

# Micropower High Efficiency DC/DC Converters with Low-Battery Detector Adjustable and Fixed 5V

## FEATURES

- **5V at 200mA from a 2V Input**
- Supply Voltage As Low As 1.8V
- Up to 88% Efficiency
- **120 $\mu$ A Quiescent Current**
- **Low-Battery Detector**
- Low  $V_{CESAT}$  Switch: 170mV at 1A Typ
- Uses Inexpensive Surface Mount Inductors
- 8-Lead PDIP or SO Package

## APPLICATIONS

- EL Panel Drivers
- 2-Cell and 3-Cell to 5V Conversion
- Palmtop Computers
- Portable Instruments
- Bar-Code Scanners
- PDAs
- Wireless Systems

## DESCRIPTION

The LT<sup>®</sup>1303/LT1303-5 are micropower step-up high efficiency DC/DC converters using Burst Mode<sup>™</sup> operation. They are ideal for use in small, low-voltage battery-operated systems. The LT1303-5 accepts an input voltage between 1.8V and 5V and converts it to a regulated 5V. The LT1303 is an adjustable version that can supply an output voltage up to 25V. Quiescent current is only 120 $\mu$ A from the battery and the shutdown pin further reduces current to 10 $\mu$ A. The low-battery detector provides an open-collector output that goes low when the input voltage drops below a preset level. The on-chip NPN power switch has a low 170mV saturation voltage at a switch current of 1A. The LT1303/LT1303-5 are available in 8-lead PDIP or SO packages, easing board space requirements.

For higher output current, please see the LT1305 or LT1302.

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 Burst Mode is a trademark of Linear Technology Corporation.

## TYPICAL APPLICATION

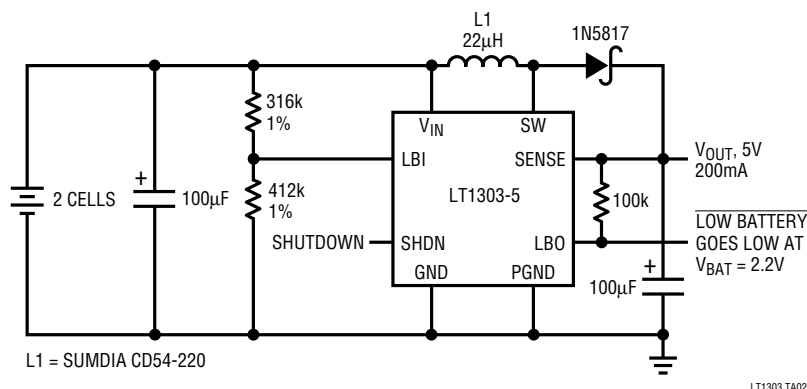
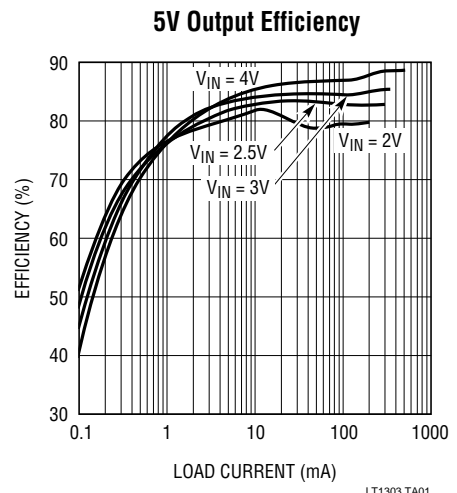


Figure 1. 2-Cell to 5V DC/DC Converter with Low-Battery Detect



## ABSOLUTE MAXIMUM RATINGS

$V_{IN}$ Voltage .....	10V
SW1 Voltage .....	25V
Sense Voltage (LT1303-5) .....	20V
FB Voltage (LT1303) .....	10V
Shutdown Voltage .....	10V
LBO Voltage .....	10V
LBI Voltage .....	10V
Maximum Power Dissipation .....	500mW
Operating Temperature Range .....	0°C to 70°C
Storage Temperature Range .....	-65°C to 150°C
Lead Temperature (Soldering, 10 sec) .....	300°C

## PACKAGE/ORDER INFORMATION

<div><p>TOP VIEW</p><p>N8 PACKAGE 8-LEAD PDIP</p><p>S8 PACKAGE 8-LEAD PLASTIC SO</p><p>*FIXED VERSION</p><p><math>T_{JMAX} = 100^{\circ}\text{C}</math>, <math>\theta_{JA} = 130^{\circ}\text{C/W}</math> (N8) <math>T_{JMAX} = 100^{\circ}\text{C}</math>, <math>\theta_{JA} = 150^{\circ}\text{C/W}</math> (S8)</p></div>	ORDER PART NUMBER	
	LT1303CN8 LT1303CS8 LT1303CN8-5 LT1303CS8-5	
	S8 PART MARKING	
	1303 13035	

Consult factory for Industrial and Military grade parts.

## ELECTRICAL CHARACTERISTICS $T_A = 25^{\circ}\text{C}$ , $V_{IN} = 2.0\text{V}$ , unless otherwise noted.

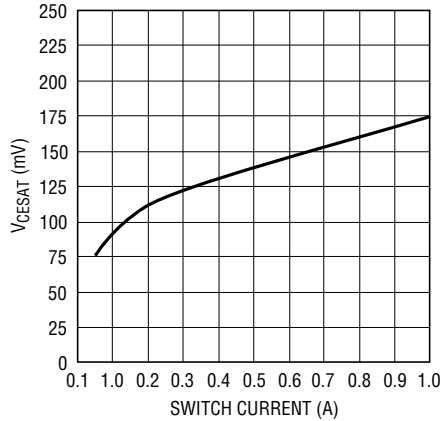
SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
$I_Q$	Quiescent Current	$V_{SHDN} = 0.5\text{V}$ , $V_{SEL} = 5\text{V}$ , $V_{SENSE} = 5.5\text{V}$ $V_{SHDN} = 1.8\text{V}$	●	120	200	$\mu\text{A}$
			●	7	15	$\mu\text{A}$
$V_{IN}$	Input Voltage Range		●	1.8	1.55	V
			●	2.0		V
	Feedback Voltage	LT1303	●	1.22	1.24	V
	Output Sense Voltage	LT1303-5	●	4.8	5.0	V
	Comparator Hysteresis	LT1303 (Note 1)	●	6	12.5	mV
	Output Hysteresis	LT1303-5 (Note 1)	●	22	50	mV
	Feedback Pin Bias Current	LT1303, $V_{FB} = 1\text{V}$	●	7	20	nA
	Oscillator Frequency	Current Limit Not Asserted		120	155	kHz
	Oscillator TC			0.2		%/ $^{\circ}\text{C}$
DC	Maximum Duty Cycle		●	75	86	%
$t_{ON}$	Switch On Time	Current Limit Not Asserted		5.6		$\mu\text{s}$
	Output Line Regulation	$1.8\text{V} < V_{IN} < 6\text{V}$	●	0.06	0.15	%/V
$V_{CESAT}$	Switch Saturation Voltage	$I_{SW} = 700\text{mA}$	●	130	200	mV
	Switch Leakage Current	$V_{SW} = 5\text{V}$ , Switch Off	●	0.1	10	$\mu\text{A}$
	Peak Switch Current	$V_{IN} = 2\text{V}$ $V_{IN} = 5\text{V}$	●	0.75 0.65	1.0 0.9	A
	LBI Trip Voltage		●	1.21	1.24	V
	LBI Input Bias Current	$V_{LBI} = 1\text{V}$	●	7	20	nA
	LBO Output Low	$I_{LOAD} = 100\mu\text{A}$	●	0.11	0.4	V
	LBO Leakage Current	$V_{LBI} = 1.3\text{V}$ , $V_{LBO} = 5\text{V}$	●	0.1	5	$\mu\text{A}$
$V_{SHDNH}$	Shutdown Pin High		●	1.8		V
$V_{SHDNL}$	Shutdown Pin Low				0.5	V
$I_{SHDN}$	Shutdown Pin Bias Current	$V_{SHDN} = 5\text{V}$ $V_{SHDN} = 2\text{V}$ $V_{SHDN} = 0\text{V}$	● ● ●	8.0 3.0 0.1	20	$\mu\text{A}$ $\mu\text{A}$ $\mu\text{A}$

The ● denotes specifications which apply over the 0°C to 70°C operating temperature range.

**Note 1:** Hysteresis specified is DC. Output ripple may be higher if output capacitance is insufficient or capacitor ESR is excessive.

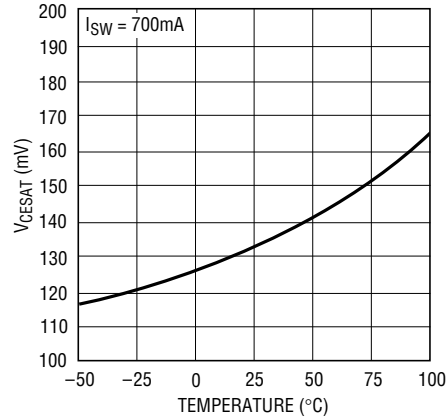
# TYPICAL PERFORMANCE CHARACTERISTICS

**V<sub>CESAT</sub> vs Switch Current**



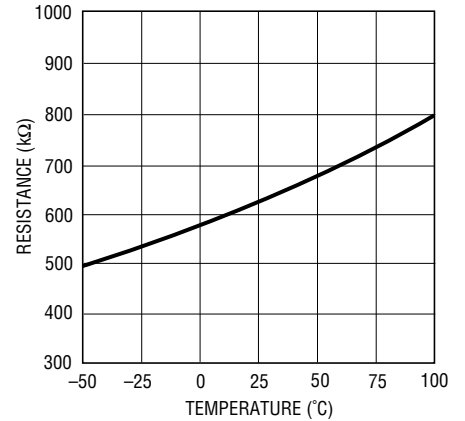
LT1303 G01

**V<sub>CESAT</sub> vs Temperature**



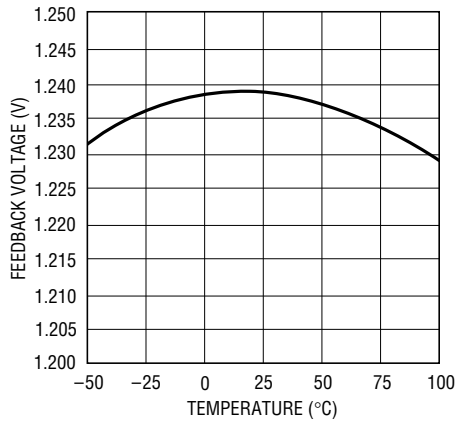
LT1303 G02

**LT1303-5 Sense Pin Resistance to Ground**



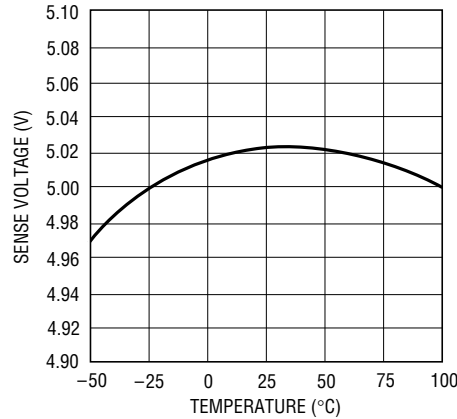
LT1303 G03

**LT1303 FB Voltage**



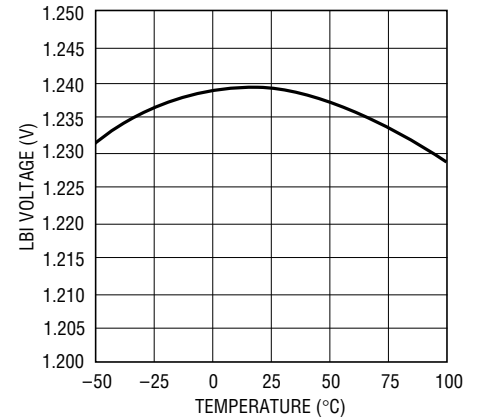
LT1303 G04

**LT1303-5 Sense Voltage**



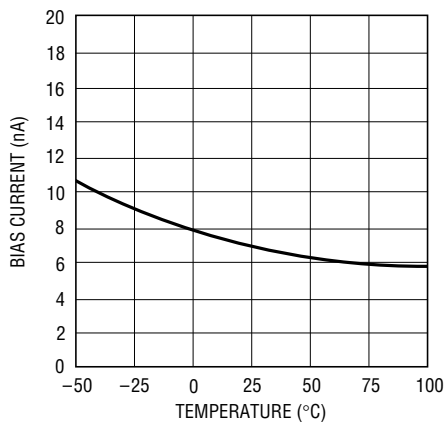
LT1303 G05

**Low Battery Detect Trip Point**



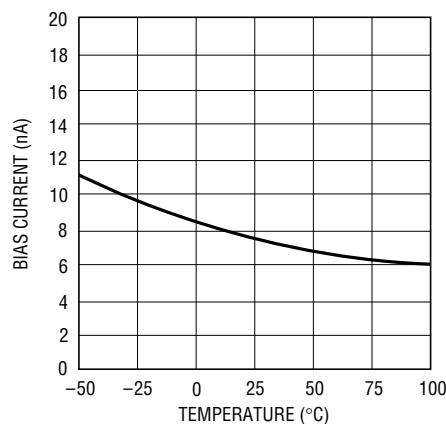
LT1303 G06

**LBI Pin Bias Current**



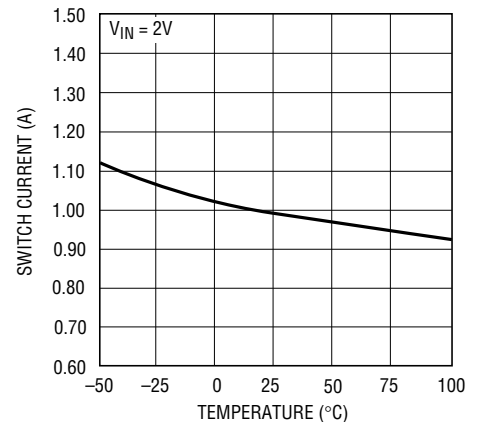
LT1303 G07

**FB Pin Bias Current**



LT1303 G08

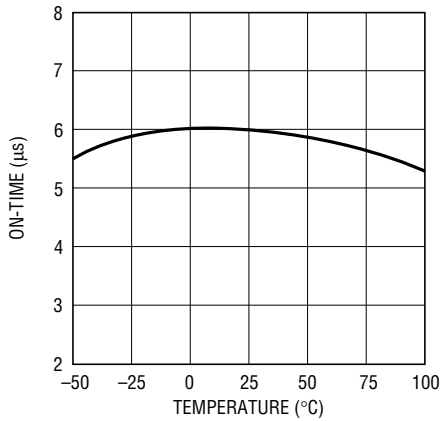
**Switch Current Limit**



LT1303 G09

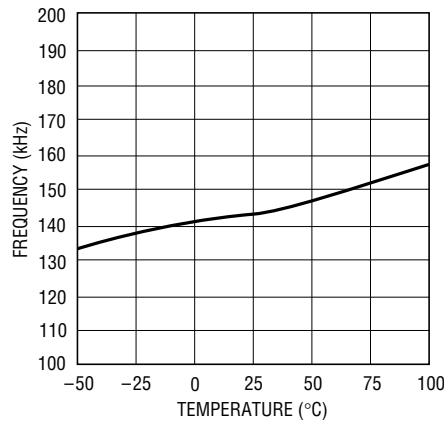
# TYPICAL PERFORMANCE CHARACTERISTICS

Switch On-Time



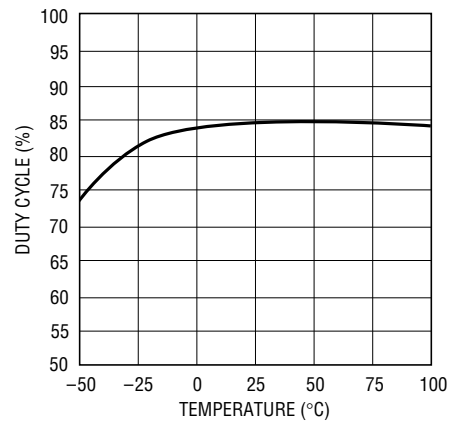
LT1303 G10

Oscillator Frequency



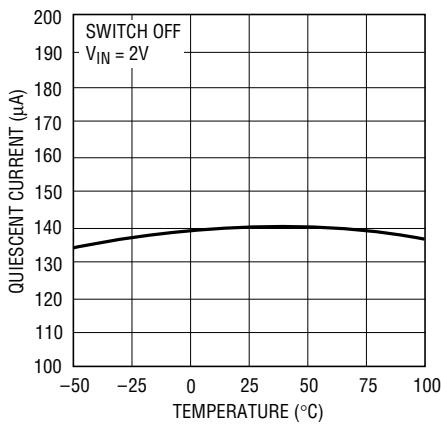
LT1303 G11

Maximum Duty Cycle



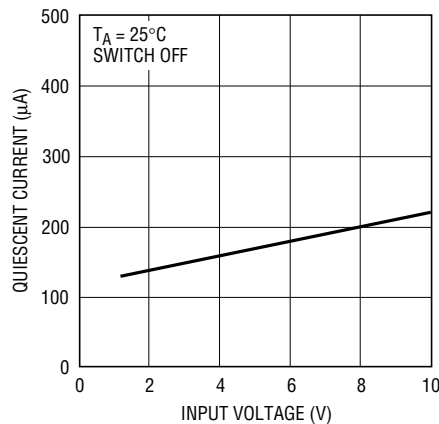
LT1303 G12

Quiescent Current



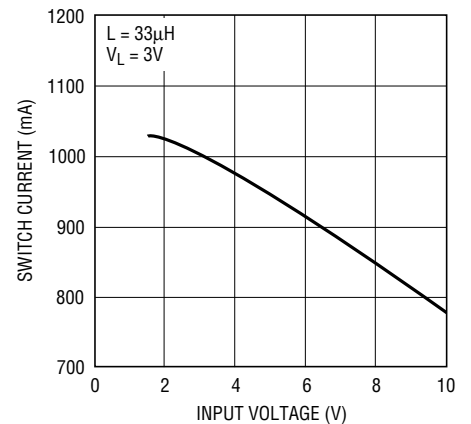
LT1303 G13

Quiescent Current



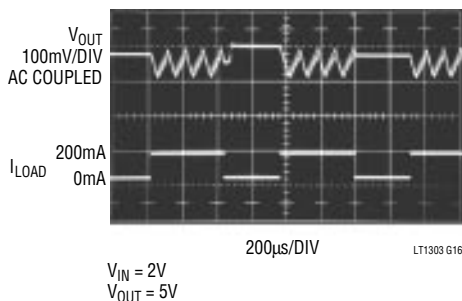
LT1303 G14

Switch Current Limit



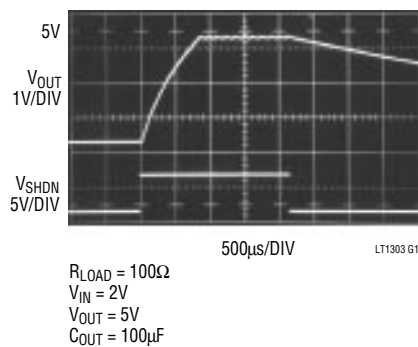
LT1303 G15

Transient Response  
Figure 1 Circuit



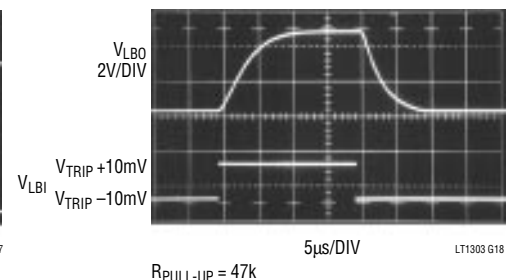
LT1303 G16

Shutdown Pin Response



LT1303 G17

Low Battery Detector Transient Response



LT1303 G18

## PIN FUNCTIONS

**GND (Pin 1):** Signal Ground. Tie to PGND under the package.

**LBO (Pin 2):** Open-Collector Output of Low-Battery Comparator. Can sink 100 $\mu$ A. Disabled when device is in shutdown.

**SHDN (Pin 3):** Shutdown. Pull high to shut down the device. Ground for normal operation.

**FB/Sense (Pin 4):** On 1303 (adjustable) this pin connects to the main comparator C1 input. On LT1303-5 this pin connects to the resistor string that sets output voltage at 5V.

**LB1 (Pin 5):** Low-Battery Comparator Input. When voltage on this pin below 1.24V, LBO is low.

**V<sub>IN</sub> (Pin 6):** Supply Pin. Must be bypassed with a large value electrolytic to ground. Keep bypass within 0.2" of the device.

**SW (Pin 7):** Switch Pin. Connect inductor and diode here. Keep layout short and direct to minimize radio frequency interference.

**PGND (Pin 8):** Power ground. Tie to signal ground (pin1) under the package. Bypass capacitor from V<sub>IN</sub> should be tied directly to PGND within 0.2" of the device.

## BLOCK DIAGRAMS

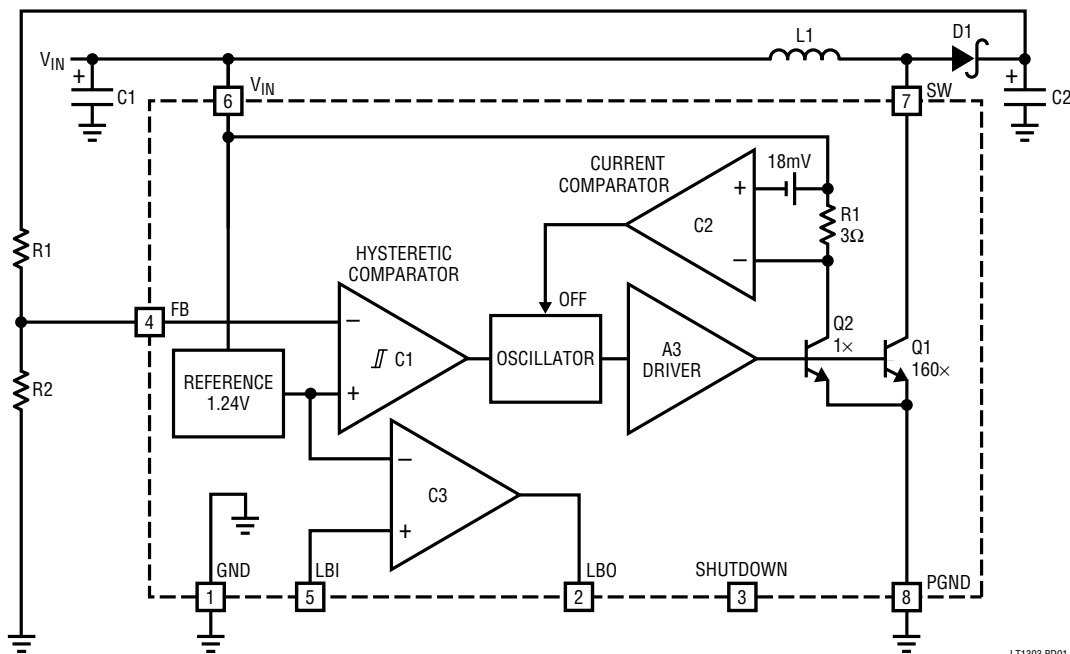


Figure 2. LT1303 Block Diagram

# BLOCK DIAGRAMS

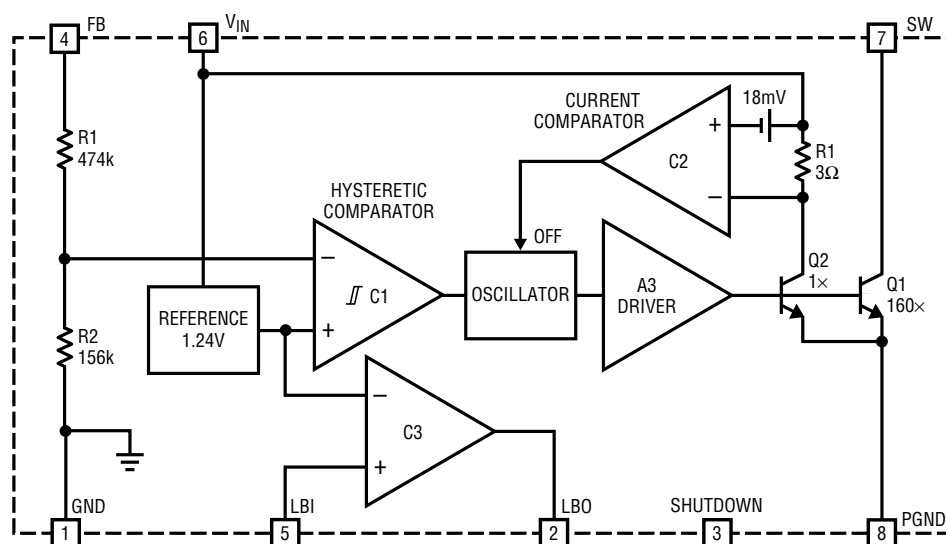


Figure 3. LT1303-5 Block Diagram

LT1303 B002

# OPERATION

Operation of the LT1303 is best understood by referring to the Block Diagram in Figure 2. When C1's negative input, related to the output voltage by the appropriate resistor-divider ratio, is higher than the 1.24V reference voltage, C1's output is low. C2, A3 and the oscillator are turned off, drawing no current. Only the reference and C1 consume current, typically 140 $\mu$ A. When C1's negative input drops below 1.24V and overcomes C1's 6mV hysteresis, C1's output goes high, enabling the oscillator, current comparator C2 and driver A3. Quiescent current increases to 2mA as the device goes into active switching mode. Q1 then turns on in controlled saturation for nominally 6 $\mu$ s or until current comparator C2 trips, whichever comes first. The switch then turns off for approximately 1.5 $\mu$ s, then turns on again. The LT1303's switching causes current to alternately build up in L1 and dump into output capacitor C4 via D1, increasing the output voltage. When the output is high enough to cause C1's output to go high, switching action ceases. Capacitor C4 is left to supply current to the load until  $V_{OUT}$  decreases enough to force C1's output high, and the entire cycle repeats. Figure 4 details relevant waveforms. C1's cycling causes low-to-mid-frequency ripple voltage on the output. Ripple can be reduced by making the

output capacitor large. The 100 $\mu$ F unit specified results in ripple of 50mV to 100mV on the 5V output. A 220 $\mu$ F capacitor will decrease ripple by approximately 50%.

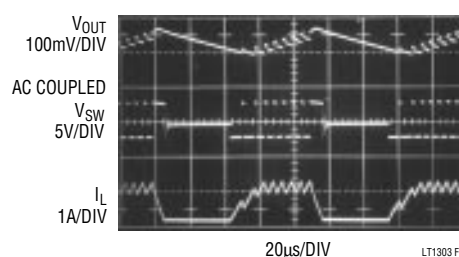


Figure 4. Burst Mode Operation in Action

If switch current reaches 1A, causing C2 to trip, switch on-time is reduced and off-time increases slightly. This allows continuous operation during bursts. C2 monitors the voltage across 3 $\Omega$  resistor R1 which is directly related to the switch current. Q2's collector current is set by the emitter-area ratio to 0.6% of Q1's collector current. When R1's voltage drop exceeds 18mV, corresponding to 1A switch current, C2's output goes high, truncating the on-time portion of the oscillator cycle and increasing off-time

## OPERATION

to about 2μs. Response time of C2, which determines minimum on-time, is approximately 300ns.

### Low Battery Detector

The low battery detector is enabled when SHDN is low and disabled when SHDN is high. The comparator has no

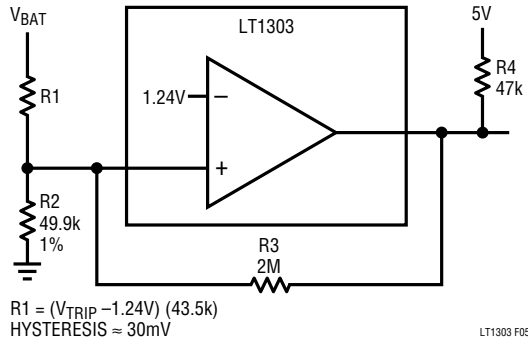


Figure 5. R3 Adds Hysteresis to Low-Battery Detector

hysteresis built in, but hysteresis can be added by connecting a high-value resistor from LBI to LBO as shown in Figure 5. The internal reference can be accessed via the comparator as shown in Figure 6.

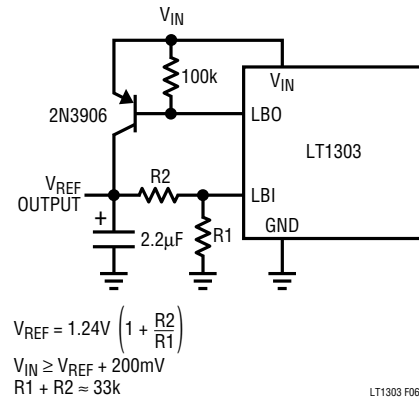


Figure 6. Accessing Internal Reference

## APPLICATIONS INFORMATION

### Inductor Section

Inductors suitable for use with the LT1303 usually fall in the 5μH to 50μH range. The inductor must: (1) handle current of 1.25A without saturating, (2) have enough inductance to provide a di/dt lower than 400mA/μs, and (3) have low enough DC resistance to avoid excessive heating or efficiency losses. Higher value inductors will deliver more power but tend to be physically larger. Most ferrite core drum or rod inductors such as those specified in Table 1 are suitable for use. It is acceptable to bias open-flux inductors (e.g. Sumida CD54) into saturation by 10 to 20% without adverse effects.

Table 1. Recommended Inductors

VENDOR	SERIES	APPROPRIATE VALUES	PHONE NUMBERS
Coilcraft	D03316 D01608	10μH to 47μH 10μH	(708) 639-6400
Coiltronics	OCTAPAK CTX20-1 CTX20-2 CTX33-4	20μH 20μH 33μH	(407) 241-7876
Sumida	CD54	10μH to 33μH	(708) 956-0666
Gowanda	GA10	10μH to 33μH	(716) 532-2234

Figure 7 shows inductor current of a suitable inductor, di/dt is controlled at all times. The rapid rise in current shown in Figure 8 results from this inductor saturating at approximately 1A. Saturation occurs when the inductor cannot hold any more magnetic energy in the core. Current then increases rapidly, limited only by the resistance of the winding. Figure 9's inductor has high DC resistance which results in the exponential time constant shape of the inductor current.

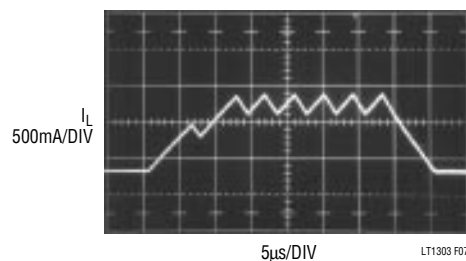
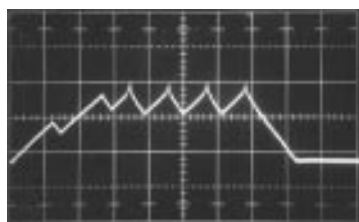
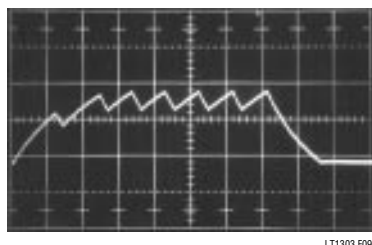


Figure 7. Properly Chosen Inductor Does Not Saturate

## APPLICATIONS INFORMATION



**Figure 8. This Inductor Saturates at  $I_L \approx 1A$ . A Poor Choice**



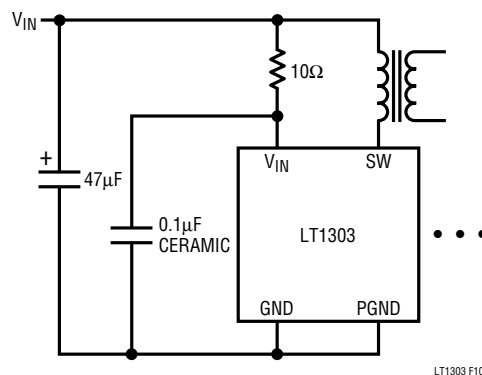
**Figure 9. Slight Exponential Shape to Inductor Current Waveform Indicates Excessive DC Resistance**

### Diode Selection

The LT1303's high switching speed demands a high speed rectifier. Schottky diodes are preferred for their low forward drop and fast recovery. Suitable choices include the 1N5817, MBR5120LT3, and MBR0520LT1. Do not use signal diodes such as 1N4148. They cannot carry 1A current. Also avoid "general-purpose" diodes such as 1N4001. These are far too slow and are unsuitable for any switching regulator application. For high temperature applications a silicon diode such as the MUR105 will have less leakage.

### Capacitor Selection

Input and output capacitors should have low ESR for best efficiency. Recommended capacitors include AVX TPS series, Sprague 595D series, and Sanyo OS-CON. The output capacitor's ESR determines the high frequency ripple amplitude. A 100 $\mu$ F capacitor is the minimum recommended for a 5V output. Higher output voltages can use lower capacitance values. For example, a 12V output can use a 33 $\mu$ F or 47 $\mu$ F capacitor. The  $V_{IN}$  pin of the LT1303 should be decoupled with a 47 $\mu$ F or 100 $\mu$ F capacitor at the pin. When driving a transformer, an additional decoupling network of 10 $\Omega$  and 0.1 $\mu$ F ceramic is recommended as shown in Figure 10.



**Figure 10. 10 $\Omega$ -1 $\mu$ F Network to LT1303  $V_{IN}$  Pin Provides Additional Decoupling. Recommended When Driving Transformers.**

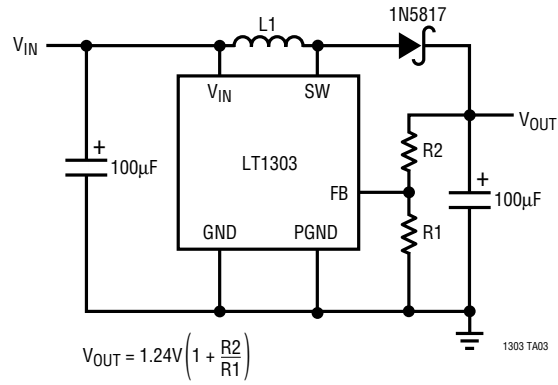
**Table 2. Recommended Capacitors**

VENDOR	SERIES	TYPE	PHONE NUMBERS
AVX	TPS	Surface Mount	(803) 448-9411
Sanyo	OS-CON	Through-Hole	(619) 661-6835
Panasonic	HFQ	Through-Hole	(201) 348-5200
Sprague	595D	Surface Mount	(603) 224-1961

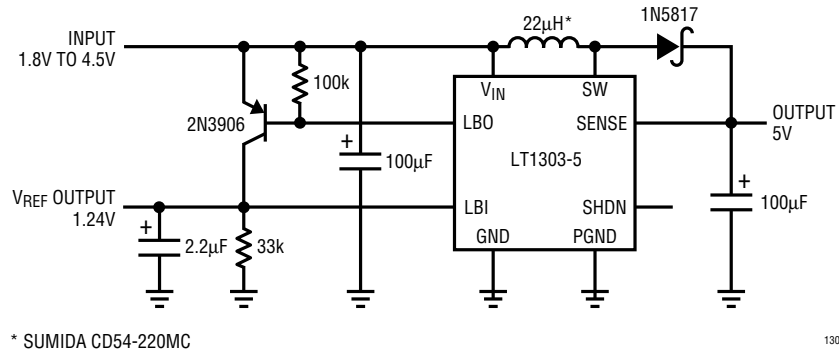


## TYPICAL APPLICATIONS

### Setting Output Voltage on LT1303

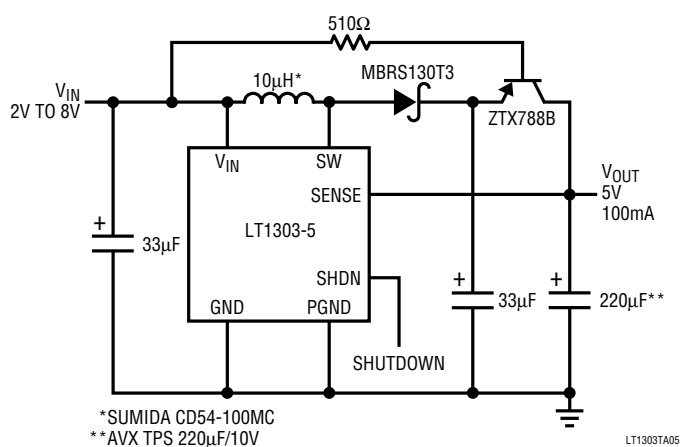


### 5V Step-Up Converter with Reference Output

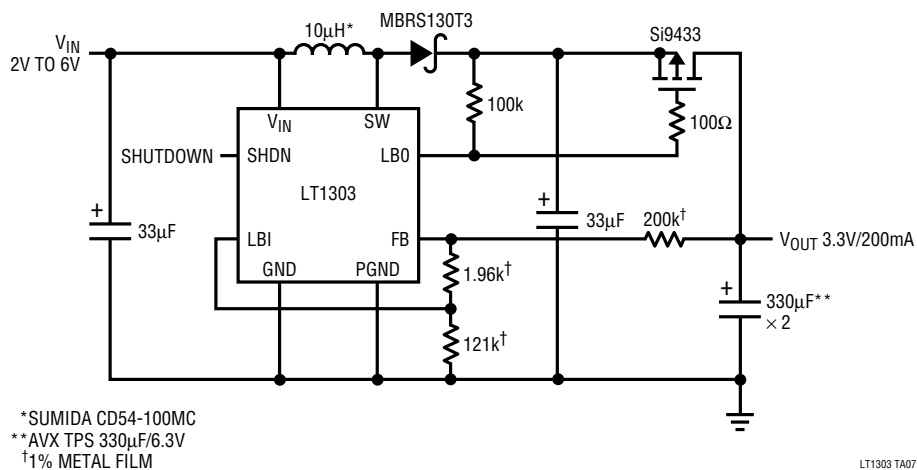


## TYPICAL APPLICATIONS

### 4-, 5-Cell to 5V Converter with Output Disconnect



### 3-Cell to 3.3V Boost/Linear Converter with Output Disconnect



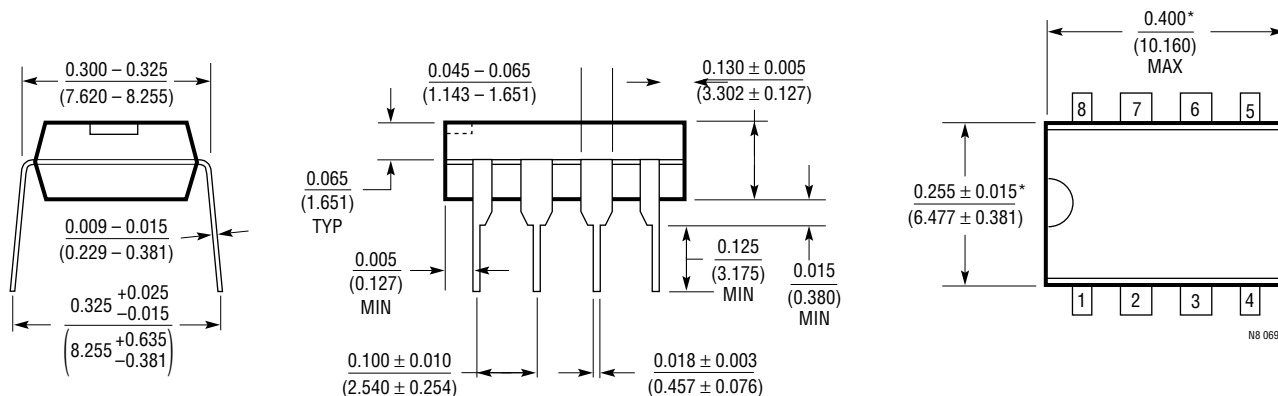
[illegible]

<sup>†</sup>R1 ADJUSTS  $V_{OUT}$  83V<sub>RMS</sub> TO 115V<sub>RMS</sub>



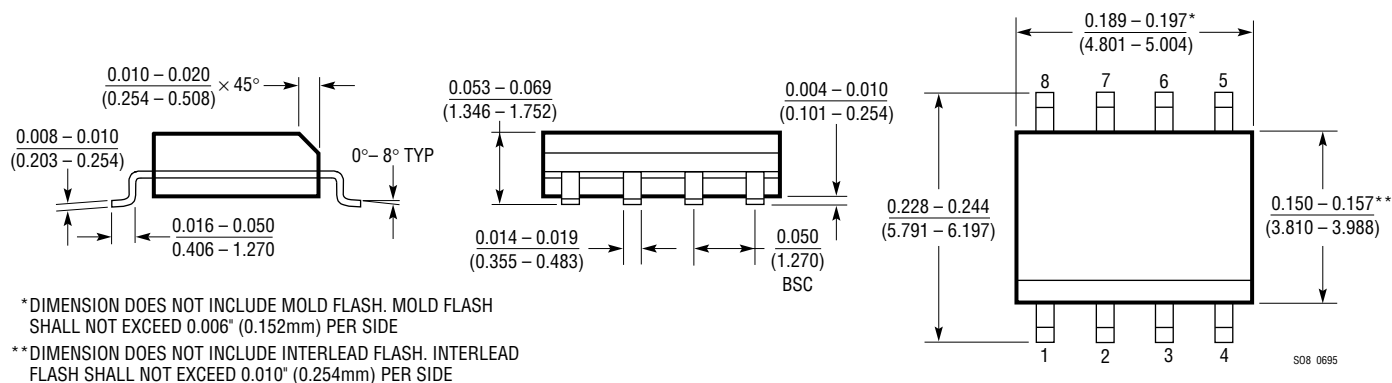
## PACKAGE DESCRIPTION Dimensions in inches (millimeters) unless otherwise noted.

### N8 Package 8-Lead Plastic DIP



\*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.  
MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

### S8 Package 8-Lead Plastic SOIC



\*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

\*\*DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

## RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1129	Micropower Low Dropout Regulator	700mA Output Current in SO-8 Package
LT1182/83/84	LCD and CCFL Backlight Controller	High Efficiency and Excellent Backlight Control Range
LT1301	5V to 12V/200mA Step-Up DC/DC Converter	120 $\mu$ A Quiescent Current
LT1302	2-Cell to 5V/600mA Step-Up DC/DC Converter	200 $\mu$ A Quiescent Current
LT1305	Micropower 2A Switch DC/DC Converter with Low-Battery Detect	2V to 5V at 400mA
LT1372	500kHz Step-Up PWM, 1.5A Switch	Low Noise, Fixed Frequency Operation
LTC <sup>®</sup> 1472	PCMCIA Host Switch with Protection	Includes Current Limit and Thermal Shutdown