



± 15 kV ESD Protected, EMC Compliant Slew Rate Limited, EIA RS-485 Transceiver

ADM483E

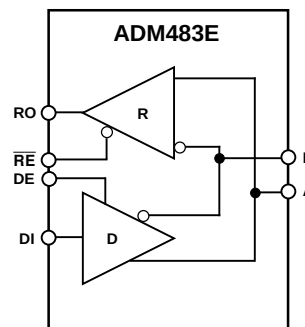
FEATURES

- Robust RS-485 Transceiver
- 15 kV ESD Protection Using HBM
- 2 kV EFT Protection Meets IEC1000-4-4
- High EM Immunity Meets IEC1000-4-3
- Reduced Slew Rate for Low EM Interference
- 250 kbps Data Rate
- Single +5 V $\pm$ 10% Supply
- 7 V to +12 V Bus Common-Mode Range
- 12 k Ω Input Impedance
- Short Circuit Protection
- Excellent Noise Immunity
- 36 μ A Supply Current
- 0.1 μ A Shutdown Current

APPLICATIONS

- Low Power RS-485 Systems
- Electrically Harsh Environments
- EMI Sensitive Applications
- DTE-DCE Interface
- Packet Switching
- Local Area Networks

FUNCTIONAL BLOCK DIAGRAM



GENERAL DESCRIPTION

The ADM483E is a robust, low power differential line transceiver suitable for communication on multipoint bus transmission lines. Internal protection against electrostatic discharge (ESD), electrical fast transient (EFT) and electromagnetic immunity (EMI) allows operation in electrically harsh environments. ESD protection on the I-O lines meets ± 15 kV when tested using the Human Body Model. EFT protection meets ± 2 kV in accordance with IEC1000-4-4, while EMI immunity is in excess of 10 V/m meeting IEC1000-4-3.

The level of unwanted emissions is also carefully controlled using slew limiting on the driver outputs. This reduces reflections with improperly terminated cables and also minimizes electromagnetic interference. The controlled slew rate limits the data rate to 250 kbps.

The ADM483E is intended for balanced data transmission and complies with both EIA Standards RS-485 and RS-422. It contains a differential line driver and a differential line receiver and is suitable for half duplex data transmission, as the driver and receiver share the same differential pins.

The input impedance on the ADM483E is 12 k Ω , allowing up to 32 transceivers on the bus.

The ADM483E operates from a single +5 V $\pm$ 10% power supply. Excessive power dissipation caused by bus contention or by output shorting is prevented by a thermal shutdown circuit. This feature forces the driver output into a high impedance state if, during fault conditions, a significant temperature increase is detected in the internal driver circuitry.

The receiver contains a fail-safe feature that results in a logic high output state if the inputs are unconnected (floating).

The ADM483E is fabricated on BiCMOS, an advanced mixed technology process combining low power CMOS with robust bipolar technology.

It is fully specified over the industrial temperature range and is available in 8-lead DIP and SOIC packages.

REV. 0

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ADM483E—SPECIFICATIONS ($V_{CC} = +5\text{ V} \pm 10\%$. All specifications T_{MIN} to T_{MAX} unless otherwise noted)

Parameter	Min	Typ	Max	Units	Test Conditions/Comments
DRIVER					
Differential Output Voltage, V_{OD}			5.0	V	$V_{CC} = 5.25\text{ V}$. $R = \infty$, Figure 1
	2.0		5.0	V	$R = 50\ \Omega$ (RS-422), Figure 1
	1.5		5.0	V	$R = 27\ \Omega$ (RS-485), Figure 1
	1.5		5.0	V	$V_{TST} = -7\text{ V}$ to $+12\text{ V}$, Figure 2, $V_{CC} \geq 4.75\text{ V}$
$\Delta V_{OD} $ for Complementary Output States			0.2	V	$R = 27\ \Omega$ or $50\ \Omega$, Figure 1
Common-Mode Output Voltage V_{OC}			3	V	$R = 27\ \Omega$ or $50\ \Omega$, Figure 1
$\Delta V_{OC} $ for Complementary Output States			0.2	V	$R = 27\ \Omega$ or $50\ \Omega$
Output Short Circuit Current ($V_{OUT} = \text{High}$)			250	mA	$-7\text{ V} \leq V_O \leq +12\text{ V}$
Output Short Circuit Current ($V_{OUT} = \text{Low}$)			250	mA	$-7\text{ V} \leq V_O \leq +12\text{ V}$
CMOS Input Logic Threshold Low, V_{INL}		1.4	0.8	V	
CMOS Input Logic Threshold High, V_{INH}	2.0	1.4		V	
Logic Input Current (DE, DI)			± 1.0	μA	
RECEIVER					
Differential Input Threshold Voltage, V_{TH}	-0.2		+0.2	V	$-7\text{ V} \leq V_{CM} \leq +12\text{ V}$
Input Voltage Hysteresis, ΔV_{TH}		70		mV	$V_{CM} = 0\text{ V}$
Input Resistance	12			k Ω	$-7\text{ V} \leq V_{CM} \leq +12\text{ V}$
Input Current (A, B)			+1	mA	$V_{IN} = 12\text{ V}$
	-0.8			mA	$V_{IN} = -7\text{ V}$
Logic Enable Input Current ($\overline{RE}$)			± 1	μA	
CMOS Output Voltage Low, V_{OL}			0.4	V	$I_{OUT} = +4.0\text{ mA}$
CMOS Output Voltage High, V_{OH}	4.0			V	$I_{OUT} = -4.0\text{ mA}$
Short Circuit Output Current	7		85	mA	$V_{OUT} = \text{GND}$ or V_{CC}
Three-State Output Leakage Current			± 1.0	μA	$0.4\text{ V} \leq V_{OUT} \leq +2.4\text{ V}$
POWER SUPPLY CURRENT					
I_{CC} (ADM483E)		36	60	μA	Outputs Unloaded, Receivers Enabled
		270	360	μA	$DE = 0\text{ V}$ (Disabled) $\overline{RE} = 0\text{ V}$
Supply Current in Shutdown		0.1	10	μA	$DE = 5\text{ V}$ (Enabled) $\overline{RE} = 0\text{ V}$
					$DE = 0\text{ V}$, $\overline{RE} = V_{CC}$
ESD/EFT IMMUNITY					
ESD Protection		± 15		kV	HBM Air Discharge. A, B Pins
		± 3.5		kV	HBM 3015.7 Contact Discharge. All Pins
EFT Protection		± 2		kV	IEC1000-4-4, A, B Pins
EMI Immunity		10		V/m	IEC1000-4-3

Specifications subject to change without notice.

TIMING SPECIFICATIONS ($V_{CC} = +5\text{ V} \pm 10\%$. All specifications T_{MIN} to T_{MAX} unless otherwise noted.)

Parameter	Min	Typ	Max	Units	Test Conditions/Comments
DRIVER					
Propagation Delay Input to Output T_{PLH} , T_{PHL}	250		2000	ns	$R_L \text{ Diff} = 54\ \Omega$ $C_{L1} = C_{L2} = 100\text{ pF}$, Figure 5
Driver O/P to $\overline{O/P}$ T_{SKEW}		100	800	ns	$R_L \text{ Diff} = 54\ \Omega$ $C_{L1} = C_{L2} = 100\text{ pF}$, Figure 5
Driver Rise/Fall Time T_R , T_F	250		2000	ns	$R_L \text{ Diff} = 54\ \Omega$ $C_{L1} = C_{L2} = 100\text{ pF}$, Figure 5
Driver Enable to Output Valid	250		2000	ns	$R_L = 500\ \Omega$, $C_L = 100\text{ pF}$, Figure 3
Driver Disable Timing	300		3000	ns	$R_L = 500\ \Omega$, $C_L = 15\text{ pF}$, Figure 3
RECEIVER					
Propagation Delay Input to Output T_{PLH} , T_{PHL}	250		2000	ns	$C_L = 15\text{ pF}$, Figure 5
Skew $ T_{PLH} - T_{PHL} $		200		ns	
Receiver Enable T_{EN1}		10	50	ns	$R_L = 1\text{ k}\Omega$, $C_L = 15\text{ pF}$, Figure 4
Receiver Disable T_{EN2}		10	50	ns	$R_L = 1\text{ k}\Omega$, $C_L = 15\text{ pF}$, Figure 4
SHUTDOWN					
Time to Shutdown	50	200	600	ns	
Driver Enable from Shutdown			2000	ns	$R_L = 500\ \Omega$, $C_L = 100\text{ pF}$, Figure 3
Receiver Enable from Shutdown			2500	ns	$R_L = 1\text{ k}\Omega$, $C_L = 15\text{ pF}$, Figure 4

Specifications subject to change without notice.

ABSOLUTE MAXIMUM RATINGS*

(T_A = +25°C unless otherwise noted)

V_{CC}+7 V

Inputs

Driver Input (DI) -0.5 V to V_{CC} + 0.5 V

Control Inputs (DE, RE) -0.5 V to V_{CC} + 0.5 V

Receiver Inputs (A, B) -14 V to +14 V

Outputs

Driver Outputs -12.5 V to +12.5 V

Receiver Output -0.5 V to V_{CC} +0.5 V

ESD Rating: Air (Human Body Model) (A, B Pins) .. ±15 kV

ESD Rating: Contact (Human Body Model)

(A, B Pins) ±8 kV

ESD Rating MIL-STD-883B Method 3015

(Except A, B) ±3.5 kV

EFT Rating (IEC1000-4-4) (A, B Pins) ±2 kV

EMI Immunity (IEC1000-4-3) 10 V/m

Power Dissipation 8-Pin DIP 727 mW

θ_{JA}, Thermal Impedance +135°C/W

Power Dissipation 8-Pin SOIC 470 mW

θ_{JA}, Thermal Impedance +110°C/W

Operating Temperature Range

Industrial (A Version) -40°C to +85°C

Storage Temperature Range -65°C to +150°C

Lead Temperature (Soldering, 10 sec) +300°C

Vapor Phase (60 sec) +215°C

Infrared (15 sec) +220°C

*Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum ratings for extended periods may affect device reliability.

ORDERING GUIDE

Model	Temperature Range	Package Option
ADM483EAN	-40°C to +85°C	N-8
ADM483EAR	-40°C to +85°C	SO-8

PIN FUNCTION DESCRIPTION

Pin	Mnemonic	Function
1	RO	Receiver Output. When enabled if A > B by 200 mV, then RO = High. If A < B by 200 mV, then RO = Low.
2	$\overline{\text{RE}}$	Receiver Output Enable. A low level enables the receiver output, RO. A high level places it in a high impedance state.
3	DE	Driver Output Enable. A high level enables the driver differential outputs, A and B. A low level places it in a high impedance state.
4	DI	Driver Input. When the driver is enabled a logic Low on DI forces A low and B high while a logic High on DI forces A high and B low.
5	GND	Ground Connection, 0 V.
6	A	Noninverting Receiver Input A/Driver Output A.
7	B	Inverting Receiver Input B/Driver Output B.
8	V _{CC}	Power Supply, 5 V ± 10%.

PIN CONFIGURATION

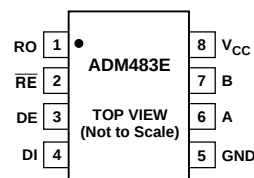


Table I. Selection Table

Part No.	Duplex	Data Rate kb/s	Low Power Shutdown	Tx/Rx Enable	I _{CC} μA	No of Tx/Rx On Bus	ESD kV	EFT kV	EMI V/m
ADM483E	Half	250	Yes	Yes	36	32	±15	±2	10

ADM483E

Test Circuits

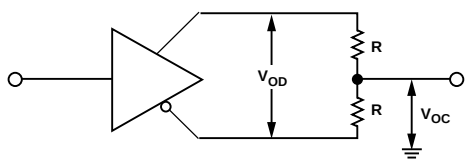


Figure 1. Driver Voltage Measurement Test Circuit

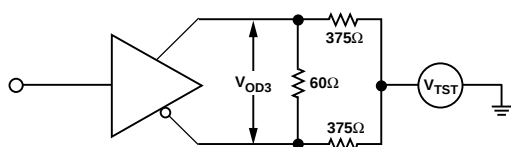


Figure 2. Driver Voltage Measurement Test Circuit 2

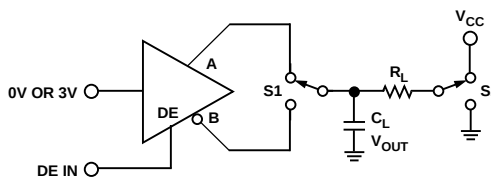


Figure 3. Driver Enable/Disable Test Circuit

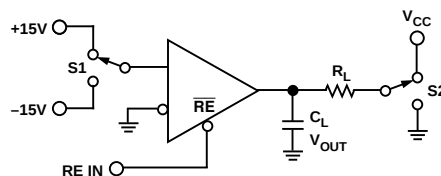


Figure 4. Receiver Enable/Disable Test Circuit

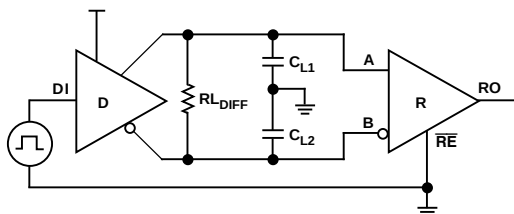


Figure 5. Receiver Propagation Delay Test Circuit

Switching Characteristics

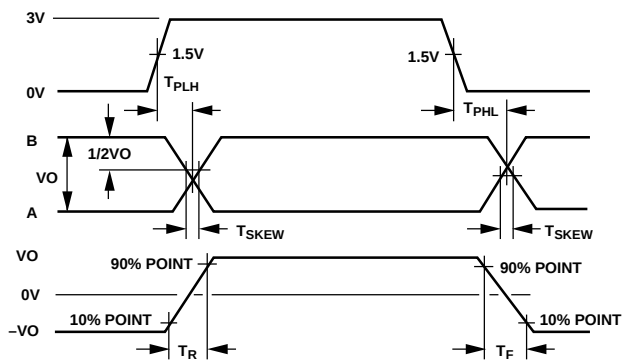


Figure 6. Driver Propagation Delay, Rise/Fall Timing

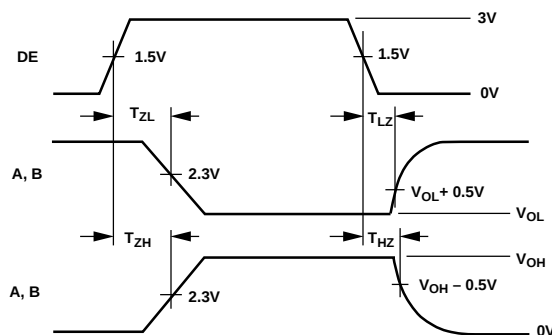


Figure 7. Driver Enable/Disable Timing

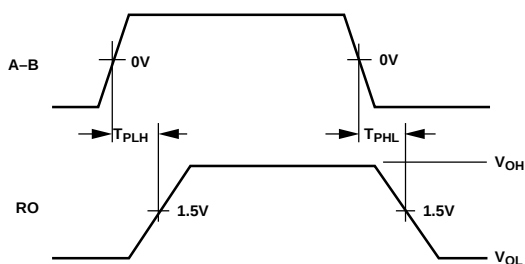


Figure 8. Receiver Propagation Delay

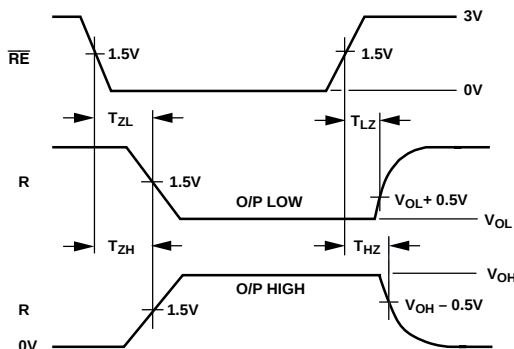


Figure 9. Receiver Enable/Disable Timing

Typical Performance Characteristics—ADM483E

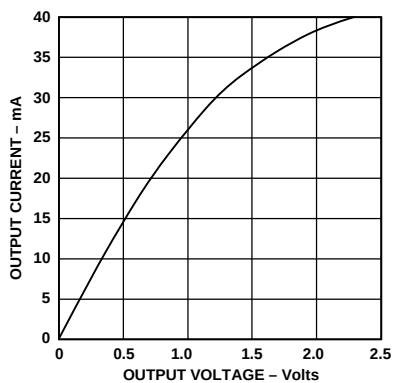


Figure 11. Receiver Output Low Voltage vs. Output Current

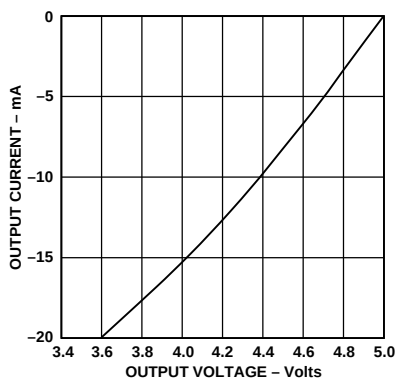


Figure 12. Receiver Output High Voltage vs. Output Current

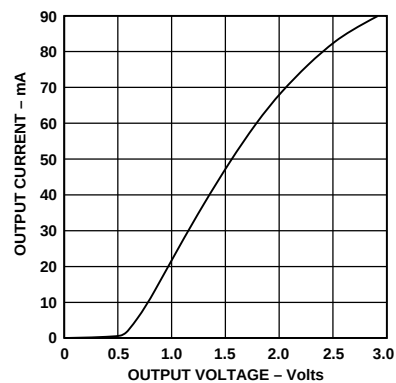


Figure 13. Driver Output Low Voltage vs. Output Current

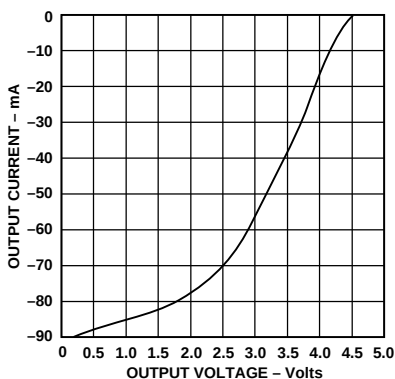


Figure 14. Driver Output High Voltage vs. Output Current

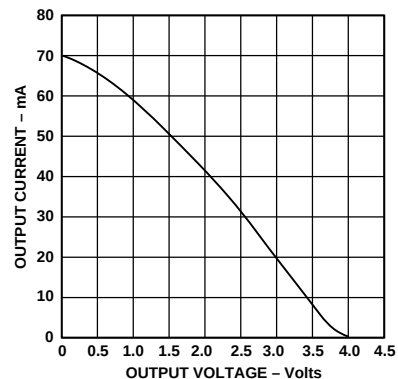


Figure 15. Driver Differential Output Voltage vs. Output Current

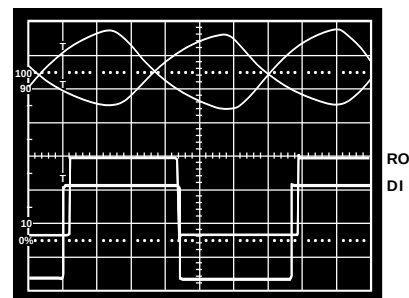


Figure 16. ADM483E Driving 4000 ft. of Cable

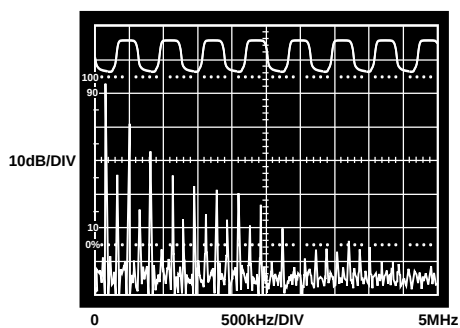


Figure 17. Driver Output Waveform and FFT Plot Transmitting @ 150 kHz

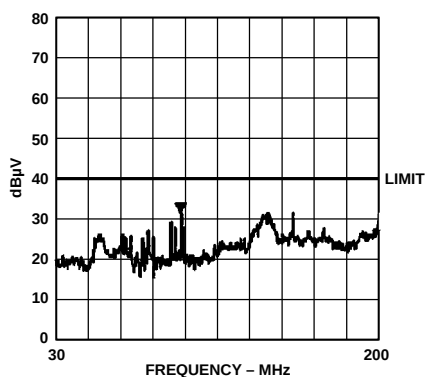


Figure 18. Radiated Emissions

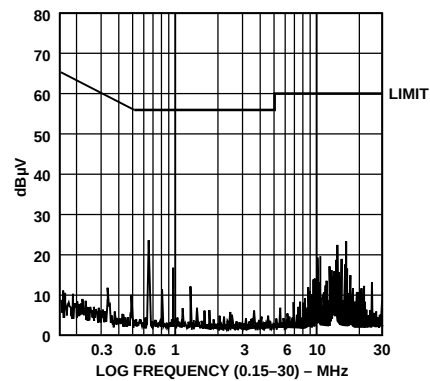


Figure 19. Conducted Emissions

ADM483E

GENERAL INFORMATION

The ADM483E is a ruggedized RS-485 transceiver that operates from a single +5 V supply.

It contains protection against radiated and conducted interference, including high levels of electrostatic discharge.

It is ideally suited for operation in electrically harsh environments or where cables may be plugged/unplugged. It is also immune to high RF field strengths without special shielding precautions. It is intended for balanced data transmission and complies with both EIA Standards RS-485 and RS-422. It contains a differential line driver and a differential line receiver, and is suitable for half duplex data transmission as the driver and receiver share the same differential pins.

The input impedance on the ADM483E is 12 k Ω , allowing up to 32 transceivers on the differential bus.

The ADM483E operates from a single +5 V $\pm$ 10% power supply. Excessive power dissipation caused by bus contention or by output shorting is prevented by a thermal shutdown circuit. This feature forces the driver output into a high impedance state if, during fault conditions, a significant temperature increase is detected in the internal driver circuitry.

The receiver contains a fail-safe feature that results in a logic high output state if the inputs are unconnected (floating).

A high level of robustness is achieved using internal protection circuitry, eliminating the need for external protection components such as tranzorbs or surge suppressors.

Low electromagnetic emissions are achieved using slew limited drivers, minimizing interference both conducted and radiated.

The ADM483 can transmit at data rates up to 250 kbps.

A typical application for the ADM483E is illustrated in Figure 20. This shows a half-duplex link where data may be transferred at rates up to 250 kbps. A terminating resistor is shown at both ends of the link. This termination is not critical since the slew rate is controlled by the ADM483E and reflections are minimized.

The communications network may be extended to include multipoint connections as shown in Figure 30. Up to 32 transceivers may be connected to the bus.

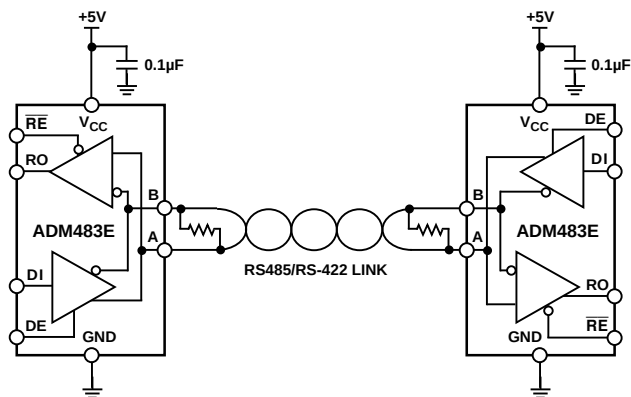


Figure 20. Typical Half-Duplex Link Application

Tables II and III show the truth tables for transmitting and receiving.

Table II. Transmitting Truth Table

Inputs			Outputs	
$\overline{RE}$	DE	DI	B	A
X	1	1	0	1
X	1	0	1	0
0	0	X	Hi-Z	Hi-Z
1	0	X	Hi-Z	Hi-Z

X = Don't Care.

Table III. Receiving Truth Table

Inputs		Outputs	
$\overline{RE}$	DE	A-B	RO
0	0	$\geq +0.2$ V	1
0	0	≤ -0.2 V	0
0	0	Inputs O/C	1
1	0	X	Hi-Z

X = Don't Care.

ESD/EFT TRANSIENT PROTECTION SCHEME

The ADM483E uses protective clamping structures on its inputs and outputs that clamp the voltage to a safe level and dissipates the energy present in ESD (Electrostatic) and EFT (Electrical Fast Transients) discharges.

The protection structure achieves ESD protection up to ± 15 kV according to the Human Body Model, and EFT protection up to ± 2 kV on all I-O lines.

ESD TESTING

Two coupling methods are used for ESD testing, contact discharge and air-gap discharge. Contact discharge calls for a direct connection to the unit being tested. Air-gap discharge uses a higher test voltage but does not make direct contact with the unit under test. With air discharge, the discharge gun is moved toward the unit under test, developing an arc across the air gap, hence the term air-discharge. This method is influenced by humidity, temperature, barometric pressure, distance and rate of closure of the discharge gun. The contact-discharge method, while less realistic, is more repeatable and is gaining acceptance and preference over the air-gap method.

Although very little energy is contained within an ESD pulse, the extremely fast rise time, coupled with high voltages, can cause failures in unprotected semiconductors. Catastrophic destruction can occur immediately as a result of arcing or heating. Even if catastrophic failure does not occur immediately, the device may suffer from parametric degradation, which may result in degraded performance. The cumulative effects of continuous exposure can eventually lead to complete failure.

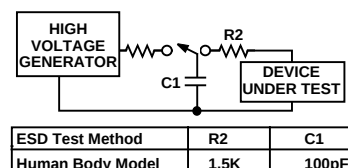


Figure 21. ESD Generator

I-O lines are particularly vulnerable to ESD damage. Simply touching or plugging in an I-O cable can result in a static discharge that can damage or completely destroy the interface product connected to the I-O port.

It is, therefore, extremely important to have high levels of ESD protection on the I-O lines.

It is possible that the ESD discharge could induce latchup in the device under test. It is therefore important that ESD testing on the I-O pins be carried out while device power is applied. This type of testing is more representative of a real world I-O discharge where the equipment is operating normally when the discharge occurs.

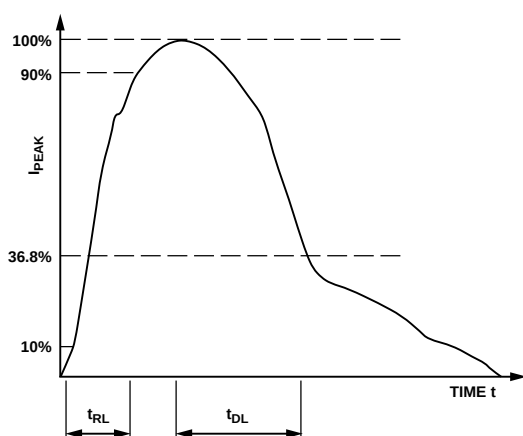


Figure 22. Human Body Model ESD Current Waveform

Table IV. ADM483E ESD Test Results

ESD Test Method	I-O Pins	Other Pins
Human Body Model: Air	±15 kV	
Human Body Model: Contact	±8 kV	±3.5 V

FAST TRANSIENT BURST IMMUNITY (IEC1000-4-4)

IEC1000-4-4 (previously 801-4) covers electrical fast-transient/burst (EFT) immunity. Electrical fast transients occur as a result of arcing contacts in switches and relays. The tests simulate the interference generated when, for example, a power relay disconnects an inductive load. A spark is generated due to the well known back EMF effect. In fact, the spark consists of a burst of sparks as the relay contacts separate. The voltage appearing on the line, therefore, consists of a burst of extremely fast transient impulses. A similar effect occurs when switching on fluorescent lights.

The fast transient burst test, defined in IEC1000-4-4, simulates this arcing and its waveform is illustrated in Figure 23. It consists of a burst of 2.5 kHz to 5 kHz transients repeating at 300 ms intervals. It is specified for both power and data lines.

Four severity levels are defined in terms of an open-circuit voltage as a function of installation environment. The installation environments are defined as

1. Well-protected
2. Protected
3. Typical Industrial
4. Severe Industrial

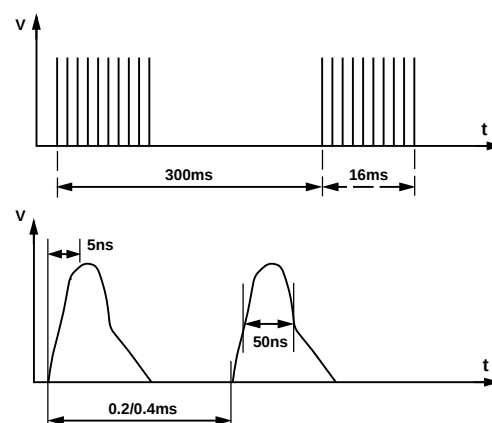


Figure 23. IEC1000-4-4 Fast Transient Waveform

Table V shows the peak voltages for each of the environments.

Table V.

Level	V _{PEAK} (kV) PSU	V _{PEAK} (kV) I-O
1	0.5	0.25
2	1	0.5
3	2	1
4	4	2

A simplified circuit diagram of the actual EFT generator is illustrated in Figure 24.

These transients are coupled onto the signal lines using an EFT coupling clamp. The clamp is 1 m long and completely surrounds the cable, providing maximum coupling capacitance (50 pF to 200 pF typ) between the clamp and the cable. High energy transients are capacitively coupled onto the signal lines. Fast rise times (5 ns) as specified by the standard result in very effective coupling. This test is very severe since high voltages are coupled onto the signal lines. The repetitive transients can often cause problems, where single pulses do not. Destructive latchup may be induced due to the high energy content of the transients. Note that this stress is applied while the interface products are powered up and are transmitting data. The EFT test applies hundreds of pulses with higher energy than ESD. Worst case transient current on an I-O line can be as high as 40 A.

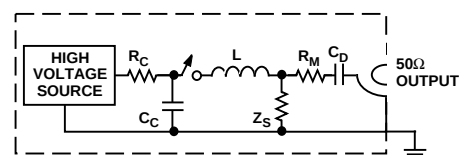


Figure 24. EFT Generator

Test results are classified according to the following

1. Normal performance within specification limits.
2. Temporary degradation or loss of performance that is self-recoverable.
3. Temporary degradation or loss of function or performance that requires operator intervention or system reset.
4. Degradation or loss of function that is not recoverable due to damage.

ADM483E

The ADM483E has been tested under worst case conditions using unshielded cables, and meets Classification 2 at severity Level 4. Data transmission during the transient condition is corrupted, but it may be resumed immediately following the EFT event without user intervention.

RADIATED IMMUNITY (IEC1000-4-3)

IEC1000-4-3 (previously IEC801-3) describes the measurement method and defines the levels of immunity to radiated electromagnetic fields. It was originally intended to simulate the electromagnetic fields generated by portable radio transceivers or any other device that generates continuous wave radiated electromagnetic energy. Its scope has since been broadened to include spurious EM energy, which can be radiated from fluorescent lights, thyristor drives, inductive loads, etc.

Testing for immunity involves irradiating the device with an EM field. There are various methods of achieving this including use of anechoic chamber, stripline cell, TEM cell and GTEM cell. These consist essentially of two parallel plates with an electric field developed between them. The device under test is placed between the plates and exposed to the electric field. There are three severity levels having field strengths ranging from 1 V to 10 V/m. Results are classified in a similar fashion to those for IEC1000-4-2.

1. Normal Operation.
2. Temporary Degradation or loss of function that is self-recoverable when the interfering signal is removed.
3. Temporary degradation or loss of function that requires operator intervention or system reset when the interfering signal is removed.
4. Degradation or loss of function that is not recoverable due to damage.

The ADM483E comfortably meets Classification 1 at the most stringent (Level 3) requirement. In fact, field strengths up to 30 V/m showed no performance degradation and error-free data transmission continued even during irradiation.

Table VI.

Level V/m	Field Strength
1	1
2	3
3	10

EMI EMISSIONS

The ADM483E contains internal slew rate limiting in order to minimize the level of electromagnetic interference generated. Figure 25 shows an FFT plot when transmitting a 150 kHz data stream.

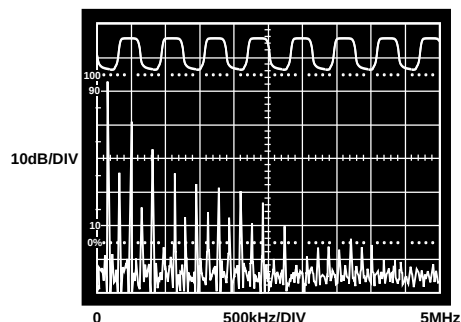


Figure 25. Driver Output Waveform and FFT Plot Transmitting @ 150 kHz

As may be seen, the slew limiting attenuates the high frequency components. EMI is therefore reduced, as are reflections due to improperly terminated cables.

EN55022, CISPR22 defines the permitted limits of radiated and conducted interference from Information Technology Equipment (ITE).

The objective is to control the level of emissions, both conducted and radiated.

For ease of measurement and analysis, conducted emissions are assumed to predominate below 30 MHz, while radiated emissions predominate above this frequency.

CONDUCTED EMISSIONS

This is a measure of noise that is conducted onto the mains power supply. The noise is measured using a LISN (Line Impedance Stabilizing Network) and a spectrum analyzer. The test setup is illustrated in Figure 26. The spectrum analyzer is set to scan the spectrum from 0 MHz to 30 MHz. Figure 27 shows that the level of conducted emissions from the ADM483E are well below the allowable limits.

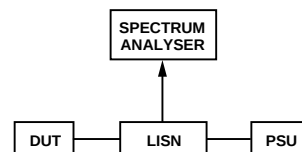


Figure 26. Conducted Emissions Test Setup

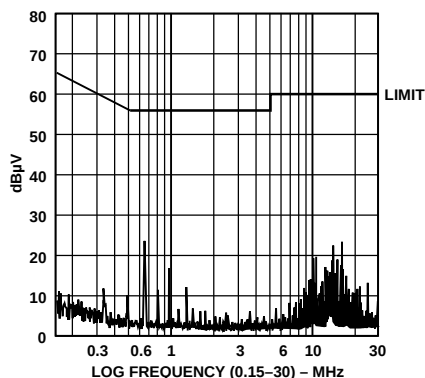


Figure 27. Conducted Emissions

RADIATED EMISSIONS

Radiated emissions are measured at frequencies in excess of 30 MHz.

A typical test setup for monitoring radiated emissions is illustrated in Figure 28.

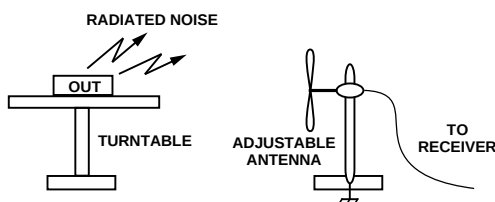


Figure 28. Radiated Emissions Test Setup

Figure 29 shows that the level of radiated emissions is also well below the allowable limit.

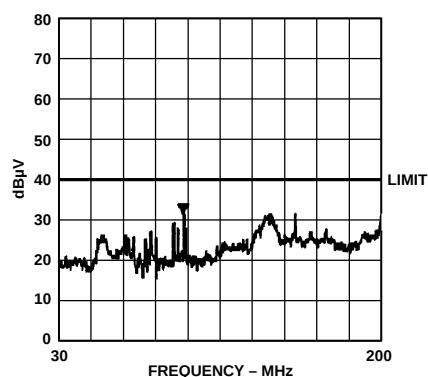


Figure 29. Radiated Emissions

APPLICATIONS INFORMATION

Differential Data Transmission

Differential data transmission is used to reliably transmit data at high rates over long distances and through noisy environments. Differential transmission nullifies the effects of ground shifts and noise signals that appear as common-mode voltages on the line. There are two main standards approved by the Electronics Industries Association (EIA) that specify the electrical characteristics of transceivers used in differential data transmission.

The RS-422 standard specifies data rates up to 10 MBaud and line lengths up to 4000 ft. A single driver can drive a transmission line with up to 10 receivers.

In order to cater for true multipoint communications, the RS-485 standard was defined. This standard meets or exceeds all the requirements of RS-422, but also allows for up to 32 drivers and 32 receivers to be connected to a single bus. An extended common-mode range of -7 V to $+12$ V is defined. The most significant difference between RS-422 and RS-485 is the fact that the drivers may be disabled, thereby allowing more than one (32 in fact) to be connected to a single line. Only one driver should be enabled at a time, but the RS-485 standard contains additional specifications to guarantee device safety in the event of line contention.

Cable and Data Rate

The transmission line of choice for RS-485 communications is a twisted pair. Twisted pair cable tends to cancel common-mode noise and also causes cancellation of the magnetic fields generated by the current flowing through each wire, thereby reducing the effective inductance of the pair.

A typical application showing a multipoint transmission network is illustrated in Figure 30. An RS-485 transmission line can have as many as 32 transceivers on the bus. Only one driver can transmit at a particular time, but multiple receivers may be enabled simultaneously.

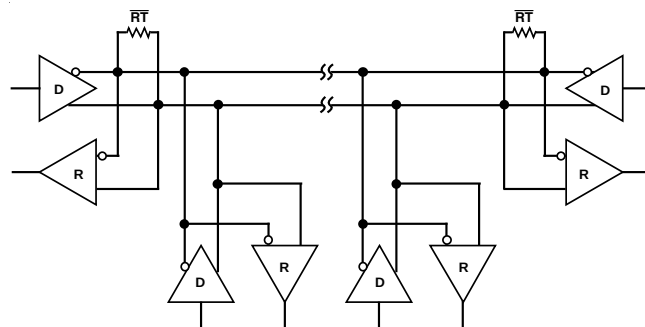
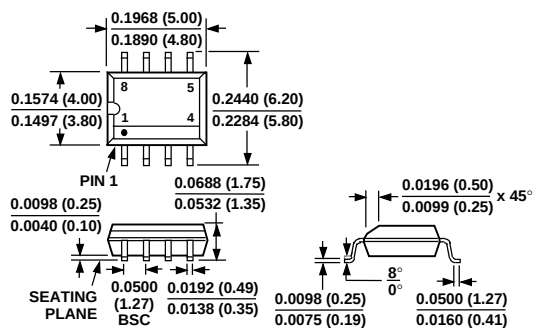


Figure 30. Typical RS-485 Network

OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

8-Lead SOIC (SO-8)



8-Pin Plastic DIP (N-8)

