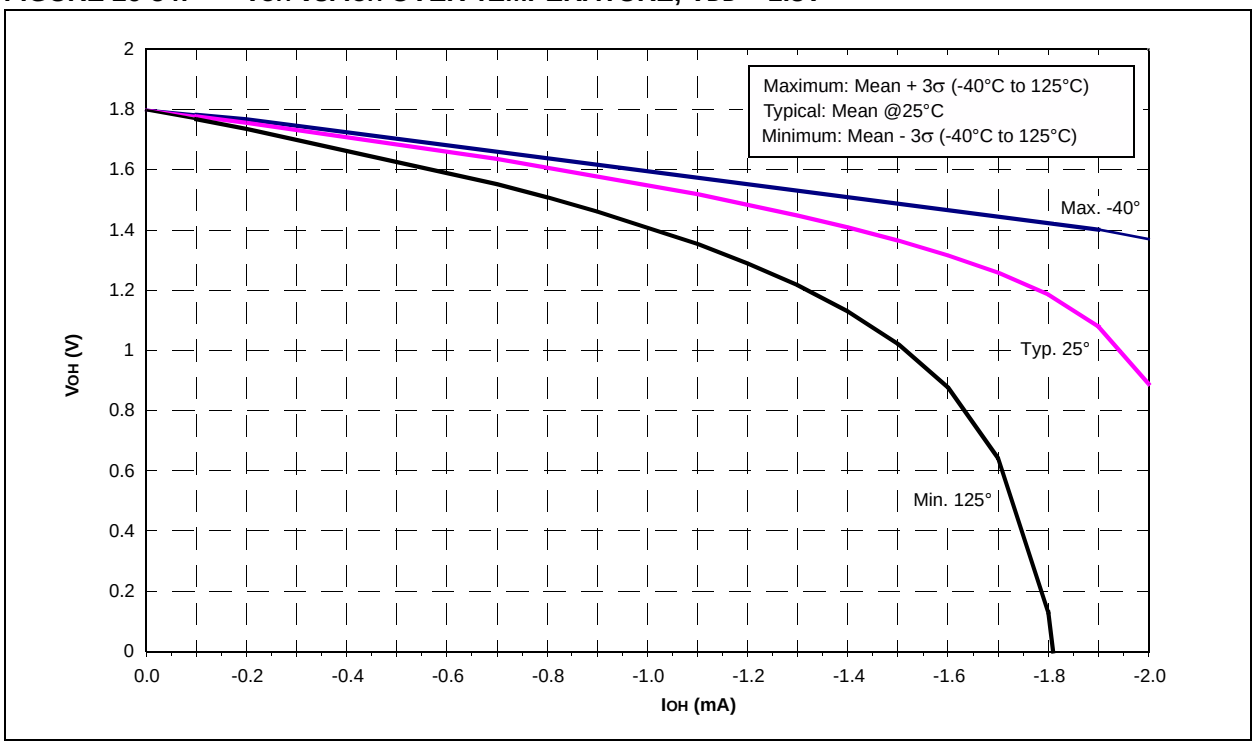

FIGURE 26-55: VOL vs. IOL OVER TEMPERATURE, VDD = 5.5V

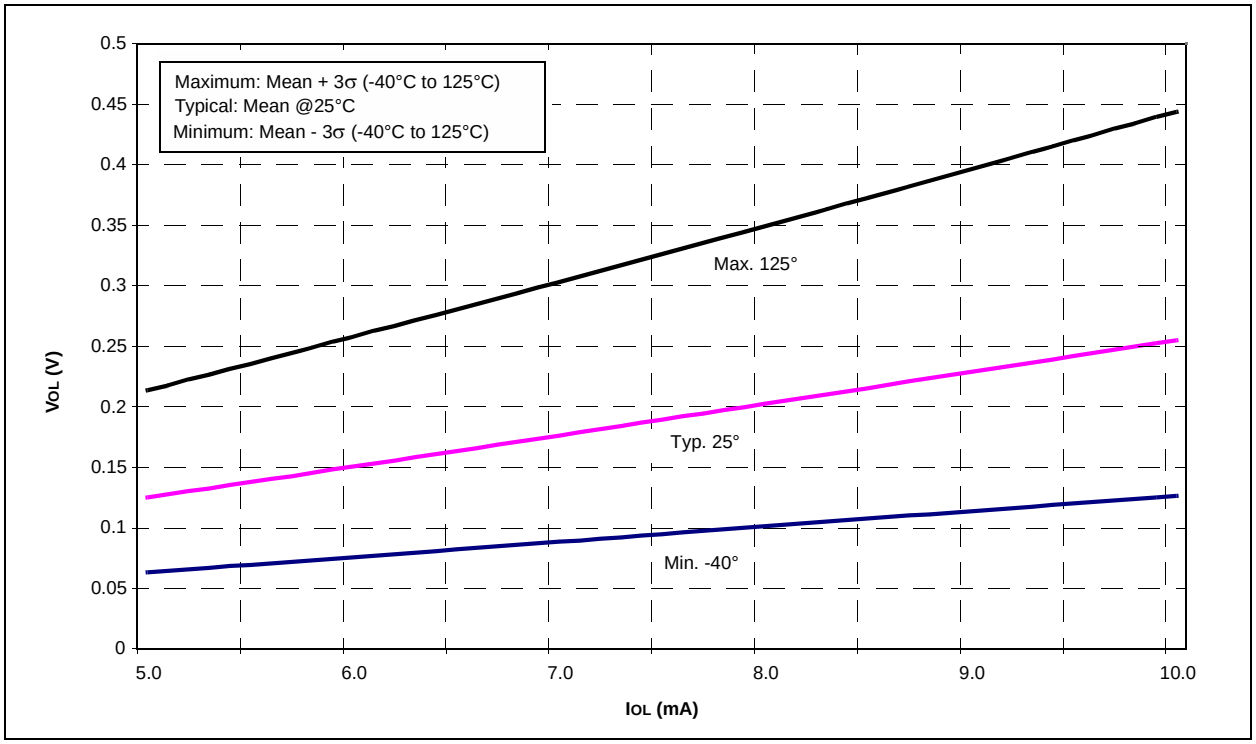

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FIGURE 26-56: VOL vs. IOL OVER TEMPERATURE, VDD = 3.6

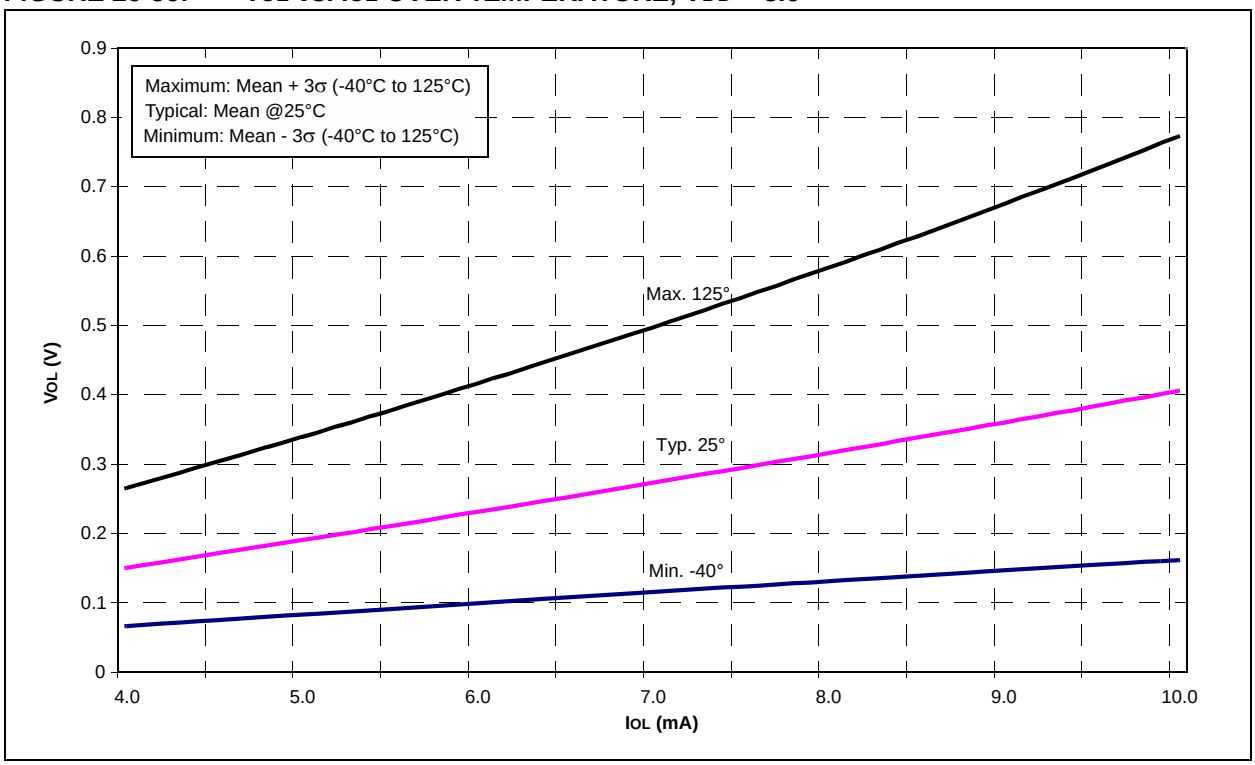

FIGURE 26-57: VOL vs. IOL OVER TEMPERATURE, VDD = 1.8V

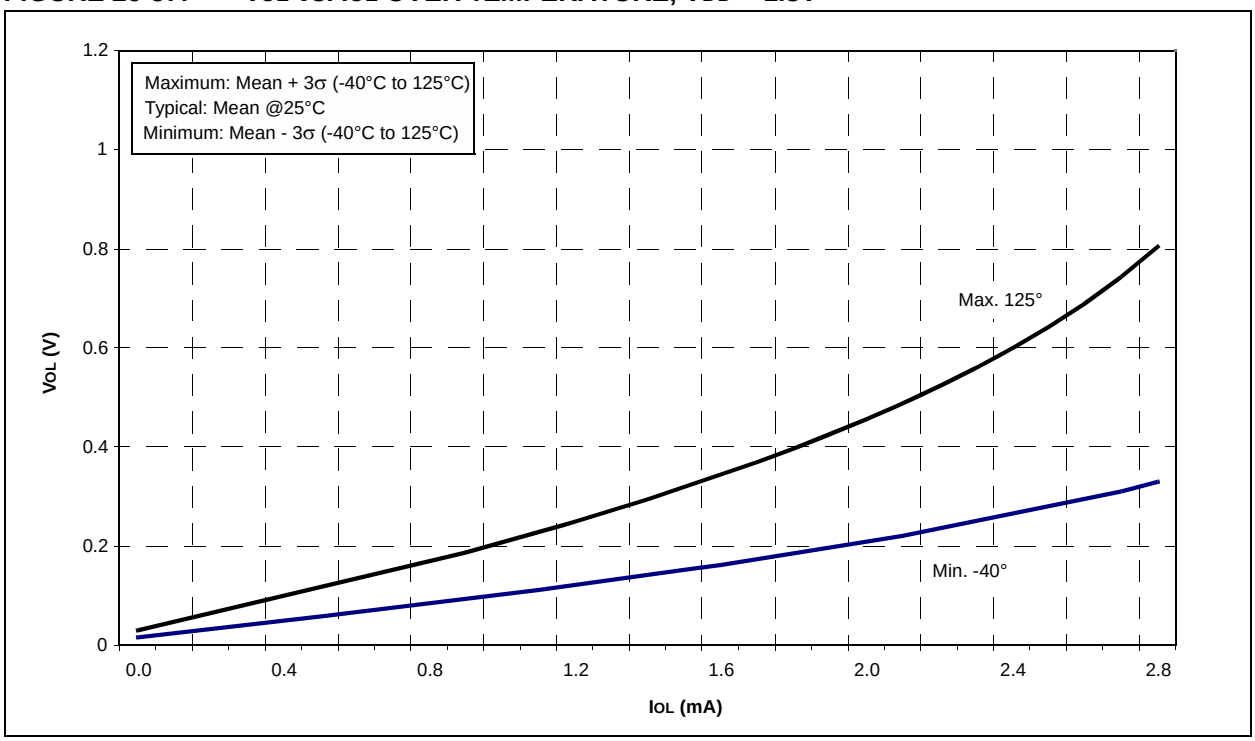

FIGURE 26-58: PIC16F707 PWRT PERIOD

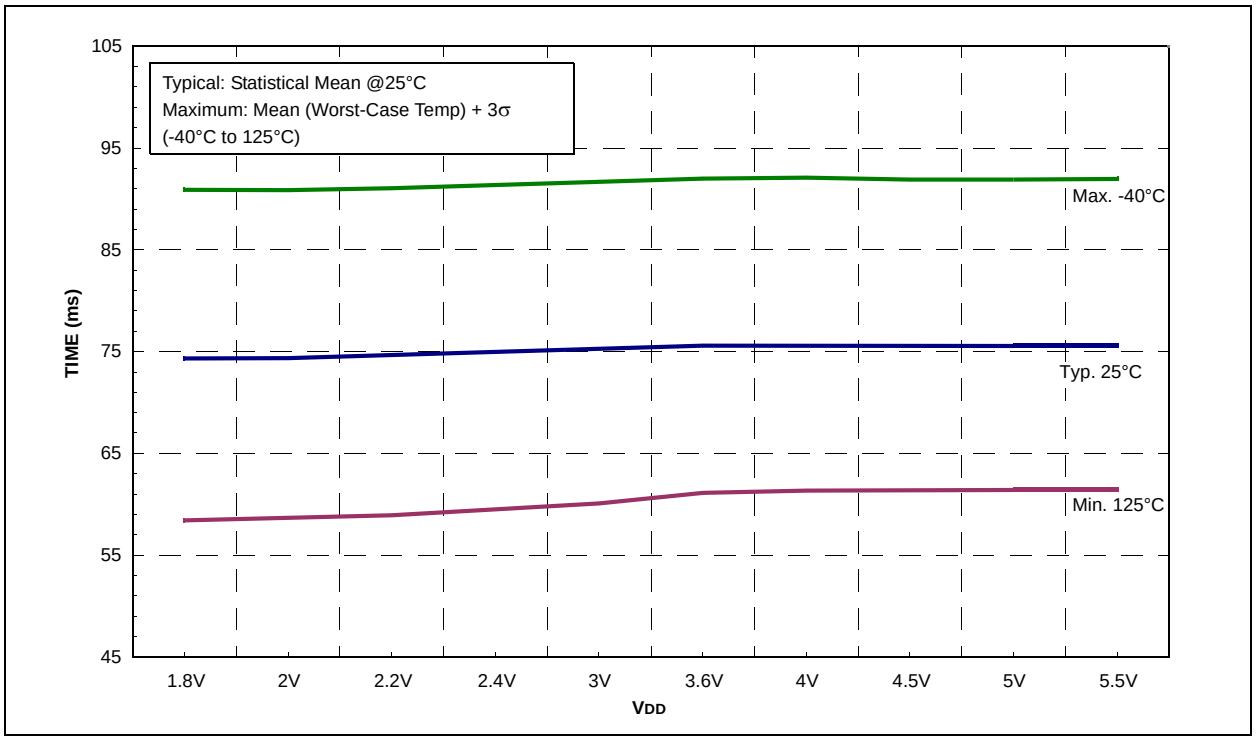

FIGURE 26-59: PIC16F707 WDT TIME-OUT PERIOD

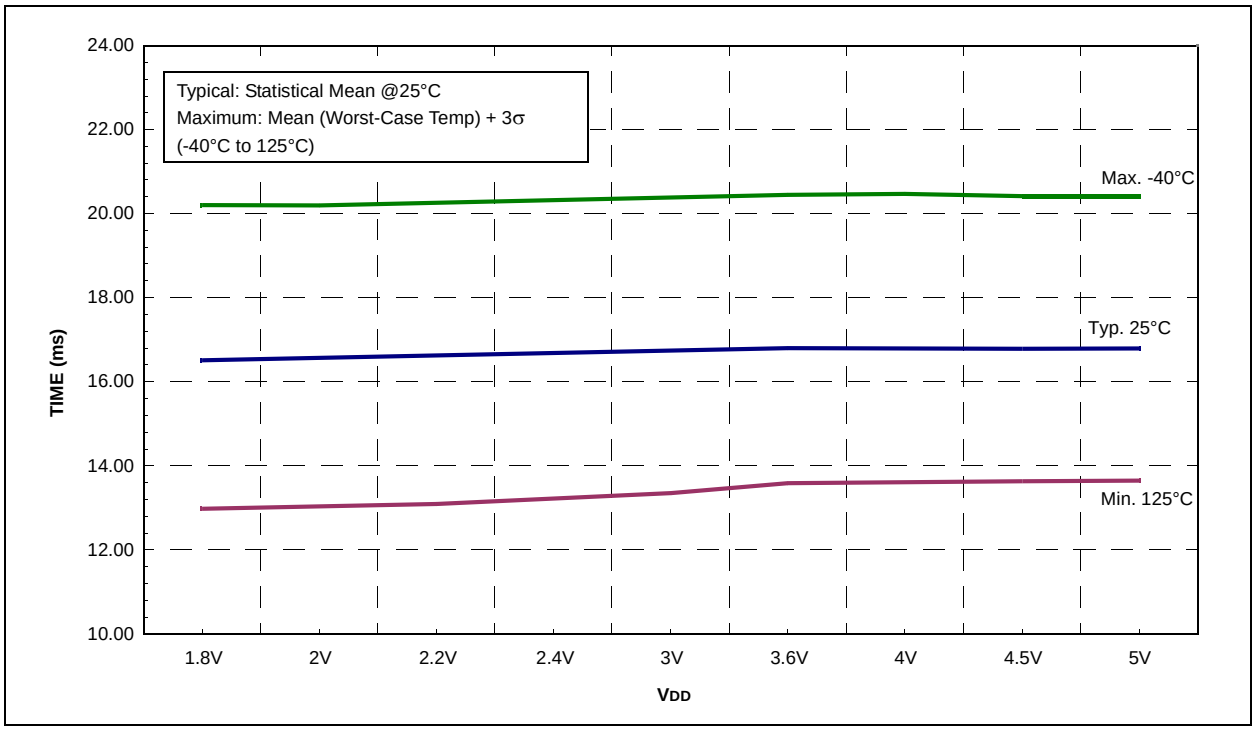

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FIGURE 26-60: PIC16F707 HFINTOSC WAKE-UP FROM SLEEP START-UP TIME

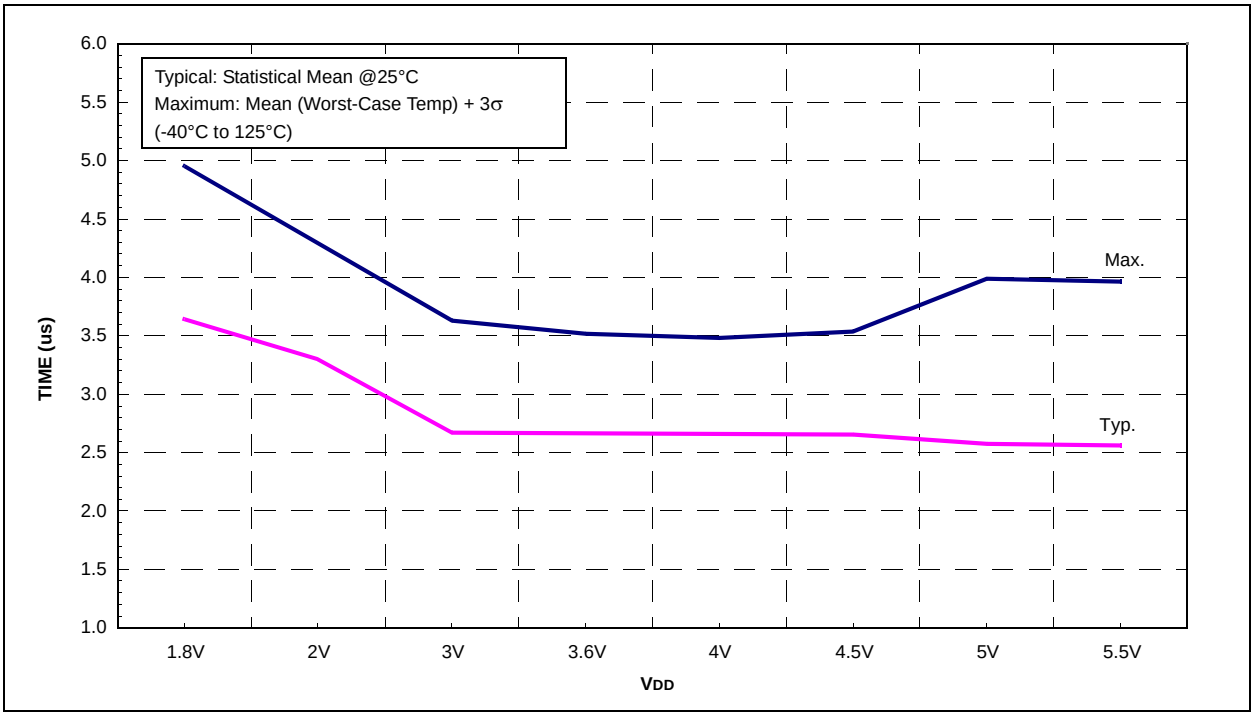

FIGURE 26-61: PIC16F707 A/D INTERNAL RC OSCILLATOR PERIOD

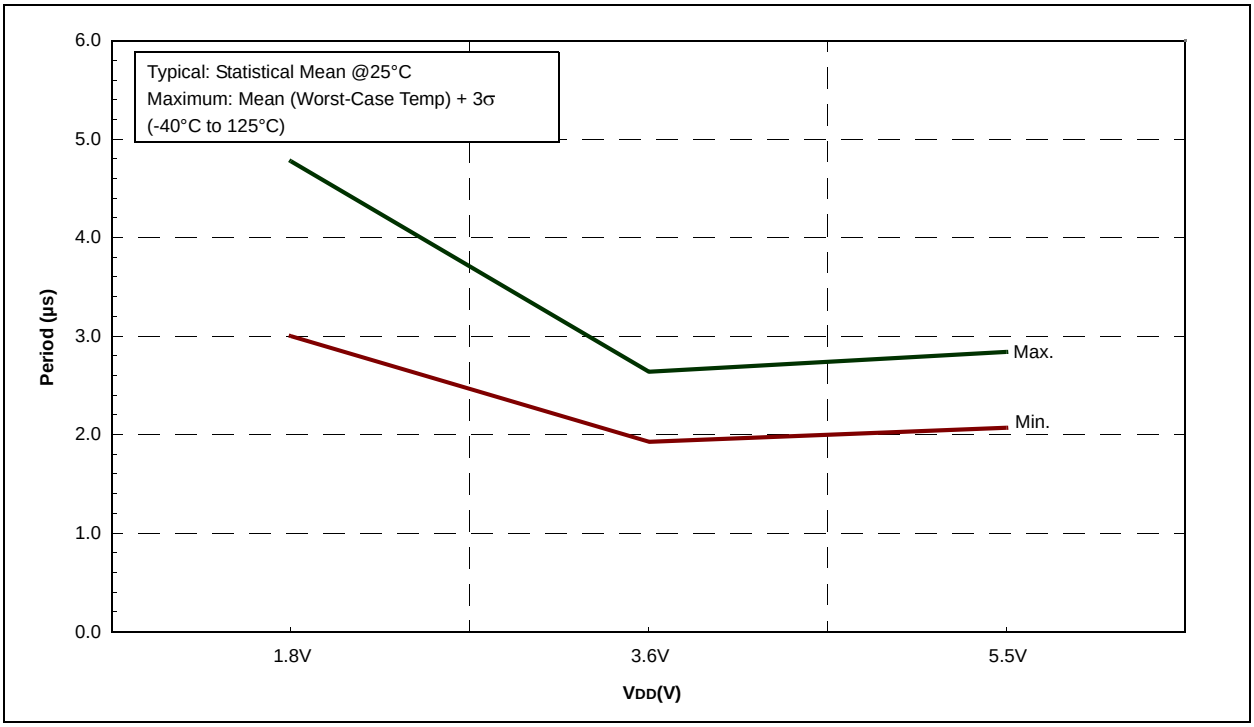

FIGURE 26-62: PIC16F707 CAP SENSE OUTPUT CURRENT, POWER MODE = HIGH

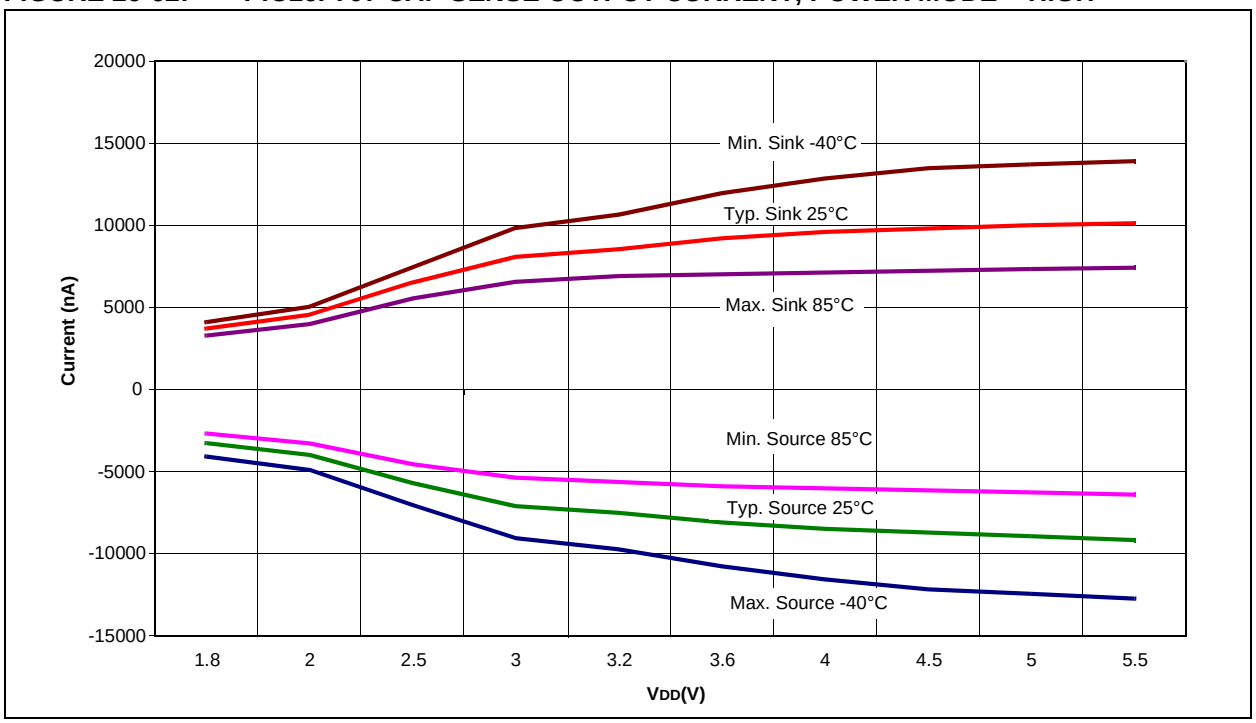

FIGURE 26-63: PIC16F707 CAP SENSE OUTPUT CURRENT, POWER MODE = MEDIUM

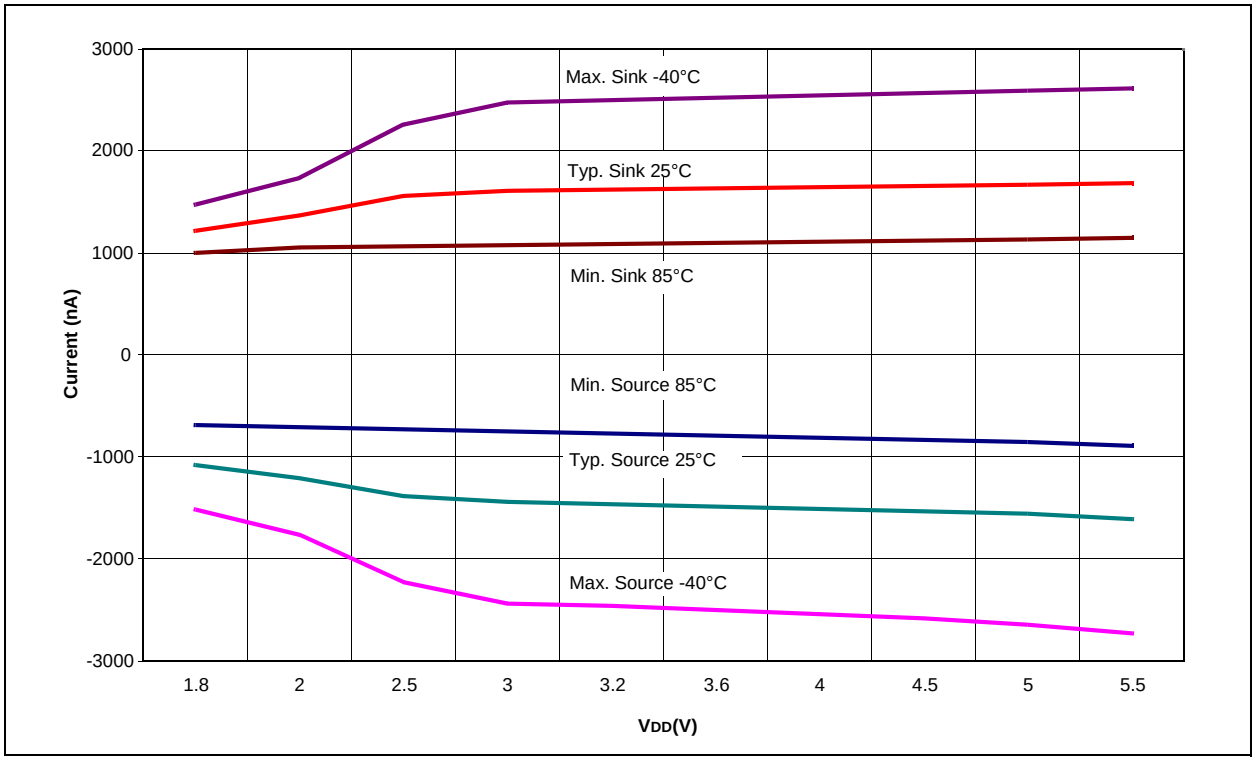

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FIGURE 26-64: PIC16F707 CAP SENSE OUTPUT CURRENT, POWER MODE = LOW

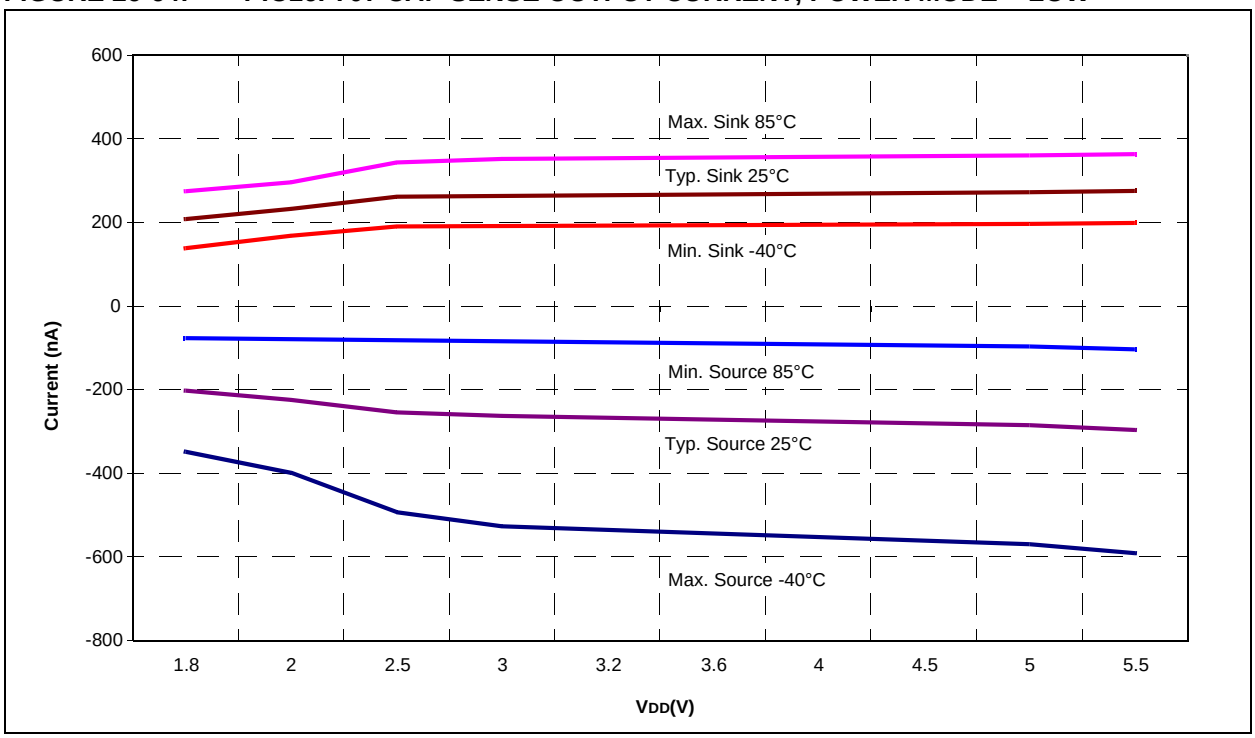

FIGURE 26-65: PIC16F707 CAP SENSOR HYSTERESIS, POWER MODE = HIGH

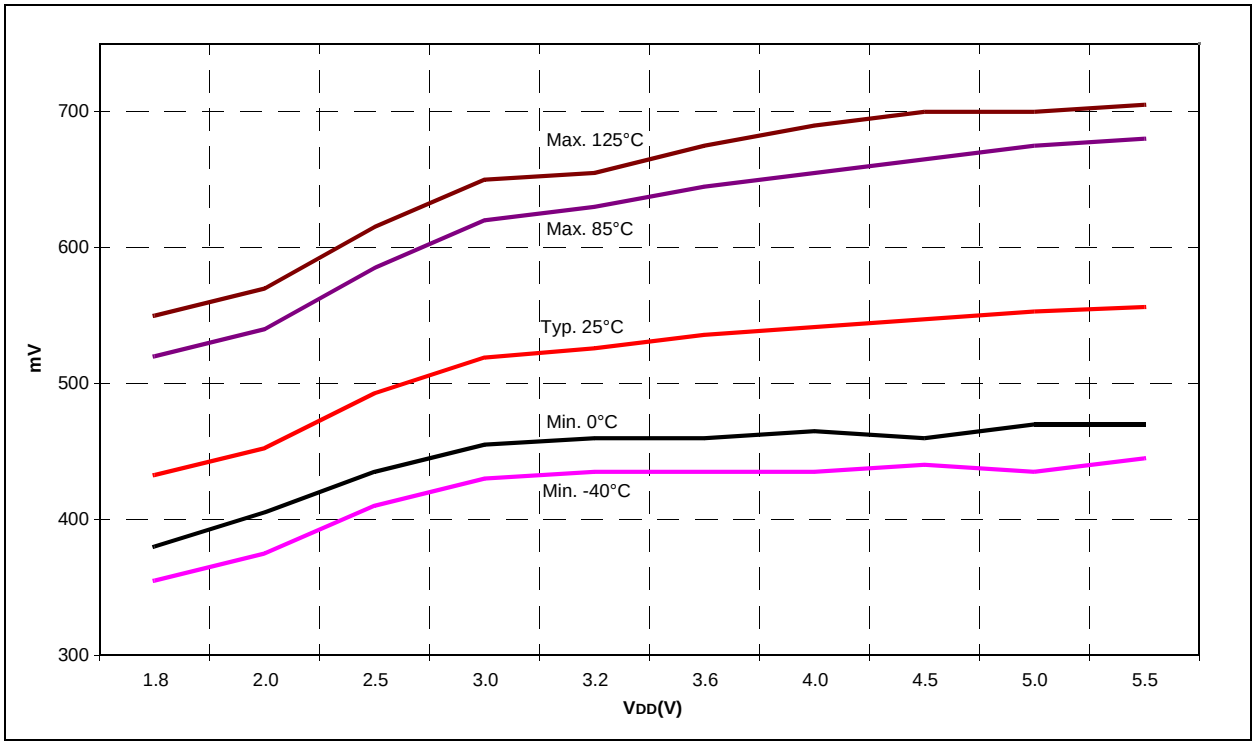

FIGURE 26-66: PIC16F707 CAP SENSOR HYSTERESIS, POWER MODE = MEDIUM

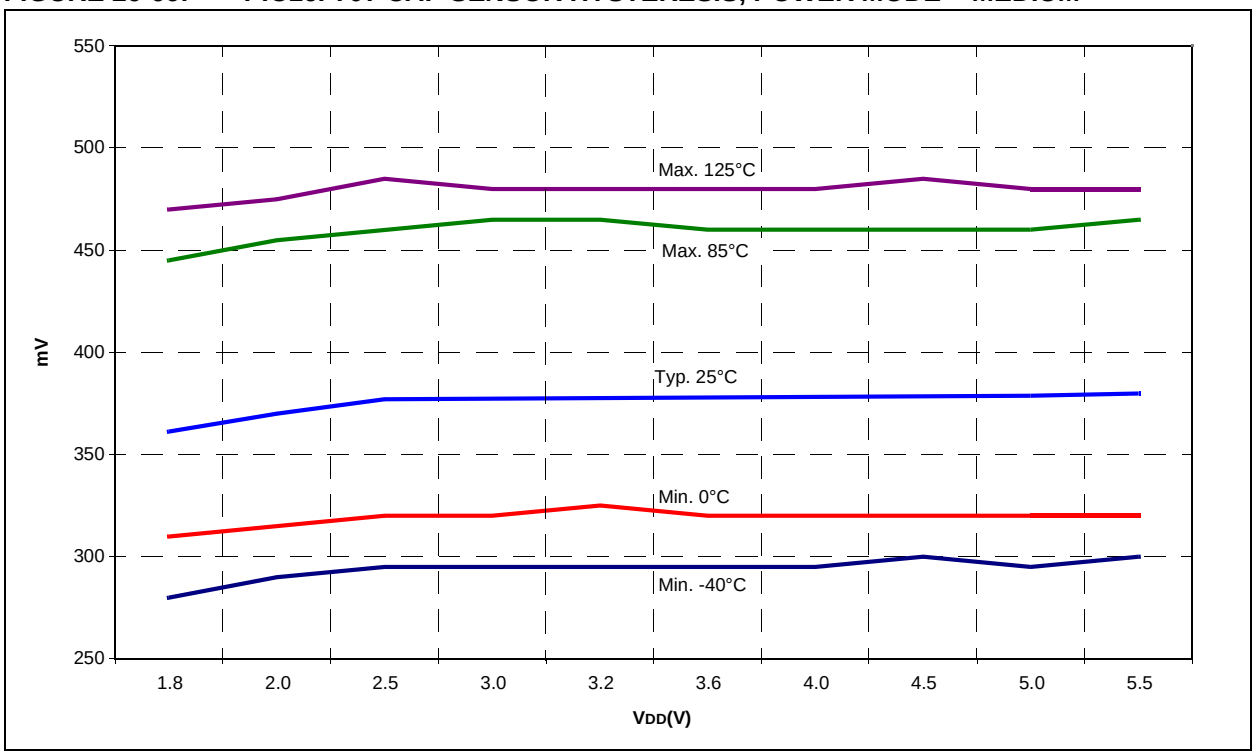

FIGURE 26-67: PIC16F707 CAP SENSOR HYSTERESIS, POWER MODE = LOW

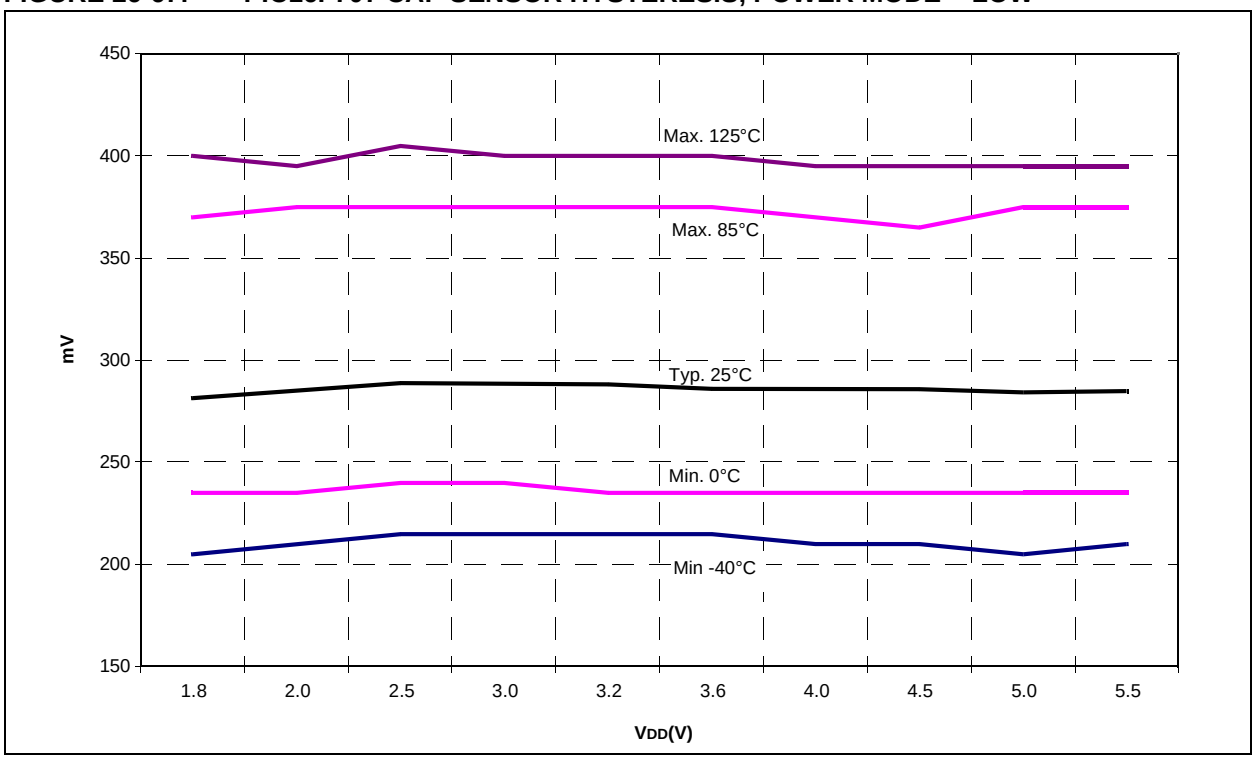

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FIGURE 26-68: TYPICAL FVR (X1 AND X2) VS. SUPPLY VOLTAGE (V) NORMALIZED AT 3.0V

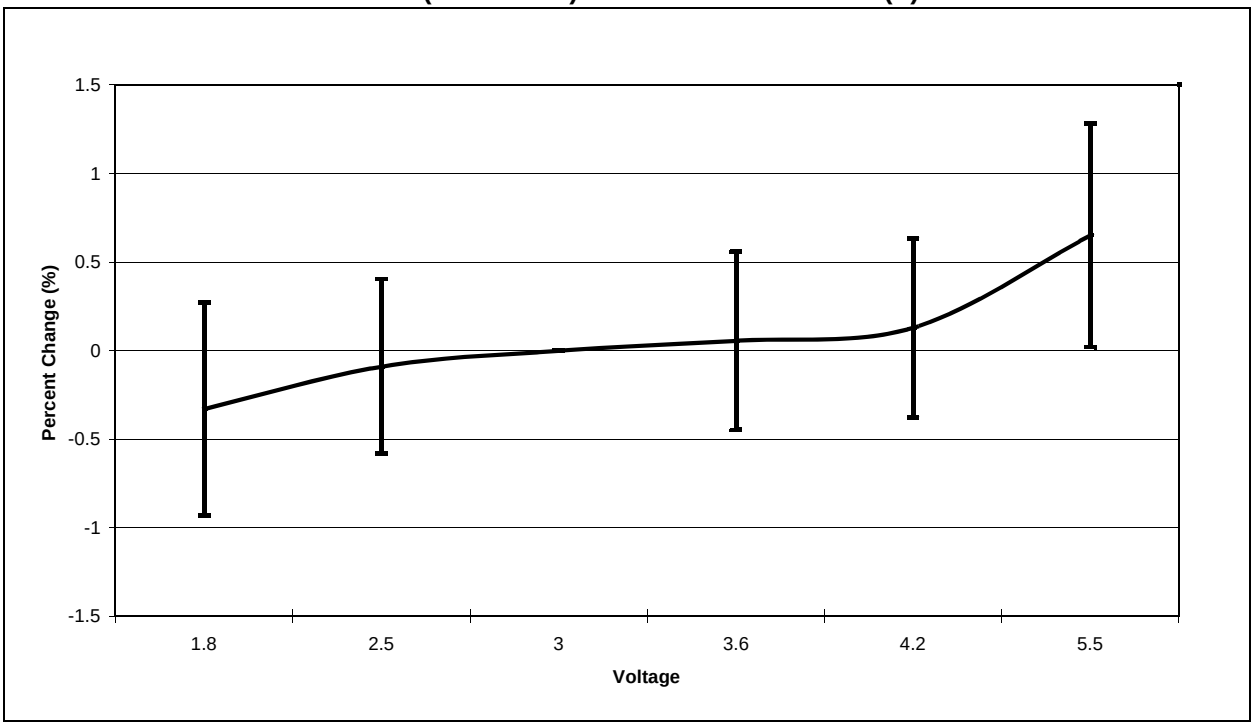
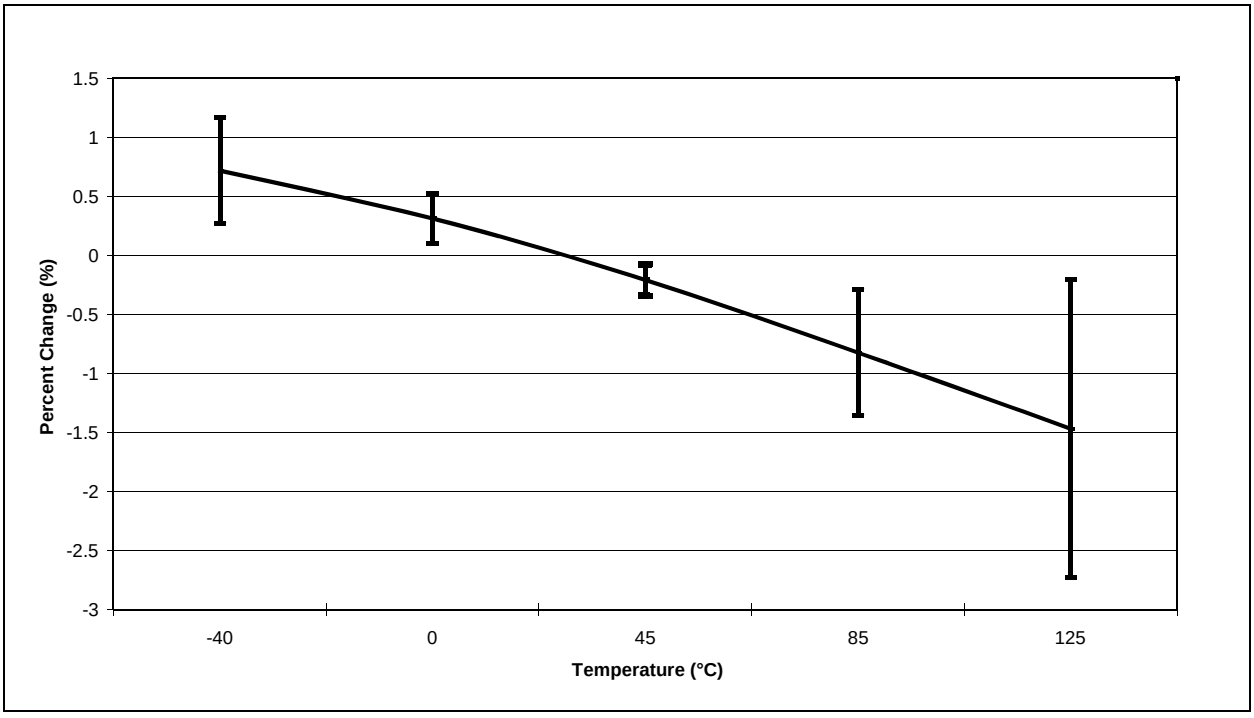


FIGURE 26-69: TYPICAL FVR CHANGE VS. TEMPERATURE NORMALIZED AT 25°C



NOTES:

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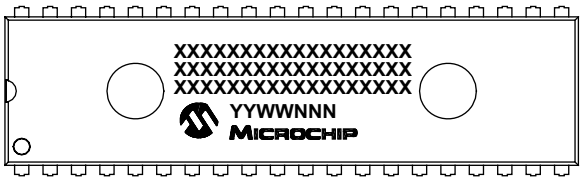
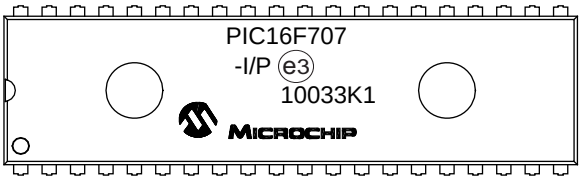
27.0 PACKAGING INFORMATION

27.1 Package Marking Information

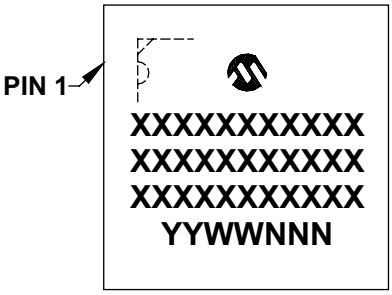
40-Lead PDIP (600 mil)



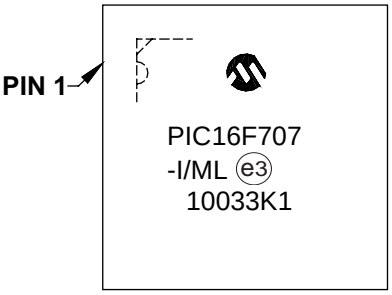
Example



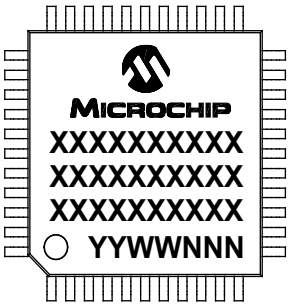
44-Lead QFN (8x8x0.9 mm)



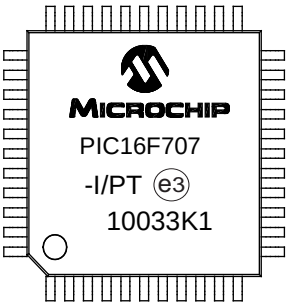
Example



44-Lead TQFP (10x10x1 mm)



Example

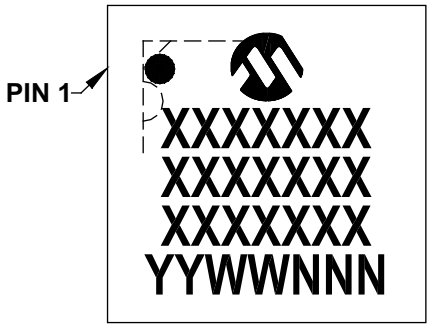


Legend:	XX...X	Customer-specific information
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week '01')
	NNN	Alphanumeric traceability code
	(e3)	Pb-free JEDEC designator for Matte Tin (Sn)
	*	This package is Pb-free. The Pb-free JEDEC designator ((e3)) can be found on the outer packaging for this package.

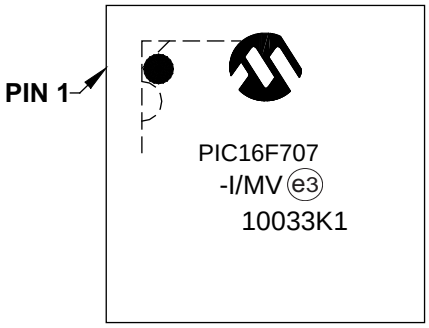
Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

***** Standard PICmicro® device marking consists of Microchip part number, year code, week code and traceability code. For PICmicro device marking beyond this, certain price adders apply. Please check with your Microchip Sales Office. For QTP devices, any special marking adders are included in QTP price.

40-Lead UQFN (5x5x0.5 mm)



Example



Legend:	XX...X	Customer-specific information
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week ‘01’)
	NNN	Alphanumeric traceability code
	(e3)	Pb-free JEDEC designator for Matte Tin (Sn)
	*	This package is Pb-free. The Pb-free JEDEC designator ((e3)) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

***** Standard PICmicro® device marking consists of Microchip part number, year code, week code and traceability code. For PICmicro device marking beyond this, certain price adders apply. Please check with your Microchip Sales Office. For QTP devices, any special marking adders are included in QTP price.

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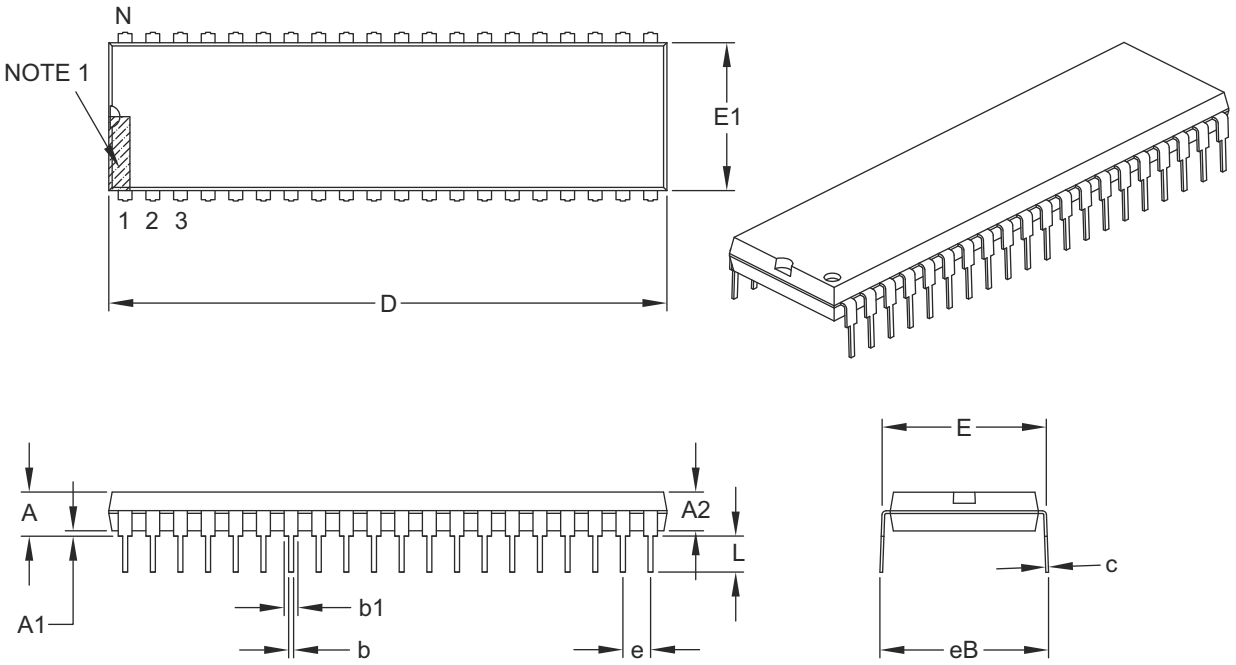
27.2 Package Details

The following sections give the technical details of the packages.

40-Lead Plastic Dual In-Line (P) – 600 mil Body [PDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		INCHES		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		40	
Pitch	e		.100 BSC	
Top to Seating Plane	A	–	–	.250
Molded Package Thickness	A2	.125	–	.195
Base to Seating Plane	A1	.015	–	–
Shoulder to Shoulder Width	E	.590	–	.625
Molded Package Width	E1	.485	–	.580
Overall Length	D	1.980	–	2.095
Tip to Seating Plane	L	.115	–	.200
Lead Thickness	c	.008	–	.015
Upper Lead Width	b1	.030	–	.070
Lower Lead Width	b	.014	–	.023
Overall Row Spacing §	eB	–	–	.700

Notes:

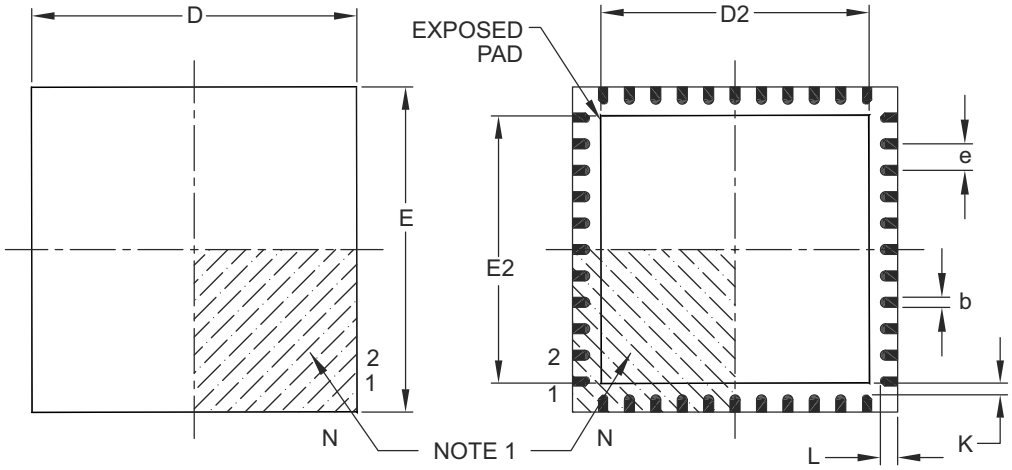
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. § Significant Characteristic.
3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
4. Dimensioning and tolerancing per ASME Y14.5M.

   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

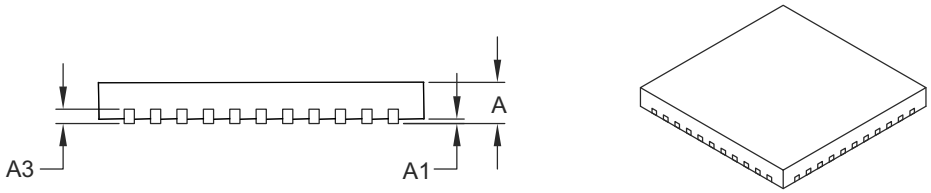
Microchip Technology Drawing C04-016B

44-Lead Plastic Quad Flat, No Lead Package (ML) – 8x8 mm Body [QFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



TOP VIEW **BOTTOM VIEW**



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		44	
Pitch	e		0.65 BSC	
Overall Height	A	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3		0.20 REF	
Overall Width	E		8.00 BSC	
Exposed Pad Width	E2	6.30	6.45	6.80
Overall Length	D		8.00 BSC	
Exposed Pad Length	D2	6.30	6.45	6.80
Contact Width	b	0.25	0.30	0.38
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	–	–

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Package is saw singulated.
3. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-103B

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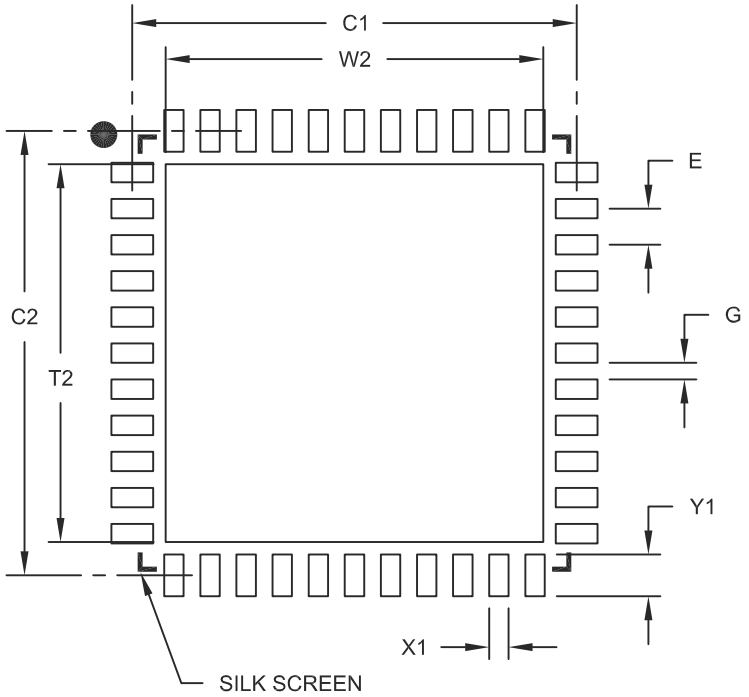
44-Lead Plastic Quad Flat, No Lead Package (ML) – 8x8 mm Body [QFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		0.65 BSC	
Optional Center Pad Width	W2			6.80
Optional Center Pad Length	T2			6.80
Contact Pad Spacing	C1		8.00	
Contact Pad Spacing	C2		8.00	
Contact Pad Width (X44)	X1			0.35
Contact Pad Length (X44)	Y1			0.80
Distance Between Pads	G	0.25		

Notes:

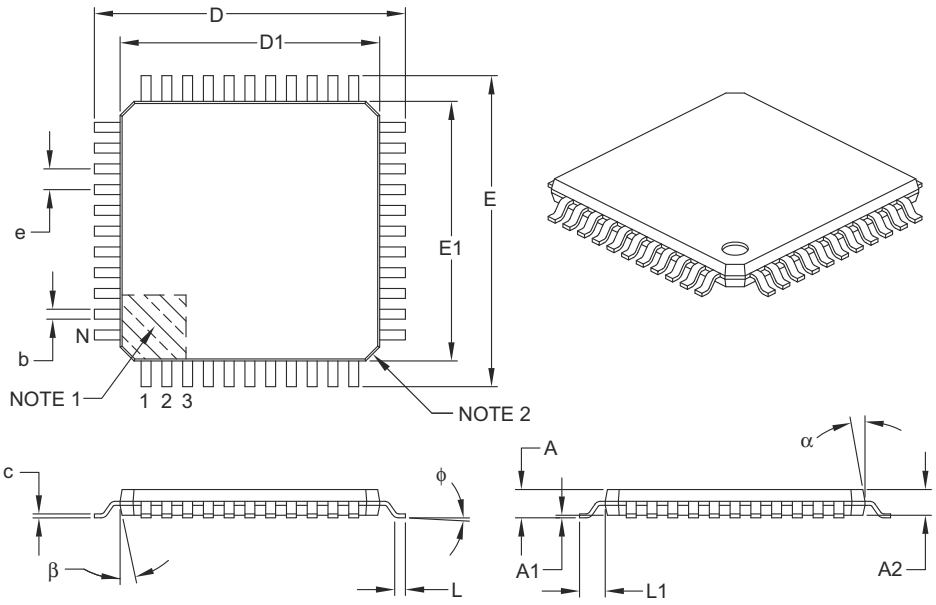
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2103A

44-Lead Plastic Thin Quad Flatpack (PT) – 10x10x1 mm Body, 2.00 mm [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Leads	N		44	
Lead Pitch	e		0.80 BSC	
Overall Height	A	–	–	1.20
Molded Package Thickness	A2	0.95	1.00	1.05
Standoff	A1	0.05	–	0.15
Foot Length	L	0.45	0.60	0.75
Footprint	L1		1.00 REF	
Foot Angle	φ	0°	3.5°	7°
Overall Width	E		12.00 BSC	
Overall Length	D		12.00 BSC	
Molded Package Width	E1		10.00 BSC	
Molded Package Length	D1		10.00 BSC	
Lead Thickness	c	0.09	–	0.20
Lead Width	b	0.30	0.37	0.45
Mold Draft Angle Top	α	11°	12°	13°
Mold Draft Angle Bottom	β	11°	12°	13°

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Chamfers at corners are optional; size may vary.
3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.
4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-076B

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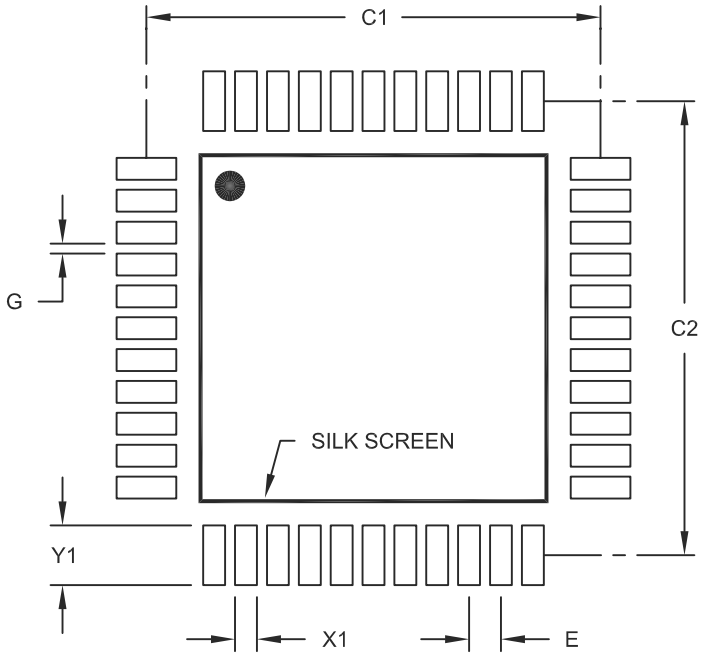
44-Lead Plastic Thin Quad Flatpack (PT) – 10x10x1 mm Body, 2.00 mm [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		0.80 BSC	
Contact Pad Spacing	C1		11.40	
Contact Pad Spacing	C2		11.40	
Contact Pad Width (X44)	X1			0.55
Contact Pad Length (X44)	Y1			1.50
Distance Between Pads	G	0.25		

Notes:

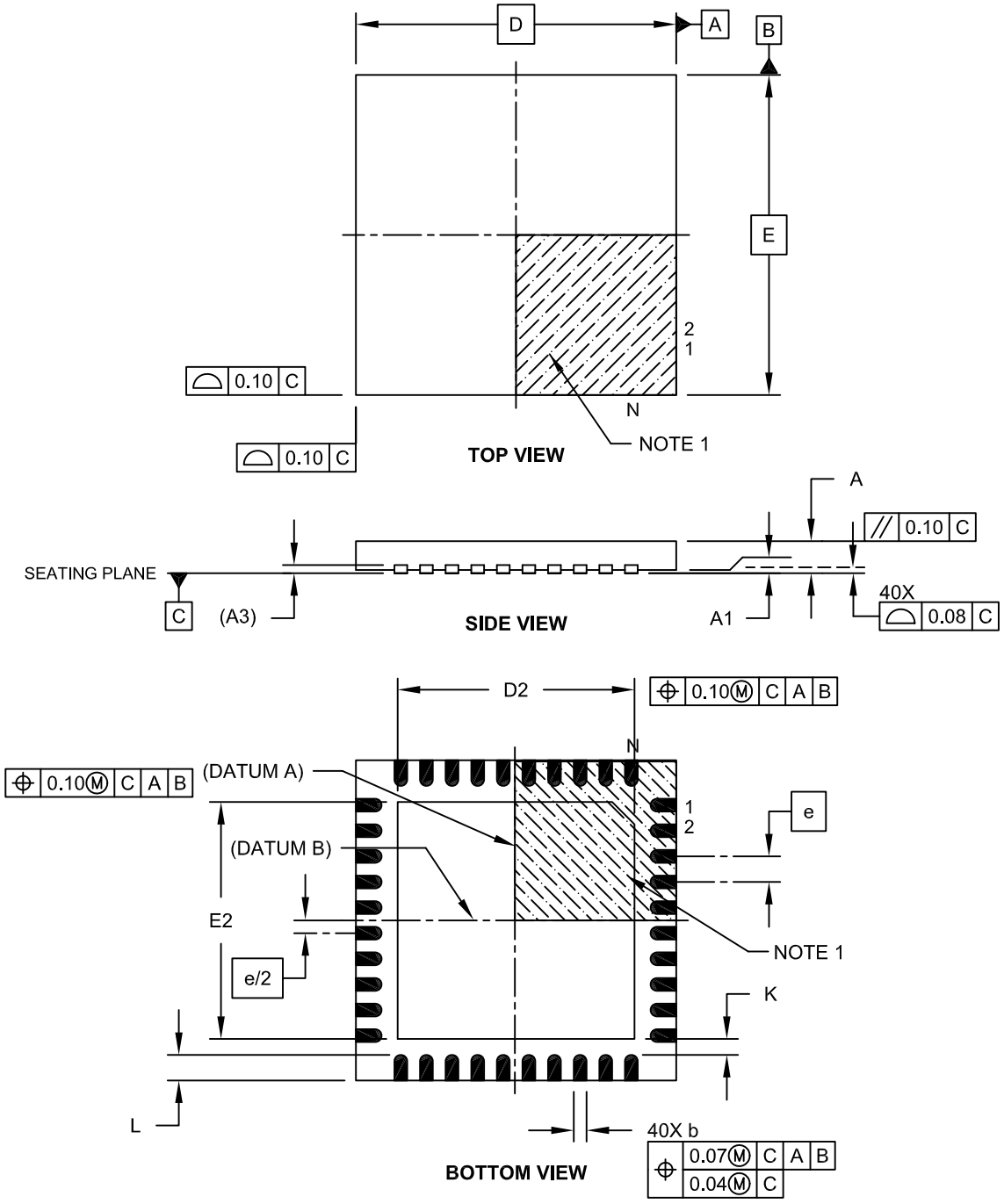
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2076A

40-Lead Ultra Thin Plastic Quad Flat, No Lead Package (MV) – 5x5x0.5 mm Body [UQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-156A Sheet 1 of 2

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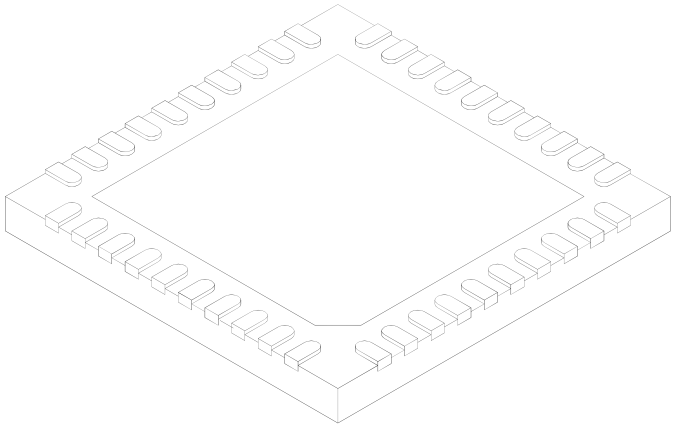
40-Lead Ultra Thin Plastic Quad Flat, No Lead Package (MV) – 5x5x0.5 mm Body [UQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		40	
Pitch	e		0.40 BSC	
Overall Height	A	0.45	0.50	0.55
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3		0.127 REF	
Overall Width	E		5.00 BSC	
Exposed Pad Width	E2	3.60	3.70	3.80
Overall Length	D		5.00 BSC	
Exposed Pad Length	D2	3.60	3.70	3.80
Contact Width	b	0.15	0.20	0.25
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	-	-

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Package is saw singulated.
3. Dimensioning and tolerancing per ASME Y14.5M.

   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

   REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-156A Sheet 2 of 2

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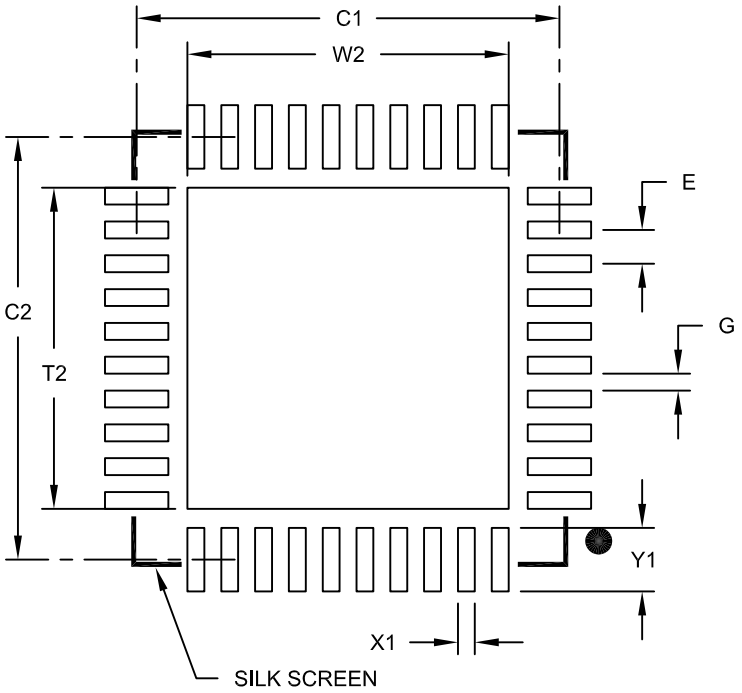
40-Lead Plastic Ultra Thin Quad Flat, No Lead Package (MV) - 5x5 mm Body [UQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		0.40 BSC	
Optional Center Pad Width	W2			3.80
Optional Center Pad Length	T2			3.80
Contact Pad Spacing	C1		5.00	
Contact Pad Spacing	C2		5.00	
Contact Pad Width (X40)	X1			0.20
Contact Pad Length (X40)	Y1			0.75
Distance Between Pads	G	0.20		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2156A

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APPENDIX A: DATA SHEET REVISION HISTORY

Revision A (4/2010)

Original release of this data sheet.

Revision B (4/2011)

Updated the data sheet to new format; Added 40-Pin UQFN diagram; Updated Table 1 and Table 25-5; Added 40-Lead UQFN Package Marking Information and Package Details; Other minor corrections.

APPENDIX B: MIGRATING FROM OTHER PIC® DEVICES

This discusses some of the issues in migrating from other PIC® devices to the PIC16F707 family of devices.

Note:	This device has been designed to perform to the parameters of its data sheet. It has been tested to an electrical specification designed to determine its conformance with these parameters. Due to process differences in the manufacture of this device, this device may have different performance characteristics than its ealier version. These differences may cause this device to perform differently in your application than the earlier version of this device.

Note:	The user should verify that the device oscillator starts and performs as expected. Adjusting the loading capacitor values and/or the oscillator mode may be required.

B.1 PIC16F77 to PIC16F707

TABLE B-1: FEATURE COMPARISON

Feature	PIC16F77	PIC16F707
Max. Operating Speed	20 MHz	20 MHz
Max. Program Memory (Words)	8K	8K
Max. SRAM (Bytes)	368	363
A/D Resolution	8-bit	8-bit
Timers (8/16-bit)	2/1	4/2
Oscillator Modes	4	8
Brown-out Reset	Y	Y
Internal Pull-ups	RB<7:0>	RB<7:0>
Interrupt-on-change	RB<7:4>	RB<7:0>
Comparator	0	0
USART	Y	Y
Extended WDT	N	N
Software Control Option of WDT/BOR	N	N
INTOSC Frequencies	None	500 kHz - 16 MHz
Clock Switching	N	N

NOTES:

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INDEX

A

A/D
- Specifications ... 223, 224

Absolute Maximum Ratings ... 203

AC Characteristics
- Industrial and Extended ... 215
- Load Conditions ... 214

ADC ... 81
- Acquisition Requirements ... 88
- Associated registers ... 90
- Block Diagram ... 81
- Calculating Acquisition Time ... 88
- Channel Selection ... 82
- Configuration ... 82
- Configuring Interrupt ... 84
- Conversion Clock ... 82
- Conversion Procedure ... 84
- Internal Sampling Switch (RSS) IMPEDANCE ... 88
- Interrupts ... 83
- Operation ... 84
- Operation During Sleep ... 84
- Port Configuration ... 82
- Reference Voltage (VREF) ... 82
- Source Impedance ... 88
- Special Event Trigger ... 84

ADCON0 Register ... 21, 86

ADCON1 Register ... 22, 87

Addressable Universal Synchronous Asynchronous
- Receiver Transmitter (AUSART) ... 139

ADRES Register ... 87

ADRESH Register ... 21

Alternate Pin Function ... 53

Analog-to-Digital Converter. *See* ADC

ANSELA Register ... 55

ANSELB Register ... 59, 62

ANSELD Register ... 65

ANSELE Register ... 68

APFCON Register ... 53

Assembler
- MPASM Assembler ... 200

AUSART ... 139
- Associated Registers
  - Baud Rate Generator ... 149
- Asynchronous Mode ... 141
  - Associated Registers
    - Receive ... 146
    - Transmit ... 143
  - Baud Rate Generator (BRG) ... 149
  - Receiver ... 143
  - Setting up 9-bit Mode with Address Detect ... 145
  - Transmitter ... 141
- Baud Rate Generator (BRG)
  - Baud Rate Error, Calculating ... 149
  - Formulas ... 149
  - High Baud Rate Select (BRGH Bit) ... 149
- Synchronous Master Mode ... 152, 156
  - Associated Registers
    - Receive ... 155
    - Transmit ... 153
  - Reception ... 154
  - Transmission ... 152
- Synchronous Slave Mode
  - Associated Registers
    - Receive ... 157
    - Transmit ... 156
  - Reception ... 157
  - Transmission ... 156

B

BF bit ... 167, 179

Block Diagrams
- (CCP) Capture Mode Operation ... 131
- ADC ... 81
- ADC Transfer Function ... 89
- Analog Input Model ... 89
- AUSART Receive ... 140
- AUSART Transmit ... 139
- CCP PWM ... 135
- Clock Source ... 71
- Compare ... 133
- Crystal Operation ... 75
- Digital-to-Analog Converter (DAC) ... 94
- External RC Mode ... 76
- Interrupt Logic ... 41
- MCLR Circuit ... 33
- On-Chip Reset Circuit ... 31
- Resonator Operation ... 76
- SPI Mode ... 160
- SSP (I^2C Mode) ... 169
- Timer1 ... 102
- Timer2 ... 117
- TMR0/WDT Prescaler ... 97, 113
- Voltage Reference ... 94
- Voltage Reference Output Buffer Example ... 94

Brown-out Reset (BOR) ... 35
- Specifications ... 221
- Timing and Characteristics ... 220

C

C Compilers
- MPLAB C18 ... 200

Capacitive Sensing ... 119

Capture Module. *See* Capture/Compare/PWM (CCP)

Capture/Compare/PWM (CCP) ... 129
- Associated registers w/ Capture ... 132
- Associated registers w/ Compare ... 134
- Associated registers w/ PWM ... 138
- Capture Mode ... 131
- CCPx Pin Configuration ... 131
- Compare Mode ... 133
  - CCPx Pin Configuration ... 133
  - Software Interrupt Mode ... 131, 133
  - Special Event Trigger ... 133
  - Timer1 Mode Selection ... 131, 133
- Interaction of Two CCP Modules (table) ... 129
- Prescaler ... 131
- PWM Mode ... 135
  - Duty Cycle ... 136
  - Effects of Reset ... 137
  - Example PWM Frequencies and Resolutions, 20 MHZ ... 137
  - Example PWM Frequencies and Resolutions, 8 MHz ... 137
  - Operation in Sleep Mode ... 137
  - Setup for Operation ... 137
  - System Clock Frequency Changes ... 137
- PWM Period ... 136

Setup for PWM Operation 137
Timer Resources 129
CCP. *See* Capture/Compare/PWM (CCP)
CCP1CON Register 21
CCP2CON Register 21
CCPR1H Register 21
CCPR1L Register 21
CCPR2H Register 21
CCPR2L Register 21
CCPxCON Register 130
CKE bit 167, 179
CKP bit 166, 178
Clock Sources
External Modes 75
EC 75
HS 75
LP 75
OST 75
RC 76
XT 75
Code Examples
A/D Conversion 85
Call of a Subroutine in Page 1 from Page 0 28
Changing Between Capture Prescalers 131
Indirect Addressing 29
Initializing PORTA 54
Initializing PORTB 57
Initializing PORTC 61
Initializing PORTD 64
Initializing PORTE 67
Loading the SSPBUF (SSPSR) Register 162
Saving W, STATUS and PCLATH Registers in RAM . 43
Compare Module. *See* Capture/Compare/PWM (CCP)
CONFIG1 Register 77, 78
Customer Change Notification Service 287
Customer Notification Service 287
Customer Support 287

D

D/$\overline{A}$ bit 179
DACCON0 (Digital-to-Analog Converter Control 0) Register 95
DACCON1 (Digital-to-Analog Converter Control 1) Register 95
Data Memory 19
Data/Address bit (D/$\overline{A}$) 179
DC and AC Characteristics 233
DC Characteristics
Extended and Industrial 211
Industrial and Extended 204
Development Support 199
Device Configuration 77
Code Protection 79
Configuration Word 77
User ID 79
Device Overview 13
Digital-to-Analog Converter (DAC) 93
Associated Registers 96
Effects of a Reset 93
Operation During Sleep 93

E

EECON1 Register 24
Effects of Reset
PWM mode 137
Electrical Specifications 203
Enhanced Capture/Compare/PWM (ECCP) Specifications 223
Errata 11

F

Firmware Instructions 189
Fixed Voltage Reference. *See* FVR
FSR Register 21, 22
Fuses. *See* Configuration Bits
FVR 91
FVRCON Register 92

G

General Purpose Register File 19

I

I^2C Mode
Associated Registers 180
INDF Register 21, 22
Indirect Addressing, INDF and FSR Registers 29
Instruction Format 189
Instruction Set 189
ADDLW 191
ADDWF 191
ANDLW 191
ANDWF 191
MOVF 194
BCF 191
BSF 191
BTFSC 191
BTFSS 192
CALL 192
CLRF 192
CLRW 192
CLRWDT 192
COMF 192
DECF 192
DECFSZ 193
GOTO 193
INCF 193
INCFSZ 193
IORLW 193
IORWF 193
MOVLW 194
MOVWF 194
NOP 194
RETFIE 195
RETLW 195
RETURN 195
RLF 196
RRF 196
SLEEP 196
SUBLW 196
SUBWF 197
SWAPF 197
XORLW 197
XORWF 197
Summary Table 190
INTCON Register 44
Internal Oscillator Block
INTOSC
Specifications 218
Internal Sampling Switch (RSS) IMPEDANCE 88
Internet Address 287
Interrupts 41
ADC 84

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Associated registers w/ Interrupts 49
Configuration Word w/ LDO 51
Interrupt-on-Change 57
Synchronous Serial Port Interrupt 48
INTOSC Specifications 218
IOCB Register 59

L

Load Conditions 214

M

MCLR 33
Internal 33
Memory Organization 19
Data 19
Program 19
Microchip Internet Web Site 287
Migrating from other PIC Microcontroller Devices 279
MPLAB ASM30 Assembler, Linker, Librarian 200
MPLAB Integrated Development Environment Software 199
MPLAB PM3 Device Programmer 202
MPLAB REAL ICE In-Circuit Emulator System 201
MPLINK Object Linker/MPLIB Object Librarian 200

O

OPCODE Field Descriptions 189
OPTION Register 26
OPTION_REG Register 99
OSCCON Register 73
Oscillator
Associated registers 76, 117
Oscillator Module
EC 71
HS 71
INTOSC 71
INTOSCIO 71
LP 71
Oscillator Tuning 74
RC 71
RCIO 71
XT 71
Oscillator Parameters 218
Oscillator Specifications 217
Oscillator Start-up Timer (OST)
Specifications 221
OSCTUNE Register 74

P

P (Stop) bit 179
Packaging 269
Marking 269
PDIP Details 271
Paging, Program Memory 28
PCL and PCLATH 28
Computed GOTO 28
Stack 28
PCL Register 21, 22
PCLATH Register 21, 22
PCON Register 22, 27, 36
PIE1 Register 22, 45
PIE2 Register 22
Pinout Descriptions
PIC16F707/PIC16LF707 15
PIR1 Register 21, 47
PIR2 Register 21, 48
PMADRH Register 183
PMADRL Register 183
PMCON1 Register 182, 183
PMDATH Register 182
PMDATL Register 182
PORTA 54
ANSELA Register 55
Associated Registers 56
Pin Descriptions and Diagrams 55
PORTA Register 21
RA0 55
RA3 56
RA4 56
RA5 56
RA6 56
RA7 56
Specifications 219
PORTA Register 54
PORTB 57
Additional Pin Functions
ANSELB Register 57, 62, 67
Weak Pull-up 57
Associated Registers 60
Interrupt-on-Change 57
P1B/P1C/P1D.*See* Enhanced Capture/Compare/
PWM+ (ECCP+) 57
Pin Descriptions and Diagrams 59
PORTB Register 21
RB0 59
RB1 59
RB2 59
RB3 60
RB4 60
RB5 60
RB6 60
RB7 60
PORTB Register 58
PORTC 61
Associated Registers 63
P1A.*See* Enhanced Capture/Compare/PWM+
(ECCP+) 61
PORTC Register 21
RC0 62
RC2 63
RC3 63
RC4 63
RC5 63
RC6 63
RC7 63
Specifications 219
PORTC Register 61
PORTD 64
Additional Pin Functions
ANSELD Register 64
Associated Registers 66
P1B/P1C/P1D.*See* Enhanced Capture/Compare/
PWM+ (ECCP+) 64
PORTD Register 21
RD6 ..... 65, 66
PORTD Register 64
PORTE 67
Associated Registers 69
PORTE Register 21
RE0 68
RE1 68
RE2 68

RE3 69
PORTE Register 67
Power-Down Mode (Sleep) 185
Associated Registers 186
Power-on Reset 33
Power-up Timer (PWRT) 33
Specifications 221
PR2 Register 22, 168
Precision Internal Oscillator Parameters 218
Prescaler
Shared WDT/Timer0 ..... 98, 114
Product Identification System 289
Program 19
Program Memory 19
Map and Stack (PIC16F707/PIC16LF707) 19
Paging 28
Program Memory Read (PMR) 181
Associated Registers 183
Programming, Device Instructions 189

R

R/$\overline{W}$ bit 179
RCREG 145
RCREG Register 21
RCSTA Register 21, 148
Reader Response 288
Read-Modify-Write Operations 189
Receive Overflow Indicator bit (SSPOV) 166, 178
Registers
ADCON0 (ADC Control 0) 86
ADCON1 (ADC Control 1) 87
ADRES (ADC Result) 87
ANSELA (PORTA Analog Select) 55
ANSELB (PORTB Analog Select) 59, 62
ANSELD (PORTD Analog Select) 65
ANSELE (PORTE Analog Select) 68
APFCON (Alternate Pin Function Control) 53
CCPxCON (CCP Operation) 130
CONFIG1 (Configuration Word Register 1) 77, 78
DACCON0 95
DACCON1 95
FVRCON (Fixed Voltage Reference Register) 92
INTCON (Interrupt Control) 44
IOCB (Interrupt-on-Change PORTB) 59
OPTION_REG (OPTION) 26
OPTION_REG (Option) 99
OSCCON (Oscillator Control) 73
OSCTUNE (Oscillator Tuning) 74
PCON (Power Control Register) 27
PCON (Power Control) 36
PIE1 (Peripheral Interrupt Enable 1) 45
PIR1 (Peripheral Interrupt Register 1) 47
PIR2 (Peripheral Interrupt Request 2) 48
PMADRH (Program Memory Address High) 183
PMADRL (Program Memory Address Low) 183
PMCON1 (Program Memory Control 1) 182
PMDATH (Program Memory Data High) 182
PMDATL (Program Memory Data Low) 182
PORTA 54
PORTB 58
PORTC 61
PORTD 64
PORTE 67
RCSTA (Receive Status and Control) 148
Reset Values 38
Reset Values (Special Registers) 40
SSPCON (Sync Serial Port Control) Register .. 166, 178
SSPSTAT (Sync Serial Port Status) Register ... 167, 179
STATUS 25
T2CON 118
TRISA (Tri-State PORTA) 54
TRISB (Tri-State PORTB) 58
TRISC (Tri-State PORTC) 61
TRISD (Tri-State PORTD) 65
TRISE (Tri-State PORTE) 68
TXSTA (Transmit Status and Control) 147
WPUB (Weak Pull-up PORTB) 58
Reset 31
Resets
Associated Registers 40
Revision History 279

S

S (Start) bit 179
SMP bit 167, 179
Software Simulator (MPLAB SIM) 201
SPBRG 149
SPBRG Register 22
Special Event Trigger 84
Special Function Registers 19
SPI Mode 165
Associated Registers 168
Typical Master/Slave Connection 159
SSP 159
I^2C Mode 169
Acknowledge 170
Addressing 171
Clock Stretching 176
Clock Synchronization 177
Firmware Master Mode 176
Hardware Setup 169
Multi-Master Mode 176
Reception 172
Sleep Operation 177
Start/Stop Conditions 170
Transmission 174
Master Mode 160
SPI Mode 159
Slave Mode 163
Typical SPI Master/Slave Connection 159
SSPADD Register 22
SSPBUF Register 21
SSPCON Register 21, 166, 178
SSPEN bit 166, 178
SSPIF 48
SSPM bits 166, 178
SSPOV bit 166, 178
SSPSTAT Register 22, 167, 179
STATUS Register 25
Synchronous Serial Port Enable bit (SSPEN) 166, 178
Synchronous Serial Port Interrupt 48
Synchronous Serial Port Mode Select bits (SSPM) .. 166, 178

T

T1CON Register 21, 22
T2CON Register 21, 118, 168
Thermal Considerations 213
Time-out Sequence 36
Timer0 97
Associated Registers 99, 115
Interrupt 99
Operation 98, 103, 114

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Specifications 222
Timer1 101
Asynchronous Counter Mode 104
Reading and Writing 104
Modes of Operation 103
Oscillator 104
Prescaler 104
Specifications 222
TMR1H Register 101
TMR1L Register 101
Timer2
Associated registers 118
Timers
Timer2
T2CON 118
Timing Diagrams
A/D Conversion 224
A/D Conversion (Sleep Mode) 225
Asynchronous Reception 146
Asynchronous Transmission 142
Asynchronous Transmission (Back-to-Back) 143
Brown-out Reset (BOR) 220
Brown-out Reset Situations 35
CLKOUT and I/O 218
Clock Synchronization 177
Clock Timing 215
Enhanced Capture/Compare/PWM (ECCP) 223
I2C Bus Data 230
I2C Bus Start/Stop Bits 229
I2C Reception (7-bit Address) 172
I2C Slave Mode with SEN = 0 (Reception, 10-bit Address) 173
I2C Transmission (7-bit Address) 174
INT Pin Interrupt 42
Reset, WDT, OST and Power-up Timer 220
Slave Select Synchronization 165
SPI Master Mode 162
SPI Master Mode (CKE = 1, SMP = 1) 227
SPI Mode (Slave Mode with CKE = 0) 164
SPI Mode (Slave Mode with CKE = 1) 164
SPI Slave Mode (CKE = 0) 227
SPI Slave Mode (CKE = 1) 228
Synchronous Reception (Master Mode, SREN) 155
Synchronous Transmission 153
Synchronous Transmission (Through TXEN) 153
Time-out Sequence
Case 1 36
Case 2 37
Case 3 37
Timer0 and Timer1 External Clock 222
USART Synchronous Receive (Master/Slave) 226
USART Synchronous Transmission (Master/Slave) 225
Wake-up from Interrupt 186
Timing Parameter Symbology 214
Timing Requirements
I2C Bus Data 231
I2C Bus Start/Stop Bits 230
SPI Mode 229
TMR0 Register 21
TMR1H Register 21, 22
TMR1L Register 21, 22
TMR2 Register 21
TMRO Register 23
TRISA 54
TRISA Register 22, 54
TRISB 57
TRISB Register 22, 58
TRISC 61
TRISC Register 22, 61
TRISD 64
TRISD Register 22, 65
TRISE 67
TRISE Register 22, 68
TXREG 141
TXREG Register 21
TXSTA Register 22, 147
BRGH Bit 149

U

UA 179
Update Address bit, UA 179
USART
Synchronous Master Mode
Requirements, Synchronous Receive 226
Requirements, Synchronous Transmission 225
Timing Diagram, Synchronous Receive 226
Timing Diagram, Synchronous Transmission 225

V

VREF. *SEE* ADC Reference Voltage

W

Wake-up Using Interrupts 186
Watchdog Timer (WDT) 33
Clock Source 33
Modes 34
Period 33
Specifications 221
WCOL bit 166, 178
WPUB Register 58
Write Collision Detect bit (WCOL) 166, 178
WWW Address 287
WWW, On-Line Support 11

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Temperature Range:	I = -40°C to +85°C E = -40°C to +125°C	
Package:	MV = Micro Lead Frame (UQFN) ML = Micro Lead Frame (QFN) P = Plastic DIP PT = TQFP (Thin Quad Flatpack)	
Pattern:	3-Digit Pattern Code for QTP (blank otherwise)	

Examples:

a) PIC16F707-E/P 301 = Extended Temp., PDIP package, QTP pattern #301

b) PIC16F707-I/ML = Industrial Temp., QFN package

Note 1: T = In tape and reel.

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